PIC16(L)F1717/8/9 microcontrollers combine Intelligent Analog integration with low cost and extreme low power (XLP) to suit a variety of general purpose applications. These 28-pin and 40-pin devices deliver on-chip op amps, Core Independent Peripherals (CLC, NCO and COG), Peripheral Pin Select and Zero-Cross Detect, providing for increased design flexibility.

**Core Features:**
- C Compiler Optimized RISC Architecture
- Only 49 Instructions
- Operating Speed:
  - 0-32 MHz clock input
  - 125 ns minimum instruction cycle
- Interrupt Capability
- 16-Level Deep Hardware Stack
- Up to Four 8-Bit Timers
- One 16-Bit Timer
- Power-on Reset (POR)
- Power-up Timer (PWRT)
- Low-Power Brown-out Reset (LPBOR)
- Programmable Watchdog Timer (WDT) up to 256s
- Programmable Code Protection

**Memory:**
- Up to 16 Kwords Flash Program Memory
- Up to 2048 Bytes Data SRAM Memory
- Direct, Indirect and Relative Addressing modes
- High-Endurance Flash (HEF):
  - 128B of nonvolatile data storage
  - 100K Erase/Write cycles

**Operating Characteristics:**
- Operating Voltage Range:
  - 1.8V to 3.6V (PIC16LF1717/8/9)
  - 2.3V to 5.5V (PIC16F1717/8/9)
- Temperature Range:
  - Industrial: -40°C to 85°C
  - Extended: -40°C to 125°C

**eXtreme Low-Power (XLP) Features:**
- Sleep mode: 50 nA @ 1.8V, typical
- Watchdog Timer: 500 nA @ 1.8V, typical
- Secondary Oscillator: 500 nA @ 32 kHz
- Operating Current:
  - 8 uA @ 32 kHz, 1.8V, typical
  - 32 uA/MHz @ 1.8V, typical

**Digital Peripherals:**
- Configurable Logic Cell (CLC):
  - Integrated combinational and sequential logic
- Complementary Output Generator (COG):
  - Rising/falling edge dead-band control/blanking
- Numerically Controlled Oscillator (NCO):
  - Generates true linear frequency control and increased frequency resolution
  - Input Clock: 0 Hz < F NCO < 32 MHz
  - Resolution: F NCO/220
- Capture/Compare/PWM (CCP) module
- PWM: Two 10-Bit Pulse-Width Modulators
- Serial Communications:
  - SPI, I2C, RS-232, RS-485, LIN compatible
  - Auto-Baud Detect, auto-wake-up on start
- Up to 35 I/O Pins and One Input Pin:
  - Individually programmable pull-ups
  - Slew rate control
  - Interrupt-on-Change with edge-select
- Peripheral Pin Select (PPS):
  - Enables pin mapping of digital I/O

**Intelligent Analog Peripherals:**
- Operational Amplifiers:
  - Two configurable rail-to-rail op amps
  - Selectable internal and external channels
  - 2 MHz gain bandwidth product
- High-Speed Comparators:
  - Up to two comparators
  - 50 ns response time
  - Rail-to-rail inputs
- 10-Bit Analog-to-Digital Converter (ADC):
  - Up to 28 external channels
  - Conversion available during Sleep
  - Temperature indicator
- Zero-Cross Detector (ZCD):
  - Detect when AC signal on pin crosses ground
- 8-Bit Digital-to-Analog Converter (DAC):
  - Output available externally
  - Internal connections to comparators, op amps, Fixed Voltage Reference (FVR) and ADC
  - Internal Voltage Reference module
Clocking Structure:
- 16 MHz Internal Oscillator Block:
  - ±1% at calibration
  - Selectable frequency range from 0 to 32 MHz
- 31 kHz Low-Power Internal Oscillator
- External Oscillator Block with:
  - Three crystal/resonator modes up to 20 MHz
  - Two external clock modes up to 20 MHz
- Fail-Safe Clock Monitor
- Two-Speed Oscillator Start-up
- Oscillator Start-up Timer (OST)

Programming/Debug Features:
- In-Circuit Debug Integrated On-Chip
- Emulation Header for Advanced Debug:
  - Provides trace, background debug and up to 32 hardware break points
- In-Circuit Serial Programming™ (ICSP™) via Two Pins

PIC16(L)F171X Family Types

<table>
<thead>
<tr>
<th>Device</th>
<th>Data Sheet Index</th>
<th>Program Memory Flash (words)</th>
<th>Data SRAM (bytes)</th>
<th>High Endurance Flash (bytes)</th>
<th>I/Os(c)</th>
<th>10-bit ADC (ch)</th>
<th>5/6-bit DAC</th>
<th>High-Speed Comparators</th>
<th>Op Amp</th>
<th>Zero Cross</th>
<th>Timers (8/16-bit)</th>
<th>CCP</th>
<th>PWM</th>
<th>COG</th>
<th>EUSART</th>
<th>MSSP (I2C/SPI)</th>
<th>CLC</th>
<th>NCO</th>
<th>PPS</th>
<th>Debug(1)</th>
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Note 1: Debugging Methods: (I) – Integrated on Chip; (H) – using Debug Header; E – using Emulation Header.

Note 2: One pin is input-only.

Data Sheet Index: (Unshaded devices are described in this document.)
1: DS40001726  PIC16(L)F1713/6 Data Sheet, 28-Pin Flash, 8-bit Microcontrollers.
2: DS40001740  PIC16(L)F1717/8/9 Data Sheet, 28/40-Pin Flash, 8-bit Microcontrollers.

Note: For other small form-factor package availability and marking information, please visit [http://www.microchip.com/packaging](http://www.microchip.com/packaging) or contact your local sales office.
Pin Diagrams

FIGURE 1: 28-PIN SPDIP, SOIC, SSOP

```
VPP/MCLR/RE3  1  28  RB7
RA0    2  27  RB6
RA1    3  26  RB5
RA2    4  25  RB4
RA3    5  24  RB3
RA4    6  23  RB2
RA5    7  22  RB1
Vss    8  21  RB0
RA7    9  20  Vdd
RA6   10  19  Vss
RC0   11  18  RC7
RC1   12  17  RC6
RC2   13  16  RC5
RC3   14  15  RC4
```

Note: See Table 1 for the pin allocation table.

FIGURE 2: 28-PIN (U)QFN

```
RA0  28  21  RB3
RA1  27  20  RB2
RA0  26  19  RB1
RA5  25  18  RB0
Vss  24  17  Vdd
RA7  23  16  Vss
RA6  22  15  RC7
RA2  21  14  RC6
RA3  20  13  RC5
RA4  19  12  RC4
RA5  18  11  RC3
RC0  17  10  RC2
RC1  16  09  RC1
RC2  15  08  RC0
RC3  14  07  VDD
RC4  13  06  VSS
RC5  12  05  VSS
RC6  11  04  VSS
RC7  10  03  VSS
RA0  09  02  RA0
RA1  08  01  RA1
```

Note: See Table 1 for the pin allocation table.
TABLE 1: 28-PIN ALLOCATION TABLE (PIC16(L)F1718)

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<th>SPDIP/SSOP</th>
<th>SSOP</th>
<th>GFN/LOPN</th>
<th>ADC</th>
<th>Reference</th>
<th>Comparator</th>
<th>Op Amp</th>
<th>DAC</th>
<th>Zero Cross</th>
<th>Timers</th>
<th>CCP</th>
<th>NCO</th>
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Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.

4: Alternate outputs are excluded from solid shaded areas.

5: Alternate inputs are excluded from dot shaded areas.
TABLE 1: 28-PIN ALLOCATION TABLE (PIC16(L)F1718) (CONTINUED)

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</table>

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.

4: Alternate outputs are excluded from solid shaded areas.

5: Alternate inputs are excluded from dot shaded areas.
Note: See Table 2 for the pin allocation table.
FIGURE 5: 44-PIN TQFP (10X10)

Note: See Table 2 for the pin allocation table.
| VO | PDIP | TQFP | UQFN | ADC | Reference | Comparator | Op Amp | DAC | Zero Cross | Timers | CCP | NCO | PWM | COG | MSSP | EUSART | CLC | Interrupt | Pullup | Basic |
| RA0 | 2 | 19 | 17 | AN0 | C1IN0- | C2IN0- |  |  |  |  |  |  |  |  |  |  |  |  |  | CLCIN0(1) |
| RA1 | 3 | 20 | 18 | AN1 | C1IN1- | C2IN1- | OPA1OUT |  |  |  |  |  |  |  |  |  |  |  |  | CLCIN1(1) |
| RA2 | 4 | 21 | 19 | AN2 | VREF- | C1IN0+ | C2IN0+ | DAC1OUT1 |  |  |  |  |  |  |  |  |  |  |  |  | |
| RA3 | 5 | 22 | 20 | AN3 | VREF+ | C1IN1+ |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RA4 | 6 | 23 | 21 | OPA1IN+ |  |  | T0CKI(1) |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RA5 | 7 | 24 | 22 | AN4 | OPA1IN- | DAC2OUT1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RA6 | 14 | 31 | 29 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RA7 | 13 | 30 | 28 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB0 | 33 | 8 | 8 | AN12 | C2IN1+ |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB1 | 34 | 9 | 9 | AN10 | C1IN3- | C2IN3- | OPA2OUT |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB2 | 35 | 10 | 10 | AN8 |  |  | OPA2IN- |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB3 | 36 | 11 | 11 | AN9 | C1IN2- | C2IN2- | OPA2IN+ |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB4 | 37 | 14 | 12 | AN11 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB5 | 38 | 15 | 13 | AN13 |  |  | T1(1) |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB6 | 39 | 16 | 14 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RB7 | 40 | 17 | 15 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RC0 | 15 | 32 | 30 |  |  |  | T1CKI(1) |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RC1 | 16 | 35 | 31 |  |  |  | SOSCI |  |  |  |  |  |  |  | CCP2(1) |  |  |  |  |  | |
| RC2 | 17 | 36 | 32 | AN14 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | |
| RC3 | 18 | 37 | 33 | AN15 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | SCL/SCK(1) |

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.

4: Alternate outputs are excluded from solid shaded areas.

5: Alternate inputs are excluded from dot shaded areas.
# Table 2: 40/44-Pin Allocation Table (PIC16(L)F1717/9) (Continued)

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<th>COG</th>
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**Note:**
1. Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2. All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
4. Alternate outputs are excluded from solid shaded areas.
5. Alternate inputs are excluded from dot shaded areas.
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1.0 DEVICE OVERVIEW

The PIC16(L)F1717/8/9 devices are described within this data sheet. They are available in the following package configurations:

- 28-pin SPDIP, SSOP, SOIC, QFN and UQFN
- 40-pin PDIP and UQFN
- 44-pin TQFP

Figure 1-1 and Figure 1-2 show block diagrams of the PIC16(L)F1717/8/9 devices. Table 1-2 shows the pinout descriptions.

Reference Table 1-1 for peripherals available per device.

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<td>Timer4</td>
<td>●●●</td>
<td>●●●</td>
<td>●●●</td>
</tr>
<tr>
<td>Timer6</td>
<td>●●●</td>
<td>●●●</td>
<td>●●●</td>
</tr>
</tbody>
</table>
1.1 Register and Bit Naming Conventions

1.1.1 REGISTER NAMES
When there are multiple instances of the same peripheral in a device, the peripheral control registers will be depicted as the concatenation of a peripheral identifier, peripheral instance and control identifier. The control registers section will show just one instance of all the register names with an ‘x’ in the place of the peripheral instance number. This naming convention may also be applied to peripherals when there is only one instance of that peripheral in the device to maintain compatibility with other devices in the family that contain more than one.

1.1.2 BIT NAMES
There are two variants for bit names:
• Short name: Bit function abbreviation
• Long name: Peripheral abbreviation + short name

1.1.2.1 Short Bit Names
Short bit names are an abbreviation for the bit function. For example, some peripherals are enabled with the EN bit. The bit names shown in the registers are the short name variant.

Short bit names are useful when accessing bits in C programs. The general format for accessing bits by the short name is `RegisterNameBits.ShortName`. For example, the enable bit, EN, in the COG1CON0 register can be set in C programs with the instruction `COG1CON0bits.EN = 1`.

Short names are generally not useful in assembly programs because the same name may be used by different peripherals in different bit positions. When this occurs, during the include file generation, all instances of that short bit name are appended with an underscore plus the name of the register in which the bit resides to avoid naming contentions.

1.1.2.2 Long Bit Names
Long bit names are constructed by adding a peripheral abbreviation prefix to the short name. The prefix is unique to the peripheral thereby making every long bit name unique. The long bit name for the COG1 enable bit is the COG1 prefix, G1, appended with the enable bit short name, EN, resulting in the unique bit name G1EN.

Long bit names are useful in both C and assembly programs. For example, in C the COG1CON0 enable bit can be set with the `G1EN = 1` instruction. In assembly, this bit can be set with the `BSF COG1CON0,G1EN` instruction.

1.1.3 REGISTER AND BIT NAMING EXCEPTIONS

1.1.3.1 Status, Interrupt and Mirror Bits
Status, interrupt enables, interrupt flags and mirror bits are contained in registers that span more than one peripheral. In these cases, the bit name shown is unique, so there is no prefix or short name variant.

1.1.3.2 Legacy Peripherals
There are some peripherals that do not strictly adhere to these naming conventions. Peripherals that have existed for many years and are present in almost every device are the exceptions. These exceptions were necessary to limit the adverse impact of the new conventions on legacy code. Peripherals that do adhere to the new convention will include a table in the registers section indicating the long name prefix for each peripheral instance. Peripherals that fall into the exception category will not have this table. These peripherals include, but are not limited to, the following:
• EUSART
• MSSP

1.1.2.3 Bit Fields
Bit fields are two or more adjacent bits in the same register. Bit fields adhere only to the short bit naming convention. For example, the three Least Significant bits of the COG1CON0 register contain the mode control bits. The short name for this field is MD. There is no long bit name variant. Bit field access is only possible in C programs. The following example demonstrates a C program instruction for setting the COG1 to the Push-Pull mode:

```c
COG1CON0bits.MD = 0x5;
```

Individual bits in a bit field can also be accessed with long and short bit names. Each bit is the field name appended with the number of the bit position within the field. For example, the Most Significant bit has the short bit name MD2, and the long bit name is G1MD2. The following two examples demonstrate assembly program sequences for setting the COG1 to Push-Pull mode:

Example 1:
```
MOVLW ~(1<<G1MD1)
ANDWF COG1CON0,F
MOVLW 1<<G1MD2 | 1<<G1MD0
IORWF COG1CON0,F
```

Example 2:
```
BSF COG1CON0,G1MD2
BCF COG1CON0,G1MD1
BSF COG1CON0,G1MD0
```
FIGURE 1-1: PIC16(L)F1718 BLOCK DIAGRAM

Note 1: See applicable chapters for more information on peripherals.
Note 1: See applicable chapters for more information on peripherals.
<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA0/AN0/C1IN0-/C2IN0-/CLCIN0(1)</td>
<td>RA0</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN0</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 0 input.</td>
</tr>
<tr>
<td></td>
<td>C1IN0-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C2 negative input.</td>
</tr>
<tr>
<td></td>
<td>C2IN0-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C3 negative input.</td>
</tr>
<tr>
<td></td>
<td>CLCIN0</td>
<td>TTL/ST</td>
<td>—</td>
<td>Configurable Logic Cell source input.</td>
</tr>
<tr>
<td>RA1/AN1/C1IN1-/C2IN1-/OPA1OUT/CLCIN1(1)</td>
<td>RA1</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN1</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 1 input.</td>
</tr>
<tr>
<td></td>
<td>C1IN1-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C1 negative input.</td>
</tr>
<tr>
<td></td>
<td>C2IN1-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C2 negative input.</td>
</tr>
<tr>
<td></td>
<td>OPA1OUT</td>
<td>—</td>
<td>AN</td>
<td>Operational Amplifier 1 output.</td>
</tr>
<tr>
<td></td>
<td>CLCIN1</td>
<td>TTL/ST</td>
<td>—</td>
<td>Configurable Logic Cell source input.</td>
</tr>
<tr>
<td>RA2/AN2/VREF-/C1IN0+/C2IN0+/DAC1OUT1</td>
<td>RA2</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN2</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 2 input.</td>
</tr>
<tr>
<td></td>
<td>VREF-</td>
<td>AN</td>
<td>—</td>
<td>ADC Negative Voltage Reference input.</td>
</tr>
<tr>
<td></td>
<td>C1IN0+</td>
<td>AN</td>
<td>—</td>
<td>Comparator C2 positive input.</td>
</tr>
<tr>
<td></td>
<td>C2IN0+</td>
<td>AN</td>
<td>—</td>
<td>Comparator C3 positive input.</td>
</tr>
<tr>
<td></td>
<td>DAC1OUT1</td>
<td>—</td>
<td>AN</td>
<td>Digital-to-Analog Converter output.</td>
</tr>
<tr>
<td>RA3/AN3/VREF+/C1IN1+</td>
<td>RA3</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN3</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 3 input.</td>
</tr>
<tr>
<td></td>
<td>VREF+</td>
<td>AN</td>
<td>—</td>
<td>ADC Voltage Reference input.</td>
</tr>
<tr>
<td></td>
<td>C1IN1+</td>
<td>AN</td>
<td>—</td>
<td>Comparator C1 positive input.</td>
</tr>
<tr>
<td>RA4/OPA1IN+/T0CKI(1)</td>
<td>RA4</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>OPA1IN+</td>
<td>AN</td>
<td>—</td>
<td>Operational Amplifier 1 non-inverting input.</td>
</tr>
<tr>
<td></td>
<td>T0CKI</td>
<td>TTL/ST</td>
<td>—</td>
<td>Timer0 gate input.</td>
</tr>
<tr>
<td>RA5/AN4/OPA1IN-/DAC2OUT1/SS(1)</td>
<td>RA5</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN4</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 4 input.</td>
</tr>
<tr>
<td></td>
<td>OPA1IN-</td>
<td>AN</td>
<td>—</td>
<td>Operational Amplifier 1 inverting input.</td>
</tr>
<tr>
<td></td>
<td>DAC2OUT1</td>
<td>—</td>
<td>AN</td>
<td>Digital-to-Analog Converter output.</td>
</tr>
<tr>
<td></td>
<td>SS</td>
<td>TTL/ST</td>
<td>—</td>
<td>Slave Select enable input.</td>
</tr>
<tr>
<td>RA6/OSC2/CLKOUT</td>
<td>RA6</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>OSC2</td>
<td>—</td>
<td>XTAL</td>
<td>Crystal/Resonator (LP, XT, HS modes).</td>
</tr>
<tr>
<td></td>
<td>CLKOUT</td>
<td>—</td>
<td>CMOS</td>
<td>Fosc/4 output.</td>
</tr>
<tr>
<td>RA7/OSC1/CLKIN</td>
<td>RA7</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>OSC1</td>
<td>—</td>
<td>XTAL</td>
<td>Crystal/Resonator (LP, XT, HS modes).</td>
</tr>
<tr>
<td></td>
<td>CLKIN</td>
<td>ST</td>
<td>—</td>
<td>External clock input (EC mode).</td>
</tr>
<tr>
<td>RB0/AN12/C2IN1+/ZCD/COGIN(1)</td>
<td>RB0</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN12</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 12 input.</td>
</tr>
<tr>
<td></td>
<td>C2IN1+</td>
<td>AN</td>
<td>—</td>
<td>Comparator C2 positive input.</td>
</tr>
<tr>
<td></td>
<td>ZCD</td>
<td>AN</td>
<td>—</td>
<td>Zero-Cross Detection Current Source/Sink.</td>
</tr>
<tr>
<td></td>
<td>COGIN</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>Complementary Output Generator input.</td>
</tr>
</tbody>
</table>

**Legend:**
- **AN** = Analog input or output
- **TTL** = TTL compatible input or output
- **CMOS** = CMOS compatible input or output
- **OD** = Open-Drain
- **ST** = Schmitt Trigger input with CMOS levels
- **HV** = High Voltage
- **XTAL** = Crystal levels

**Note:**
1. Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2. All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
TABLE 1-2: PIC16(L)F1718 PINOUT DESCRIPTION (CONTINUED)

<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RB1/AN10/C1IN3-/C2IN3-/OPA2OUT</td>
<td>RB1</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN10</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 10 input.</td>
<td></td>
</tr>
<tr>
<td>C1IN3-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C1 negative input.</td>
<td></td>
</tr>
<tr>
<td>C2IN3-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C2 negative input.</td>
<td></td>
</tr>
<tr>
<td>OPA2OUT</td>
<td>—</td>
<td>AN</td>
<td>Operational Amplifier 2 output.</td>
<td></td>
</tr>
<tr>
<td>RB2/AN8/OPA2IN-</td>
<td>RB2</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN8</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 8 input.</td>
<td></td>
</tr>
<tr>
<td>OPA2IN-</td>
<td>AN</td>
<td>—</td>
<td>Operational Amplifier 2 inverting input.</td>
<td></td>
</tr>
<tr>
<td>RB3/AN9/C1IN2-/C2IN2-/OPA2IN+</td>
<td>RB3</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN9</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 9 input.</td>
<td></td>
</tr>
<tr>
<td>C1IN2-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C1 negative input.</td>
<td></td>
</tr>
<tr>
<td>C2IN2-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C2 negative input.</td>
<td></td>
</tr>
<tr>
<td>OPA2IN+</td>
<td>AN</td>
<td>—</td>
<td>Operational Amplifier 2 non-inverting input.</td>
<td></td>
</tr>
<tr>
<td>RB4/AN11</td>
<td>RB4</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN11</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 11 input.</td>
<td></td>
</tr>
<tr>
<td>RB5/AN13/T1G(1)</td>
<td>RB5</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN13</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 13 input.</td>
<td></td>
</tr>
<tr>
<td>T1G</td>
<td>TTL/ST</td>
<td>—</td>
<td>Timer 1 gate input.</td>
<td></td>
</tr>
<tr>
<td>RB6/CLCIN2(1)/ICSPCLK</td>
<td>RB6</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>CLCIN2</td>
<td>TTL/ST</td>
<td>—</td>
<td>Configurable Logic Cell source input.</td>
<td></td>
</tr>
<tr>
<td>ICSPCLK</td>
<td>ST</td>
<td>—</td>
<td>Serial Programming Clock.</td>
<td></td>
</tr>
<tr>
<td>RB7/DAC1OUT2/DAC2OUT2/CLCIN3(1)/ICSPDAT</td>
<td>RB7</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>DAC1OUT2</td>
<td>—</td>
<td>AN</td>
<td>Digital-to-Analog Converter output.</td>
<td></td>
</tr>
<tr>
<td>DAC2OUT2</td>
<td>—</td>
<td>AN</td>
<td>Digital-to-Analog Converter output.</td>
<td></td>
</tr>
<tr>
<td>CLCIN3</td>
<td>TTL/ST</td>
<td>—</td>
<td>Configurable Logic Cell source input.</td>
<td></td>
</tr>
<tr>
<td>ICSPDAT</td>
<td>ST</td>
<td>CMOS</td>
<td>ICSP™ Data I/O.</td>
<td></td>
</tr>
<tr>
<td>RC0/T1CK(1)/SOSCO</td>
<td>RC0</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>T1CKI</td>
<td>TTL/ST</td>
<td>—</td>
<td>Timer 1 clock input.</td>
<td></td>
</tr>
<tr>
<td>SOSCO</td>
<td>XTAL</td>
<td>XTAL</td>
<td>Secondary Oscillator connection.</td>
<td></td>
</tr>
<tr>
<td>RC1/SOSCI/CCP2(1)</td>
<td>RC1</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>SOSCI</td>
<td>XTAL</td>
<td>XTAL</td>
<td>Secondary Oscillator connection.</td>
<td></td>
</tr>
<tr>
<td>CCP2</td>
<td>TTL/ST</td>
<td>—</td>
<td>Capture input</td>
<td></td>
</tr>
<tr>
<td>RC2/AN14/CCP1(1)</td>
<td>RC2</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN14</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 14 input.</td>
<td></td>
</tr>
<tr>
<td>CCP1</td>
<td>TTL/ST</td>
<td>—</td>
<td>Capture input</td>
<td></td>
</tr>
<tr>
<td>RC3/AN15/SCK(1)/SCL(1)</td>
<td>RC3</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN15</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 15 input.</td>
<td></td>
</tr>
<tr>
<td>SCK</td>
<td>TTL/ST</td>
<td>—</td>
<td>SPI clock input</td>
<td></td>
</tr>
<tr>
<td>SCL</td>
<td>TTL/ST</td>
<td>—</td>
<td>I²C clock input</td>
<td></td>
</tr>
</tbody>
</table>

Legend: AN = Analog input or output  CMOS = CMOS compatible input or output  OD = Open-Drain
TTL = TTL compatible input  ST = Schmitt Trigger input with CMOS levels  I²C = Schmitt Trigger input with I²C
HV = High Voltage  XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RC4/AN16/SDI(1)/SDA(1)</td>
<td>RC4</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN16</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 16 input.</td>
</tr>
<tr>
<td></td>
<td>SDI</td>
<td>TTL/ST</td>
<td>—</td>
<td>SPI Data input.</td>
</tr>
<tr>
<td></td>
<td>SDA</td>
<td>I²C</td>
<td>—</td>
<td>I²C Data input.</td>
</tr>
<tr>
<td>RC5/AN17</td>
<td>RC5</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN17</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 17 input.</td>
</tr>
<tr>
<td>RC6/AN18/CK(4)</td>
<td>RC6</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN18</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 18 input.</td>
</tr>
<tr>
<td></td>
<td>CK</td>
<td>TTL/ST</td>
<td>—</td>
<td>EUSART synchronous clock.</td>
</tr>
<tr>
<td>RC7/AN19/RX(4)</td>
<td>RC7</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>AN19</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 19 input.</td>
</tr>
<tr>
<td></td>
<td>RX</td>
<td>TTL/ST</td>
<td>—</td>
<td>EUSART receive.</td>
</tr>
<tr>
<td>RE3/MCLR/VPP</td>
<td>RE3</td>
<td>TTL/ST</td>
<td>—</td>
<td>General purpose input.</td>
</tr>
<tr>
<td></td>
<td>MCLR</td>
<td>ST</td>
<td>—</td>
<td>Master clear input.</td>
</tr>
<tr>
<td></td>
<td>VPP</td>
<td>HV</td>
<td>—</td>
<td>Programming enable.</td>
</tr>
<tr>
<td>VDD</td>
<td>VDD</td>
<td>Power</td>
<td>—</td>
<td>Positive supply.</td>
</tr>
<tr>
<td>Vss</td>
<td>Vss</td>
<td>Power</td>
<td>—</td>
<td>Ground reference.</td>
</tr>
<tr>
<td>OUT(2)</td>
<td>C1OUT</td>
<td>CMOS</td>
<td></td>
<td>Comparator 1 output.</td>
</tr>
<tr>
<td></td>
<td>C2OUT</td>
<td>CMOS</td>
<td></td>
<td>Comparator 2 output.</td>
</tr>
<tr>
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<td>Compare/PWM1 output.</td>
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<td>CCP2</td>
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<td>Compare/PWM2 output.</td>
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<td>CMOS</td>
<td></td>
<td>Numerically controlled oscillator output.</td>
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<tr>
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<td>PWM3OUT</td>
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<td>PWM3 output.</td>
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</tr>
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<td>PWM4OUT</td>
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<td>Complementary output generator output B.</td>
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<td>COG1C</td>
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<td>Complementary output generator output C.</td>
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<td>COG1D</td>
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<td></td>
<td>Complementary output generator output D.</td>
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<tr>
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<td>SDA(3)</td>
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<td>I²C</td>
<td>I²C Data output.</td>
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<td>SCK</td>
<td>CMOS</td>
<td></td>
<td>SPI clock output.</td>
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<tr>
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<td>SCL(3)</td>
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<td>I²C</td>
<td>I²C clock output.</td>
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<td>SDO</td>
<td>CMOS</td>
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<td>SPI data output.</td>
</tr>
<tr>
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<td>TX/CK</td>
<td>CMOS</td>
<td></td>
<td>EUSART asynchronous TX data/synchronous clock out.</td>
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<tr>
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<td>DT(3)</td>
<td>CMOS</td>
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<td>EUSART synchronous data output.</td>
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<td>Configurable Logic Cell 1 output.</td>
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<td>CMOS</td>
<td></td>
<td>Configurable Logic Cell 2 output.</td>
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<td>CMOS</td>
<td></td>
<td>Configurable Logic Cell 3 output.</td>
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<tr>
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<td>CLC4OUT</td>
<td>CMOS</td>
<td></td>
<td>Configurable Logic Cell 4 output.</td>
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</tbody>
</table>

Legend:  
AN = Analog input or output  
CMOS = CMOS compatible input or output  
OD = Open-Drain  
TTL = TTL compatible input  
ST = Schmitt Trigger input with CMOS levels  
I²C = Schmitt Trigger input with I²C levels  
HV = High Voltage  
XTAL = Crystal levels

Note  
1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.  
2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.  
3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA0/AN0/C1IN0-/C2IN0-/CLCIN0(1)</td>
<td>RA0</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
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<td>AN0</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 0 input.</td>
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<tr>
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<td>AN</td>
<td>—</td>
<td>Comparator C2 negative input.</td>
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<tr>
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<td>C2IN0-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C3 negative input.</td>
</tr>
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<td>TTL/ST</td>
<td>—</td>
<td>Configurable Logic Cell source input.</td>
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<tr>
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<td>CMOS</td>
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<tr>
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<td>CLCIN1</td>
<td>TTL/ST</td>
<td>—</td>
<td>Configurable Logic Cell source input.</td>
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<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
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<td>AN2</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 2 input.</td>
</tr>
<tr>
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<td>VREF-</td>
<td>AN</td>
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<td>ADC Negative Voltage Reference input.</td>
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<td>C1IN0+</td>
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<td>Comparator C2 positive input.</td>
</tr>
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<td>C2IN0+</td>
<td>AN</td>
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<tr>
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<td>DAC1OUT1</td>
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<td>AN</td>
<td>Digital-to-Analog Converter output.</td>
</tr>
<tr>
<td>RA3/AN3/VREF+/C1IN1+</td>
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<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
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<td>AN3</td>
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<td>—</td>
<td>ADC Channel 3 input.</td>
</tr>
<tr>
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<td>VREF+</td>
<td>AN</td>
<td>—</td>
<td>ADC Voltage Reference input.</td>
</tr>
<tr>
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<td>C1IN1+</td>
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<td>Comparator C1 positive input.</td>
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<tr>
<td>RA4/OPA1IN+/T0CKI(1)</td>
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<td>TTL/ST</td>
<td>CMOS</td>
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</tr>
<tr>
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<td>OPA1IN+</td>
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<td>—</td>
<td>Operational Amplifier 1 non-inverting input.</td>
</tr>
<tr>
<td></td>
<td>T0CKI</td>
<td>TTL/ST</td>
<td>—</td>
<td>Timer0 gate input.</td>
</tr>
<tr>
<td>RA5/AN4/OPA1IN-/DAC2OUT1/SS(1)</td>
<td>RA5</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
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<td>AN</td>
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<td>ADC Channel 4 input.</td>
</tr>
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<td>OPA1IN-</td>
<td>AN</td>
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<td>Operational Amplifier 1 inverting input.</td>
</tr>
<tr>
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<td>DAC2OUT1</td>
<td>—</td>
<td>AN</td>
<td>Digital-to-Analog Converter output.</td>
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<tr>
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<td>SS</td>
<td>—</td>
<td>—</td>
<td>Slave Select enable input.</td>
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<tr>
<td>RA6/OSC2/CLKOUT</td>
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<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td></td>
<td>OSC2</td>
<td>—</td>
<td>XTAL</td>
<td>Crystal/Resonator (LP, XT, HS modes).</td>
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<tr>
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<td>CLKOUT</td>
<td>—</td>
<td>CMOS</td>
<td>Fosc/4 output.</td>
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<tr>
<td>RA7/OSC1/CLKIN</td>
<td>RA7</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
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<tr>
<td></td>
<td>OSC1</td>
<td>—</td>
<td>XTAL</td>
<td>Crystal/Resonator (LP, XT, HS modes).</td>
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<tr>
<td></td>
<td>CLKIN</td>
<td>ST</td>
<td>—</td>
<td>External clock input (EC mode).</td>
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<tr>
<td>RB0/AN12/C2IN1+/ZCD/COGIN(1)</td>
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<td>CMOS</td>
<td>General purpose I/O.</td>
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<td>Comparator C2 positive input.</td>
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<tr>
<td></td>
<td>ZCD</td>
<td>AN</td>
<td>—</td>
<td>Zero-Cross Detection Current Source/Sink.</td>
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<tr>
<td></td>
<td>COGIN</td>
<td>TTL/ST</td>
<td>—</td>
<td>Complementary Output Generator input.</td>
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</tbody>
</table>

**Legend:**
- AN = Analog input or output
- CMOS = CMOS compatible input or output
- TTL = TTL compatible input
- ST = Schmitt Trigger input with CMOS levels
- OD = Open-Drain
- HV = High Voltage
- XTAL = Crystal levels
- I^2C = Schmitt Trigger input with I^2C

**Note:**
1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
### PIC16(L)F1717/9 PINOUT DESCRIPTION (CONTINUED)

<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
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</thead>
<tbody>
<tr>
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<td>C2IN3-</td>
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<td>Comparator C2 negative input.</td>
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<td>RB2/AN8/OPA2IN-</td>
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<td>Operational Amplifier 2 inverting input.</td>
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<td>RB3/AN9/C1IN2-/C2IN2-/OPA2IN+</td>
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<td>Comparator C2 negative input.</td>
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<td>General purpose I/O.</td>
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<td>Timer1 gate input.</td>
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<td>TTL/ST CMOS</td>
<td>General purpose I/O.</td>
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<td>ICSPCLK</td>
<td>ST</td>
<td>Serial Programming Clock.</td>
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</tr>
<tr>
<td>RB7/DAC1OUT2/DAC2OUT2/CLCIN3(1)/ICSPDAT</td>
<td>RB7</td>
<td>TTL/ST CMOS</td>
<td>General purpose I/O.</td>
<td></td>
</tr>
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<td>DAC2OUT2</td>
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<td>Digital-to-Analog Converter output.</td>
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</tr>
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<td>ST</td>
<td>Timer1 clock input.</td>
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<td>XTL</td>
<td>XTAL Secondary Oscillator Connection.</td>
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<td>TTL/ST CMOS</td>
<td>General purpose I/O.</td>
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<td>XTL</td>
<td>XTAL Secondary Oscillator Connection.</td>
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<td>TTL/ST</td>
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<td>TTL/ST CMOS</td>
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<td>Capture input.</td>
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</tr>
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<tr>
<td></td>
<td>SCL</td>
<td>I²C OD</td>
<td>I²C clock.</td>
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</tbody>
</table>

**Legend:**
- **AN** = Analog input or output
- **CMOS** = CMOS compatible input or output
- **TTL** = TTL compatible input
- **ST** = Schmitt Trigger input with CMOS levels
- **HV** = High Voltage
- **I²C** = Schmitt Trigger input with I²C levels
- **XTAL** = Crystal levels
- **OD** = Open-Drain

**Note:**
1. Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2. All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
### TABLE 1-3: PIC16(L)F1717/9 PINOUT DESCRIPTION (CONTINUED)

<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RC4/AN16/SDI(1)/SDA(1)</td>
<td>RC4</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
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<td>TTL/ST</td>
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<td>SPI Data input.</td>
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<tr>
<td>SDA(3)</td>
<td>I^2C</td>
<td>—</td>
<td>I^2C Data input.</td>
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</tr>
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<td>TTL/ST</td>
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<td>EUSART synchronous clock.</td>
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<tr>
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<td>TTL/ST</td>
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<td>EUSART receive.</td>
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</tr>
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</tr>
<tr>
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</tr>
<tr>
<td>AN22</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 22 input.</td>
<td></td>
</tr>
<tr>
<td>RD3/AN23</td>
<td>RD3</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN23</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 23 input.</td>
<td></td>
</tr>
<tr>
<td>RD4/AN24</td>
<td>RD4</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN24</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 24 input.</td>
<td></td>
</tr>
<tr>
<td>RD5/AN25</td>
<td>RD5</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN25</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 25 input.</td>
<td></td>
</tr>
<tr>
<td>RD6/AN26</td>
<td>RD6</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN26</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 26 input.</td>
<td></td>
</tr>
<tr>
<td>RD7/AN27</td>
<td>RD7</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN27</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 27 input.</td>
<td></td>
</tr>
<tr>
<td>RE0/AN5</td>
<td>RE0</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN5</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 5 input.</td>
<td></td>
</tr>
<tr>
<td>RE1/AN6</td>
<td>RE1</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN6</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 6 input.</td>
<td></td>
</tr>
<tr>
<td>RE2/AN7</td>
<td>RE2</td>
<td>TTL/ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN7</td>
<td>AN</td>
<td>—</td>
<td>ADC Channel 7 input.</td>
<td></td>
</tr>
<tr>
<td>RE3/MCLR/VPP</td>
<td>RE3</td>
<td>TTL/ST</td>
<td>—</td>
<td>General purpose input.</td>
</tr>
<tr>
<td>MCLR</td>
<td>ST</td>
<td>—</td>
<td>Master clear input.</td>
<td></td>
</tr>
<tr>
<td>VPP</td>
<td>HV</td>
<td>—</td>
<td>Programming voltage.</td>
<td></td>
</tr>
<tr>
<td>VDD</td>
<td>Power</td>
<td>—</td>
<td>Positive supply.</td>
<td></td>
</tr>
<tr>
<td>VSS</td>
<td>Power</td>
<td>—</td>
<td>Ground reference.</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- AN = Analog input or output
- CMOS = CMOS compatible input or output
- TTL = TTL compatible input
- ST = Schmitt Trigger input with CMOS levels
- OD = Open-Drain
- I^2C = Schmitt Trigger input with I^2C levels
- HV = High Voltage
- XTAL = Crystal levels

**Note:**
1. Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2. All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>OUT(2)</td>
<td>C1OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Comparator 1 output.</td>
</tr>
<tr>
<td></td>
<td>C2OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Comparator 2 output.</td>
</tr>
<tr>
<td></td>
<td>CCP1</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Compare/PWM1 output.</td>
</tr>
<tr>
<td></td>
<td>CCP2</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Compare/PWM2 output.</td>
</tr>
<tr>
<td></td>
<td>NCO1OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Numerically controlled oscillator output.</td>
</tr>
<tr>
<td></td>
<td>PWM3OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>PWM3 output.</td>
</tr>
<tr>
<td></td>
<td>PWM4OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>PWM4 output.</td>
</tr>
<tr>
<td></td>
<td>COG1A</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Complementary output generator output A.</td>
</tr>
<tr>
<td></td>
<td>COG1B</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Complementary output generator output B.</td>
</tr>
<tr>
<td></td>
<td>COG1C</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Complementary output generator output C.</td>
</tr>
<tr>
<td></td>
<td>COG1D</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Complementary output generator output D.</td>
</tr>
<tr>
<td></td>
<td>SDA(3)</td>
<td>OD</td>
<td>I^2C</td>
<td>I^2C Data output.</td>
</tr>
<tr>
<td></td>
<td>SCK</td>
<td>CMOS</td>
<td>CMOS</td>
<td>SPI clock output.</td>
</tr>
<tr>
<td></td>
<td>SCL(3)</td>
<td>OD</td>
<td>I^2C</td>
<td>I^2C clock output.</td>
</tr>
<tr>
<td></td>
<td>SDO</td>
<td>CMOS</td>
<td>CMOS</td>
<td>SPI data output.</td>
</tr>
<tr>
<td></td>
<td>TX/CK</td>
<td>CMOS</td>
<td>CMOS</td>
<td>EUSART asynchronous TX data/synchronous clock out.</td>
</tr>
<tr>
<td></td>
<td>DT(3)</td>
<td>CMOS</td>
<td>CMOS</td>
<td>EUSART synchronous data output.</td>
</tr>
<tr>
<td></td>
<td>CLC1OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Configurable Logic Cell 1 output.</td>
</tr>
<tr>
<td></td>
<td>CLC2OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Configurable Logic Cell 2 output.</td>
</tr>
<tr>
<td></td>
<td>CLC3OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Configurable Logic Cell 3 output.</td>
</tr>
<tr>
<td></td>
<td>CLC4OUT</td>
<td>CMOS</td>
<td>CMOS</td>
<td>Configurable Logic Cell 4 output.</td>
</tr>
</tbody>
</table>

**Legend:**
- AN = Analog input or output
- CMOS = CMOS compatible input or output
- OD = Open-Drain
- TTL = TTL compatible input
- ST = Schmitt Trigger input with CMOS levels
- I^2C = Schmitt Trigger input with I^2C levels
- HV = High Voltage
- XTL = Crystal levels

**Note:**
1. Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2. All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.
2.0 ENHANCED MID-RANGE CPU

This family of devices contains an enhanced mid-range 8-bit CPU core. The CPU has 49 instructions. Interrupt capability includes automatic context saving. The hardware stack is 16 levels deep and has Overflow and Underflow Reset capability. Direct, Indirect and Relative addressing modes are available. Two File Select Registers (FSRs) provide the ability to read program and data memory.

- Automatic Interrupt Context Saving
- 16-level Stack with Overflow and Underflow
- File Select Registers
- Instruction Set
2.1 Automatic Interrupt Context Saving

During interrupts, certain registers are automatically saved in shadow registers and restored when returning from the interrupt. This saves stack space and user code. See Section 7.5 “Automatic Context Saving” for more information.

2.2 16-Level Stack with Overflow and Underflow

These devices have a hardware stack memory 15 bits wide and 16 words deep. A Stack Overflow or Underflow will set the appropriate bit (STKOVF or STKUNF) in the PCON register, and if enabled, will cause a software Reset. See Section 3.6 “Stack” for more details.

2.3 File Select Registers

There are two 16-bit File Select Registers (FSR). FSRs can access all file registers and program memory, which allows one Data Pointer for all memory. When an FSR points to program memory, there is one additional instruction cycle in instructions using INDF to allow the data to be fetched. General purpose memory can now also be addressed linearly, providing the ability to access contiguous data larger than 80 bytes. There are also new instructions to support the FSRs. See Section 3.7 “Indirect Addressing” for more details.

2.4 Instruction Set

There are 49 instructions for the enhanced mid-range CPU to support the features of the CPU. See Section 33.0 “Instruction Set Summary” for more details.
3.0 MEMORY ORGANIZATION

These devices contain the following types of memory:

- **Program Memory**
  - Configuration Words
  - Device ID
  - User ID
  - Flash Program Memory
- **Data Memory**
  - Core Registers
  - Special Function Registers
  - General Purpose RAM
  - Common RAM

3.1 Program Memory Organization

The enhanced mid-range core has a 15-bit program counter capable of addressing a 32K x 14 program memory space. Table 3-1 shows the memory sizes implemented for the PIC16(L)F1717/8/9 family. Accessing a location above these boundaries will cause a wrap-around within the implemented memory space. The Reset vector is at 0000h and the interrupt vector is at 0004h (see Figure 3-1).

3.2 High-Endurance Flash

This device has a 128-byte section of high-endurance program Flash memory (PFM) in lieu of data EEPROM. This area is especially well suited for nonvolatile data storage that is expected to be updated frequently over the life of the end product. See Section 10.2 “Flash Program Memory Overview” for more information on writing data to PFM. See Section 3.2.1.2 “Indirect Read with FSR” for more information about using the FSR registers to read byte data stored in PFM.

---

**Note 1:** The method to access Flash memory through the PMCON registers is described in Section 10.0 “Flash Program Memory Control”.

The following features are associated with access and control of program memory and data memory:

- PCL and PCLATH
- Stack
- Indirect Addressing

---

**TABLE 3-1: DEVICE SIZES AND ADDRESSES**

<table>
<thead>
<tr>
<th>Device</th>
<th>Program Memory Space (Words)</th>
<th>Last Program Memory Address</th>
<th>High-Endurance Flash Memory Address Range(1)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC16(L)F1717</td>
<td>8,192</td>
<td>1FFFh</td>
<td>1F80h-1FFFh</td>
</tr>
<tr>
<td>PIC16(L)F1718/9</td>
<td>16,384</td>
<td>3FFFh</td>
<td>3F80h-3FFFh</td>
</tr>
</tbody>
</table>

**Note 1:** High-endurance Flash applies to the low byte of each address in the range.
FIGURE 3-1: PROGRAM MEMORY MAP AND STACK FOR PIC16(L)F1717

FIGURE 3-2: PROGRAM MEMORY MAP AND STACK FOR PIC16(L)F1718/9

CALL, CALLW
RETURN, RETLW
Interrupt, RETFIE

0000h

Reset Vector

Interrupt Vector

0004h
0005h
07FFh
0800h
0FFFh
1000h
1800h
17FFh
1800h
1FFFh
2000h
7FFFh

On-chip Program Memory

Rollover to Page 0

Page 0

Page 1

Page 2

Page 3

Rollover to Page 0

Page 4

Page 5

Page 6

Page 7

Rollover to Page 1

7FFFh

0000h

Stack Level 0

Stack Level 1

Stack Level 15

Stack Level 0

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

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Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1

Stack Level 15

Stack Level 1
3.2.1 READING PROGRAM MEMORY AS DATA

There are two methods of accessing constants in program memory. The first method is to use tables of RETLW instructions. The second method is to set an FSR to point to the program memory.

3.2.1.1 RETLW Instruction

The RETLW instruction can be used to provide access to tables of constants. The recommended way to create such a table is shown in Example 3-1.

**EXAMPLE 3-1: RETLW INSTRUCTION**

```
constants
  BRW ;Add Index in W to ;program counter to
  RETLW DATA0 ;Index0 data
  RETLW DATA1 ;Index1 data
  RETLW DATA2
  RETLW DATA3

my_function
  ;... LOTS OF CODE...
  MOVLW DATA_INDEX
call constants
  ;... THE CONSTANT IS IN W
```

The BRW instruction makes this type of table very simple to implement. If your code must remain portable with previous generations of microcontrollers, then the BRW instruction is not available, so the older table read method must be used.

3.2.1.2 Indirect Read with FSR

The program memory can be accessed as data by setting bit 7 of the FSRxH register and reading the matching INDFx register. The MOVIW instruction will place the lower eight bits of the addressed word in the W register. Writes to the program memory cannot be performed via the INDF registers. Instructions that access the program memory via the FSR require one extra instruction cycle to complete. Example 3-2 demonstrates accessing the program memory via an FSR.

The high directive will set bit<7> if a label points to a location in program memory.

**EXAMPLE 3-2: ACCESSING PROGRAM MEMORY VIA FSR**

```
constants
  DW DATA0 ;First constant
  DW DATA1 ;Second constant
  DW DATA2
  DW DATA3
my_function
  ;... LOTS OF CODE...
  MOVLW DATA_INDEX
  ADDLW LOW constants
  MOVWF FSR1L
  MOVFLW HIGH constants;MSb sets
  INCF FSR1H, f ;yes
  MOVIW 0[FSR1]
  ;THE PROGRAM MEMORY IS IN W
```

3.3 Data Memory Organization

The data memory is partitioned in 32 memory banks with 128 bytes in a bank. Each bank consists of (see Figure 3-3):

- 12 core registers
- 20 Special Function Registers (SFR)
- Up to 80 bytes of General Purpose RAM (GPR)
- 16 bytes of common RAM

The active bank is selected by writing the bank number into the Bank Select Register (BSR). Unimplemented memory will read as '0'. All data memory can be accessed either directly (via instructions that use the file registers) or indirectly via the two File Select Registers (FSR). See Section 3.7 “Indirect Addressing” for more information.

Data memory uses a 12-bit address. The upper five bits of the address define the Bank address and the lower seven bits select the registers/RAM in that bank.

3.3.1 CORE REGISTERS

The core registers contain the registers that directly affect the basic operation. The core registers occupy the first 12 addresses of every data memory bank (addresses x00h/x08h through x08h/x88h). These registers are listed below in Table 3-2. For detailed information, see Table 3-11.
TABLE 3-2: CORE REGISTERS

<table>
<thead>
<tr>
<th>Addresses</th>
<th>BANKx</th>
</tr>
</thead>
<tbody>
<tr>
<td>x00h or x80h</td>
<td>INDF0</td>
</tr>
<tr>
<td>x01h or x81h</td>
<td>INDF1</td>
</tr>
<tr>
<td>x02h or x82h</td>
<td>STATUS</td>
</tr>
<tr>
<td>x03h or x83h</td>
<td>PCL</td>
</tr>
<tr>
<td>x04h or x84h</td>
<td>FSR0L</td>
</tr>
<tr>
<td>x05h or x85h</td>
<td>FSR0H</td>
</tr>
<tr>
<td>x06h or x86h</td>
<td>FSR1L</td>
</tr>
<tr>
<td>x07h or x87h</td>
<td>FSR1H</td>
</tr>
<tr>
<td>x08h or x88h</td>
<td>BSR</td>
</tr>
<tr>
<td>x09h or x89h</td>
<td>WREG</td>
</tr>
<tr>
<td>x0Ah or x8Ah</td>
<td>PCLATH</td>
</tr>
<tr>
<td>x0Bh or x8Bh</td>
<td>INTCON</td>
</tr>
</tbody>
</table>

3.3.1.1 STATUS Register

The STATUS register, shown in Register 3-1, contains:

- The arithmetic status of the ALU
- The Reset status

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS will clear the upper three bits and set the Z bit. This leaves the STATUS register as '000u uuu' (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits (Refer to Section 33.0 “Instruction Set Summary”).

Note: The C and DC bits operate as Borrow and Digit Borrow out bits, respectively, in subtraction.
3.4 Register Definitions: Status

**REGISTER 3-1: STATUS: STATUS REGISTER**

<table>
<thead>
<tr>
<th>bit</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R-1/q</th>
<th>R-1/q</th>
<th>R/W-0/u</th>
<th>R/W-0/u</th>
<th>R/W-0/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC</td>
<td>C</td>
</tr>
<tr>
<td>0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **q** = Value depends on condition

**bit 7-5**  
**Unimplemented:** Read as '0'

**bit 4**  
**TO:** Time-Out bit  
1 = After power-up, **CLRWDT** instruction or **SLEEP** instruction  
0 = A WDT Time-out occurred

**bit 3**  
**PD:** Power-Down bit  
1 = After power-up or by the **CLRWDT** instruction  
0 = By execution of the **SLEEP** instruction

**bit 2**  
**Z:** Zero bit  
1 = The result of an arithmetic or logic operation is zero  
0 = The result of an arithmetic or logic operation is not zero

**bit 1**  
**DC:** Digit Carry/Digit Borrow bit (ADDWF, ADDLW, SUBLW, SUBWF instructions)(1)  
1 = A carry-out from the 4th low-order bit of the result occurred  
0 = No carry-out from the 4th low-order bit of the result

**bit 0**  
**C:** Carry/Borrow bit(1) (ADDWF, ADDLW, SUBLW, SUBWF instructions)(1)  
1 = A carry-out from the Most Significant bit of the result occurred  
0 = No carry-out from the Most Significant bit of the result occurred

**Note 1:** For **Borrow**, the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand.
3.4.1 SPECIAL FUNCTION REGISTER
The Special Function Registers are registers used by the application to control the desired operation of peripheral functions in the device. The Special Function Registers occupy the 20 bytes after the core registers of every data memory bank (addresses x0Ch/x8Ch through x1Fh/x9Fh). The registers associated with the operation of the peripherals are described in the appropriate peripheral chapter of this data sheet.

3.4.2 GENERAL PURPOSE RAM
There are up to 80 bytes of GPR in each data memory bank. The GPR occupies the 80 bytes after the SFR registers of selected data memory banks.

3.4.2.1 Linear Access to GPR
The general purpose RAM can be accessed in a non-banked method via the FSRs. This can simplify access to large memory structures. See Section 3.7.2 “Linear Data Memory” for more information.

3.4.3 COMMON RAM
There are 16 bytes of common RAM accessible from all banks.

FIGURE 3-3: BANKED MEMORY PARTITIONING

3.4.4 DEVICE MEMORY MAPS
The memory maps for the device family are as shown in Table 3-3 through Table 3-10.
### TABLE 3-3: PIC16(L)F1718 MEMORY MAP (BANKS 0-7)

<table>
<thead>
<tr>
<th>BANK 0</th>
<th>BANK 1</th>
<th>BANK 2</th>
<th>BANK 3</th>
<th>BANK 4</th>
<th>BANK 5</th>
<th>BANK 6</th>
<th>BANK 7</th>
</tr>
</thead>
<tbody>
<tr>
<td>000h</td>
<td>008h</td>
<td>100h</td>
<td>180h</td>
<td>200h</td>
<td>280h</td>
<td>300h</td>
<td>380h</td>
</tr>
<tr>
<td>00Ch</td>
<td>008h</td>
<td>10Ch</td>
<td>18Ch</td>
<td>20Ch</td>
<td>28Ch</td>
<td>30Ch</td>
<td>38Ch</td>
</tr>
<tr>
<td>00Dh</td>
<td>008h</td>
<td>10Dh</td>
<td>18Dh</td>
<td>20Dh</td>
<td>28Dh</td>
<td>30Dh</td>
<td>38Dh</td>
</tr>
<tr>
<td>00eh</td>
<td>008h</td>
<td>10eh</td>
<td>18eh</td>
<td>20eh</td>
<td>28eh</td>
<td>30eh</td>
<td>38eh</td>
</tr>
<tr>
<td>00Fh</td>
<td>00Fh</td>
<td>10Fh</td>
<td>18Fh</td>
<td>20Fh</td>
<td>28Fh</td>
<td>30Fh</td>
<td>38Fh</td>
</tr>
<tr>
<td>010h</td>
<td>010h</td>
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**Legend:**  
- Unimplemented data memory locations, read as ‘0’.

**Note 1:** Unimplemented on PIC16LF1717/8/9.
### TABLE 3-4: PIC16(L)F1717/9 MEMORY MAP (BANKS 0-7)

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Legend: = Unimplemented data memory locations, read as ‘0’.

### TABLE 3-5: PIC16(L)F1717 MEMORY MAP, BANK 8-23

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**Legend:**
- Unimplemented data memory locations, read as ‘0’.

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© 2014-2018 Microchip Technology Inc.
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<th>BANK 13</th>
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**Legend:**
- = Unimplemented data memory locations, read as '0'.

**TABLE 3-6:** PIC16(L)F1718/9 MEMORY MAP, BANK 8-23
### TABLE 3-7: PIC16(L)F1717 MEMORY MAP, BANK 24-31

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**Legend:**
- Unimplemented data memory locations, read as '0'.
- *Unimplemented Read as '0'.

See Table 3-9 for register mapping details.
### Table 3-8: PIC16(L)F1718/9 Memory Map, Bank 24-31

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**Legend:**
- = Unimplemented data memory locations, read as '0'.
TABLE 3-9: PIC16(L)F1717/8/9 MEMORY MAP, BANK 28-30

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<td>CKPPS</td>
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</tr>
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<td>E20h</td>
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<td>EA6h</td>
</tr>
<tr>
<td>E21h</td>
<td>SSPDATPPS</td>
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</tr>
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<td>E22h</td>
<td>SSPSSPPS</td>
<td>EA8h</td>
</tr>
<tr>
<td>E23h</td>
<td>—</td>
<td>EA9h</td>
</tr>
<tr>
<td>E24h</td>
<td>RXPPS</td>
<td>EAAh</td>
</tr>
<tr>
<td>E25h</td>
<td>CKPPS</td>
<td>EA Bh</td>
</tr>
<tr>
<td>E26h</td>
<td>—</td>
<td>EACh</td>
</tr>
<tr>
<td>E27h</td>
<td>—</td>
<td>EADh</td>
</tr>
<tr>
<td>E28h</td>
<td>CLCIN0PPS</td>
<td>EA8h</td>
</tr>
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<td>E29h</td>
<td>CLCIN1PPS</td>
<td>EA9h</td>
</tr>
<tr>
<td>E2Ah</td>
<td>CLCIN2PPS</td>
<td>EAAh</td>
</tr>
<tr>
<td>E2Bh</td>
<td>CLCIN3PPS</td>
<td>EA Bh</td>
</tr>
<tr>
<td>E2Ch</td>
<td>—</td>
<td>EACh</td>
</tr>
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</tr>
<tr>
<td>E2Eh</td>
<td>—</td>
<td>EA Eh</td>
</tr>
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<td>E2Fh</td>
<td>—</td>
<td>EAFh</td>
</tr>
<tr>
<td>E30h</td>
<td>—</td>
<td>EB0h</td>
</tr>
<tr>
<td>E31h</td>
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</tr>
<tr>
<td>E32h</td>
<td>—</td>
<td>EB2h</td>
</tr>
<tr>
<td>E33h</td>
<td>—</td>
<td>EB3h</td>
</tr>
<tr>
<td>E34h</td>
<td>—</td>
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<tr>
<td>E37h</td>
<td>—</td>
<td>EB7h</td>
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<tr>
<td>E38h</td>
<td>—</td>
<td>EB8h</td>
</tr>
<tr>
<td>E39h</td>
<td>—</td>
<td>EB9h</td>
</tr>
<tr>
<td>E3Ah</td>
<td>—</td>
<td>EB Ah</td>
</tr>
<tr>
<td>E3Bh</td>
<td>—</td>
<td>EBBh</td>
</tr>
<tr>
<td>E3Ch</td>
<td>—</td>
<td>EBCh</td>
</tr>
<tr>
<td>E3Dh</td>
<td>—</td>
<td>EBDh</td>
</tr>
<tr>
<td>E3Eh</td>
<td>—</td>
<td>EBEh</td>
</tr>
<tr>
<td>E3Fh</td>
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<td>EB Fh</td>
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<tr>
<td>E40h</td>
<td>—</td>
<td>ECFh</td>
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</tbody>
</table>

Legend:  = Unimplemented data memory locations, read as ‘0’.

Note 1: Only available on PIC16(L)F1717/9 devices.
TABLE 3-10: PIC16(L)F1717/8/9 MEMORY MAP, BANK 31

<table>
<thead>
<tr>
<th>Address</th>
<th>Description</th>
</tr>
</thead>
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</tr>
<tr>
<td>FE3h</td>
<td></td>
</tr>
<tr>
<td>FE4h</td>
<td>STATUS_SHAD</td>
</tr>
<tr>
<td>FE5h</td>
<td>WREG_SHAD</td>
</tr>
<tr>
<td>FE6h</td>
<td>BSR_SHAD</td>
</tr>
<tr>
<td>FE7h</td>
<td>PCLATH_SHAD</td>
</tr>
<tr>
<td>FE8h</td>
<td>FSR0L_SHAD</td>
</tr>
<tr>
<td>FE9h</td>
<td>FSR0H_SHAD</td>
</tr>
<tr>
<td>FEAh</td>
<td>FSR1L_SHAD</td>
</tr>
<tr>
<td>FEBh</td>
<td>FSR1H_SHAD</td>
</tr>
<tr>
<td>FECh</td>
<td></td>
</tr>
<tr>
<td>FE Dh</td>
<td>STKPTR</td>
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<tr>
<td>FEEh</td>
<td>TOSL</td>
</tr>
<tr>
<td>FE Fh</td>
<td>TOSH</td>
</tr>
<tr>
<td>FF0h</td>
<td></td>
</tr>
<tr>
<td>FFFh</td>
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</table>

Legend:  = Unimplemented data memory locations, read as '0'.
3.4.5 CORE FUNCTION REGISTERS

SUMMARY

The Core Function registers listed in Table 3-11 can be addressed from any Bank.

**TABLE 3-11: CORE FUNCTION REGISTERS SUMMARY (1)**

<table>
<thead>
<tr>
<th>Addr.</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
</tr>
</thead>
<tbody>
<tr>
<td>Bank 0-31</td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x00h or x80h</td>
<td>INDF0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x01h or x81h</td>
<td>INDF1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x02h or x82h</td>
<td>PCL</td>
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<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x03h or x83h</td>
<td>STATUS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC</td>
<td>C</td>
<td>---1 1000</td>
<td>---q quuu</td>
</tr>
<tr>
<td>x04h or x84h</td>
<td>FSR0L</td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x05h or x85h</td>
<td>FSR0H</td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x06h or x86h</td>
<td>FSR1L</td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
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<tr>
<td>x07h or x87h</td>
<td>FSR1H</td>
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<td></td>
</tr>
<tr>
<td>x08h or x88h</td>
<td>BSR</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>BSR4</td>
<td>BSR3</td>
<td>BSR2</td>
<td>BSR1</td>
<td>BSR0</td>
<td>---0 0000</td>
<td>---0 0000</td>
</tr>
<tr>
<td>x09h or x89h</td>
<td>WREG</td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x0Ah or x8Ah</td>
<td>PCLATH</td>
<td>—</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>x0Bh or x8Bh</td>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCF</td>
<td>0000 0000</td>
<td>0000 0000</td>
</tr>
</tbody>
</table>

Legend:  
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved.  
Shaded locations are unimplemented, read as '0'.

**Note 1:** These registers can be addressed from any bank.
### TABLE 3-12: SPECIAL FUNCTION REGISTER SUMMARY

<table>
<thead>
<tr>
<th>Addr.</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
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<tr>
<td>00Ch</td>
<td>PORTA</td>
<td>RA7</td>
<td>RA6</td>
<td>RA5</td>
<td>RA4</td>
<td>RA3</td>
<td>RA2</td>
<td>RA1</td>
<td>RA0</td>
<td>xxxx xxxx</td>
<td>uuuu uuuu</td>
</tr>
<tr>
<td>00Dh</td>
<td>PORTB</td>
<td>RB7</td>
<td>RB6</td>
<td>RB5</td>
<td>RB4</td>
<td>RB3</td>
<td>RB2</td>
<td>RB1</td>
<td>RB0</td>
<td>xxxx xxxx</td>
<td>uuuu uuuu</td>
</tr>
<tr>
<td>00Eh</td>
<td>PORTC</td>
<td>RC7</td>
<td>RC6</td>
<td>RC5</td>
<td>RC4</td>
<td>RC3</td>
<td>RC2</td>
<td>RC1</td>
<td>RC0</td>
<td>xxxx xxxx</td>
<td>uuuu uuuu</td>
</tr>
<tr>
<td>00Fh</td>
<td>PORTD(1)</td>
<td>RD7</td>
<td>RD6</td>
<td>RD5</td>
<td>RD4</td>
<td>RD3</td>
<td>RD2</td>
<td>RD1</td>
<td>RD0</td>
<td>xxxx xxxx</td>
<td>- - - - - -</td>
</tr>
<tr>
<td>010h</td>
<td>PORTE</td>
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<td>—</td>
<td>—</td>
<td>RE3</td>
<td>RE2(1)</td>
<td>RE1(1)</td>
<td>RE0(1)</td>
<td>---- ----</td>
<td>---- ----</td>
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<tr>
<td>011h</td>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>0000 0000</td>
<td>0000 0000 0-0-0</td>
</tr>
<tr>
<td>012h</td>
<td>PIR2</td>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR8IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td>00- 0000</td>
<td>00- 00-00-0</td>
</tr>
<tr>
<td>013h</td>
<td>PIR3</td>
<td>—</td>
<td>NCOIF</td>
<td>COGIF</td>
<td>ZCIF</td>
<td>CLC4IF</td>
<td>CLC3IF</td>
<td>CLC2IF</td>
<td>CLC1IF</td>
<td>-000 0000</td>
<td>--00 00-00</td>
</tr>
<tr>
<td>014h</td>
<td>—</td>
<td>Unimplemented</td>
<td>—</td>
<td>—</td>
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<td>—</td>
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</tr>
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<td>015h</td>
<td>TMR0</td>
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<tr>
<td>017h</td>
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<td>—</td>
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</tr>
<tr>
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<td>T1CON</td>
<td>TMR1CS&lt;1:0&gt;</td>
<td>T1CKPS&lt;1:0&gt;</td>
<td>T1OSCEN</td>
<td>T1SYNC</td>
<td>—</td>
<td>TMR1ON</td>
<td>0000 00-0</td>
<td>uuuu uu-u</td>
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</tr>
<tr>
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<td>T1GCON</td>
<td>TMR1GE</td>
<td>T1GPOL</td>
<td>T1GM</td>
<td>T1GSPM</td>
<td>T1GGO/</td>
<td>T1VAL</td>
<td>T1GSS&lt;1:0&gt;</td>
<td>0000 0x00</td>
<td>uuuu uuuu</td>
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</tr>
<tr>
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<td>TMR2</td>
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<td>—</td>
<td>—</td>
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<tr>
<td>01Bh</td>
<td>PR2</td>
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<td>01Dh</td>
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#### Bank 1

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<th>Addr.</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
</tr>
</thead>
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<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
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<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
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<td>1111 1111</td>
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<tr>
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<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>1111 1111</td>
<td>1111 1111</td>
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<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
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<td>1111 1111</td>
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<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
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<td>TRISD0</td>
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<td>TRISE1(1)</td>
<td>TRISE0(1)</td>
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<td>---- ----</td>
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<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
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<td>0000 0000</td>
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<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
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<td>00- 00-00-0</td>
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<td>NCOIE</td>
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<td>ZCIE</td>
<td>CLC4IE</td>
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<td>CLC2IE</td>
<td>CLC1IE</td>
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<td>--00 00-00</td>
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<td>095h</td>
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<td>INTEDG</td>
<td>TMROCS</td>
<td>TMROSE</td>
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<td></td>
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<td>STKUNF</td>
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<td>RWDT</td>
<td>RMCCLR</td>
<td>RI</td>
<td>POR</td>
<td>BOR</td>
<td>00-1 11qg</td>
<td>qqq-q qqq</td>
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#### Legend:
- **x** = unknown,
- **u** = unchanged,
- **q** = value depends on condition,
- **-** = unimplemented, read as ‘0’,
- **r** = reserved.

**Shaded locations are unimplemented, read as ‘0’.**

**Note 1:** Unimplemented on PIC16(L)F1718.

**Note 2:** Unimplemented on PIC16LF1717/8/9
### TABLE 3-12: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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**Legend:**
- `x` = unknown, `u` = unchanged, `q` = value depends on condition, `-` = unimplemented, read as `0`
- *Shaded locations are unimplemented, read as `0`.

**Note:**
1. Unimplemented on PIC16L1718.
2. Unimplemented on PIC16L1717/8/9
### TABLE 3-12: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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<th>Bit 4</th>
<th>Bit 3</th>
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</table>

**Legend:**
- x = unknown,
- u = unchanged,
- q = value depends on condition,
- - = unimplemented, read as '0',
- r = reserved.

**Shaded locations are unimplemented, read as '0'**

**Note 1:** Unimplemented on PIC16(L)F1718.
**Note 2:** Unimplemented on PIC16LF1717/8/9
### TABLE 3-12: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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<th>Name</th>
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<th>Bit 4</th>
<th>Bit 3</th>
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<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
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Legend:  
- u = unchanged, q = value depends on condition, x = unimplemented, read as '0', rz = reserved.

Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented on PIC16LF1718.
Note 2: Unimplemented on PIC16LF1717/9
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**Legend:**
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved.

**Shaded locations are unimplemented, read as '0'**.

**Note**:
1. Unimplemented on PIC16LF1718.
2. Unimplemented on PIC16LF1717/8/9
### TABLE 3-12: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

<table>
<thead>
<tr>
<th>Addr.</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
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<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
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</table>

**Bank 14-27**

- xOCn/x8Ch  
- x1Fh/x9Fh  
  - Unimplemented

**Bank 28**

- E0Ch  
- E0Fh  
  - Unimplemented

- E0Fh  
  - PPSLOCK

- E10h  
  - INTPPS

- E11h  
  - T0CKIPPS

- E12h  
  - T1CKIPPS

- E13h  
  - T1GPPS

- E14h  
  - CCP1PPS

- E15h  
  - CCP2PPS

- E16h  
  - Unimplemented

- E17h  
  - COGINPPS

- E18h  
  - Unimplemented

- E19h  
  - Unimplemented

- E1Ah  
  - Unimplemented

- E1FH  
  - Unimplemented

- E20h  
  - SSPCLKPPS

- E21h  
  - SSPDATPPS

- E22h  
  - SSPSSPPS

- E23h  
  - Unimplemented

- E24h  
  - RXPPS

- E25h  
  - CKPPS

- E26h  
  - Unimplemented

- E27h  
  - Unimplemented

- E28h  
  - CLCIN0PPS

- E29h  
  - CLCIN1PPS

- E2Ah  
  - CLCIN2PPS

- E2 Bh  
  - CLCIN3PPS

- E2Ch to E6Fh  
  - Unimplemented

**Legend:**  
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as ‘0’, r = reserved.  
- Shaded locations are unimplemented, read as ‘0’.

**Note**

1. Unimplemented on PIC16L/F1718.  
2. Unimplemented on PIC16L/F1717/8/9
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**Legend:**
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as ‘0’, r = reserved.

**Note**
1: Unimplemented on PIC16LRF1718.
2: Unimplemented on PIC16LF1717/8/9
### TABLE 3-12: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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**Legend:**
- x = unknown,
- u = unchanged,
- q = value depends on condition,
- - = unimplemented, read as '0',
- r = reserved.

**Note 1:** Unimplemented on PIC16LF1718.
**Note 2:** Unimplemented on PIC16LF1717/9
### TABLE 3-12: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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<tr>
<td>FE9h</td>
<td>FSR0H_SHAD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>FSR0H_SHAD</td>
<td></td>
<td></td>
</tr>
<tr>
<td>FEAh</td>
<td>FSR1L_SHAD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>FSR1L_SHAD</td>
<td></td>
<td></td>
</tr>
<tr>
<td>FE Bh</td>
<td>FSR1H_SHAD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>FSR1H_SHAD</td>
<td></td>
<td></td>
</tr>
<tr>
<td>FExh</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>Unimplemented</td>
<td></td>
<td></td>
</tr>
<tr>
<td>FEEd</td>
<td>STKPTR</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>STKPTR</td>
<td></td>
<td></td>
</tr>
<tr>
<td>FEEh</td>
<td>TOSL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TOSL</td>
<td></td>
<td></td>
</tr>
<tr>
<td>FEFh</td>
<td>TOSH</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TOSH</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- \(x\) = unknown,
- \(u\) = unchanged,
- \(q\) = value depends on condition,
- \(-\) = unimplemented, read as '0',
- \(r\) = reserved.

Shaded locations are unimplemented, read as '0'.

**Note:**
1. Unimplemented on PIC16(L)F1718.
3.5  PCL and PCLATH

The Program Counter (PC) is 15 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<14:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 3-4 shows the five situations for the loading of the PC.

**FIGURE 3-4: LOADING OF PC IN DIFFERENT SITUATIONS**

- **3.5.1 MODIFYING PCL**
  Executing any instruction with the PCL register as the destination simultaneously causes the Program Counter PC<14:8> bits (PCH) to be replaced by the contents of the PCLATH register. This allows the entire contents of the program counter to be changed by writing the desired upper seven bits to the PCLATH register. When the lower eight bits are written to the PCL register, all 15 bits of the program counter will change to the values contained in the PCLATH register and those being written to the PCL register.

- **3.5.2 COMPUTED GOTO**
  A computed **GOTO** is accomplished by adding an offset to the program counter (**ADDWF PCL**). When performing a table read using a computed **GOTO** method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to Application Note AN556, “Implementing a Table Read” (DS00556).

- **3.5.3 COMPUTED FUNCTION CALLS**
  A computed function **CALL** allows programs to maintain tables of functions and provide another way to execute state machines or look-up tables. When performing a table read using a computed function **CALL**, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block).

  If using the **CALL** instruction, the PCH<2:0> and PCL registers are loaded with the operand of the **CALL** instruction. PCH<6:3> is loaded with PCLATH<6:3>.

  The **CALLW** instruction enables computed calls by combining PCLATH and W to form the destination address. A computed **CALLW** is accomplished by loading the W register with the desired address and executing **CALLW**. The PCL register is loaded with the value of W and PCH is loaded with PCLATH.

- **3.5.4 BRANCHING**
  The branching instructions add an offset to the PC. This allows relocatable code and code that crosses page boundaries. There are two forms of branching, **BRW** and **BRA**. The PC will have incremented to fetch the next instruction in both cases. When using either branching instruction, a PCL memory boundary may be crossed.

  If using **BRW**, load the W register with the desired unsigned address and execute **BRW**. The entire PC will be loaded with the address PC + 1 + W.

  If using **BRA**, the entire PC will be loaded with PC + 1 +, the signed value of the operand of the **BRA** instruction.
3.6 Stack

All devices have a 16-level x 15-bit wide hardware stack (refer to Figure 3-1). The stack space is not part of either program or data space. The PC is pushed onto the stack when CALL or CALLW instructions are executed or an interrupt causes a branch. The stack is popped in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a push or pop operation.

The stack operates as a circular buffer if the STVREN bit is programmed to ‘0’ (Configuration Words). This means that after the stack has been pushed sixteen times, the seventeenth push overwrites the value that was stored from the first push. The eighteenth push overwrites the second push (and so on). The STKOVF and STKUNF flag bits will be set on an Overflow/Underflow, regardless of whether the Reset is enabled.

3.6.1 Accessing the Stack

The stack is available through the TOSH, TOSL and STKPTR registers. STKPTR is the current value of the Stack Pointer. TOSH:TOSL register pair points to the top of the stack. Both registers are read/writeable. TOS is split into TOSH and TOSL due to the 15-bit size of the PC. To access the stack, adjust the value of STKPTR, which will position TOSH:TOSL, then read/write to TOSH:TOSL. STKPTR is five bits to allow detection of overflow and underflow.

Note: Care should be taken when modifying the STKPTR while interrupts are enabled.

During normal program operation, CALL, CALLW and interrupts will increment STKPTR while RETLW, RETURN, and RETFIE will decrement STKPTR. At any time, STKPTR can be inspected to see how much stack is left. The STKPTR always points at the currently used place on the stack. Therefore, a CALL or CALLW will increment the STKPTR and then write the PC, and a return will unload the PC and then decrement the STKPTR.

Reference Figure 3-5 through Figure 3-8 for examples of accessing the stack.

FIGURE 3-5: ACCESSING THE STACK EXAMPLE 1

Note: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, CALLW, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

Initial Stack Configuration:

After Reset, the stack is empty. The empty stack is initialized so the Stack Pointer is pointing at 0x1F. If the Stack Overflow/Underflow Reset is enabled, the TOSH/TOSL registers will return ‘0’. If the Stack Overflow/Underflow Reset is disabled, the TOSH/TOSL registers will return the contents of stack address 0x0F.
FIGURE 3-6: ACCESSING THE STACK EXAMPLE 2

This figure shows the stack configuration after the first CALL or a single interrupt. If a RETURN instruction is executed, the return address will be placed in the Program Counter and the Stack Pointer decremented to the empty state (0x1F).

FIGURE 3-7: ACCESSING THE STACK EXAMPLE 3

After seven CALLs or six CALLs and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
3.6.2 OVERFLOW/UNDERFLOW RESET

If the STVREN bit in Configuration Words is programmed to ‘1’, the device will be reset if the stack is PUSHed beyond the sixteenth level or POPed beyond the first level, setting the appropriate bits (STKOVF or STKUNF, respectively) in the PCON register.

3.7 Indirect Addressing

The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the File Select Registers (FSR). If the FSRn address specifies one of the two INDFn registers, the read will return ‘0’ and the write will not occur (though Status bits may be affected). The FSRn register value is created by the pair FSRnH and FSRnL.

The FSR registers form a 16-bit address that allows an addressing space with 65536 locations. These locations are divided into three memory regions:

- Traditional Data Memory
- Linear Data Memory
- Program Flash Memory
FIGURE 3-9: INDIRECT ADDRESSING

Note: Not all memory regions are completely implemented. Consult device memory tables for memory limits.
3.7.1 TRADITIONAL DATA MEMORY

The traditional data memory is a region from FSR address 0x000 to FSR address 0xFFF. The addresses correspond to the absolute addresses of all SFR, GPR and common registers.

FIGURE 3-10: TRADITIONAL DATA MEMORY MAP
3.7.2 LINEAR DATA MEMORY

The linear data memory is the region from FSR address 0x2000 to FSR address 0x29AF. This region is a virtual region that points back to the 80-byte blocks of GPR memory in all the banks.

Unimplemented memory reads as 0x00. Use of the linear data memory region allows buffers to be larger than 80 bytes because incrementing the FSR beyond one bank will go directly to the GPR memory of the next bank.

The 16 bytes of common memory are not included in the linear data memory region.

3.7.3 PROGRAM FLASH MEMORY

To make constant data access easier, the entire program Flash memory is mapped to the upper half of the FSR address space. When the MSB of FSRnH is set, the lower 15 bits are the address in program memory which will be accessed through INDF. Only the lower eight bits of each memory location is accessible via INDF. Writing to the program Flash memory cannot be accomplished via the FSR/INDF interface. All instructions that access program Flash memory via the FSR/INDF interface will require one additional instruction cycle to complete.
4.0 DEVICE CONFIGURATION

Device configuration consists of Configuration Words, Code Protection and Device ID.

4.1 Configuration Words

There are several Configuration Word bits that allow different oscillator and memory protection options. These are implemented as Configuration Word 1 at 8007h and Configuration Word 2 at 8008h.

Note: The DEBUG bit in Configuration Words is managed automatically by device development tools including debuggers and programmers. For normal device operation, this bit should be maintained as a ‘1’.

4.2 Register Definitions: Configuration Words

REGISTER 4-1: CONFIG1: CONFIGURATION WORD 1

<table>
<thead>
<tr>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>U-1</th>
</tr>
</thead>
<tbody>
<tr>
<td>FCMEN</td>
<td>IESO</td>
<td>CLKOUTEN</td>
<td>BOREN&lt;1:0&gt;</td>
<td>___</td>
<td></td>
</tr>
</tbody>
</table>

bit 13

Legend:

R = Readable bit
P = Programmable bit
U = Unimplemented bit, read as ‘1’

‘0’ = Bit is cleared
‘1’ = Bit is set
-n = Value when blank or after Bulk Erase

bit 13 FCMEN: Fail-Safe Clock Monitor Enable bit

1 = Fail-Safe Clock Monitor and internal/external switchover are both enabled
0 = Fail-Safe Clock Monitor is disabled

bit 12 IESO: Internal External Switchover bit

1 = Internal/External Switchover mode is enabled
0 = Internal/External Switchover mode is disabled

bit 11 CLKOUTEN: Clock Out Enable bit

This bit is ignored, CLKOUT function is disabled. Oscillator function on the CLKOUT pin.

All other FOSC modes:

1 = CLKOUT function is disabled. I/O function on the CLKOUT pin.
0 = CLKOUT function is enabled on the CLKOUT pin

bit 10-9 BOREN<1:0>: Brown-out Reset Enable bits

11 = BOR enabled
10 = BOR enabled during operation and disabled in Sleep
01 = BOR controlled by SBoren bit of the BORCON register
00 = BOR disabled

bit 8 Unimplemented: Read as ‘1’
### REGISTER 4-1: CONFIG1: CONFIGURATION WORD 1 (CONTINUED)

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td><strong>CP</strong>: Code Protection bit&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>1</td>
<td>Program memory code protection is disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0</td>
<td>Program memory code protection is enabled</td>
</tr>
<tr>
<td>6</td>
<td><strong>MCLRE</strong>: MCLR/VPP Pin Function Select bit</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>If LVP bit = 1:</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>This bit is ignored</td>
</tr>
<tr>
<td></td>
<td>If LVP bit = 0:</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = MCLR/VPP pin function is MCLR; Weak pull-up enabled</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = MCLR/VPP pin function is digital input; MCLR internally disabled; Weak pull-up under control of</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>WPUE&lt;3&gt; bit.</td>
<td></td>
<td></td>
</tr>
<tr>
<td>5</td>
<td><strong>PWRTE</strong>: Power-up Timer Enable bit</td>
<td>1</td>
<td>PWRT disabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0</td>
<td>PWRT enabled</td>
</tr>
<tr>
<td>4-3</td>
<td><strong>WDTE&lt;1:0&gt;</strong>: Watchdog Timer Enable bit</td>
<td>11</td>
<td>WDT enabled</td>
</tr>
<tr>
<td></td>
<td></td>
<td>10</td>
<td>WDT enabled while running and disabled in Sleep</td>
</tr>
<tr>
<td></td>
<td></td>
<td>01</td>
<td>WDT controlled by the SWDTEN bit in the WDTCON register</td>
</tr>
<tr>
<td></td>
<td></td>
<td>00</td>
<td>WDT disabled</td>
</tr>
<tr>
<td>2-0</td>
<td><strong>FOSC&lt;2:0&gt;</strong>: Oscillator Selection bits</td>
<td>111</td>
<td>ECH: External Clock, High-Power mode (4-20 MHz): device clock supplied to CLKin pin</td>
</tr>
<tr>
<td></td>
<td></td>
<td>110</td>
<td>ECM: External Clock, Medium Power mode (0.5-4 MHz): device clock supplied to CLKin pin</td>
</tr>
<tr>
<td></td>
<td></td>
<td>101</td>
<td>ECL: External Clock, Low-Power mode (0-0.5 MHz): device clock supplied to CLKin pin</td>
</tr>
<tr>
<td></td>
<td></td>
<td>100</td>
<td>INTOSC oscillator: I/O function on CLKin pin</td>
</tr>
<tr>
<td></td>
<td></td>
<td>011</td>
<td>EXTRC oscillator: External RC circuit connected to CLKin pin</td>
</tr>
<tr>
<td></td>
<td></td>
<td>010</td>
<td>HS oscillator: High-speed crystal/resonator connected between OSC1 and OSC2 pins</td>
</tr>
<tr>
<td></td>
<td></td>
<td>001</td>
<td>XT oscillator: Crystal/resonator connected between OSC1 and OSC2 pins</td>
</tr>
<tr>
<td></td>
<td></td>
<td>000</td>
<td>LP oscillator: Low-power crystal connected between OSC1 and OSC2 pins</td>
</tr>
</tbody>
</table>

**Note 1:** The entire Flash program memory will be erased when the code protection is turned off during an erase. When a Bulk Erase Program Memory command is executed, the entire program Flash memory and configuration memory will be erased.
REGISTER 4-2: CONFIG2: CONFIGURATION WORD 2

<table>
<thead>
<tr>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
</tr>
</thead>
<tbody>
<tr>
<td>R/P-1</td>
<td>U-1</td>
<td>U-1</td>
<td>U-1</td>
<td>U-1</td>
<td>PPS1WAY</td>
</tr>
</tbody>
</table>

bit 13 \(R/P-1\) R/P-1 R/P-1 R/P-1 R/P-1 R/P-1

bit 12 \(LVP(1)\) DEBUG(2) LPBOR \(\text{BORV(3)}\) STVREN PLLEN

bit 7 \(ZCDDIS\) — — — — PPS1WAY WRT<1:0>

Legend:

- **R** = Readable bit
- **P** = Programmable bit
- **U** = Unimplemented bit, read as ‘1’
- ‘0’ = Bit is cleared
- ‘1’ = Bit is set
- ‘n’ = Value when blank or after Bulk Erase

**bit 13** \(LVP\): Low-Voltage Programming Enable bit(1)

- 1 = Low-voltage programming enabled
- 0 = High-voltage on MCLR must be used for programming

**bit 12** \(DEBUG\): In-Circuit Debugger Mode bit(2)

- 1 = In-Circuit Debugger disabled, ICSPCLK and ICSPDAT are general purpose I/O pins
- 0 = In-Circuit Debugger enabled, ICSPCLK and ICSPDAT are dedicated to the debugger

**bit 11** \(LPBOR\): Low-Power BOR Enable bit

- 1 = Low-Power Brown-out Reset is disabled
- 0 = Low-Power Brown-out Reset is enabled

**bit 10** \(BORV\): Brown-out Reset Voltage Selection bit(3)

- 1 = Brown-out Reset voltage (\(V_{\text{BOR}}\)), low trip point selected.
- 0 = Brown-out Reset voltage (\(V_{\text{BOR}}\)), high trip point selected.

**bit 9** \(STVREN\): Stack Overflow/Underflow Reset Enable bit

- 1 = Stack Overflow or Underflow will cause a Reset
- 0 = Stack Overflow or Underflow will not cause a Reset

**bit 8** \(PLLEN\): PLL Enable bit

- 1 = 4xPLL enabled
- 0 = 4xPLL disabled

**bit 7** \(ZCDDIS\): ZCD Disable bit

- 1 = ZCD disabled. ZCD can be enabled by setting the ZCDSEN bit of ZCDCON
- 0 = ZCD always enabled

**bit 6-3** Unimplemented: Read as ‘1’

**bit 2** \(PPS1WAY\): PPSLOCK Bit One-Way Set Enable bit

- 1 = The PPSLOCK bit can only be set once after an unlocking sequence is executed; once PPSLOCK is set, all future changes to PPS registers are prevented
- 0 = The PPSLOCK bit can be set and cleared as needed (provided an unlocking sequence is executed)

**bit 1-0** \(WRT<1:0>\): Flash Memory Self-Write Protection bits

- 11 = Write protection off
- 10 = 0000h to 01FFh write-protected, 0200h to 1FFFh may be modified by PMCON control
- 01 = 0000h to 0FFh write-protected, 1000h to 1FFFh may be modified by PMCON control
- 00 = 0000h to 1FFFh write-protected, no addresses may be modified by PMCON control

16 kW Flash memory (PIC16(L)F1718/9)

- 11 = Write protection off
- 10 = 0000h to 01FFh write-protected, 0200h to 3FFFh may be modified by PMCON control
- 01 = 0000h to 1FFh write-protected, 2000h to 3FFFh may be modified by PMCON control
- 00 = 0000h to 3FFFh write-protected, no addresses may be modified by PMCON control

**Note**

1: The LVP bit cannot be programmed to ‘0’ when Programming mode is entered via LVP

2: The \(DEBUG\) bit in Configuration Words is managed automatically by device development tools including debuggers and programmers. For normal device operation, this bit should be maintained as a ‘1’.

3: See \(V_{\text{BOR}}\) parameter for specific trip point voltages.
4.3 Code Protection

Code protection allows the device to be protected from unauthorized access. Program memory protection is controlled independently. Internal access to the program memory is unaffected by any code protection setting.

4.3.1 PROGRAM MEMORY PROTECTION

The entire program memory space is protected from external reads and writes by the CP bit in Configuration Words. When CP = 0, external reads and writes of program memory are inhibited and a read will return all ‘0’s. The CPU can continue to read program memory, regardless of the protection bit settings. Writing the program memory is dependent upon the write protection setting. See Section 4.4 “Write Protection” for more information.

4.4 Write Protection

Write protection allows the device to be protected from unintended self-writes. Applications, such as boot loader software, can be protected while allowing other regions of the program memory to be modified.

The WRT<1:0> bits in Configuration Words define the size of the program memory block that is protected.

4.5 User ID

Four memory locations (8000h-8003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are readable and writable during normal execution. See Section 10.4 “User ID, Device ID and Configuration Word Access” for more information on accessing these memory locations. For more information on checksum calculation, see the “PIC16(L)F170X Memory Programming Specification” (DS41683).

4.6 Device ID and Revision ID

The 14-bit device ID word is located at 8006h and the 14-bit revision ID is located at 8005h. These locations are read-only and cannot be erased or modified. See Section 10.4 “User ID, Device ID and Configuration Word Access” for more information on accessing these memory locations.

Development tools, such as device programmers and debuggers, may be used to read the Device ID and Revision ID.
4.7 Register Definitions: Device and Revision

REGISTER 4-3: DEVID: DEVICE ID REGISTER

<table>
<thead>
<tr>
<th>bit 13</th>
<th>bit 8</th>
</tr>
</thead>
<tbody>
<tr>
<td>DEV&lt;13:8&gt;</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
'1' = Bit is set
'0' = Bit is cleared

bit 13-0 DEV<13:0>: Device ID bits

<table>
<thead>
<tr>
<th>Device</th>
<th>DEVID&lt;13:0&gt; Values</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC16F1717</td>
<td>11 0000 0101 1100 (305Ch)</td>
</tr>
<tr>
<td>PIC16LF1717</td>
<td>11 0000 0101 1111 (305Fh)</td>
</tr>
<tr>
<td>PIC16F1718</td>
<td>11 0000 0101 1011 (305Bh)</td>
</tr>
<tr>
<td>PIC16LF1718</td>
<td>11 0000 0101 1110 (305Eh)</td>
</tr>
<tr>
<td>PIC16F1719</td>
<td>11 0000 0101 1010 (305Ah)</td>
</tr>
<tr>
<td>PIC16LF1719</td>
<td>11 0000 0101 1101 (305Dh)</td>
</tr>
</tbody>
</table>

REGISTER 4-4: REVID: REVISION ID REGISTER

<table>
<thead>
<tr>
<th>bit 13</th>
<th>bit 8</th>
</tr>
</thead>
<tbody>
<tr>
<td>REV&lt;13:8&gt;</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
'1' = Bit is set
'0' = Bit is cleared

bit 13-0 REV<13:0>: Revision ID bits
5.0  RESETS

There are multiple ways to reset this device:

- Power-on Reset (POR)
- Brown-out Reset (BOR)
- Low-Power Brown-out Reset (LPBOR)
- MCLR Reset
- WDT Reset
- RESET instruction
- Stack Overflow
- Stack Underflow
- Programming mode exit

To allow VDD to stabilize, an optional Power-up Timer can be enabled to extend the Reset time after a BOR or POR event.

A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 5-1.

**FIGURE 5-1:  SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT**

Note 1:  See Table 5-1 for BOR active conditions.
5.1 Power-on Reset (POR)

The POR circuit holds the device in Reset until VDD has reached an acceptable level for minimum operation. Slow rising VDD, fast operating speeds or analog performance may require greater than minimum VDD. The PWRT, BOR or MCLR features can be used to extend the start-up period until all device operation conditions have been met.

5.1.1 POWER-UP TIMER (PWRT)

The Power-up Timer provides a nominal 64 ms time-out on POR or Brown-out Reset. The device is held in Reset as long as PWRT is active. The PWRT delay allows additional time for the VDD to rise to an acceptable level. The Power-up Timer is enabled by clearing the PWRTEN bit in Configuration Words.

The Power-up Timer starts after the release of the POR and BOR.

For additional information, refer to Application Note AN607, “Power-up Trouble Shooting” (DS00607).

TABLE 5-1: BOR OPERATING MODES

<table>
<thead>
<tr>
<th>BOREN&lt;1:0&gt;</th>
<th>SBOREN</th>
<th>Device Mode</th>
<th>BOR Mode</th>
<th>Instruction Execution upon: Release of POR or Wake-up from Sleep</th>
</tr>
</thead>
<tbody>
<tr>
<td>11</td>
<td>X</td>
<td>X</td>
<td>Active</td>
<td>Waits for BOR ready(1) (BORRDY = 1)</td>
</tr>
<tr>
<td>10</td>
<td>X</td>
<td>Awake</td>
<td>Active</td>
<td>Waits for BOR ready (BORRDY = 1)</td>
</tr>
<tr>
<td>01</td>
<td>1</td>
<td>Sleep</td>
<td>Disabled</td>
<td></td>
</tr>
<tr>
<td>00</td>
<td>X</td>
<td>Disabled</td>
<td>Disabled</td>
<td>Begins immediately (BORRDY = x)</td>
</tr>
</tbody>
</table>

Note 1: In these specific cases, "Release of POR" and "Wake-up from Sleep", there is no delay in start-up. The BOR ready flag, (BORRDY = 1), will be set before the CPU is ready to execute instructions because the BOR circuit is forced on by the BOREN<1:0> bits.

5.2 Brown-out Reset (BOR)

The BOR circuit holds the device in Reset when VDD reaches a selectable minimum level. Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

The Brown-out Reset module has four operating modes controlled by the BOREN<1:0> bits in Configuration Words. The four operating modes are:

- BOR is always on
- BOR is off when in Sleep
- BOR is controlled by software
- BOR is always off

Refer to Table 5-1 for more information.

The Brown-out Reset voltage level is selectable by configuring the BORV bit in Configuration Words.

A VDD noise rejection filter prevents the BOR from triggering on small events. If VDD falls below VBOR for a duration greater than parameter TBORDC, the device will reset. See Figure 5-2 for more information.

5.2.1 BOR IS ALWAYS ON

When the BOREN bits of Configuration Words are programmed to ‘11’, the BOR is always on. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is active during Sleep. The BOR does not delay wake-up from Sleep.

5.2.2 BOR IS OFF IN SLEEP

When the BOREN bits of Configuration Words are programmed to ‘10’, the BOR is on, except in Sleep. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is not active during Sleep. The device wake-up will be delayed until the BOR is ready.

5.2.3 BOR CONTROLLED BY SOFTWARE

When the BOREN bits of Configuration Words are programmed to ‘01’, the BOR is controlled by the SBOREN bit of the BORCON register. The device start-up is not delayed by the BOR ready condition or the VDD level.

BOR protection begins as soon as the BOR circuit is ready. The status of the BOR circuit is reflected in the BORRDY bit of the BORCON register.

BOR protection is unchanged by Sleep.
5.3 Register Definitions: BOR Control

REGISTER 5-1: BORCON: BROWN-OUT RESET CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-1/u</th>
<th>R/W-0/u</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R-q/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>SBOREN</td>
<td>BORFS(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>BORRDY</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- q = Value depends on condition

bit 7 SBOREN: Software Brown-out Reset Enable bit
- If BOREN <1:0> in Configuration Words ≠ 01:
  - SBOREN is read/write, but has no effect on the BOR.
- If BOREN <1:0> in Configuration Words = 01:
  - 1 = BOR Enabled
  - 0 = BOR Disabled

bit 6 BORFS: Brown-out Reset Fast Start bit(1)
- If BOREN<1:0> = 11 (Always on) or BOREN<1:0> = 00 (Always off)
  - BORFS is Read/Write, but has no effect.
- If BOREN<1:0> = 10 (Disabled in Sleep) or BOREN<1:0> = 01 (Under software control):
  - 1 = Band gap is forced on always (covers sleep/wake-up/operating cases)
  - 0 = Band gap operates normally, and may turn off

bit 5-1 Unimplemented: Read as ‘0’

bit 0 BORRDY: Brown-out Reset Circuit Ready Status bit
- 1 = The Brown-out Reset circuit is active
- 0 = The Brown-out Reset circuit is inactive

Note 1: BOREN<1:0> bits are located in Configuration Words.
5.4 Low-Power Brown-out Reset (LPBOR)

The Low-Power Brown-out Reset (LPBOR) is an essential part of the Reset subsystem. Refer to Figure 5-1 to see how the BOR interacts with other modules.

The LPBOR is used to monitor the external VDD pin. When too low of a voltage is detected, the device is held in Reset. When this occurs, a register bit (BOR) is changed to indicate that a BOR Reset has occurred. The same bit is set for both the BOR and the LPBOR. Refer to Register 5-2.

5.4.1 ENABLING LPBOR

The LPBOR is controlled by the LPBOR bit of Configuration Words. When the device is erased, the LPBOR module defaults to disabled.

5.4.1.1 LPBOR Module Output

The output of the LPBOR module is a signal indicating whether or not a Reset is to be asserted. This signal is OR’d together with the Reset signal of the BOR module to provide the generic BOR signal, which goes to the PCON register and to the power control block.

5.5 MCLR

The MCLR is an optional external input that can reset the device. The MCLR function is controlled by the MCLRE bit of Configuration Words and the LVP bit of Configuration Words (Table 5-2).

<table>
<thead>
<tr>
<th>MCLRE</th>
<th>LVP</th>
<th>MCLR</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>Disabled</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Enabled</td>
</tr>
<tr>
<td>x</td>
<td>1</td>
<td>Enabled</td>
</tr>
</tbody>
</table>

**Table 5-2: MCLR Configuration**

**Note:** A Reset does not drive the MCLR pin low.

5.5.1 MCLR ENABLED

When MCLR is enabled and the pin is held low, the device is held in Reset. The MCLR pin is connected to VDD through an internal weak pull-up.

The device has a noise filter in the MCLR Reset path. The filter will detect and ignore small pulses.

5.5.2 MCLR DISABLED

When MCLR is disabled, the pin functions as a general purpose input and the internal weak pull-up is under software control. See Section 11.1 "PORTA Registers" for more information.

5.6 Watchdog Timer (WDT) Reset

The Watchdog Timer generates a Reset if the firmware does not issue a CLRWDT instruction within the time-out period. The TO and PD bits in the STATUS register are changed to indicate the WDT Reset. See Section 9.0 “Watchdog Timer (WDT)” for more information.

5.7 RESET Instruction

A RESET instruction will cause a device Reset. The RI bit in the PCON register will be set to ‘0’. See Table 5-4 for default conditions after a RESET instruction has occurred.

5.8 Stack Overflow/Underflow Reset

The device can reset when the Stack Overflows or Underflows. The STKOVF or STKUNF bits of the PCON register indicate the Reset condition. These Resets are enabled by setting the STVREN bit in Configuration Words. See Section 3.6.2 “Overflow/Underflow Reset” for more information.

5.9 Programming Mode Exit

Upon exit of Programming mode, the device will behave as if a POR had just occurred.

5.10 Power-up Timer

The Power-up Timer optionally delays device execution after a BOR or POR event. This timer is typically used to allow VDD to stabilize before allowing the device to start running.

The Power-up Timer is controlled by the PWRT bit of Configuration Words.

5.11 Start-up Sequence

Upon the release of a POR or BOR, the following must occur before the device will begin executing:

1. Power-up Timer runs to completion (if enabled).
2. Oscillator start-up timer runs to completion (if required for oscillator source).
3. MCLR must be released (if enabled).

The total time-out will vary based on oscillator configuration and Power-up Timer configuration. See Section 6.0 “Oscillator Module (with Fail-Safe Clock Monitor)” for more information.

The Power-up Timer and oscillator start-up timer run independently of MCLR Reset. If MCLR is kept low long enough, the Power-up Timer and oscillator start-up timer will expire. Upon bringing MCLR high, the device will begin execution after 10 FOSC cycles (see Figure 5-3). This is useful for testing purposes or to synchronize more than one device operating in parallel.
FIGURE 5-3:  RESET START-UP SEQUENCE

- VDD
- Internal POR
- Power-up Timer
- MCLR
- Internal RESET

---

Oscillator Modes

- Oscillator Start-up Timer
  - Oscillator
  - Fosc

---

External Crystal

- Oscillator Start-up Timer
  - Oscillator
  - Fosc

---

Internal Oscillator

- Oscillator Start-up Timer
  - Oscillator
  - Fosc

---

External Clock (EC)

- CLkin
  - Fosc

---

TSTOP

---

TMCLR

---

TPWRT
5.12 Determining the Cause of a Reset

Upon any Reset, multiple bits in the STATUS and PCON register are updated to indicate the cause of the Reset. Table 5-3 and Table 5-4 show the Reset conditions of these registers.

### Table 5-3: Reset Status Bits and Their Significance

<table>
<thead>
<tr>
<th>STKOVF</th>
<th>STKUNF</th>
<th>RWDT</th>
<th>RMCLR</th>
<th>RI</th>
<th>POR</th>
<th>BOR</th>
<th>TO</th>
<th>PD</th>
<th>Condition</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>x</td>
<td>1</td>
<td>1</td>
<td>Power-on Reset</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>x</td>
<td>0</td>
<td>x</td>
<td>Illegal, TO is set on POR</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>u</td>
<td>1</td>
<td>1</td>
<td>u</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>Brown-out Reset</td>
</tr>
<tr>
<td>u</td>
<td>u</td>
<td>0</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>0</td>
<td>0</td>
<td>WDT Reset</td>
</tr>
<tr>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>0</td>
<td>0</td>
<td>WDT Wake-up from Sleep</td>
</tr>
<tr>
<td>u</td>
<td>u</td>
<td>0</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>0</td>
<td>0</td>
<td>Interrupt Wake-up from Sleep</td>
</tr>
<tr>
<td>u</td>
<td>u</td>
<td>u</td>
<td>1</td>
<td>0</td>
<td>u</td>
<td>u</td>
<td>1</td>
<td>0</td>
<td>MCLR Reset during normal operation</td>
</tr>
<tr>
<td>u</td>
<td>u</td>
<td>u</td>
<td>0</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>MCLR Reset during Sleep</td>
</tr>
<tr>
<td>1</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>RESET Instruction Executed</td>
</tr>
<tr>
<td>u</td>
<td>1</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>Stack Overflow Reset (STVREN = 1)</td>
</tr>
<tr>
<td>u</td>
<td>1</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>u</td>
<td>Stack Underflow Reset (STVREN = 1)</td>
</tr>
</tbody>
</table>

### Table 5-4: Reset Condition for Special Registers

<table>
<thead>
<tr>
<th>Condition</th>
<th>Program Counter</th>
<th>STATUS Register</th>
<th>PCON Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>Power-on Reset</td>
<td>0000h</td>
<td>---1 1000</td>
<td>00-- 110x</td>
</tr>
<tr>
<td>MCLR Reset during normal operation</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>uu-- 0uuu</td>
</tr>
<tr>
<td>MCLR Reset during Sleep</td>
<td>0000h</td>
<td>---1 0uuu</td>
<td>uu-- 0uuu</td>
</tr>
<tr>
<td>WDT Reset</td>
<td>0000h</td>
<td>---0 uuuu</td>
<td>uu-- uuuu</td>
</tr>
<tr>
<td>WDT Wake-up from Sleep</td>
<td>PC + 1</td>
<td>---0 0uuu</td>
<td>uu-- uuuu</td>
</tr>
<tr>
<td>Brown-out Reset</td>
<td>0000h</td>
<td>---1 1uuu</td>
<td>00-- 11u0</td>
</tr>
<tr>
<td>Interrupt Wake-up from Sleep</td>
<td>PC + 1(1)</td>
<td>---1 0uuu</td>
<td>uu-- uuuu</td>
</tr>
<tr>
<td>RESET Instruction Executed</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>uu-- u0uuu</td>
</tr>
<tr>
<td>Stack Overflow Reset (STVREN = 1)</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>1u-- uuuu</td>
</tr>
<tr>
<td>Stack Underflow Reset (STVREN = 1)</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>u1-- uuuu</td>
</tr>
</tbody>
</table>

**Legend:**  
- u = unchanged,  
- x = unknown,  
- = unimplemented bit, reads as ‘0’.  

**Note 1:** When the wake-up is due to an interrupt and Global Enable bit (GIE) is set, the return address is pushed on the stack and PC is loaded with the interrupt vector (0004h) after execution of PC + 1.
5.13 Power Control (PCON) Register

The Power Control (PCON) register contains flag bits to differentiate between a:

- Power-on Reset (POR)
- Brown-out Reset (BOR)
- Reset Instruction Reset (RI)
- MCLR Reset (RMCLR)
- Watchdog Timer Reset (RWDT)
- Stack Underflow Reset (STKUNF)
- Stack Overflow Reset (STKOVF)

The PCON register bits are shown in Register 5-2.

5.14 Register Definitions: Power Control

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 7</th>
<th>STKOVF: Stack Overflow Flag bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>A Stack Overflow occurred</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>A Stack Overflow has not occurred or cleared by firmware</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 6</th>
<th>STKUNF: Stack Underflow Flag bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>A Stack Underflow occurred</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>A Stack Underflow has not occurred or cleared by firmware</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 5</th>
<th>Unimplemented: Read as '0'</th>
</tr>
</thead>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 4</th>
<th>RWDT: Watchdog Timer Reset Flag bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>A Watchdog Timer Reset has not occurred or set to '1' by firmware</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>A Watchdog Timer Reset has occurred (cleared by hardware)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 3</th>
<th>RMCLR: MCLR Reset Flag bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>A MCLR Reset has not occurred or set to '1' by firmware</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>A MCLR Reset has occurred (cleared by hardware)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 2</th>
<th>RI: RESET Instruction Flag bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>A RESET instruction has not been executed or set to '1' by firmware</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>A RESET instruction has been executed (cleared by hardware)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 1</th>
<th>POR: Power-on Reset Status bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>No Power-on Reset occurred</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 0</th>
<th>BOR: Brown-out Reset Status bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>No Brown-out Reset occurred</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>A Brown-out Reset occurred (must be set in software after a Power-on Reset or Brown-out Reset occurs)</td>
</tr>
</tbody>
</table>
### TABLE 5-5: SUMMARY OF REGISTERS ASSOCIATED WITH RESETS

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BORCON</td>
<td>SBOREN</td>
<td>BORFS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>BORRDY 63</td>
</tr>
<tr>
<td>PCON</td>
<td>STKOVF</td>
<td>STKUNF</td>
<td>—</td>
<td>RWDT</td>
<td>RMCLR</td>
<td>RI</td>
<td>POR</td>
<td>BOR</td>
<td>67</td>
</tr>
<tr>
<td>STATUS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC</td>
<td>C</td>
<td>29</td>
</tr>
<tr>
<td>WDTCN</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>WDTPS&lt;4:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>SWDTEN 105</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used by Resets.
6.0 OSCILLATOR MODULE (WITH FAIL-SAFE CLOCK MONITOR)

6.1 Overview

The oscillator module has a wide variety of clock sources and selection features that allow it to be used in a wide range of applications while maximizing performance and minimizing power consumption. Figure 6-1 illustrates a block diagram of the oscillator module.

Clock sources can be supplied from external oscillators, quartz crystal resonators, ceramic resonators and Resistor-Capacitor (RC) circuits. In addition, the system clock source can be supplied from one of two internal oscillators and PLL circuits, with a choice of speeds selectable via software. Additional clock features include:

- Selectable system clock source between external or internal sources via software
- Two-Speed Start-up mode, which minimizes latency between external oscillator start-up and code execution
- Fail-Safe Clock Monitor (FSCM) designed to detect a failure of the external clock source (LP, XT, HS, ECH, ECM, ECL or EXTRC modes) and switch automatically to the internal oscillator
- Oscillator Start-up Timer (OST), which ensures stability of crystal oscillator sources

The oscillator module can be configured in one of the following clock modes:

1. ECL – External Clock Low-Power mode (0 MHz to 0.5 MHz)
2. ECM – External Clock Medium Power mode (0.5 MHz to 4 MHz)
3. ECH – External Clock High-Power mode (4 MHz to 32 MHz)
4. LP – 32 kHz Low-Power Crystal mode.
5. XT – Medium Gain Crystal or Ceramic Resonator Oscillator mode (up to 4 MHz)
6. HS – High Gain Crystal or Ceramic Resonator mode (4 MHz to 20 MHz)
7. EXTRC – External Resistor-Capacitor
8. INTOSC – Internal oscillator (31 kHz to 32 MHz)

Clock Source modes are selected by the FOSC<2:0> bits in the Configuration Words. The FOSC bits determine the type of oscillator that will be used when the device is first powered.

The ECH, ECM, and ECL clock modes rely on an external logic level signal as the device clock source. The LP, XT, and HS clock modes require an external crystal or resonator to be connected to the device.

Each mode is optimized for a different frequency range. The EXTRC clock mode requires an external resistor and capacitor to set the oscillator frequency.

The INTOSC internal oscillator block produces low, medium, and high-frequency clock sources, designated LFINTOSC, MFINTOSC and HFINTOSC. (see Internal Oscillator Block, Figure 6-1). A wide selection of device clock frequencies may be derived from these three clock sources.
FIGURE 6-1: SIMPLIFIED PIC® MCU CLOCK SOURCE BLOCK DIAGRAM

<table>
<thead>
<tr>
<th>Inputs</th>
<th>Outputs</th>
</tr>
</thead>
<tbody>
<tr>
<td>SCS</td>
<td>FOSC&lt;2:0&gt;</td>
</tr>
<tr>
<td>=00</td>
<td>=100</td>
</tr>
<tr>
<td></td>
<td>1</td>
</tr>
<tr>
<td></td>
<td>≠1110</td>
</tr>
<tr>
<td>≠00</td>
<td>=100</td>
</tr>
<tr>
<td></td>
<td>1</td>
</tr>
<tr>
<td>≠00</td>
<td>x</td>
</tr>
</tbody>
</table>

Notes:
- SCS = Sleep Control Select
- FOSC<2:0> = Oscillator Source Select
- PLLEN or SPLLEN = PLL Enable or Slow PLL Enable
- IRCF = Internal Reference Clock Frequency
- PRIMUX = Peripheral Multiplexer
- PLLMUX = PLL Multiplexer

- 16 MHz
- 8 MHz
- 4 MHz
- 2 MHz
- 1 MHz
- 500 kHz
- 250 kHz
- 125 kHz
- 62.5 kHz
- 31.25 kHz
- 31 kHz

- WDT, PWRT, Fail-Safe Clock Monitor
- Two-Speed Start-up and other modules
6.2 Clock Source Types

Clock sources can be classified as external or internal. External clock sources rely on external circuitry for the clock source to function. Examples are: oscillator modules (ECH, ECM, ECL mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (EXTRC) mode circuits.

Internal clock sources are contained within the oscillator module. The internal oscillator block has two internal oscillators and a dedicated Phase-Lock Loop (HFPLL) that are used to generate three internal system clock sources: the 16 MHz High-Frequency Internal Oscillator (HFINTOSC), 500 kHz (MFINTOSC) and the 31 kHz Low-Frequency Internal Oscillator (LFINTOSC).

The system clock can be selected between external or internal clock sources via the System Clock Select (SCS) bits in the OSCCON register. See Section 6.3 “Clock Switching” for additional information.

6.2.1 EXTERNAL CLOCK SOURCES

An external clock source can be used as the device system clock by performing one of the following actions:

- Program the FOSC<2:0> bits in the Configuration Words to select an external clock source that will be used as the default system clock upon a device Reset.
- Write the SCS<1:0> bits in the OSCCON register to switch the system clock source to:
  - Secondary oscillator during run-time, or
  - An external clock source determined by the value of the FOSC bits.

See Section 6.3 “Clock Switching” for more information.

6.2.1.1 EC Mode

The External Clock (EC) mode allows an externally generated logic level signal to be the system clock source. When operating in this mode, an external clock source is connected to the OSC1 input. OSC2/CLKOUT is available for general purpose I/O or CLKOUT. Figure 6-2 shows the pin connections for EC mode.

EC mode has three power modes to select from through Configuration Words:

- ECH – High power, 4-32 MHz
- ECM – Medium power, 0.5-4 MHz
- ECL – Low power, 0-0.5 MHz

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC® MCU design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.

FIGURE 6-2: EXTERNAL CLOCK (EC) MODE OPERATION

6.2.1.2 LP, XT, HS Modes

The LP, XT and HS modes support the use of quartz crystal resonators or ceramic resonators connected to OSC1 and OSC2 (see Figure 6-3). The three modes select a low, medium or high gain setting of the internal inverter-amplifier to support various resonator types and speed.

**LP Oscillator mode** selects the lowest gain setting of the internal inverter-amplifier. LP mode current consumption is the least of the three modes. This mode is designed to drive only 32.768 kHz tuning-fork type crystals (watch crystals).

**XT Oscillator mode** selects the intermediate gain setting of the internal inverter-amplifier. XT mode current consumption is the medium of the three modes. This mode is best suited to drive resonators with a medium drive level specification.

**HS Oscillator mode** selects the highest gain setting of the internal inverter-amplifier. HS mode current consumption is the highest of the three modes. This mode is best suited for resonators that require a high drive setting.

Figure 6-3 and Figure 6-4 show typical circuits for quartz crystal and ceramic resonators, respectively.
6.2.1.3 Oscillator Start-up Timer (OST)

If the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended, unless either FSCM or Two-Speed Start-Up are enabled. In this case, code will continue to execute at the selected INTOSC frequency while the OST is counting. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the oscillator module.

In order to minimize latency between external oscillator start-up and code execution, the Two-Speed Clock Start-up mode can be selected (see Section 6.4 “Two-Speed Clock Start-up Mode”).
6.2.1.4 4x PLL

The oscillator module contains a 4x PLL that can be used with both external and internal clock sources to provide a system clock source. The input frequency for the 4x PLL must fall within specifications. See the PLL Clock Timing Specifications in Table 34-9: PLL Clock Timing Specifications.

The 4x PLL may be enabled for use by one of two methods:

1. Program the PLLEN bit in Configuration Words to a ‘1’.
2. Write the SPLLEN bit in the OSCCON register to a ‘1’. If the PLLEN bit in Configuration Words is programmed to a ‘1’, then the value of SPLLEN is ignored.

6.2.1.5 Secondary Oscillator

The secondary oscillator is a separate crystal oscillator that is associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the SOSCO and SOSCI device pins.

The secondary oscillator can be used as an alternate system clock source and can be selected during run-time using clock switching. Refer to Section 6.3 “Clock Switching” for more information.

Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.

2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

3: For oscillator design assistance, reference the following Microchip Application Notes:
   - AN826, “Crystal Oscillator Basics and Crystal Selection for rfPIC® and PIC® Devices” (DS00826)
   - AN849, “Basic PIC® Oscillator Design” (DS00849)
   - AN943, “Practical PIC® Oscillator Analysis and Design” (DS00943)
   - AN949, “Making Your Oscillator Work” (DS00949)
   - TB097, “Interfacing a Micro Crystal MS1V-T1K 32.768 kHz Tuning Fork Crystal to a PIC16F690/SS” (DS91097)
   - AN1288, “Design Practices for Low-Power External Oscillators” (DS01288)
6.2.1.6 External RC Mode

The external Resistor-Capacitor (EXTRC) mode supports the use of an external RC circuit. This allows the designer maximum flexibility in frequency choice while keeping costs to a minimum when clock accuracy is not required.

The RC circuit connects to OSC1. OSC2/CLKOUT is available for general purpose I/O or CLKOUT. The function of the OSC2/CLKOUT pin is determined by the CLKOUTEN bit in Configuration Words. Figure 6-6 shows the external RC mode connections.

**FIGURE 6-6: EXTERNAL RC MODES**

The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. Other factors affecting the oscillator frequency are:

- Threshold voltage variation
- Component tolerances
- Packaging variations in capacitance

The user also needs to take into account variation due to tolerance of external RC components used.

6.2.2 INTERNAL CLOCK SOURCES

The device may be configured to use the internal oscillator block as the system clock by performing one of the following actions:

- Program the FOSC<2:0> bits in Configuration Words to select the INTOSC clock source, which will be used as the default system clock upon a device Reset.
- Write the SCS<1:0> bits in the OSCCON register to switch the system clock source to the internal oscillator during run-time. See Section 6.3 “Clock Switching” for more information.

In INTOSC mode, OSC1/CLKIN is available for general purpose I/O. OSC2/CLKOUT is available for general purpose I/O or CLKOUT.

The function of the OSC2/CLKOUT pin is determined by the CLKOUTEN bit in Configuration Words.

The internal oscillator block has two independent oscillators and a dedicated Phase-Lock Loop, HFPLL that can produce one of three internal system clock sources.

1. The **HFINTOSC** (High-Frequency Internal Oscillator) is factory calibrated and operates at 16 MHz. The HFINTOSC source is generated from the 500 kHz MFINTOSC source and the dedicated Phase-Lock Loop, HFPLL. The frequency of the HFINTOSC can be user-adjusted via software using the OSCTUNE register (Register 6-3).

2. The **MFINTOSC** (Medium Frequency Internal Oscillator) is factory calibrated and operates at 500 kHz. The frequency of the MFINTOSC can be user-adjusted via software using the OSCTUNE register (Register 6-3).

3. The **LFINTOSC** (Low-Frequency Internal Oscillator) is uncalibrated and operates at 31 kHz.
6.2.2.1 HFINTOSC

The High-Frequency Internal Oscillator (HFINTOSC) is a factory calibrated 16 MHz internal clock source. The frequency of the HFINTOSC can be altered via software using the OSCTUNE register (Register 6-3).

The output of the HFINTOSC connects to a postscaler and multiplexer (see Figure 6-1). One of multiple frequencies derived from the HFINTOSC can be selected via software using the IRCF<3:0> bits of the OSCCON register. See Section 6.2.2.7 “Internal Oscillator Clock Switch Timing” for more information.

The HFINTOSC is enabled by:

• Configure the IRCF<3:0> bits of the OSCCON register for the desired HF frequency, and
• FOSC<2:0> = 100, or
• Set the System Clock Source (SCS) bits of the OSCCON register to ‘1x’

A fast start-up oscillator allows internal circuits to power up and stabilize before switching to HFINTOSC.

The High-Frequency Internal Oscillator Ready bit (HFIOFR) of the OSCSTAT register indicates when the HFINTOSC is running.

The High-Frequency Internal Oscillator Status Locked bit (HFIOFL) of the OSCSTAT register indicates when the HFINTOSC is running within 2% of its final value.

The High-Frequency Internal Oscillator Stable bit (HFIOFS) of the OSCSTAT register indicates when the HFINTOSC is running within 0.5% of its final value.

6.2.2.2 MFINTOSC

The Medium Frequency Internal Oscillator (MFINTOSC) is a factory calibrated 500 kHz internal clock source. The frequency of the MFINTOSC can be altered via software using the OSCTUNE register (Register 6-3).

The output of the MFINTOSC connects to a postscaler and multiplexer (see Figure 6-1). One of nine frequencies derived from the MFINTOSC can be selected via software using the IRCF<3:0> bits of the OSCCON register. See Section 6.2.2.7 “Internal Oscillator Clock Switch Timing” for more information. The MFINTOSC is also the frequency for the Power-up Timer (PWRT), Watchdog Timer (WDT) and Fail-Safe Clock Monitor (FSCM).

The MFINTOSC is enabled by:

• Configure the IRCF<3:0> bits of the OSCCON register for the desired LF frequency, and
• FOSC<2:0> = 100, or
• Set the System Clock Source (SCS) bits of the OSCCON register to ‘1x’

The Medium Frequency Internal Oscillator Ready bit (MFIOFR) of the OSCSTAT register indicates when the MFINTOSC is running.

6.2.2.3 Internal Oscillator Frequency Adjustment

The 500 kHz internal oscillator is factory calibrated. This internal oscillator can be adjusted in software by writing to the OSCTUNE register (Register 6-3). Since the HFINTOSC and MFINTOSC clock sources are derived from the 500 kHz internal oscillator a change in the OSCTUNE register value will apply to both.

The default value of the OSCTUNE register is ‘0’. The value is a 6-bit two’s complement number. A value of 1Fh will provide an adjustment to the maximum frequency. A value of 20h will provide an adjustment to the minimum frequency.

When the OSCTUNE register is modified, the oscillator frequency will begin shifting to the new frequency. Code execution continues during this shift. There is no indication that the shift has occurred.

OSCTUNE does not affect the LFINTOSC frequency. Operation of features that depend on the LFINTOSC clock source frequency, such as the Power-up Timer (PWRT), Watchdog Timer (WDT), Fail-Safe Clock Monitor (FSCM) and peripherals, are not affected by the change in frequency.

6.2.2.4 LFINTOSC

The Low-Frequency Internal Oscillator (LFINTOSC) is an uncalibrated 31 kHz internal clock source.

The output of the LFINTOSC connects to a multiplexer (see Figure 6-1). Select 31 kHz, via software, using the IRCF<3:0> bits of the OSCCON register. See Section 6.2.2.7 “Internal Oscillator Clock Switch Timing” for more information. The LFINTOSC is also the frequency for the Power-up Timer (PWRT), Watchdog Timer (WDT) and Fail-Safe Clock Monitor (FSCM).

The LFINTOSC is enabled by selecting 31 kHz (IRCF<3:0> bits of the OSCCON register = 000) as the system clock source (SCS bits of the OSCCON register = 1x), or when any of the following are enabled:

• Configure the IRCF<3:0> bits of the OSCCON register for the desired LF frequency, and
• FOSC<2:0> = 100, or
• Set the System Clock Source (SCS) bits of the OSCCON register to ‘1x’

Peripherals that use the LFINTOSC are:

• Power-up Timer (PWRT)
• Watchdog Timer (WDT)
• Fail-Safe Clock Monitor (FSCM)

The Low-Frequency Internal Oscillator Ready bit (LFIOR) of the OSCSTAT register indicates when the LFINTOSC is running.
6.2.2.5 Internal Oscillator Frequency Selection

The system clock speed can be selected via software using the Internal Oscillator Frequency Select bits IRCF<3:0> of the OSCCON register.

The postscaled output of the 16 MHz HFINTOSC, 500 kHz MFINTOSC, and 31 kHz LFINTOSC connect to a multiplexer (see Figure 6-1). The Internal Oscillator Frequency Select bits IRCF<3:0> of the OSCCON register select the frequency output of the internal oscillators. One of the following frequencies can be selected via software:

- 32 MHz (requires 4 x PLL)
- 16 MHz
- 8 MHz
- 4 MHz
- 2 MHz
- 1 MHz
- 500 kHz (default after Reset)
- 250 kHz
- 125 kHz
- 62.5 kHz
- 31.25 kHz
- 31 kHz (LFINTOSC)

The IRCF<3:0> bits of the OSCCON register allow duplicate selections for some frequencies. These duplicate choices can offer system design trade-offs. Lower power consumption can be obtained when changing oscillator sources for a given frequency. Faster transition times can be obtained between frequency changes that use the same oscillator source.

Note: Following any Reset, the IRCF<3:0> bits of the OSCCON register are set to '0111' and the frequency selection is set to 500 kHz. The user can modify the IRCF bits to select a different frequency.

6.2.2.6 32 MHz Internal Oscillator Frequency Selection

The Internal Oscillator Block can be used with the 4x PLL associated with the External Oscillator Block to produce a 32 MHz internal system clock source. The following settings are required to use the 32 MHz internal clock source:

- The FOSC bits in Configuration Words must be set to use the INTOSC source as the device system clock (FOSC<2:0> = 100).
- The SCS bits in the OSCCON register must be cleared to use the clock determined by FOSC<2:0> in Configuration Words (SCS<1:0> = 00).
- The IRCF bits in the OSCCON register must be set to the 8 MHz HFINTOSC set to use (IRCF<3:0> = 1110).
- The SPLLEN bit in the OSCCON register must be set to enable the 4x PLL, or the PLLEN bit of the Configuration Words must be programmed to a '1'.

Note: When using the PLLEN bit of the Configuration Words, the 4x PLL cannot be disabled by software and the SPLLEN option will not be available.

The 4x PLL is not available for use with the internal oscillator when the SCS bits of the OSCCON register are set to '1x'. The SCS bits must be set to '00' to use the 4x PLL with the internal oscillator.
6.2.2.7 Internal Oscillator Clock Switch Timing

When switching between the HFINTOSC, MFINTOSC and the LFINTOSC, the new oscillator may already be shut down to save power (see Figure 6-7). If this is the case, there is a delay after the IRCF<3:0> bits of the OSCCON register are modified before the frequency selection takes place. The OSCSTAT register will reflect the current active status of the HFINTOSC, MFINTOSC and LFINTOSC oscillators. The sequence of a frequency selection is as follows:

1. IRCF<3:0> bits of the OSCCON register are modified.
2. If the new clock is shut down, a clock start-up delay is started.
3. Clock switch circuitry waits for a falling edge of the current clock.
4. The current clock is held low and the clock switch circuitry waits for a rising edge in the new clock.
5. The new clock is now active.
6. The OSCSTAT register is updated as required.
7. Clock switch is complete.

See Figure 6-7 for more details.

If the internal oscillator speed is switched between two clocks of the same source, there is no start-up delay before the new frequency is selected. Clock switching time delays are shown in Table 6-1.

Start-up delay specifications are located in the oscillator tables of Section 34.0 “Electrical Specifications”.
FIGURE 6-7: INTERNAL OSCILLATOR SWITCH TIMING

HFINTOSC/→ LFINTOSC (FSCM and WDT disabled)
MFINTOSC

HFINTOSC/→ LFINTOSC (Either FSCM or WDT enabled)
MFINTOSC

LFINTOSC → HFINTOSC/MFINTOSC

HFINTOSC/→ LFINTOSC (Either FSCM or WDT enabled)
MFINTOSC

LFINTOSC turns off unless WDT or FSCM is enabled
6.3 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the System Clock Select (SCS) bits of the OSCCON register. The following clock sources can be selected using the SCS bits:

- Default system oscillator determined by FOSC bits in Configuration Words
- Timer1 32 kHz crystal oscillator
- Internal Oscillator Block (INTOSC)

6.3.1 SYSTEM CLOCK SELECT (SCS) BITS

The System Clock Select (SCS) bits of the OSCCON register select the system clock source that is used for the CPU and peripherals.

- When the SCS bits of the OSCCON register = 00, the system clock source is determined by the value of the FOSC<2:0> bits in the Configuration Words.
- When the SCS bits of the OSCCON register = 01, the system clock source is the secondary oscillator.
- When the SCS bits of the OSCCON register = 1x, the system clock source is chosen by the internal oscillator frequency selected by the IRCF<3:0> bits of the OSCCON register. After a Reset, the SCS bits of the OSCCON register are always cleared.

Note: Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bits of the OSCCON register. The user can monitor the OSTS bit of the OSCSTAT register to determine the current system clock source.

When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 6-1.

6.3.2 OSCILLATOR START-UP TIMER STATUS (OSTS) BIT

The Oscillator Start-up Timer Status (OSTS) bit of the OSCSTAT register indicates whether the system clock is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Words, or from the internal clock source. In particular, OSTS indicates that the Oscillator Start-up Timer (OST) has timed out for LP, XT or HS modes. The OST does not reflect the status of the secondary oscillator.

6.3.3 SECONDARY OSCILLATOR

The secondary oscillator is a separate crystal oscillator associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the SOSCO and SOSCI device pins.

The secondary oscillator is enabled using the T1OSCEN control bit in the T1CON register. See Section 27.0 “Timer1 Module with Gate Control” for more information about the Timer1 peripheral.

6.3.4 SECONDARY OSCILLATOR READY (SOSCR) BIT

The user must ensure that the secondary oscillator is ready to be used before it is selected as a system clock source. The Secondary Oscillator Ready (SOSCR) bit of the OSCSTAT register indicates whether the secondary oscillator is ready to be used. After the SOSCR bit is set, the SCS bits can be configured to select the secondary oscillator.

6.3.5 CLOCK SWITCHING BEFORE SLEEP

When clock switching from an old clock to a new clock is requested just prior to entering Sleep mode, it is necessary to confirm that the switch is complete before the SLEEP instruction is executed. Failure to do so may result in an incomplete switch and consequential loss of the system clock altogether. Clock switching is confirmed by monitoring the clock status bits in the OSCSTAT register. Switch confirmation can be accomplished by sensing that the ready bit for the new clock is set or the ready bit for the old clock is cleared.

For example, when switching between the internal oscillator with the PLL and the internal oscillator without the PLL, monitor the PLLR bit. When PLLR is set, the switch to 32 MHz operation is complete. Conversely, when PLLR is cleared, the switch from 32 MHz operation to the selected internal clock is complete.

Note: Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bits of the OSCCON register. The user can monitor the OSTS bit of the OSCSTAT register to determine the current system clock source.
6.4 Two-Speed Clock Start-up Mode

Two-Speed Start-up mode provides additional power savings by minimizing the latency between external oscillator start-up and code execution. In applications that make heavy use of the Sleep mode, Two-Speed Start-up will remove the external oscillator start-up time from the time spent awake and can reduce the overall power consumption of the device. This mode allows the application to wake-up from Sleep, perform a few instructions using the INTOSC internal oscillator block as the clock source and go back to Sleep without waiting for the external oscillator to become stable.

Two-Speed Start-up provides benefits when the oscillator module is configured for LP, XT or HS modes. The Oscillator Start-up Timer (OST) is enabled for these modes and must count 1024 oscillations before the oscillator can be used as the system clock source.

If the oscillator module is configured for any mode other than LP, XT or HS mode, then Two-Speed Start-up is disabled. This is because the external clock oscillator does not require any stabilization time after POR or an exit from Sleep.

If the OST count reaches 1024 before the device enters Sleep mode, the OSTS bit of the OSCSTAT register is set and program execution switches to the external oscillator. However, the system may never operate from the external oscillator if the time spent awake is very short.

Note: Executing a SLEEP instruction will abort the oscillator start-up time and will cause the OSTS bit of the OSCSTAT register to remain clear.

---

### TABLE 6-1: OSCILLATOR SWITCHING DELAYS

<table>
<thead>
<tr>
<th>Switch From</th>
<th>Switch To</th>
<th>Frequency</th>
<th>Oscillator Delay</th>
</tr>
</thead>
<tbody>
<tr>
<td>Sleep</td>
<td>LFINTOSC(^{(1)})</td>
<td>31 kHz</td>
<td>Oscillator Warm-up Delay (TWARM)(^{(2)})</td>
</tr>
<tr>
<td></td>
<td>MFINTOSC(^{(1)})</td>
<td>31.25 kHz-500 kHz</td>
<td></td>
</tr>
<tr>
<td></td>
<td>HFINTOSC(^{(1)})</td>
<td>31.25 kHz-16 MHz</td>
<td></td>
</tr>
<tr>
<td>Sleep/POR</td>
<td>EC, RC(^{(1)})</td>
<td>DC – 32 MHz</td>
<td>2 cycles</td>
</tr>
<tr>
<td>LFINTOSC</td>
<td>EC, RC(^{(1)})</td>
<td>DC – 32 MHz</td>
<td>1 cycle of each</td>
</tr>
<tr>
<td>Sleep/POR</td>
<td>Secondary Oscillator</td>
<td>32 kHz-20 MHz</td>
<td>1024 Clock Cycles (OST)</td>
</tr>
<tr>
<td>Any clock source</td>
<td>MFINTOSC(^{(1)})</td>
<td>31.25 kHz-500 kHz</td>
<td>2 (\mu) s (approx.)</td>
</tr>
<tr>
<td></td>
<td>HFINTOSC(^{(1)})</td>
<td>31.25 kHz-16 MHz</td>
<td></td>
</tr>
<tr>
<td>Any clock source</td>
<td>LFINTOSC(^{(1)})</td>
<td>31 kHz</td>
<td>1 cycle of each</td>
</tr>
<tr>
<td>Any clock source</td>
<td>Secondary Oscillator</td>
<td>32 kHz</td>
<td>1024 Clock Cycles (OST)</td>
</tr>
<tr>
<td>PLL inactive</td>
<td>PLL active</td>
<td>16-32 MHz</td>
<td>2 ms (approx.)</td>
</tr>
</tbody>
</table>

---

Note 1: PLL inactive.  
Note 2: See Section 34.0 “Electrical Specifications”.

---

6.4.1 TWO-SPEED START-UP MODE CONFIGURATION

Two-Speed Start-up mode is configured by the following settings:

- IESO (of the Configuration Words) = 1; Internal/External Switchover bit (Two-Speed Start-up mode enabled).
- SCS (of the OSCCON register) = 00.
- FOSC\(<2:0>\) bits in the Configuration Words configured for LP, XT or HS mode.

Two-Speed Start-up mode is entered after:

- Power-on Reset (POR) and, if enabled, after Power-up Timer (PWRT) has expired, or
- Wake-up from Sleep.
6.4.2 TWO-SPEED START-UP SEQUENCE

1. Wake-up from Power-on Reset or Sleep.
2. Instructions begin execution by the internal oscillator at the frequency set in the IRCF<3:0> bits of the OSCCON register.
3. OST enabled to count 1024 clock cycles.
4. OST timed out, wait for falling edge of the internal oscillator.
5. OSTS is set.
6. System clock held low until the next falling edge of new clock (LP, XT or HS mode).
7. System clock is switched to external clock source.

6.4.3 CHECKING TWO-SPEED CLOCK STATUS

Checking the state of the OSTS bit of the OSCSTAT register will confirm if the microcontroller is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Words, or the internal oscillator.

FIGURE 6-8: TWO-SPEED START-UP
6.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the device to continue operating should the external oscillator fail. The FSCM can detect oscillator failure any time after the Oscillator Start-up Timer (OST) has expired. The FSCM is enabled by setting the FCMEN bit in the Configuration Words. The FSCM is applicable to all external Oscillator modes (LP, XT, HS, EC, Secondary Oscillator and RC).

**FIGURE 6-9: FSCM BLOCK DIAGRAM**

![FSCM Block Diagram]

6.5.1 FAIL-SAFE DETECTION

The FSCM module detects a failed oscillator by comparing the external oscillator to the FSCM sample clock. The sample clock is generated by dividing the LFINTOSC by 64. See Figure 6-9. Inside the fail detector block is a latch. The external clock sets the latch on each falling edge of the external clock. The sample clock clears the latch on each rising edge of the sample clock. A failure is detected when an entire half-cycle of the sample clock elapses before the external clock goes low.

6.5.2 FAIL-SAFE OPERATION

When the external clock fails, the FSCM switches the device clock to an internal clock source and sets the bit flag OSFIF of the PIR2 register. Setting this flag will generate an interrupt if the OSFIE bit of the PIE2 register is also set. The device firmware can then take steps to mitigate the problems that may arise from a failed clock. The system clock will continue to be sourced from the internal clock source until the device firmware successfully restarts the external oscillator and switches back to external operation.

The internal clock source chosen by the FSCM is determined by the IRCF<3:0> bits of the OSCCON register. This allows the internal oscillator to be configured before a failure occurs.

6.5.3 FAIL-SAFE CONDITION CLEARING

The Fail-Safe condition is cleared after a Reset, executing a SLEEP instruction or changing the SCS bits of the OSCCON register. When the SCS bits are changed, the OST is restarted. While the OST is running, the device continues to operate from the INTOSC selected in OSCCON. When the OST times out, the Fail-Safe condition is cleared after successfully switching to the external clock source. The OSFIF bit should be cleared prior to switching to the external clock source. If the Fail-Safe condition still exists, the OSFIF flag will again become set by hardware.

6.5.4 RESET OR WAKE-UP FROM SLEEP

The FSCM is designed to detect an oscillator failure after the Oscillator Start-up Timer (OST) has expired. The OST is used after waking up from Sleep and after any type of Reset. The OST is not used with the EC or RC Clock modes so that the FSCM will be active as soon as the Reset or wake-up has completed. When the FSCM is enabled, the Two-Speed Start-up is also enabled. Therefore, the device will always be executing code while the OST is operating.

**Note:** Due to the wide range of oscillator start-up times, the Fail-Safe circuit is not active during oscillator start-up (i.e., after exiting Reset or Sleep). After an appropriate amount of time, the user should check the Status bits in the OSCSTAT register to verify the oscillator start-up and that the system clock switchover has successfully completed.
FIGURE 6-10: FSCM TIMING DIAGRAM

Note: The system clock is normally at a much higher frequency than the sample clock. The relative frequencies in this example have been chosen for clarity.
## 6.6 Register Definitions: Oscillator Control

### REGISTER 6-1: OSCCON: OSCILLATOR CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>SPLLEN</td>
<td>IRCF&lt;3:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

#### bit 7 SPLLEN: Software PLL Enable bit
- If PLL in Configuration Words = 1:
  - SPLLEN bit is ignored. 4x PLL is always enabled (subject to oscillator requirements)
- If PLL in Configuration Words = 0:
  - 1 = 4x PLL is enabled
  - 0 = 4x PLL is disabled

#### bit 6-3 IRCF<3:0>: Internal Oscillator Frequency Select bits
- 1111 = 16 MHz HF
- 1110 = 8 MHz or 32 MHz HF(2)
- 1101 = 4 MHz HF
- 1100 = 2 MHz HF
- 1011 = 1 MHz HF
- 1010 = 500 kHz HF(1)
- 1001 = 250 kHz HF(1)
- 1000 = 125 kHz HF(1)
- 0111 = 500 kHz MF (default upon Reset)
- 0110 = 250 kHz MF
- 0101 = 125 kHz MF
- 0100 = 62.5 kHz MF
- 0011 = 31.25 kHz HF(1)
- 0010 = 31.25 kHz MF
- 000x = 31 kHz LF

#### bit 2 Unimplemented: Read as ‘0’

#### bit 1-0 SCS<1:0>: System Clock Select bits
- 1x = Internal oscillator block
- 01 = Secondary oscillator
- 00 = Clock determined by FOSC<2:0> in Configuration Words

**Note 1:** Duplicate frequency derived from HFINTOSC.

**Note 2:** 32 MHz when SPLLEN bit is set. Refer to Section 6.2.2.6 “32 MHz Internal Oscillator Frequency Selection”.

---

**Note:**
- Duplicate frequency derived from HFINTOSC.
- 32 MHz when SPLLEN bit is set. Refer to Section 6.2.2.6 “32 MHz Internal Oscillator Frequency Selection”.

---

DS40001740C-page 84 © 2014-2018 Microchip Technology Inc.
## REGISTER 6-2: OSCSTAT: OSCILLATOR STATUS REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>SOSCR</td>
<td>Secondary Oscillator Ready bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>If T1OSCEN = 1:</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = Secondary oscillator is ready</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = Secondary oscillator is not ready</td>
</tr>
<tr>
<td></td>
<td></td>
<td>If T1OSCEN = 0:</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = Secondary clock source is always ready</td>
</tr>
<tr>
<td>6</td>
<td>PLLR</td>
<td>4x PLL Ready bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = 4x PLL is ready</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = 4x PLL is not ready</td>
</tr>
<tr>
<td>5</td>
<td>OSTS</td>
<td>Oscillator Start-up Timer Status bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = Running from the clock defined by the FOSC&lt;2:0&gt; bits of the Configuration Words</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = Running from an internal oscillator (FOSC&lt;2:0&gt; = 100)</td>
</tr>
<tr>
<td>4</td>
<td>HFIOFR</td>
<td>High-Frequency Internal Oscillator Ready bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = HFINTOSC is ready</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = HFINTOSC is not ready</td>
</tr>
<tr>
<td>3</td>
<td>HFIOFL</td>
<td>High-Frequency Internal Oscillator Locked bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = HFINTOSC is at least 2% accurate</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = HFINTOSC is not 2% accurate</td>
</tr>
<tr>
<td>2</td>
<td>MFIOFR</td>
<td>Medium Frequency Internal Oscillator Ready bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = MFINTOSC is ready</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = MFINTOSC is not ready</td>
</tr>
<tr>
<td>1</td>
<td>LFIOFR</td>
<td>Low-Frequency Internal Oscillator Ready bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = LFINTOSC is ready</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = LFINTOSC is not ready</td>
</tr>
<tr>
<td>0</td>
<td>HFIOFS</td>
<td>High-Frequency Internal Oscillator Stable bit</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1 = HFINTOSC is at least 0.5% accurate</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0 = HFINTOSC is not 0.5% accurate</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **q** = Conditional
REGISTER 6-3:  OSCTUNE: OSCILLATOR TUNING REGISTER

| Bit 7-6 | Unimplemented: Read as ‘0’ |
| Bit 5-0 | TUN<5:0>: Frequency Tuning bits |
| 100000 | Minimum frequency |
| • • • • | |
| 111111 | Oscillator module is running at the factory-calibrated frequency |
| 000000 | |
| • • • • | |
| 011110 | Maximum frequency |
| 011111 | |

TABLE 6-2:  SUMMARY OF REGISTERS ASSOCIATED WITH CLOCK SOURCES

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>OSCCON</td>
<td>SPLLEN</td>
<td>IRCF&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
<td>SCS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>OSCSTAT</td>
<td>SOSCR</td>
<td>PLLR</td>
<td>OSTS</td>
<td>HFIOFR</td>
<td>HFIOFL</td>
<td>MFOFR</td>
<td>LFIOFR</td>
<td>HFIOFS</td>
<td>85</td>
</tr>
<tr>
<td>OSCTUNE</td>
<td>—</td>
<td>—</td>
<td>TUN&lt;5:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td>96</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
<td>93</td>
</tr>
<tr>
<td>T1CON</td>
<td>TM1CS&lt;1:0&gt;</td>
<td>T1CKPS&lt;1:0&gt;</td>
<td>T1OSCEN</td>
<td>T1SYNC</td>
<td>—</td>
<td>TMR1ON</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:  — = unimplemented location, read as ‘0’. Shaded cells are not used by clock sources.

TABLE 6-3:  SUMMARY OF CONFIGURATION WORD WITH CLOCK SOURCES

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit -7</th>
<th>Bit -6</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG1</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>FCMEN</td>
<td>IESO</td>
<td>CLKOUTEN</td>
<td>BOREN&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>56</td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>CP</td>
<td>MCLRE</td>
<td>PWRT</td>
<td>WDTE&lt;1:0&gt;</td>
<td>—</td>
<td>FOSC&lt;2:0&gt;</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:  — = unimplemented location, read as ‘0’. Shaded cells are not used by clock sources.
7.0 INTERRUPTS

The interrupt feature allows certain events to preempt normal program flow. Firmware is used to determine the source of the interrupt and act accordingly. Some interrupts can be configured to wake the MCU from Sleep mode.

This chapter contains the following information for Interrupts:
- Operation
- Interrupt Latency
- Interrupts During Sleep
- INT Pin
- Automatic Context Saving

Many peripherals produce interrupts. Refer to the corresponding chapters for details.

A block diagram of the interrupt logic is shown in Figure 7-1.

FIGURE 7-1: INTERRUPT LOGIC
7.1 Operation

Interrupts are disabled upon any device Reset. They are enabled by setting the following bits:

- GIE bit of the INTCON register
- Interrupt Enable bit(s) for the specific interrupt event(s)
- PEIE bit of the INTCON register (if the Interrupt Enable bit of the interrupt event is contained in the PIE1 or PIE2 registers)

The INTCON, PIR1 and PIR2 registers record individual interrupts via interrupt flag bits. Interrupt flag bits will be set, regardless of the status of the GIE, PEIE and individual interrupt enable bits.

The following events happen when an interrupt event occurs while the GIE bit is set:

- Current prefetched instruction is flushed
- GIE bit is cleared
- Current Program Counter (PC) is pushed onto the stack
- Critical registers are automatically saved to the shadow registers (See "Section 7.5 Automatic Context Saving")
- PC is loaded with the interrupt vector 0004h

The firmware within the Interrupt Service Routine (ISR) should determine the source of the interrupt by polling the interrupt flag bits. The interrupt flag bits must be cleared before exiting the ISR to avoid repeated interrupts. Because the GIE bit is cleared, any interrupt that occurs while executing the ISR will be recorded through its interrupt flag, but will not cause the processor to redirect to the interrupt vector.

The RETFIE instruction exits the ISR by popping the previous address from the stack, restoring the saved context from the shadow registers and setting the GIE bit.

For additional information on a specific interrupt's operation, refer to its peripheral chapter.

Note 1: Individual interrupt flag bits are set, regardless of the state of any other enable bits.

2: All interrupts will be ignored while the GIE bit is cleared. Any interrupt occurring while the GIE bit is clear will be serviced when the GIE bit is set again.

7.2 Interrupt Latency

Interrupt latency is defined as the time from when the interrupt event occurs to the time code execution at the interrupt vector begins. The latency for synchronous interrupts is three or four instruction cycles. For asynchronous interrupts, the latency is three to five instruction cycles, depending on when the interrupt occurs. See Figure 7-2 and Figure 7-3 for more details.
FIGURE 7-2: INTERRUPT LATENCY

osc1

q1 q2 q3 q4

CLKR

q1 q2 q3 q4

Interrupt Sampled during Q1

GIE

pc

1 cycle instruction at PC

1 Cycle Instruction at PC

inst(pc)

nop

nop

inst(0004h)

execute

interrupt

GIE

pc

2 cycle instruction at PC

2 Cycle Instruction at PC

inst(pc)

nop

nop

inst(0004h)

execute

interrupt

GIE

pc

3 cycle instruction at PC

3 Cycle Instruction at PC

inst(pc)

nop

nop

nop

inst(0004h) inst(0005h)

execute

interrupt

GIE

pc

3 cycle instruction at PC

3 Cycle Instruction at PC

inst(pc)

nop

nop

nop

nop

inst(0004h)
7.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR. Refer to Section 8.0 “Power-Down Mode (Sleep)” for more details.

7.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. This interrupt is enabled by setting the INTE bit of the INTCON register. The INTEDG bit of the OPTION_REG register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector.

7.5 Automatic Context Saving

Upon entering an interrupt, the return PC address is saved on the stack. Additionally, the following registers are automatically saved in the shadow registers:

- W register
- STATUS register (except for TO and PD)
- BSR register
- FSR registers
- PCLATH register

Upon exiting the Interrupt Service Routine, these registers are automatically restored. Any modifications to these registers during the ISR will be lost. If modifications to any of these registers are desired, the corresponding shadow register should be modified and the value will be restored when exiting the ISR. The shadow registers are available in Bank 31 and are readable and writable. Depending on the user’s application, other registers may also need to be saved.
7.6 Register Definitions: Interrupt Control

REGISTER 7-1: INTCON: INTERRUPT CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>GIE: Global Interrupt Enable bit</td>
<td>1 = Enables all active interrupts 0 = Disables all interrupts</td>
</tr>
<tr>
<td>6</td>
<td>PEIE: Peripheral Interrupt Enable bit</td>
<td>1 = Enables all active peripheral interrupts 0 = Disables all peripheral interrupts</td>
</tr>
<tr>
<td>5</td>
<td>TMR0IE: Timer0 Overflow Interrupt Enable bit</td>
<td>1 = Enables the Timer0 interrupt 0 = Disables the Timer0 interrupt</td>
</tr>
<tr>
<td>4</td>
<td>INTE: INT External Interrupt Enable bit</td>
<td>1 = Enables the INT external interrupt 0 = Disables the INT external interrupt</td>
</tr>
<tr>
<td>3</td>
<td>IOCIE: Interrupt-on-Change Enable bit</td>
<td>1 = Enables the interrupt-on-change 0 = Disables the interrupt-on-change</td>
</tr>
<tr>
<td>2</td>
<td>TMR0IF: Timer0 Overflow Interrupt Flag bit</td>
<td>1 = TMR0 register has overflowed 0 = TMR0 register did not overflow</td>
</tr>
<tr>
<td>1</td>
<td>INTF: INT External Interrupt Flag bit</td>
<td>1 = The INT external interrupt occurred 0 = The INT external interrupt did not occur</td>
</tr>
<tr>
<td>0</td>
<td>IOCIF: Interrupt-on-Change Interrupt Flag bit(1)</td>
<td>1 = When at least one of the interrupt-on-change pins changed state 0 = None of the interrupt-on-change pins have changed state</td>
</tr>
</tbody>
</table>

Note 1: The IOCIF Flag bit is read-only and cleared when all the interrupt-on-change flags in the IOCxF registers have been cleared by software.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.
## REGISTER 7-2: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>TMR1GIE: Timer1 Gate Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the Timer1 gate acquisition interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the Timer1 gate acquisition interrupt</td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>ADIE: Analog-to-Digital Converter (ADC) Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the ADC interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the ADC interrupt</td>
<td></td>
</tr>
<tr>
<td>5</td>
<td>RCIE: USART Receive Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the USART receive interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the USART receive interrupt</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>TXIE: USART Transmit Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the USART transmit interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the USART transmit interrupt</td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>SSP1IE: Synchronous Serial Port (MSSP) Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the MSSP interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the MSSP interrupt</td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>CCP1IE: CCP1 Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the CCP1 interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the CCP1 interrupt</td>
<td></td>
</tr>
<tr>
<td>1</td>
<td>TMR2IE: TMR2 to PR2 Match Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the Timer2 to PR2 match interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the Timer2 to PR2 match interrupt</td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>TMR1IE: Timer1 Overflow Interrupt Enable bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables the Timer1 overflow interrupt</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables the Timer1 overflow interrupt</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**Note:** Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.
### REGISTER 7-3: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>OSFIE: Oscillator Fail Interrupt Enable bit</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>U-0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>6</td>
<td>C2IE: Comparator C2 Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>-</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>5</td>
<td>C1IE: Comparator C1 Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>-</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>4</td>
<td>Unimplemented: Read as ‘0’</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>-</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>BCL1IE: MSSP Bus Collision Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>-</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>2</td>
<td>TMR6IE: TMR6 to PR6 Match Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>-</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>TMR4IE: TMR4 to PR4 Match Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>-</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>CCP2IE: CCP2 Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>-</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

**Note:** Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.
REGISTER 7-4: PIE3: PERIPHERAL INTERRUPT ENABLE REGISTER 3

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>Unimplemented</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>6</td>
<td>NCOIE: NCO Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>5</td>
<td>COGIE: COG Auto-Shutdown Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>4</td>
<td>ZCDIE: Zero-Cross Detection Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>CLC4IE: CLC4 Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>2</td>
<td>CLC3IE: CLC3 Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>CLC2IE: CLC2 Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>CLC1IE: CLC1 Interrupt Enable bit</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
</tbody>
</table>

Legend:

R = Readable bit
W = Writable bit
U = Unimplemented bit, read as '0'
u = Bit is unchanged
x = Bit is unknown
-\(n/n\) = Value at POR and BOR/Value at all other Resets

'1' = Bit is set
'0' = Bit is cleared

Note: Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.
### REGISTER 7-5:  PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

<table>
<thead>
<tr>
<th>Bit</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **1** = Bit is set
- **0** = Bit is cleared

- **bit 7** TMR1GIF: Timer1 Gate Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 6** ADIF: Analog-to-Digital Converter (ADC) Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 5** RCIF: USART Receive Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 4** TXIF: USART Transmit Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 3** SSP1IF: Synchronous Serial Port (MSSP) Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 2** CCP1IF: CCP1 Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 1** TMR2IF: Timer2 to PR2 Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 0** TMR1IF: Timer1 Overflow Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

**Note:** Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.
### REGISTER 7-6: PIR2: PERIPHERAL INTERRUPT REQUEST REGISTER 2

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>___</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
</tr>
</tbody>
</table>

#### Legend:
- **R** = Readable bit
- **W** =Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **'1'** = Bit is set
- **'0'** = Bit is cleared

#### bit 7
- **OSFIF**: Oscillator Fail Interrupt Flag bit
  - **1** = Interrupt is pending
  - **0** = Interrupt is not pending

#### bit 6
- **C2IF**: Comparator C2 Interrupt Flag bit
  - **1** = Interrupt is pending
  - **0** = Interrupt is not pending

#### bit 5
- **C1IF**: Comparator C1 Interrupt Flag bit
  - **1** = Interrupt is pending
  - **0** = Interrupt is not pending

#### bit 4
- **Unimplemented**: Read as ‘0’

#### bit 3
- **BCL1IF**: MSSP Bus Collision Interrupt Flag bit
  - **1** = Interrupt is pending
  - **0** = Interrupt is not pending

#### bit 2
- **TMR6IF**: Timer6 to PR6 Interrupt Flag bit
  - **1** = Interrupt is pending
  - **0** = Interrupt is not pending

#### bit 1
- **TMR4IF**: Timer4 to PR4 Interrupt Flag bit
  - **1** = Interrupt is pending
  - **0** = Interrupt is not pending

#### bit 0
- **CCP2IF**: CCP2 Interrupt Flag bit
  - **1** = Interrupt is pending
  - **0** = Interrupt is not pending

#### Note:
Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.
## REGISTER 7-7: PIR3: PERIPHERAL INTERRUPT REQUEST REGISTER 3

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Description</th>
<th>Value at POR and BOR/Value at all other Resets</th>
<th>User Software</th>
<th>Note</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>NCOIF</td>
<td>NCO Interrupt Flag bit</td>
<td>-n/n</td>
<td></td>
<td>Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.</td>
</tr>
<tr>
<td>6</td>
<td>COGIF</td>
<td>COG Auto-Shutdown Interrupt Flag bit</td>
<td>-n/n</td>
<td></td>
<td></td>
</tr>
<tr>
<td>5</td>
<td>ZCDIF</td>
<td>Zero-Cross Detection Interrupt Flag bit</td>
<td>-n/n</td>
<td></td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>CLC4IF</td>
<td>CLC4 Interrupt Flag bit</td>
<td>-n/n</td>
<td></td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>CLC3IF</td>
<td>CLC3 Interrupt Flag bit</td>
<td>-n/n</td>
<td></td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>CLC2IF</td>
<td>CLC2 Interrupt Flag bit</td>
<td>-n/n</td>
<td></td>
<td></td>
</tr>
<tr>
<td>1</td>
<td>CLC1IF</td>
<td>CLC1 Interrupt Flag bit</td>
<td>-n/n</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0</td>
<td></td>
<td></td>
<td>-n/n</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Legend:</th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>R =</td>
<td>Readable bit</td>
<td>W = Writable bit</td>
<td>U = Unimplemented bit, read as '0'</td>
</tr>
<tr>
<td>u =</td>
<td>Bit is unchanged</td>
<td>x = Bit is unknown</td>
<td>-n/n = Value at POR and BOR/Value at all other Resets</td>
</tr>
<tr>
<td>'1' =</td>
<td>Bit is set</td>
<td>'0' = Bit is cleared</td>
<td></td>
</tr>
</tbody>
</table>

**Note:** Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.
## TABLE 7-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
<td>OPTION_REG</td>
<td>WPUEN</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS&lt;2:0&gt;</td>
<td></td>
<td></td>
<td>273</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
<td>93</td>
</tr>
<tr>
<td>PIE3</td>
<td>—</td>
<td>NCOIE</td>
<td>COGIE</td>
<td>ZCDIE</td>
<td>CLC4IE</td>
<td>CLC3IE</td>
<td>CLC2IE</td>
<td>CLC1IE</td>
<td>94</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>95</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td>96</td>
</tr>
<tr>
<td>PIR3</td>
<td>—</td>
<td>NCOIF</td>
<td>COGIF</td>
<td>ZCIF</td>
<td>CLC4IF</td>
<td>CLC3IF</td>
<td>CLC2IF</td>
<td>CLC1IF</td>
<td>97</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used by interrupts.
8.0 POWER-DOWN MODE (SLEEP)

The Power-down mode is entered by executing a SLEEP instruction.

Upon entering Sleep mode, the following conditions exist:

1. WDT will be cleared but keeps running, if
   enabled for operation during Sleep.
2. PD bit of the STATUS register is cleared.
3. TO bit of the STATUS register is set.
4. CPU clock is disabled.
5. 31 kHz LFINTOSC is unaffected and peripherals
   that operate from it may continue operation in
   Sleep.
6. Timer1 and peripherals that operate from
   Timer1 continue operation in Sleep when the
   Timer1 clock source selected is:
   • LFINTOSC
   • T1CKI
   • Secondary oscillator
7. ADC is unaffected, if the dedicated FRC
   oscillator is selected.
8. I/O ports maintain the status they had before
   SLEEP was executed (driving high, low or
   high-impedance).
9. Resets other than WDT are not affected by
   Sleep mode.

Refer to individual chapters for more details on
peripheral operation during Sleep.

To minimize current consumption, the following
conditions should be considered:

• I/O pins should not be floating
• External circuitry sinking current from I/O pins
• Internal circuitry sourcing current from I/O pins
• Current draw from pins with internal weak pull-ups
• Modules using 31 kHz LFINTOSC
• Modules using secondary oscillator

I/O pins that are high-impedance inputs should be
pulled to VDD or VSS externally to avoid switching
currents caused by floating inputs.

Examples of internal circuitry that might be sourcing
current include modules such as the DAC and FVR
modules. See Section 22.0 “Operational Amplifier
(OPA) Modules” and Section 14.0 “Fixed Voltage
Reference (FVR)” for more information on these
modules.

8.1 Wake-up from Sleep

The device can wake-up from Sleep through one of the
following events:

1. External Reset input on MCLR pin, if enabled
2. BOR Reset, if enabled
3. POR Reset
4. Watchdog Timer, if enabled
5. Any external interrupt
6. Interrupts by peripherals capable of running
   during Sleep (see individual peripheral for more
   information)

The first three events will cause a device Reset. The
last three events are considered a continuation of
program execution. To determine whether a device
Reset or wake-up event occurred, refer to
Section 5.12 “Determining the Cause of a Reset”.

When the SLEEP instruction is being executed, the next
instruction (PC + 1) is prefetched. For the device to
wake-up through an interrupt event, the corresponding
interrupt enable bit must be enabled. Wake-up will
occur regardless of the state of the GIE bit. If the GIE
bit is disabled, the device continues execution at the
instruction after the SLEEP instruction. If the GIE bit is
enabled, the device executes the instruction after the
SLEEP instruction, the device will then call the Interrupt
Service Routine. In cases where the execution of the
instruction following SLEEP is not desirable, the user
should have a NOP after the SLEEP instruction.

The WDT is cleared when the device wakes up from
Sleep, regardless of the source of wake-up.
8.1.1 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction
  - SLEEP instruction will execute as a NOP
  - WDT and WDT prescaler will not be cleared
  - TO bit of the STATUS register will not be set
  - PD bit of the STATUS register will not be cleared
- If the interrupt occurs **during or after** the execution of a SLEEP instruction
  - SLEEP instruction will be completely executed
  - Device will immediately wake-up from Sleep
  - WDT and WDT prescaler will be cleared
  - TO bit of the STATUS register will be set
  - PD bit of the STATUS register will be cleared

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the PD bit. If the PD bit is set, the SLEEP instruction was executed as a NOP.

**FIGURE 8-1: WAKE-UP FROM SLEEP THROUGH INTERRUPT**

<table>
<thead>
<tr>
<th>PC</th>
<th>Instruction Executed</th>
<th>Instruction Fetched</th>
<th>Interrupt flag</th>
<th>GIE bit (INTCON reg.)</th>
<th>Interrupt Latency</th>
<th>Processor in Sleep</th>
</tr>
</thead>
<tbody>
<tr>
<td>0004h</td>
<td>Inst(PC - 1)</td>
<td>Sleep</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0005h</td>
<td>Inst(PC + 1)</td>
<td>Forced NOP</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Note:
1: External clock. High, Medium, Low mode assumed.
2: CLKOUT is shown here for timing reference.
3: TOST = 1024 TOSC. This delay does not apply to EC, RC and INTOSC Oscillator modes or Two-Speed Start-up (see Section 6.4 “Two-Speed Clock Start-up Mode”)
4: GIE = 1 assumed. In this case after wake-up, the processor calls the ISR at 0004h. If GIE = 0, execution will continue in-line.
8.2 Low-Power Sleep Mode

The PIC16F1717/8/9 device contains an internal Low Dropout (LDO) voltage regulator, which allows the device I/O pins to operate at voltages up to 5.5V while the internal device logic operates at a lower voltage. The LDO and its associated reference circuitry must remain active when the device is in Sleep mode. The PIC16F1717/8/9 allows the user to optimize the operating current in Sleep, depending on the application requirements.

A Low-Power Sleep mode can be selected by setting the VREGPM bit of the VREGCON register. With this bit set, the LDO and reference circuitry are placed in a low-power state when the device is in Sleep.

8.2.1 SLEEP CURRENT VS. WAKE-UP TIME

In the default operating mode, the LDO and reference circuitry remain in the normal configuration while in Sleep. The device is able to exit Sleep mode quickly since all circuits remain active. In Low-Power Sleep mode, when waking up from Sleep, an extra delay time is required for these circuits to return to the normal configuration and stabilize.

The Low-Power Sleep mode is beneficial for applications that stay in Sleep mode for long periods of time. The Normal mode is beneficial for applications that need to wake from Sleep quickly and frequently.

8.2.2 PERIPHERAL USAGE IN SLEEP

Some peripherals that can operate in Sleep mode will not operate properly with the Low-Power Sleep mode selected. The Low-Power Sleep mode is intended for use only with the following peripherals:

- Brown-out Reset (BOR)
- Watchdog Timer (WDT)
- External interrupt pin/Interrupt-on-Change pins
- Timer1 (with external clock source < 100 kHz)

Note: The PIC16LF1717/8/9 does not have a configurable Low-Power Sleep mode. PIC16LF1717/8/9 is an unregulated device and is always in the lowest power state when in Sleep, with no wake-up time penalty. This device has a lower maximum VDD and I/O voltage than the PIC16F1717/8/9. See Section 34.0 “Electrical Specifications” for more information.
8.3 Register Definitions: Voltage Regulator Control

REGISTER 8-1: VREGCON: VOLTAGE REGULATOR CONTROL REGISTER(1)

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>STATUS</td>
</tr>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>VREGCON(1)</td>
</tr>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>WDTCON</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as ‘0’
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
’1’ = Bit is set  ‘0’ = Bit is cleared

bit 7-2  Unimplemented: Read as ‘0’
bit 1  VREGPM: Voltage Regulator Power Mode Selection bit
  1 = Low-Power Sleep mode enabled in Sleep(2)
  Draws lowest current in Sleep, slower wake-up
  0 = Normal-Power mode enabled in Sleep(2)
  Draws higher current in Sleep, faster wake-up
bit 0  Reserved: Read as ‘1’. Maintain this bit set.

Note 1: PIC16F1717/8/9 only.
2: See Section 34.0 “Electrical Specifications”.

TABLE 8-1: SUMMARY OF REGISTERS ASSOCIATED WITH POWER-DOWN MODE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>STATUS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC</td>
<td>C</td>
</tr>
<tr>
<td>VREGCON(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>VREGPM</td>
</tr>
<tr>
<td>WDTCON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>WDTPS&lt;4:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>SWDTEN</td>
<td>105</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used in Power-Down mode.
Note 1: PIC16F1717/8/9 only.
9.0 WATCHDOG TIMER (WDT)

The Watchdog Timer is a system timer that generates a Reset if the firmware does not issue a `CLRWDT` instruction within the time-out period. The Watchdog Timer is typically used to recover the system from unexpected events.

The WDT has the following features:

- Independent clock source
- Multiple operating modes
  - WDT is always on
  - WDT is off when in Sleep
  - WDT is controlled by software
  - WDT is always off
- Configurable time-out period is from 1 ms to 256 seconds (nominal)
- Multiple Reset conditions
- Operation during Sleep

FIGURE 9-1: WATCHDOG TIMER BLOCK DIAGRAM

9.1 Independent Clock Source

The WDT derives its time base from the 31 kHz LFINTOSC internal oscillator. Time intervals in this chapter are based on a nominal interval of 1 ms. See Table 34-8: Oscillator Parameters for the LFINTOSC specification.

9.2 WDT Operating Modes

The Watchdog Timer module has four operating modes controlled by the `WDTE<1:0>` bits in Configuration Words. See Table 9-1.

9.2.1 WDT IS ALWAYS ON

When the WDTE bits of Configuration Words are set to '11', the WDT is always on.
WDT protection is active during Sleep.

9.2.2 WDT IS OFF IN SLEEP

When the WDTE bits of Configuration Words are set to '10', the WDT is on, except in Sleep.
WDT protection is not active during Sleep.

9.2.3 WDT CONTROLLED BY SOFTWARE

When the WDTE bits of Configuration Words are set to '01', the WDT is controlled by the SWDTEN bit of the WDTCM register.
WDT protection is unchanged by Sleep. See Table 9-1 for more details.

TABLE 9-1: WDT OPERATING MODES

<table>
<thead>
<tr>
<th>WDTE&lt;1:0&gt;</th>
<th>SWDTEN</th>
<th>Device Mode</th>
<th>WDT Mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>11</td>
<td>X</td>
<td>X</td>
<td>Active</td>
</tr>
<tr>
<td>10</td>
<td>X</td>
<td>Awake</td>
<td>Active</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Sleep</td>
<td>Disabled</td>
</tr>
<tr>
<td>01</td>
<td>1</td>
<td>X</td>
<td>Active</td>
</tr>
<tr>
<td>00</td>
<td>X</td>
<td>X</td>
<td>Disabled</td>
</tr>
</tbody>
</table>
9.3 Time-out Period

The WDTPS bits of the WDTCON register set the time-out period from 1 ms to 256 seconds (nominal). After a Reset, the default time-out period is two seconds.

9.4 Clearing the WDT

The WDT is cleared when any of the following conditions occur:
- Any Reset
- CLRWDT instruction is executed
- Device enters Sleep
- Device wakes up from Sleep
- Oscillator fail
- WDT is disabled
- Oscillator Start-up Timer (OST) is running

See Table 9-2 for more information.

9.5 Operation During Sleep

When the device enters Sleep, the WDT is cleared. If the WDT is enabled during Sleep, the WDT resumes counting.

When the device exits Sleep, the WDT is cleared again. The WDT remains clear until the OST, if enabled, completes. See Section 6.0 “Oscillator Module (with Fail-Safe Clock Monitor)” for more information on the OST.

When a WDT time-out occurs while the device is in Sleep, no Reset is generated. Instead, the device wakes up and resumes operation. The TO and PD bits in the STATUS register are changed to indicate the event. See STATUS Register (Register 3-1) for more information.

<table>
<thead>
<tr>
<th>Conditions</th>
<th>WDT</th>
</tr>
</thead>
<tbody>
<tr>
<td>WDTE&lt;1:0&gt; = 00</td>
<td>Cleared</td>
</tr>
<tr>
<td>WDTE&lt;1:0&gt; = 01 and SWDTEN = 0</td>
<td>Cleared</td>
</tr>
<tr>
<td>WDTE&lt;1:0&gt; = 10 and enter Sleep</td>
<td>Cleared</td>
</tr>
<tr>
<td>CLRWDT Command</td>
<td>Cleared until the end of OST</td>
</tr>
<tr>
<td>Oscillator Fail Detected</td>
<td>Unaffected</td>
</tr>
<tr>
<td>Exit Sleep + System Clock = T1OSC, EXTRC, INTOSC, EXTCLK</td>
<td></td>
</tr>
<tr>
<td>Exit Sleep + System Clock = XT, HS, LP</td>
<td></td>
</tr>
<tr>
<td>Change INTOSC divider (IRCF bits)</td>
<td></td>
</tr>
</tbody>
</table>
9.6 Register Definitions: Watchdog Control

**REGISTER 9-1: WDTCON: WATCHDOG TIMER CONTROL REGISTER**

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
<th>SWDTEN</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>bit 0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -m/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-6 Unimplemented:** Read as ‘0’

**bit 5-1 WDTPS<4:0>:** Watchdog Timer Period Select bits \(^{(1)}\)

<table>
<thead>
<tr>
<th>Bit Value = Prescale Rate</th>
</tr>
</thead>
<tbody>
<tr>
<td>11111 = Reserved. Results in minimum interval (1:32)</td>
</tr>
<tr>
<td>10011 = Reserved. Results in minimum interval (1:32)</td>
</tr>
<tr>
<td>10010 = 1:8388608 (2(^{23})) (Interval 256s nominal)</td>
</tr>
<tr>
<td>10001 = 1:4194304 (2(^{22})) (Interval 128s nominal)</td>
</tr>
<tr>
<td>10000 = 1:2097152 (2(^{21})) (Interval 64s nominal)</td>
</tr>
<tr>
<td>01111 = 1:1048576 (2(^{20})) (Interval 32s nominal)</td>
</tr>
<tr>
<td>01110 = 1:524288 (2(^{19})) (Interval 16s nominal)</td>
</tr>
<tr>
<td>01101 = 1:262144 (2(^{18})) (Interval 8s nominal)</td>
</tr>
<tr>
<td>01100 = 1:131072 (2(^{17})) (Interval 4s nominal)</td>
</tr>
<tr>
<td>01011 = 1:65536 (Interval 2s nominal) (Reset value)</td>
</tr>
<tr>
<td>01010 = 1:32768 (Interval 1s nominal)</td>
</tr>
<tr>
<td>01001 = 1:16384 (Interval 512 ms nominal)</td>
</tr>
<tr>
<td>01000 = 1:8192 (Interval 256 ms nominal)</td>
</tr>
<tr>
<td>00111 = 1:4096 (Interval 128 ms nominal)</td>
</tr>
<tr>
<td>00110 = 1:2048 (Interval 64 ms nominal)</td>
</tr>
<tr>
<td>00101 = 1:1024 (Interval 32 ms nominal)</td>
</tr>
<tr>
<td>00100 = 1:512 (Interval 16 ms nominal)</td>
</tr>
<tr>
<td>00011 = 1:256 (Interval 8 ms nominal)</td>
</tr>
<tr>
<td>00010 = 1:128 (Interval 4 ms nominal)</td>
</tr>
<tr>
<td>00001 = 1:64 (Interval 2 ms nominal)</td>
</tr>
<tr>
<td>00000 = 1:32 (Interval 1 ms nominal)</td>
</tr>
</tbody>
</table>

**bit 0 SWDTEN: Software Enable/Disable for Watchdog Timer bit**

- If WDTE<1:0> = 1x:
  - This bit is ignored.
- If WDTE<1:0> = 01:
  - 1 = WDT is turned on
  - 0 = WDT is turned off
- If WDTE<1:0> = 00:
  - This bit is ignored.

**Note 1:** Times are approximate. WDT time is based on 31 kHz LFINTOSC.
### TABLE 9-3: SUMMARY OF REGISTERS ASSOCIATED WITH WATCHDOG TIMER

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>OSCCON</td>
<td>SPLLEN</td>
<td>IRCF&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
<td>SCS&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>84</td>
</tr>
<tr>
<td>STATUS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC</td>
<td>29</td>
</tr>
<tr>
<td>WDTCN</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>WDTPS&lt;4:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>105</td>
</tr>
</tbody>
</table>

Legend:  
- $\times$ = unknown,  
- $\_\_\_$ = unchanged,  
- $-$ = unimplemented locations read as '0'. Shaded cells are not used by Watchdog Timer.

### TABLE 9-4: SUMMARY OF CONFIGURATION WORD WITH WATCHDOG TIMER

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit -/7</th>
<th>Bit -/6</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG1</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>FCMEN</td>
<td>IESO</td>
<td>CLKOUTEN</td>
<td>BOREN&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>56</td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>CP</td>
<td>MCLRE</td>
<td>PWRT&lt;1:0&gt;</td>
<td>WDTE&lt;1:0&gt;</td>
<td>FOSC&lt;2:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

Legend:  
- $-$ = unimplemented location, read as '0'. Shaded cells are not used by Watchdog Timer.
10.0 FLASH PROGRAM MEMORY CONTROL

The Flash program memory is readable and writable during normal operation over the full VDD range. Program memory is indirectly addressed using Special Function Registers (SFRs). The SFRs used to access program memory are:

- PMCON1
- PMCON2
- PMDATL
- PMDATH
- PMADRL
- PMADRH

When accessing the program memory, the PMDATH:PMDATL register pair forms a 2-byte word that holds the 14-bit data for read/write, and the PMADRHP:PMADRL register pair forms a 2-byte word that holds the 15-bit address of the program memory location being read.

The write time is controlled by an on-chip timer. The write/erase voltages are generated by an on-chip charge pump rated to operate over the operating voltage range of the device.

The Flash program memory can be protected in two ways: by code protection (CP bit in Configuration Words) and write protection (WRT<1:0> bits in Configuration Words).

Code protection (CP = 0)\(^{(1)}\), disables access, reading and writing, to the Flash program memory via external device programmers. Code protection does not affect the self-write and erase functionality. Code protection can only be reset by a device programmer performing a Bulk Erase to the device, clearing all Flash program memory, Configuration bits and User IDs.

Write protection prohibits self-write and erase to a portion or all of the Flash program memory as defined by the bits WRT<1:0>. Write protection does not affect a device programmers ability to read, write or erase the device.

**Note 1:** Code protection of the entire Flash program memory array is enabled by clearing the CP bit of Configuration Words.

10.1 PMADRL and PMADRH Registers

The PMADRHP:PMADRL register pair can address up to a maximum of 32K words of program memory. When selecting a program address value, the MSB of the address is written to the PMADRHP register and the LSB is written to the PMADRL register.

10.1.1 PMCON1 AND PMCON2 REGISTERS

PMCON1 is the control register for Flash program memory accesses.

Control bits RD and WR initiate read and write, respectively. These bits cannot be cleared, only set, in software. They are cleared by hardware at completion of the read or write operation. The inability to clear the WR bit in software prevents the accidental, premature termination of a write operation.

The WREN bit, when set, will allow a write operation to occur. On power-up, the WREN bit is clear. The WRERR bit is set when a write operation is interrupted by a Reset during normal operation. In these situations, following Reset, the user can check the WRERR bit and execute the appropriate error handling routine.

The PMCON2 register is a write-only register. Attempting to read the PMCON2 register will return all '0's.

To enable writes to the program memory, a specific pattern (the unlock sequence), must be written to the PMCON2 register. The required unlock sequence prevents inadvertent writes to the program memory write latches and Flash program memory.

10.2 Flash Program Memory Overview

It is important to understand the Flash program memory structure for erase and programming operations. Flash program memory is arranged in rows. A row consists of a fixed number of 14-bit program memory words. A row is the minimum size that can be erased by user software.

After a row has been erased, the user can reprogram all or a portion of this row. Data to be written into the program memory row is written to 14-bit wide data write latches. These write latches are not directly accessible to the user, but may be loaded via sequential writes to the PMDATH:PMDATL register pair.

**Note:** If the user wants to modify only a portion of a previously programmed row, then the contents of the entire row must be read and saved in RAM prior to the erase. Then, new data and retained data can be written into the write latches to reprogram the row of Flash program memory. However, any unprogrammed locations can be written without first erasing the row. In this case, it is not necessary to save and rewrite the other previously programmed locations.
See Table 10-1 for Erase Row size and the number of write latches for Flash program memory.

**TABLE 10-1: FLASH MEMORY ORGANIZATION BY DEVICE**

<table>
<thead>
<tr>
<th>Device</th>
<th>Row Erase (words)</th>
<th>Write Latches (words)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC16(L)F1717</td>
<td>32</td>
<td>32</td>
</tr>
<tr>
<td>PIC16(L)F1718</td>
<td>32</td>
<td>32</td>
</tr>
<tr>
<td>PIC16(L)F1719</td>
<td>32</td>
<td>32</td>
</tr>
</tbody>
</table>

10.2.1 READING THE FLASH PROGRAM MEMORY

To read a program memory location, the user must:
1. Write the desired address to the PMADR:PMADRL register pair.
2. Clear the CFGS bit of the PMCON1 register.
3. Then, set control bit RD of the PMCON1 register.

Once the read control bit is set, the program memory Flash controller will use the second instruction cycle to read the data. This causes the second instruction immediately following the "BSF PMCON1,RD" instruction to be ignored. The data is available in the very next cycle, in the PMDAT:PMDATL register pair; therefore, it can be read as two bytes in the following instructions.

PMDATH:PMDATL register pair will hold this value until another read or until it is written to by the user.

**Note:** The two instructions following a program memory read are required to be NOPs. This prevents the user from executing a 2-cycle instruction on the next instruction after the RD bit is set.
**EXAMPLE 10-1: FLASH PROGRAM MEMORY READ**

* This code block will read 1 word of program
* memory at the memory address:
  
PROG_ADDR_HI : PROG_ADDR_LO
* data will be returned in the variables;
* PROG_DATA_HI, PROG_DATA_LO

```assembly
BANKSEL PMADRL ; Select Bank for PMCON registers
MOVLW PROG_ADDR_LO ;
MOVF PMADRL ; Store LSB of address
MOVLW PROG_ADDR_HI ;
MOVF PMADRH ; Store MSB of address
BCF PMCON1,CFG5 ; Do not select Configuration Space
BSF PMCON1,RD ; Initiate read
NOP ; Ignored (Figure 10-1)
NOP ; Ignored (Figure 10-1)

MOVF PMDATL,W ; Get LSB of word
MOVWF PROG_DATA_LO ; Store in user location
MOVF PMDATH,W ; Get MSB of word
MOVWF PROG_DATA_HI ; Store in user location
```
10.2.2 FLASH MEMORY UNLOCK SEQUENCE

The unlock sequence is a mechanism that protects the Flash program memory from unintended self-write programming or erasing. The sequence must be executed and completed without interruption to successfully complete any of the following operations:

- Row Erase
- Load program memory write latches
- Write of program memory write latches to program memory
- Write of program memory write latches to User IDs

The unlock sequence consists of the following steps:

1. Write 55h to PMCON2
2. Write AAh to PMCON2
3. Set the WR bit in PMCON1
4. NOP instruction
5. NOP instruction

Once the WR bit is set, the processor will always force two NOP instructions. When an Erase Row or Program Row operation is being performed, the processor will stall internal operations (typical 2 ms), until the operation is complete and then resume with the next instruction. When the operation is loading the program memory write latches, the processor will always force the two NOP instructions and continue uninterrupted with the next instruction.

Since the unlock sequence must not be interrupted, global interrupts should be disabled prior to the unlock sequence and re-enabled after the unlock sequence is completed.
10.2.3 ERASING FLASH PROGRAM MEMORY

While executing code, program memory can only be erased by rows. To erase a row:

1. Load the PMADRH:PMADRL register pair with any address within the row to be erased.
2. Clear the CFGS bit of the PMCON1 register.
3. Set the FREE and WREN bits of the PMCON1 register.
4. Write 55h, then AAh, to PMCON2 (Flash programming unlock sequence).
5. Set control bit WR of the PMCON1 register to begin the erase operation.

See Example 10-2.

After the “BSF PMCON1, WR” instruction, the processor requires two cycles to set up the erase operation. The user must place two NOP instructions immediately following the WR bit set instruction. The processor will halt internal operations for the typical 2 ms erase time. This is not Sleep mode as the clocks and peripherals will continue to run. After the erase cycle, the processor will resume operation with the third instruction after the PMCON1 write instruction.
EXAMPLE 10-2: ERASING ONE ROW OF PROGRAM MEMORY

; This row erase routine assumes the following:
; 1. A valid address within the erase row is loaded in ADDRH:ADDRL
; 2. ADDRH and ADDRL are located in shared data memory 0x70 - 0x7F (common RAM)

BCF INTCON,GIE ; Disable ints so required sequences will execute properly
BANKSEL PMADRL
MOVF ADDRL,W ; Load lower 8 bits of erase address boundary
MOVWF PMADRL
MOVF ADDRH,W ; Load upper 6 bits of erase address boundary
MOVWF PMADRH
BCF PMCON1,CFGS ; Not configuration space
BSF PMCON1,FREE ; Specify an erase operation
BSF PMCON1,WREN ; Enable writes

MOVLW 55h ; Start of required sequence to initiate erase
MOVWF PMCON2 ; Write 55h
MOVLW 0AAh ;
MOVWF PMCON2 ; Write AAh
BSF PMCON1,WR ; Set WR bit to begin erase
NOP ; NOP instructions are forced as processor starts
NOP ; row erase of program memory.
; The processor stalls until the erase process is complete
; after erase processor continues with 3rd instruction

BCF PMCON1,WREN ; Disable writes
BSF INTCON,GIE ; Enable interrupts
10.2.4 WRITING TO FLASH PROGRAM MEMORY

Program memory is programmed using the following steps:

1. Load the address in PMADRH:PMADRL of the row to be programmed.
2. Load each write latch with data.
3. Initiate a programming operation.
4. Repeat steps 1 through 3 until all data is written.

Before writing to program memory, the word(s) to be written must be erased or previously unwritten. Program memory can only be erased one row at a time. No automatic erase occurs upon the initiation of the write.

Program memory can be written one or more words at a time. The maximum number of words written at one time is equal to the number of write latches. See Figure 10-5 (row writes to program memory with 32 write latches) for more details.

The write latches are aligned to the Flash row address boundary defined by the upper ten bits of PMADRH:PMADRL, (PMADRH<6:0>:PMADRL<7:5>) with the lower five bits of PMADRL, (PMADRL<4:0>) determining the write latch being loaded. Write operations do not cross these boundaries. At the completion of a program memory write operation, the data in the write latches is reset to contain 0x3FFF.

The following steps should be completed to load the write latches and program a row of program memory. These steps are divided into two parts. First, each write latch is loaded with data from the PMDATH:PMDATL using the unlock sequence with LWLO = 1. When the last word to be loaded into the write latch is ready, the LWLO bit is cleared and the unlock sequence executed. This initiates the programming operation, writing all the latches into Flash program memory.

**Note:** The special unlock sequence is required to load a write latch with data or initiate a Flash programming operation. If the unlock sequence is interrupted, writing to the latches or program memory will not be initiated.

1. Set the WREN bit of the PMCON1 register.
2. Clear the CFGS bit of the PMCON1 register.
3. Set the LWLO bit of the PMCON1 register.

When the LWLO bit of the PMCON1 register is ‘1’, the write sequence will only load the write latches and will not initiate the write to Flash program memory.

4. Load the PMADRH:PMADRL register pair with the address of the location to be written.
5. Load the PMDATH:PMDATL register pair with the program memory data to be written.
6. Execute the unlock sequence (Section 10.2.2 “Flash Memory Unlock Sequence”). The write latch is now loaded.
7. Increment the PMADRH:PMADRL register pair to point to the next location.
8. Repeat steps 5 through 7 until all but the last write latch has been loaded.
9. Clear the LWLO bit of the PMCON1 register.

When the LWLO bit of the PMCON1 register is ‘0’, the write sequence will initiate the write to Flash program memory.

10. Load the PMDATH:PMDATL register pair with the program memory data to be written.
11. Execute the unlock sequence (Section 10.2.2 “Flash Memory Unlock Sequence”). The entire program memory latch content is now written to Flash program memory.

**Note:** The program memory write latches are reset to the blank state (0x3FFF) at the completion of every write or erase operation. As a result, it is not necessary to load all the program memory write latches. Unloaded latches will remain in the blank state.

An example of the complete write sequence is shown in Example 10-3. The initial address is loaded into the PMADRH:PMADRL register pair; the data is loaded using indirect addressing.
FIGURE 10-5: BLOCK WRITES TO FLASH PROGRAM MEMORY WITH 32 WRITE LATCHES

- PMADRHI
- PMADRL<4:0>
- PMADRL<7:5>
- PMADRH<6:0>

Program Memory Write Latches

Write Latch #0
Write Latch #1
Write Latch #30
Write Latch #31

Row
Addr
Addr
Addr
Addr

000h
0000h
0001h
001Eh
001Fh

001h
0020h
0021h
003Eh
003Fh

002h
0040h
0041h
005Eh
005Fh

3FEh
7FC0h
7FC1h
7FDEh
7FDFh

3FFh
7FE0h
7FE1h
7FFEh
7FFFh

Flash Program Memory

400h
8000h - 8003h
8004h - 8005h
8006h
8007h - 8008h
8009h - 801Fh

USER ID 0 - 3
reserved
DEVICE ID
Dev / Rev
Configuration
Words
reserved

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FIGURE 10-6: FLASH PROGRAM MEMORY WRITE FLOWCHART

- Start
- Write Operation

- Disable Interrupts (GIE = 0)

- Select Program or Config. Memory (CFGS)

- Select Row Address (PMADRH:PMADRL)

- Select Write Operation (FREE = 0)

- Load Write Latches Only (LWLO = 1)

- Determine number of words to be written into Program or Configuration Memory. The number of words cannot exceed the number of words per row. (word_cnt)

- Enable Write/Erase Operation (WREN = 1)

- Load the value to write (PMDATH:PMDATL)

- Update the word counter (word_cnt--)

- Last word to write?
  - Yes
  - Unlock Sequence (Figure 10-3)
  - CPU stalls while Write operation completes (2ms typical)
  - Disable Write/Erase Operation (WREN = 0)
  - Re-enable Interrupts (GIE = 1)

- No delay when writing to Program Memory Latches

- Increment Address (PMADRH:PMADRL++)

- Write Latches to Flash (LWLO = 0)

- Unlock Sequence (Figure 10-3)

- End Write Operation
EXAMPLE 10-3: WRITING TO FLASH PROGRAM MEMORY

; This write routine assumes the following:
; 1. 64 bytes of data are loaded, starting at the address in DATA_ADDR
; 2. Each word of data to be written is made up of two adjacent bytes in DATA_ADDR,
; stored in little endian format
; 3. A valid starting address (the least significant bits = 00000) is loaded in ADDRH:ADDRL
; 4. ADDRH and ADDRL are located in shared data memory 0x70 - 0x7F (common RAM)
;
; BCF INTCON,GIE ; Disable ints so required sequences will execute properly
BANKSEL PMADRH ; Bank 3
MOVF ADDRH,W ; Load initial address
MOVF PMADRH ;
MOVWF PMADRL ;
MOVLW LOW DATA_ADDR ; Load initial data address
MOVF FSR0L ;
MOVLW HIGH DATA_ADDR ; Load initial data address
MOVF FSR0H ;
BCF PMCON1,CFGS ; Not configuration space
BSF PMCON1,WREN ; Enable writes
BSF PMCON1,LWLO ; Only Load Write Latches

LOOP
MOVIW FSR0++ ; Load first data byte into lower
MOVWF PMDATL ;
MOVIW FSR0++ ; Load second data byte into upper
MOVWF PMDATH ;

MOVF PMADRL,W ; Check if lower bits of address are '00000'
XORLW 0x1F ; Check if we're on the last of 32 addresses
ANDLW 0x1F ;
BTFSC STATUS,Z ; Exit if last of 32 words,
GOTO START_WRITE ;

START_WRITE
BCF PMCON1,LWLO ; No more loading latches - Actually start Flash program
; memory write

MOVLW 55h ; Start of required write sequence:
MOVF PMCON2 ; Write 55h
MOVWF 0AAh ;
MOVF PMCON2 ; Write AAh
BSF PMCON1,WR ; Set WR bit to begin write
NOP ; NOP instructions are forced as processor
; loads program memory write latches

NOP ;

INCF PMADRL,Y ; Still loading latches Increment address
GOTO LOOP ; Write next latches

BCF PMCON1,LWLO ; No more loading latches - Actually start Flash program
; memory write

MOVLW 55h ; Start of required write sequence:
MOVF PMCON2 ; Write 55h
MOVWF 0AAh ;
MOVF PMCON2 ; Write AAh
BSF PMCON1,WR ; Set WR bit to begin write
NOP ; NOP instructions are forced as processor writes
; all the program memory write latches simultaneously

NOP ; to program memory.
; After NOPS, the processor
; stalls until the self-write process in complete
; after write processor continues with 3rd instruction
BCF PMCON1,WREN ; Disable writes
BSF INTCON,GIE ; Enable interrupts
10.3 Modifying Flash Program Memory

When modifying existing data in a program memory row, and data within that row must be preserved, it must first be read and saved in a RAM image. Program memory is modified using the following steps:

1. Load the starting address of the row to be modified.
2. Read the existing data from the row into a RAM image.
3. Modify the RAM image to contain the new data to be written into program memory.
4. Load the starting address of the row to be rewritten.
5. Erase the program memory row.
6. Load the write latches with data from the RAM image.
7. Initiate a programming operation.

FIGURE 10-7: FLASH PROGRAM MEMORY MODIFY FLOWCHART
10.4 User ID, Device ID and Configuration Word Access

Instead of accessing program memory, the User ID’s, Device ID/Revision ID and Configuration Words can be accessed when CFGS = 1 in the PMCON1 register. This is the region that would be pointed to by PC<15> = 1, but not all addresses are accessible. Different access may exist for reads and writes. Refer to Table 10-2.

When read access is initiated on an address outside the parameters listed in Table 10-2, the PMDATH:PMDATL register pair is cleared, reading back ‘0’s.

TABLE 10-2: USER ID, DEVICE ID AND CONFIGURATION WORD ACCESS (CFGS = 1)

<table>
<thead>
<tr>
<th>Address</th>
<th>Function</th>
<th>Read Access</th>
<th>Write Access</th>
</tr>
</thead>
<tbody>
<tr>
<td>8000h-8003h</td>
<td>User IDs</td>
<td>Yes</td>
<td>Yes</td>
</tr>
<tr>
<td>8005h-8006h</td>
<td>Device ID/Revision ID</td>
<td>Yes</td>
<td>No</td>
</tr>
<tr>
<td>8007h-8008h</td>
<td>Configuration Words 1 and 2</td>
<td>Yes</td>
<td>No</td>
</tr>
</tbody>
</table>

EXAMPLE 10-4: CONFIGURATION WORD AND DEVICE ID ACCESS

* This code block will read 1 word of program memory at the memory address:
* PROG_ADDR_LO (must be 00h-08h) data will be returned in the variables;
* PROG_DATA_HI, PROG_DATA_LO

```
BANKSEL PMADRL        ; Select correct Bank
MOVLW PROG_ADDR_LO    ;
MOVWF PMADRL          ; Store LSB of address
CLRF PMADRH           ; Clear MSB of address
BSF PMCON1,CFGS       ; Select Configuration Space
BCF INTCON,GIE        ; Disable interrupts
BSF PMCON1,RD         ; Initiate read
NOP                   ; Executed (See Figure 10-2)
NOP                   ; Ignored (See Figure 10-2)
BSF INTCON,GIE        ; Restore interrupts
MOVF PMDATL,W         ; Get LSB of word
MOVF PMDATL,W         ; Store in user location
```

MOVF PMDATH,W         ; Get MSB of word
MOVF PROG_DATA_HI     ; Store in user location
MOVF PROG_DATA_LO     ; Store in user location
10.5 Write Verify

It is considered good programming practice to verify that program memory writes agree with the intended value. Since program memory is stored as a full page then the stored program memory contents are compared with the intended data stored in RAM after the last write is complete.

**FIGURE 10-8: FLASH PROGRAM MEMORY VERIFY FLOWCHART**

This routine assumes that the last row of data written was from an image saved in RAM. This image will be used to verify the data currently stored in Flash Program Memory.

- **Start Verify Operation**
- **Read Operation** (Figure 10-1)
- **PMDAT = RAM image?**
  - **Yes**
    - **Fail Verify Operation**
  - **No**
- **Last Word?**
  - **Yes**
    - **End Verify Operation**
  - **No**
### 10.6 Register Definitions: Flash Program Memory Control

#### REGISTER 10-1: PMDATL: PROGRAM MEMORY DATA LOW BYTE REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>PMDAT&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**bit 7-0 PMDAT<7:0>:** Read/write value for Least Significant bits of program memory

#### REGISTER 10-2: PMDATH: PROGRAM MEMORY DATA HIGH BYTE REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>PMDAT&lt;13:8&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**bit 7-0 PMDAT<13:8>:** Read/write value for Most Significant bits of program memory

#### REGISTER 10-3: PMADRL: PROGRAM MEMORY ADDRESS LOW BYTE REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>PMADR&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**bit 7-0 PMADR<7:0>:** Specifies the Least Significant bits for program memory address
### PIC16(L)F1717/8/9

#### REGISTER 10-4: PMADRH: PROGRAM MEMORY ADDRESS HIGH BYTE REGISTER

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6-0</th>
<th>PMADR&lt;14:8&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>U-1</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
</tr>
<tr>
<td>—(1)</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set

**bit 7** **Unimplemented:** Read as ‘1’

**bit 6-0** **PMADR<14:8>:** Specifies the Most Significant bits for program memory address

**Note 1:** Unimplemented, read as ‘1’.

#### REGISTER 10-5: PMCON1: PROGRAM MEMORY CONTROL 1 REGISTER

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6-0</th>
<th>CFGS</th>
<th>LWLO(3)</th>
<th>FREE</th>
<th>WRERR</th>
<th>WREN</th>
<th>WR</th>
<th>RD</th>
</tr>
</thead>
<tbody>
<tr>
<td>U-1</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W/HC-0/0</td>
<td>R/W/HC-x/q(2)</td>
<td>R/W-0/0</td>
<td>R/S/HC-0/0</td>
<td>R/S/HC-0/0</td>
<td></td>
</tr>
<tr>
<td>—(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **S** = Bit can only be set
- **x** = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- **HC** = Bit is cleared by hardware

**bit 7** **Unimplemented:** Read as ‘1’

**bit 6** **CFGS:** Configuration Select bit
- **1** = Access Configuration, User ID and Device ID Registers
- **0** = Access Flash program memory

**bit 5** **LWLO:** Load Write Latches Only bit(3)
- **1** = Only the addressed program memory write latch is loaded/updated on the next WR command
- **0** = The addressed program memory write latch is loaded/updated and a write of all program memory write latches will be initiated on the next WR command

**bit 4** **FREE:** Program Flash Erase Enable bit
- **1** = Performs an erase operation on the next WR command (hardware cleared upon completion)
- **0** = Performs a write operation on the next WR command

**bit 3** **WRERR:** Program/Erase Error Flag bit
- **1** = Condition indicates an improper program or erase sequence attempt or termination (bit is set automatically on any set attempt (write ‘1’) of the WR bit)
- **0** = The program or erase operation completed normally

**bit 2** **WREN:** Program/Erase Enable bit
- **1** = Allows program/erase cycles
- **0** = Inhibits programming/erasing of program Flash

**bit 1** **WR:** Write Control bit
- **1** = Initiates a program Flash program/erase operation.
  - The operation is self-timed and the bit is cleared by hardware once operation is complete.
  - The WR bit can only be set (not cleared) in software.
- **0** = Program/erase operation to the Flash is complete and inactive

**bit 0** **RD:** Read Control bit
- **1** = Initiates a program Flash read. Read takes one cycle. RD is cleared in hardware. The RD bit can only be set (not cleared) in software.
- **0** = Does not initiate a program Flash read

**Note 1:** Unimplemented bit, read as ‘1’.

2: The WRERR bit is automatically set by hardware when a program memory write or erase operation is started (WR = 1).
3: The LWLO bit is ignored during a program memory erase operation (FREE = 1).
REGISTER 10-6: PMCON2: PROGRAM MEMORY CONTROL 2 REGISTER

<table>
<thead>
<tr>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Bit 7</td>
<td>Bit 0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Program Memory Control Register 2

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- S = Bit can only be set
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

Flash Memory Unlock Pattern bits

To unlock writes, a 55h must be written first, followed by an AAh, before setting the WR bit of the PMCON1 register. The value written to this register is used to unlock the writes. There are specific timing requirements on these writes.

TABLE 10-3: SUMMARY OF REGISTERS ASSOCIATED WITH FLASH PROGRAM MEMORY

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMROIE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMROIF</td>
<td>INTF</td>
<td>IO CIF</td>
<td>91</td>
</tr>
<tr>
<td>PMCON1</td>
<td>CFGS</td>
<td>LWLO</td>
<td>FREE</td>
<td>WRERR</td>
<td>WREN</td>
<td>WR</td>
<td>RD</td>
<td></td>
<td>121</td>
</tr>
<tr>
<td>PMCON2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>122</td>
</tr>
<tr>
<td>PMADRL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>PMADRL&lt;7:0&gt;</td>
<td></td>
<td></td>
<td>120</td>
</tr>
<tr>
<td>PMADRHO</td>
<td>CFGS</td>
<td>L WLO</td>
<td>FREE</td>
<td></td>
<td></td>
<td>PMADRHO&lt;6:0&gt;</td>
<td></td>
<td></td>
<td>121</td>
</tr>
<tr>
<td>PMDATL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>PMDATL&lt;7:0&gt;</td>
<td></td>
<td></td>
<td>120</td>
</tr>
<tr>
<td>PMDAI TH</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>PMDAI TH&lt;5:0&gt;</td>
<td></td>
<td></td>
<td>120</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used by Flash program memory.

Note 1: Unimplemented, read as ‘1’.

TABLE 10-4: SUMMARY OF CONFIGURATION WORD WITH FLASH PROGRAM MEMORY

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit -/7</th>
<th>Bit -/6</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG1</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>CLKOUTEN</td>
<td>BOREN&lt;1:0&gt;</td>
<td>—</td>
<td>56</td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>CP</td>
<td>MCLRE</td>
<td>PWRT</td>
<td>WDTE&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>FOSC&lt;1:0&gt;</td>
<td></td>
<td></td>
</tr>
<tr>
<td>CONFIG2</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>LVP</td>
<td>DEBUG</td>
<td>LPBOR</td>
<td>BORV</td>
<td>STVREN</td>
<td>PLLN</td>
<td>58</td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>ZCDDIS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>PPS1WAY</td>
<td>WRT&lt;1:0&gt;</td>
<td></td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used by Flash program memory.
11.0 I/O PORTS

Each port has six standard registers for its operation. These registers are:
- TRISx registers (data direction)
- PORTx registers (reads the levels on the pins of the device)
- LATx registers (output latch)
- INLVx (input level control)
- ODCONx registers (open-drain)
- SLRCONx registers (slew rate)

Some ports may have one or more of the following additional registers. These registers are:
- ANSELx (analog select)
- WPUx (weak pull-up)

In general, when a peripheral is enabled on a port pin, that pin cannot be used as a general purpose output. However, the pin can still be read.

The Data Latch (LATx registers) is useful for read-modify-write operations on the value that the I/O pins are driving.

A write operation to the LATx register has the same effect as a write to the corresponding PORTx register. A read of the LATx register reads of the values held in the I/O PORT latches, while a read of the PORTx register reads the actual I/O pin value.

Ports that support analog inputs have an associated ANSELx register. When an ANSEL bit is set, the digital input buffer associated with that bit is disabled. Disabling the input buffer prevents analog signal levels on the pin between a logic high and low from causing excessive current in the logic input circuitry. A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 11-1.

<table>
<thead>
<tr>
<th>Device</th>
<th>PORTA</th>
<th>PORTB</th>
<th>PORTC</th>
<th>PORTD</th>
<th>PORTE</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC16(L)F1717</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
</tr>
<tr>
<td>PIC16(L)F1718</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
</tr>
<tr>
<td>PIC16(L)F1719</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
<td>●●●●●</td>
</tr>
</tbody>
</table>

The Data Latch (LATx registers) is useful for read-modify-write operations on the value that the I/O pins are driving.

A write operation to the LATx register has the same effect as a write to the corresponding PORTx register. A read of the LATx register reads of the values held in the I/O PORT latches, while a read of the PORTx register reads the actual I/O pin value.

11.1 PORTA Registers

11.1.1 DATA REGISTER

PORTA is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISA (Register 11-2). Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., disable the output driver). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). Example 11-1 shows how to initialize PORTA.

Reading the PORTA register (Register 11-1) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATA).

11.1.2 DIRECTION CONTROL

The TRISA register (Register 11-2) controls the PORTA pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISA register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read ‘0’.

11.1.3 OPEN-DRRAIN CONTROL

The ODCONx register (Register 11-6) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONx bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONx bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.
11.1.4 SLEW RATE CONTROL
The SLRCONA register (Register 11-7) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONA bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONA bit is cleared, the corresponding port pin drive slews at the maximum rate possible.

11.1.5 INPUT THRESHOLD CONTROL
The INLVLA register (Register 11-8) controls the input voltage threshold for each of the available PORTA input pins. A selection between the Schmitt Trigger CMOS or the TTL compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTA register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 34-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.1.6 ANALOG CONTROL
The ANSELA register (Register 11-4) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELA bit high will cause all digital reads on the pin to be read as ‘0’ and allow analog functions on the pin to operate correctly.

The state of the ANSELA bits has no effect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note: The ANSELA bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to ‘0’ by user software.

EXAMPLE 11-1: INITIALIZING PORTA

; This code example illustrates
; initializing the PORTA register. The
; other ports are initialized in the same
; manner.

BANKSEL PORTA ;
CLRF PORTA ; Init PORTA
BANKSEL LATA ; Data Latch
CLRF LATA ;
BANKSEL ANSELA ;
CLRF ANSELA ; digital I/O
BANKSEL TRISA ;
MOVlw B'00111000' ; Set RA<5:3> as inputs
MOVF TRISA ; and set RA<2:0> as
; outputs

11.1.7 PORTA FUNCTIONS AND OUTPUT PRIORITIES
Each PORTA pin is multiplexed with other functions. Each pin defaults to the PORT latch data after Reset. Other functions are selected with the peripheral pin select logic. See Section 12.0 “Peripheral Pin Select (PPS) Module” for more information.

Analog input functions, such as ADC and comparator inputs are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELA register. Digital output functions may continue to control the pin when it is in Analog mode.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

Note: The ANSELA bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to ‘0’ by user software.
11.2 Register Definitions: PORTA

REGISTER 11-1: PORTA: PORTA REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA7</td>
<td>RA6</td>
<td>RA5</td>
<td>RA4</td>
<td>RA3</td>
<td>RA2</td>
<td>RA1</td>
<td>RA0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0  RA<7:0>: PORTA I/O Value bits\(^{(1)}\)
- 1 = Port pin is > \(V_{IH}\)
- 0 = Port pin is < \(V_{IL}\)

**Note 1:** Writes to PORTA are actually written to corresponding LATA register. Reads from PORTA register is return of actual I/O pin values.

REGISTER 11-2: TRISA: PORTA TRI-STATE REGISTER

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0  TRISA<7:0>: PORTA Tri-State Control bit
- 1 = PORTA pin configured as an input (tri-stated)
- 0 = PORTA pin configured as an output
REGISTER 11-3: LATA: PORTA DATA LATCH REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>LATA7</td>
<td>LATA6</td>
<td>LATA5</td>
<td>LATA4</td>
<td>LATA3</td>
<td>LATA2</td>
<td>LATA1</td>
<td>LATA0</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as '0'
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set  '0' = Bit is cleared

bit 7-0  LATA<7:0>: RA<7:0> Output Latch Value bits(1)

Note 1: Writes to PORTA are actually written to corresponding LATA register. Reads from PORTA register is return of actual I/O pin values.

REGISTER 11-4: ANSELA: PORTA ANALOG SELECT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as '0'
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set  '0' = Bit is cleared

bit 7-6  Unimplemented: Read as '0'
bit 5-0  ANSA<5:0>: Analog Select between Analog or Digital Function on Pins RA<5:0>, respectively
1 = Analog input. Pin is assigned as analog input(1). Digital input buffer disabled.
0 = Digital I/O. Pin is assigned to port or digital special function.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.
### REGISTER 11-5: WPUA: WEAK PULL-UP PORTA REGISTER\(^{(1,2)}\)

<table>
<thead>
<tr>
<th>Bit 7-0</th>
<th>WPUA&lt;7:0&gt;</th>
<th>Weak Pull-up Register bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Pull-up enabled</td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>Pull-up disabled</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

**Note:**
1. Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.
2. The weak pull-up device is automatically disabled if the pin is configured as an output.

### REGISTER 11-6: ODCONA: PORTA OPEN-DRAIN CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit 7-0</th>
<th>ODA&lt;7:0&gt;</th>
<th>PORTA Open-Drain Enable bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Port pin operates as open-drain drive (sink current only)</td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>Port pin operates as standard push-pull drive (source and sink current)</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

For RA<7:0> pins, respectively.
**REGISTER 11-7: SLRCONA: PORTA SLEW RATE CONTROL REGISTER**

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>SLRA7</td>
<td>SLRA6</td>
<td>SLRA5</td>
<td>SLRA4</td>
<td>SLRA3</td>
<td>SLRA2</td>
<td>SLRA1</td>
<td>SLRA0</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **'-n/n'** = Value at POR and BOR/Value at all other Resets

bit 7-0  **SLRA<7:0>:** PORTA Slew Rate Enable bits
For RA<7:0> pins, respectively
- **1** = Port pin slew rate is limited
- **0** = Port pin slews at maximum rate

---

**REGISTER 11-8: INLVLA: PORTA INPUT LEVEL CONTROL REGISTER**

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>INLVLA7</td>
<td>INLVLA6</td>
<td>INLVLA5</td>
<td>INLVLA4</td>
<td>INLVLA3</td>
<td>INLVLA2</td>
<td>INLVLA1</td>
<td>INLVLA0</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **'-n/n'** = Value at POR and BOR/Value at all other Resets

bit 7-0  **INLVLA<7:0>:** PORTA Input Level Select bits
For RA<7:0> pins, respectively
- **1** = Port pin digital input operates with ST thresholds
- **0** = Port pin digital input operates with TTL thresholds
### TABLE 11-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>126</td>
</tr>
<tr>
<td>INLVL A</td>
<td>INL L A</td>
<td>—</td>
<td>—</td>
<td>INLVL A</td>
<td>INLVL A</td>
<td>INLVL A</td>
<td>INLVL A</td>
<td>INLVL A</td>
<td>INLVL A</td>
</tr>
<tr>
<td>LATA</td>
<td>LATA7</td>
<td>LATA6</td>
<td>LATA5</td>
<td>LATA4</td>
<td>LATA3</td>
<td>LATA2</td>
<td>LATA1</td>
<td>LATA0</td>
<td>126</td>
</tr>
<tr>
<td>ODCONA</td>
<td>ODA7</td>
<td>ODA6</td>
<td>ODA5</td>
<td>ODA4</td>
<td>ODA3</td>
<td>ODA2</td>
<td>ODA1</td>
<td>ODA0</td>
<td>127</td>
</tr>
<tr>
<td>OPTION_REG</td>
<td>WPUEN</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS&lt;2:0&gt;</td>
<td></td>
<td></td>
<td>273</td>
</tr>
<tr>
<td>PORTA</td>
<td>RA7</td>
<td>RA6</td>
<td>RA5</td>
<td>RA4</td>
<td>RA3</td>
<td>RA2</td>
<td>RA1</td>
<td>RA0</td>
<td>125</td>
</tr>
<tr>
<td>SLRCON A</td>
<td>SLRA7</td>
<td>SLRA6</td>
<td>SLRA5</td>
<td>SLRA4</td>
<td>SLRA3</td>
<td>SLRA2</td>
<td>SLRA1</td>
<td>SLRA0</td>
<td>128</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
<tr>
<td>WPUA</td>
<td>WPUA7</td>
<td>WPUA6</td>
<td>WPUA5</td>
<td>WPUA4</td>
<td>WPUA3</td>
<td>WPUA2</td>
<td>WPUA1</td>
<td>WPUA0</td>
<td>127</td>
</tr>
</tbody>
</table>

**Legend:** x = unknown, u = unchanged, – = unimplemented locations read as ‘0’. Shaded cells are not used by PORTA.

### TABLE 11-3: SUMMARY OF CONFIGURATION WORD WITH PORTA

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit 7/6</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG 1</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>FCMEN</td>
<td>IESO</td>
<td>CLKOUTEN</td>
<td>BOREN&lt;1:0&gt;</td>
<td>—</td>
<td>56</td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>CP</td>
<td>MCLRE</td>
<td>PWRTE</td>
<td>WDTE&lt;1:0&gt;</td>
<td>FOSC&lt;2:0&gt;</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used by PORTA.
11.3 PORTB Registers

11.3.1 DATA REGISTER
PORTB is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISB (Register 11-10). Setting a TRISB bit (\(=\,1\)) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISB bit (\(=\,0\)) will make the corresponding PORTB pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTB register (Register 11-9) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATB).

11.3.2 DIRECTION CONTROL
The TRISB register (Register 11-10) controls the PORTB pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISB register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

11.3.3 OPEN-DRAIN CONTROL
The ODCONB register (Register 11-14) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONB bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONB bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.3.4 SLEW RATE CONTROL
The SLRCONB register (Register 11-15) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONB bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONB bit is cleared, the corresponding port pin drive slews at the maximum rate possible.

11.3.5 INPUT THRESHOLD CONTROL
The INLVLB register (Register 11-16) controls the input voltage threshold for each of the available PORTB input pins. A selection between the Schmitt Trigger CMOS or the TTL compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTB register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 34-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.3.6 ANALOG CONTROL
The ANSELB register (Register 11-12) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELB bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELB bits has no effect on digital output functions. A pin with TRIS clear and ANSELB set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note: The ANSELB bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to '0' by user software.

11.3.7 PORTB FUNCTIONS AND OUTPUT PRIORITIES
Each pin defaults to the PORT latch data after Reset. Other functions are selected with the peripheral pin select logic. See Section 12.0 “Peripheral Pin Select (PPS) Module” for more information. Analog input functions, such as ADC and op amp inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELB register. Digital output functions continue to may continue to control the pin when it is in Analog mode.
11.4 Register Definitions: PORTB

**REGISTER 11-9: PORTB: PORTB REGISTER**

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>RB7</td>
<td>RB6</td>
<td>RB5</td>
<td>RB4</td>
<td>RB3</td>
<td>RB2</td>
<td>RB1</td>
<td>RB0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- n/n = Value at POR and BOR/Value at all other Resets

**bit 7-0**: RB<7:0>: PORTB General Purpose I/O Pin bits

- 1 = Port pin is $> V_{IH}$
- 0 = Port pin is $< V_{IL}$

**Note 1**: Writes to PORTB are actually written to corresponding LATB register. Reads from PORTB register is return of actual I/O pin values.

**REGISTER 11-10: TRISB: PORTB TRI-STATE REGISTER**

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- n/n = Value at POR and BOR/Value at all other Resets

**bit 7-0**: TRISB<7:0>: PORTB Tri-State Control bits

- 1 = PORTB pin configured as an input (tri-stated)
- 0 = PORTB pin configured as an output

**REGISTER 11-11: LATB: PORTB DATA LATCH REGISTER**

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>LATB7</td>
<td>LATB6</td>
<td>LATB5</td>
<td>LATB4</td>
<td>LATB3</td>
<td>LATB2</td>
<td>LATB1</td>
<td>LATB0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- n/n = Value at POR and BOR/Value at all other Resets

**bit 7-0**: LATB<7:0>: PORTB Output Latch Value bits

**Note 1**: Writes to PORTB are actually written to corresponding LATB register. Reads from PORTB register is return of actual I/O pin values.
REGISTER 11-12: ANSELB: PORTB ANALOG SELECT REGISTER

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th></th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

bit 7-6 Unimplemented: Read as ‘0’

bit 5-0 ANSB<5:0>: Analog Select between Analog or Digital Function on Pins RB<5:0>, respectively
- 0 = Digital I/O. Pin is assigned to port or digital special function.
- 1 = Analog input. Pin is assigned as analog input(1). Digital input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 11-13: WPUB: WEAK PULL-UP PORTB REGISTER(1,2)

<table>
<thead>
<tr>
<th>bit 7-0</th>
<th></th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>WPUB7</td>
<td>WPUB6</td>
<td>WPUB5</td>
<td>WPUB4</td>
<td>WPUB3</td>
<td>WPUB2</td>
<td>WPUB1</td>
<td>WPUB0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

bit 7-0 WPUB<7:0>: Weak Pull-up Register bits
- 1 = Pull-up enabled
- 0 = Pull-up disabled

Note 1: Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.

2: The weak pull-up device is automatically disabled if the pin is configured as an output.
### REGISTER 11-14: ODCONB: PORTB OPEN-DRAIN CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit 7-0</th>
<th>ODB&lt;7:0&gt;: PORTB Open-Drain Enable bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>For RB&lt;7:0&gt; pins, respectively</td>
<td></td>
</tr>
<tr>
<td>1 = Port pin operates as open-drain drive (sink current only)</td>
<td></td>
</tr>
<tr>
<td>0 = Port pin operates as standard push-pull drive (source and sink current)</td>
<td></td>
</tr>
</tbody>
</table>

### REGISTER 11-15: SLRCONB: PORTB SLEW RATE CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit 7-0</th>
<th>SLRB&lt;7:0&gt;: PORTB Slew Rate Enable bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>For RB&lt;7:0&gt; pins, respectively</td>
<td></td>
</tr>
<tr>
<td>1 = Port pin slew rate is limited</td>
<td></td>
</tr>
<tr>
<td>0 = Port pin slews at maximum rate</td>
<td></td>
</tr>
</tbody>
</table>

### REGISTER 11-16: INLVLB: PORTB INPUT LEVEL CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit 7-0</th>
<th>INLVLB&lt;7:0&gt;: PORTB Input Level Select bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>For RB&lt;7:0&gt; pins, respectively</td>
<td></td>
</tr>
<tr>
<td>1 = Port pin digital input operates with ST thresholds</td>
<td></td>
</tr>
<tr>
<td>0 = Port pin digital input operates with TTL thresholds</td>
<td></td>
</tr>
</tbody>
</table>
### TABLE 11-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>INLVLB</td>
<td>INLVLB7</td>
<td>INLVLB6</td>
<td>INLVLB5</td>
<td>INLVLB4</td>
<td>INLVLB3</td>
<td>INLVLB2</td>
<td>INLVLB1</td>
<td>INLVLB0</td>
<td>133</td>
</tr>
<tr>
<td>LATB</td>
<td>LATB7</td>
<td>LATB6</td>
<td>LATB5</td>
<td>LATB4</td>
<td>LATB3</td>
<td>LATB2</td>
<td>LATB1</td>
<td>LATB0</td>
<td>131</td>
</tr>
<tr>
<td>ODCONB</td>
<td>ODB7</td>
<td>ODB6</td>
<td>ODB5</td>
<td>ODB4</td>
<td>ODB3</td>
<td>ODB2</td>
<td>ODB1</td>
<td>ODB0</td>
<td>133</td>
</tr>
<tr>
<td>PORTB</td>
<td>RB7</td>
<td>RB6</td>
<td>RB5</td>
<td>RB4</td>
<td>RB3</td>
<td>RB2</td>
<td>RB1</td>
<td>RB0</td>
<td>131</td>
</tr>
<tr>
<td>SLRCONB</td>
<td>SLRB7</td>
<td>SLRB6</td>
<td>SLRB5</td>
<td>SLRB4</td>
<td>SLRB3</td>
<td>SLRB2</td>
<td>SLRB1</td>
<td>SLRB0</td>
<td>133</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>WPUB</td>
<td>WPUB7</td>
<td>WPUB6</td>
<td>WPUB5</td>
<td>WPUB4</td>
<td>WPUB3</td>
<td>WPUB2</td>
<td>WPUB1</td>
<td>WPUB0</td>
<td>132</td>
</tr>
</tbody>
</table>

**Legend:**  
- x = unknown, u = unchanged, - = unimplemented locations read as ‘0’. Shaded cells are not used by PORTB.
11.5 PORTC Registers

11.5.1 DATA REGISTER

PORTC is an 8-bit wide bidirectional port. The corresponding data direction register is TRISC (Register 11-18). Setting a TRISC bit (= 1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISC bit (= 0) will make the corresponding PORTC pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTC register (Register 11-17) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATC).

11.5.2 DIRECTION CONTROL

The TRISC register (Register 11-18) controls the PORTC pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISC register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read ‘0’.

11.5.3 INPUT THRESHOLD CONTROL

The INLVLC register (Register 11-24) controls the input voltage threshold for each of the available PORTC input pins. A selection between the Schmitt Trigger CMOS or the TTL compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTC register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 34-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.5.4 OPEN-DRAIN CONTROL

The ODCONC register (Register 11-22) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONC bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONC bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.5.5 SLEW RATE CONTROL

The SLRCONC register (Register 11-23) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONC bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONC bit is cleared, the corresponding port pin drive slews at the maximum rate possible.

11.5.6 ANALOG CONTROL

The ANSELC register (Register 11-20) is used to configure the input mode of an I/O pin to analog. Setting the appropriate ANSELC bit high will cause all digital reads on the pin to be read as ‘0’ and allow analog functions on the pin to operate correctly.

The state of the ANSELC bits has no effect on digital output functions. A pin with TRIS clear and ANSELC set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note: The ANSELC bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to ‘0’ by user software.

11.5.7 PORTC FUNCTIONS AND OUTPUT PRIORITIES

Each pin defaults to the PORT latch data after Reset. Other functions are selected with the peripheral pin select logic. See Section 12.0 “Peripheral Pin Select (PPS) Module” for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELC register. Digital output functions may continue to control the pin when it is in Analog mode.
11.6 Register Definitions: PORTC

REGISTER 11-17: PORTC: PORTC REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>RC7</td>
<td>RC6</td>
<td>RC5</td>
<td>RC4</td>
<td>RC3</td>
<td>RC2</td>
<td>RC1</td>
<td>RC0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0  \text{RC<7:0>: PORTC General Purpose I/O Pin bits}$^{(1)}$
- 1 = Port pin is \geq \text{VIH}
- 0 = Port pin is \leq \text{VIL}

Note 1: Writes to PORTC are actually written to corresponding LATC register. Reads from PORTC register is return of actual I/O pin values.

REGISTER 11-18: TRISC: PORTC TRI-STATE REGISTER

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0  \text{TRISC<7:0>: PORTC Tri-State Control bits}
- 1 = PORTC pin configured as an input (tri-stated)
- 0 = PORTC pin configured as an output

REGISTER 11-19: LATC: PORTC DATA LATCH REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>LATC7</td>
<td>LATC6</td>
<td>LATC5</td>
<td>LATC4</td>
<td>LATC3</td>
<td>LATC2</td>
<td>LATC1</td>
<td>LATC0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0  \text{LATC<7:0>: PORTC Output Latch Value bits}
### REGISTER 11-20: ANSELC: PORTC ANALOG SELECT REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>U-0</th>
<th>U-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-2**  
**ANSC<7:2>:** Analog Select between Analog or Digital Function on Pins RC<7:2>, respectively\(^{1}\)
- 0 = Digital I/O. Pin is assigned to port or digital special function.
- 1 = Analog input. Pin is assigned as analog input\(^{1}\). Digital input buffer disabled.

**bit 1-0**  
**Unimplemented:** Read as ‘0’

**Note 1:** When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

### REGISTER 11-21: WPUC: WEAK PULL-UP PORTC REGISTER\(^{1,2}\)

<table>
<thead>
<tr>
<th>Bit</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>WPUC7</td>
<td>WPUC6</td>
<td>WPUC5</td>
<td>WPUC4</td>
<td>WPUC3</td>
<td>WPUC2</td>
<td>WPUC1</td>
<td>WPUC0</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-0**  
**WPUC<7:0>:** Weak Pull-up Register bits
- 1 = Pull-up enabled
- 0 = Pull-up disabled

**Note 1:** Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.

**Note 2:** The weak pull-up device is automatically disabled if the pin is configured as an output.
### REGISTER 11-22: ODCNC: PORTC OPEN-DRAIN CONTROL REGISTER

<table>
<thead>
<tr>
<th></th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>ODC7</td>
<td>ODC6</td>
<td>ODC5</td>
<td>ODC4</td>
<td>ODC3</td>
<td>ODC2</td>
<td>ODC1</td>
<td>ODC0</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**bit 7-0**  
**ODC<7:0>:** PORTC Open-Drain Enable bits  
For RC<7:0> pins, respectively  
1 = Port pin operates as open-drain drive (sink current only)  
0 = Port pin operates as standard push-pull drive (source and sink current)

### REGISTER 11-23: SLRCONC: PORTC SLEW RATE CONTROL REGISTER

<table>
<thead>
<tr>
<th></th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>SLRC7</td>
<td>SLRC6</td>
<td>SLRC5</td>
<td>SLRC4</td>
<td>SLRC3</td>
<td>SLRC2</td>
<td>SLRC1</td>
<td>SLRC0</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**bit 7-0**  
**SLRC<7:0>:** PORTC Slew Rate Enable bits  
For RC<7:0> pins, respectively  
1 = Port pin slew rate is limited  
0 = Port pin slews at maximum rate

### REGISTER 11-24: INLVLC: PORTC INPUT LEVEL CONTROL REGISTER

<table>
<thead>
<tr>
<th></th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>INLVLC7</td>
<td>INLVLC6</td>
<td>INLVLC5</td>
<td>INLVLC4</td>
<td>INLVLC3</td>
<td>INLVLC2</td>
<td>INLVLC1</td>
<td>INLVLC0</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**bit 7-0**  
**INLVLC<7:0>:** PORTC Input Level Select bits  
For RC<7:0> pins, respectively  
1 = Port pin digital input operates with ST thresholds  
0 = Port pin digital input operates with TTL thresholds
<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
</tr>
<tr>
<td>INLVLC</td>
<td>INLVLC7</td>
<td>INLVLC6</td>
<td>INLVLC5</td>
<td>INLVLC4</td>
<td>INLVLC3</td>
<td>INLVLC2</td>
<td>INLVLC1</td>
<td>INLVLC0</td>
<td>138</td>
</tr>
<tr>
<td>LATC</td>
<td>LATC7</td>
<td>LATC6</td>
<td>LATC5</td>
<td>LATC4</td>
<td>LATC3</td>
<td>LATC2</td>
<td>LATC1</td>
<td>LATC0</td>
<td>136</td>
</tr>
<tr>
<td>ODCONC</td>
<td>ODC7</td>
<td>ODC6</td>
<td>ODC5</td>
<td>ODC4</td>
<td>ODC3</td>
<td>ODC2</td>
<td>ODC1</td>
<td>ODC0</td>
<td>138</td>
</tr>
<tr>
<td>PORTC</td>
<td>RC7</td>
<td>RC6</td>
<td>RC5</td>
<td>RC4</td>
<td>RC3</td>
<td>RC2</td>
<td>RC1</td>
<td>RC0</td>
<td>136</td>
</tr>
<tr>
<td>SLRCONC</td>
<td>SLRC7</td>
<td>SLRC6</td>
<td>SLRC5</td>
<td>SLRC4</td>
<td>SLRC3</td>
<td>SLRC2</td>
<td>SLRC1</td>
<td>SLRC0</td>
<td>138</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>WPUC</td>
<td>WPUC7</td>
<td>WPUC6</td>
<td>WPUC5</td>
<td>WPUC4</td>
<td>WPUC3</td>
<td>WPUC2</td>
<td>WPUC1</td>
<td>WPUC0</td>
<td>137</td>
</tr>
</tbody>
</table>

**Legend:**  
- x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTC.
11.7 PORTD Registers
(PIC16(L)F1717/9 only)

11.7.1 DATA REGISTER
PORTD is an 8-bit wide bidirectional port. The corresponding data direction register is TRISD (Register 11-26). Setting a TRISD bit (\(=1\)) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISD bit (\(=0\)) will make the corresponding PORTD pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTD register (Register 11-25) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATD).

11.7.2 DIRECTION CONTROL
The TRISD register (Register 11-26) controls the PORTD pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISD register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read ‘0’.

11.7.3 INPUT THRESHOLD CONTROL
The INLVLD register (Register 11-32) controls the input voltage threshold for each of the available PORTD input pins. A selection between the Schmitt Trigger CMOS or the TTL compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTD register. See Table 34-4: I/O Ports for more information on threshold levels.

**Note:** Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.7.4 OPEN-DRAIN CONTROL
The ODCOND register (Register 11-30) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCOND bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCOND bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.7.5 SLEW RATE CONTROL
The SLRCOND register (Register 11-31) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCOND bit is set, the corresponding port pin drive is slew rate limited. When an SLRCOND bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

11.7.6 ANALOG CONTROL
The ANSELD register (Register 11-28) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELD bit high will cause all digital reads on the pin to be read as ‘0’ and allow analog functions on the pin to operate correctly.

The state of the ANSELD bits has no effect on digital output functions. A pin with TRIS clear and ANSELD set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

**Note:** The ANSELD bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to ‘0’ by user software.

11.7.7 PORTD FUNCTIONS AND OUTPUT PRIORITIES
Each pin defaults to the PORT latch data after Reset. Other functions are selected with the peripheral pin select logic. See Section 12.0 “Peripheral Pin Select (PPS) Module" for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELD register. Digital output functions may continue to control the pin when it is in Analog mode.
## 11.8 Register Definitions: PORTD

### REGISTER 11-25: PORTD: PORTD REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>RD7</td>
<td>RD6</td>
<td>RD5</td>
<td>RD4</td>
<td>RD3</td>
<td>RD2</td>
<td>RD1</td>
<td>RD0</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **’0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

*Note 1:* Writes to PORTD are actually written to corresponding LATD register. Reads from PORTD register is return of actual I/O pin values.

### REGISTER 11-26: TRISD: PORTD TRI-STATE REGISTER

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>TRISD7</td>
<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **’0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

*Note 1:* PORTD pin configured as an input (tri-stated)

### REGISTER 11-27: LATD: PORTD DATA LATCH REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>LATD7</td>
<td>LATD6</td>
<td>LATD5</td>
<td>LATD4</td>
<td>LATD3</td>
<td>LATD2</td>
<td>LATD1</td>
<td>LATD0</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **’0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

*Note 1:* PORTD Output Latch Value bits
REGISTER 11-28: ANSELD: PORTD ANALOG SELECT REGISTER

<table>
<thead>
<tr>
<th>bit 7-0</th>
<th>ANSD&lt;7:0&gt;</th>
<th>Analog Select between Analog or Digital Function on Pins RD&lt;7:0&gt;, respectively(^{(1)})</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>Digital I/O. Pin is assigned to port or digital special function. (^{(1)})</td>
<td></td>
</tr>
<tr>
<td>1</td>
<td>Analog input. Pin is assigned as analog input(^{(1)}). Digital input buffer disabled.</td>
<td></td>
</tr>
</tbody>
</table>

**Note 1:** When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 11-29: WPUD: WEAK PULL-UP PORTD REGISTER\(^{(1,2)}\)

<table>
<thead>
<tr>
<th>bit 7-0</th>
<th>WPUD&lt;7:0&gt;</th>
<th>Weak Pull-up Register bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Pull-up enabled</td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>Pull-up disabled</td>
<td></td>
</tr>
</tbody>
</table>

**Note 1:** Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.

**Note 2:** The weak pull-up device is automatically disabled if the pin is configured as an output.
## REGISTER 11-30: ODCOND: PORTD OPEN-DRAIN CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>ODD7</td>
<td>ODD6</td>
<td>ODD5</td>
<td>ODD4</td>
<td>ODD3</td>
<td>ODD2</td>
<td>ODD1</td>
<td>ODD0</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-0**
- **ODD<7:0>: PORTD Open-Drain Enable bits**
  For RD<7:0> pins, respectively
  - 1 = Port pin operates as open-drain drive (sink current only)
  - 0 = Port pin operates as standard push-pull drive (source and sink current)

## REGISTER 11-31: SLRCOND: PORTD SLEW RATE CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>SLRD7</td>
<td>SLRD6</td>
<td>SLRD5</td>
<td>SLRD4</td>
<td>SLRD3</td>
<td>SLRD2</td>
<td>SLRD1</td>
<td>SLRD0</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-0**
- **SLRD<7:0>: PORTD Slew Rate Enable bits**
  For RD<7:0> pins, respectively
  - 1 = Port pin slew rate is limited
  - 0 = Port pin slews at maximum rate

## REGISTER 11-32: INLVLD: PORTD INPUT LEVEL CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>INLVLD7</td>
<td>INLVLD6</td>
<td>INLVLD5</td>
<td>INLVLD4</td>
<td>INLVLD3</td>
<td>INLVLD2</td>
<td>INLVLD1</td>
<td>INLVLD0</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-0**
- **INLVLD<7:0>: PORTD Input Level Select bits**
  For RD<7:0> pins, respectively
  - 1 = Port pin digital input operates with ST thresholds
  - 0 = Port pin digital input operates with TTL thresholds
TABLE 11-6: SUMMARY OF REGISTERS ASSOCIATED WITH PORTD

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELD</td>
<td>ANSD7</td>
<td>ANSD6</td>
<td>ANSD5</td>
<td>ANSD4</td>
<td>ANSD3</td>
<td>ANSD2</td>
<td>ANSD1</td>
<td>ANSD0</td>
<td>142</td>
</tr>
<tr>
<td>INLVLD</td>
<td>INLVLD7</td>
<td>INLVLD6</td>
<td>INLVLD5</td>
<td>INLVLD4</td>
<td>INLVLD3</td>
<td>INLVLD2</td>
<td>INLVLD1</td>
<td>INLVLD0</td>
<td>143</td>
</tr>
<tr>
<td>LATD</td>
<td>LATD7</td>
<td>LATD6</td>
<td>LATD5</td>
<td>LATD4</td>
<td>LATD3</td>
<td>LATD2</td>
<td>LATD1</td>
<td>LATD0</td>
<td>141</td>
</tr>
<tr>
<td>ODCOND</td>
<td>ODD7</td>
<td>ODD6</td>
<td>ODD5</td>
<td>ODD4</td>
<td>ODD3</td>
<td>ODD2</td>
<td>ODD1</td>
<td>ODD0</td>
<td>143</td>
</tr>
<tr>
<td>PORTD</td>
<td>RD7</td>
<td>RD6</td>
<td>RD5</td>
<td>RD4</td>
<td>RD3</td>
<td>RD2</td>
<td>RD1</td>
<td>RD0</td>
<td>141</td>
</tr>
<tr>
<td>SLRCOND</td>
<td>SLRD7</td>
<td>SLRD6</td>
<td>SLRD5</td>
<td>SLRD4</td>
<td>SLRD3</td>
<td>SLRD2</td>
<td>SLRD1</td>
<td>SLRD0</td>
<td>143</td>
</tr>
<tr>
<td>TRISD</td>
<td>TRISD7</td>
<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
<td>141</td>
</tr>
<tr>
<td>WPUUD</td>
<td>WPUUD7</td>
<td>WPUUD6</td>
<td>WPUUD5</td>
<td>WPUUD4</td>
<td>WPUUD3</td>
<td>WPUUD2</td>
<td>WPUUD1</td>
<td>WPUUD0</td>
<td>142</td>
</tr>
</tbody>
</table>

**Legend:**  
\(x\) = unknown, \(u\) = unchanged, \(-\) = unimplemented locations read as ‘0’. Shaded cells are not used by PORTD.
11.9 PORTE Registers

11.9.1 DATA REGISTER (RE<2:0>  
PIC16(L)F1717/9 ONLY)

PORTE is a 4-bit wide input port and 3-bit wide output port. The corresponding data direction register is TRISE (Register 11-34). Setting a TRISE bit (= 1) will make the corresponding PORTE pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISE bit (= 0) will make the corresponding PORTE pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTE register (Register 11-33) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATE).

11.9.2 DIRECTION CONTROL (TRISE<2:0>  
PIC16(L)F1717/9 ONLY)

The TRISE register (Register 11-34) controls the PORTE pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISE register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

11.9.3 INPUT THRESHOLD CONTROL  
(PIC16(L)F1717/9 ONLY)

The INLVLE register (Register 11-40) controls the input voltage threshold for each of the available PORTE input pins. A selection between the Schmitt Trigger CMOS or the TTL compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTE register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 34-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.9.4 OPEN-DRAIN CONTROL  
(PIC16(L)F1717/9 ONLY)

The ODCONE register (Register 11-38) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONE bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONE bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.9.5 SLEW RATE CONTROL  
(PIC16(L)F1717/9 ONLY)

The SLRCONE register (Register 11-39) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONE bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONE bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

11.9.6 ANALOG CONTROL  
(PIC16(L)F1717/9 ONLY)

The ANSELE register (Register 11-36) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELE bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELE bits has no effect on digital output functions. A pin with TRIS clear and ANSELE set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note: The ANSELE bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to '0' by user software.

11.9.7 PORTE FUNCTIONS AND OUTPUT PRIORITIES (PIC16(L)F1717/9 ONLY)

Each pin defaults to the PORT latch data after Reset. Other functions are selected with the peripheral pin select logic. See Section 12.0 “Peripheral Pin Select (PPS) Module” for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELE register. Digital output functions may continue to control the pin when it is in Analog mode.
11.10 Register Definitions: PORTE

**REGISTER 11-33: PORTE: PORTE REGISTER**

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7-4</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RE3</td>
<td>RE2(2)</td>
<td>RE1(2)</td>
<td>RE0(2)</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **U** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

- **Unimplemented**: Read as '0'
- **RE<3:0>**: PORTE General Purpose I/O Pin bits(1)
  - 1 = Port pin is ≥ VIH
  - 0 = Port pin is ≤ VIL

**Note 1:** Writes to PORTE are actually written to corresponding LATE register. Reads from PORTE register is return of actual I/O pin values.

**Note 2:** PIC16(L)F1717/9 only.

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R-1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7-4</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TRISE3</td>
<td>TRISE2(1)</td>
<td>TRISE1(1)</td>
<td>TRISE0(1)</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **U** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

- **Unimplemented**: Read as '0'
- **TRISE<3:0>**: PORTE Tri-State Control bits
  - 1 = PORTE pin configured as an input (tri-stated)
  - 0 = PORTE pin configured as an output

**Note 1:** PIC16(L)F1717/9 only.
### REGISTER 11-35: LATE: PORTE DATA LATCH REGISTER\(^{(1)}\)

<table>
<thead>
<tr>
<th>bit 7-3</th>
<th>bit 2-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>LATE2</td>
</tr>
<tr>
<td>—</td>
<td>LATE1</td>
</tr>
<tr>
<td>—</td>
<td>LATE0</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-3** Unimplemented: Read as ‘0’

**bit 2-0** LATE2<2:0>: PORTE Output Latch Value bits

**Note 1:** PIC16(L)F1717/9 only.

### REGISTER 11-36: ANSELE: PORTE ANALOG SELECT REGISTER\(^{(2)}\)

<table>
<thead>
<tr>
<th>bit 7-3</th>
<th>bit 2-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>ANSE2</td>
</tr>
<tr>
<td>—</td>
<td>ANSE1</td>
</tr>
<tr>
<td>—</td>
<td>ANSE0</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-3** Unimplemented: Read as ‘0’

**bit 2-0** ANSE2<2:0>: Analog Select between Analog or Digital Function on Pins RE2<2:0>, respectively\(^{(1)}\)

- 0 = Digital I/O. Pin is assigned to port or digital special function.
- 1 = Analog input. Pin is assigned as analog input\(^{(1)}\). Digital input buffer disabled.

**Note 1:** When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

**Note 2:** PIC16(L)F1717/9 only.
## REGISTER 11-37: WPUE: WEAK PULL-UP PORTE REGISTER\(^{(1,2)}\)

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>WPUE3</td>
<td>WPUE2(^{(3)})</td>
<td>WPUE1(^{(3)})</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

**bit 7**
- **Unimplemented**: Read as ‘0’

**bit 3-0**
- **WPUE<3:0>:** Weak Pull-up Register bits
  - 1 = Pull-up enabled
  - 0 = Pull-up disabled

**Note 1:** Global WPUE\(^n\) bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.
**Note 2:** The weak pull-up device is automatically disabled if the pin is configured as an output.
**Note 3:** PIC16(L)F1717/9 only.

## REGISTER 11-38: ODCONE: PORTE OPEN-DRAIN CONTROL REGISTER\(^{(1)}\)

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ODE2</td>
<td>ODE1</td>
<td>ODE0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

**bit 7-3**
- **Unimplemented**: Read as ‘0’

**bit 2-0**
- **ODE<2:0>:** PORTE Open-Drain Enable bits
  - For RE<2:0> pins, respectively
  - 1 = Port pin operates as open-drain drive (sink current only)
  - 0 = Port pin operates as standard push-pull drive (source and sink current)

**Note 1:** PIC16(L)F1717/9 only.
**REGISTER 11-39: SLRCONE: PORTE SLEW RATE CONTROL REGISTER\(^{(1)}\)

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>SLRE2</td>
<td>SLRE1</td>
<td>SLRE0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

- **bit 7-3** Unimplemented: Read as ‘0’
- **bit 2-0** SLRE<2:0>: PORTE Slew Rate Enable bits
  - For RE<2:0> pins, respectively
  - 1 = Port pin slew rate is limited
  - 0 = Port pin slews at maximum rate

**Note 1:** PIC16(L)F1717/9 only.

---

**REGISTER 11-40: INLVLE: PORTE INPUT LEVEL CONTROL REGISTER\(^{(1)}\)

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>INLVLE3</td>
<td>INLVLE2</td>
<td>INLVLE1</td>
<td>INLVLE0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

- **bit 7-4** Unimplemented: Read as ‘0’
- **bit 3-0** INLVLE<3:0>: PORTE Input Level Select bits
  - For RE<3:0> pins, respectively
  - 1 = Port pin digital input operates with ST thresholds
  - 0 = Port pin digital input operates with TTL thresholds

**Note 1:** PIC16(L)F1717/9 only.
### TABLE 11-7: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELE(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSE2</td>
<td>ANSE1</td>
<td>ANSE0</td>
</tr>
<tr>
<td>INLVLE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>INLVLE3</td>
<td>INLVLE2(1)</td>
<td>INLVLE1(1)</td>
</tr>
<tr>
<td>LATE(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>LATE2</td>
<td>LATE1</td>
<td>LATE0</td>
</tr>
<tr>
<td>ODCONE(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ODE2</td>
<td>ODE1</td>
<td>ODE0</td>
</tr>
<tr>
<td>PORTE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RE3</td>
<td>RE2(1)</td>
<td>RE1(1)</td>
</tr>
<tr>
<td>SLRCONE(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>SLRE2</td>
<td>SLRE1</td>
<td>SLRE0</td>
</tr>
<tr>
<td>TRISE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TRISE3</td>
<td>TRISE2(1)</td>
<td>TRISE1(1)</td>
</tr>
<tr>
<td>WPUE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>WPUE3</td>
<td>WPUE2(1)</td>
<td>WPUE1(1)</td>
</tr>
</tbody>
</table>

**Legend:**  
- `x` = unknown,  
- `u` = unchanged,  
- `=` = unimplemented locations read as ‘0’. Shaded cells are not used by PORTE.

**Note 1:** PIC16(L)F1717/9 only.
12.0 PERIPHERAL PIN SELECT (PPS) MODULE

The Peripheral Pin Select (PPS) module connects peripheral inputs and outputs to the device I/O pins. Only digital signals are included in the selections. All analog inputs and outputs remain fixed to their assigned pins. Input and output selections are independent as shown in the simplified block diagram Figure 12-1.

12.1 PPS Inputs

Each peripheral has a PPS register with which the inputs to the peripheral are selected. Inputs include the device pins.

Multiple peripherals can operate from the same source simultaneously. Port reads always return the pin level regardless of peripheral PPS selection. If a pin also has associated analog functions, the ANSEL bit for that pin must be cleared to enable the digital input buffer.

Although every peripheral has its own PPS input selection register, the selections are identical for every peripheral as shown in Register 12-1.

Note: The notation “xxx” in the register name is a place holder for the peripheral identifier. For example, CLC1PPS.

FIGURE 12-1: SIMPLIFIED PPS BLOCK DIAGRAM

12.2 PPS Outputs

Each I/O pin has a PPS register with which the pin output source is selected. With few exceptions, the port TRIS control associated with that pin retains control over the pin output driver. Peripherals that control the pin output driver as part of the peripheral operation will override the TRIS control as needed. These peripherals include:

- EUSART (synchronous operation)
- MSSP (I²C)
- COG (auto-shutdown)

Although every pin has its own PPS peripheral selection register, the selections are identical for every pin as shown in Register 12-2.

Note: The notation "Rxy" is a place holder for the pin identifier. For example, RA0PPS.
12.3 Bidirectional Pins

PPS selections for peripherals with bidirectional signals on a single pin must be made so that the PPS input and PPS output select the same pin. Peripherals that have bidirectional signals include:

- EUSART (synchronous operation)
- MSSP (I^2C)

**Note:** The I^2C default input pins are I^2C and SMBus compatible and are the only pins on the device with this compatibility.

12.4 PPS Lock

The PPS includes a mode in which all input and output selections can be locked to prevent inadvertent changes. PPS selections are locked by setting the PPSLOCKED bit of the PPSLOCK register. Setting and clearing this bit requires a special sequence as an extra precaution against inadvertent changes. Examples of setting and clearing the PPSLOCKED bit are shown in Example 12-1.

**EXAMPLE 12-1: PPS LOCK/UNLOCK SEQUENCE**

```assembly
; suspend interrupts
  bcf INTCON,GIE
; BANKSEL PPSLOCK ; set bank
; required sequence, next 5 instructions
  movlw 0x55
  movwf PPSLOCK
  movlw 0xAA
  movwf PPSLOCK
; Set PPSLOCKED bit to disable writes or
; Clear PPSLOCKED bit to enable writes
  bsf PPSLOCK,PPSLOCKED
; restore interrupts
  bsf INTCON,GIE
```

12.5 PPS Permanent Lock

The PPS can be permanently locked by setting the PPS1WAY Configuration bit. When this bit is set, the PPSLOCKED bit can only be cleared and set one time after a device Reset. This allows for clearing the PPSLOCKED bit so that the input and output selections can be made during initialization. When the PPSLOCKED bit is set after all selections have been made, it will remain set and cannot be cleared until after the next device Reset event.

12.6 Operation During Sleep

PPS input and output selections are unaffected by Sleep.

12.7 Effects of a Reset

A device Power-On-Reset (POR) clears all PPS input and output selections to their default values. All other Resets leave the selections unchanged. Default input selections are shown in pin allocation Table 1.

Note:
The I^2C default input pins are I^2C and SMBus compatible and are the only pins on the device with this compatibility.

12.8 Operation During Sleep

PPS input and output selections are unaffected by Sleep.

12.7 Effects of a Reset

A device Power-On-Reset (POR) clears all PPS input and output selections to their default values. All other Resets leave the selections unchanged. Default input selections are shown in pin allocation Table 1.

Note:
The I^2C default input pins are I^2C and SMBus compatible and are the only pins on the device with this compatibility.
12.8 Register Definitions: PPS Input Selection

REGISTER 12-1: xxxPPS: PERIPHERAL xxx INPUT SELECTION

| bit 7-0 | U-0  | U-0  | U-0  | R/W-q/u | R/W-q/u | R/W-q/u | R/W-q/u | xxxPPS<4:0> |
|-----------------|-----------------|-----------------|-----------------|-----------------|-----------------|-----------------|-----------------|

Legend:

R = Readable bit  W = Writable bit  U = Unimplemented bit, read as ‘0’

u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets

‘1’ = Bit is set  ‘0’ = Bit is cleared  q = value depends on peripheral

bit 7-5  Unimplemented: Read as ‘0’

bit 4-3  xxxPPS<4:3>: Peripheral xxx Input PORTx Selection bits

See Table 12-1 for the list of available ports for each peripheral.

11 = Peripheral input is from PORTD (PIC16(L)F1717/9 only)
10 = Peripheral input is from PORTC
01 = Peripheral input is from PORTB
00 = Peripheral input is from PORTA

bit 2-0  xxxPPS<2:0>: Peripheral xxx Input PORTx Bit Selection bits

111 = Peripheral input is from PORTx Bit 7 (Rx7)
110 = Peripheral input is from PORTx Bit 6 (Rx6)
101 = Peripheral input is from PORTx Bit 5 (Rx5)
100 = Peripheral input is from PORTx Bit 4 (Rx4)
011 = Peripheral input is from PORTx Bit 3 (Rx3)
010 = Peripheral input is from PORTx Bit 2 (Rx2)
001 = Peripheral input is from PORTx Bit 1 (Rx1)
000 = Peripheral input is from PORTx Bit 0 (Rx0)

TABLE 12-1: AVAILABLE PORTS FOR INPUT BY PERIPHERAL

<table>
<thead>
<tr>
<th>Peripheral</th>
<th>Register</th>
<th>PIC16(L)F1717/8/9</th>
<th>PIC16(L)F1718</th>
<th>PIC16(L)F1717/9</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>PORTA</td>
<td>PORTB</td>
<td>PORTC</td>
</tr>
<tr>
<td>PIN interrupt</td>
<td>INTPPS</td>
<td>•</td>
<td>•</td>
<td></td>
</tr>
<tr>
<td>Timer0 clock</td>
<td>T0CKIPPS</td>
<td>•</td>
<td>•</td>
<td></td>
</tr>
<tr>
<td>Timer1 clock</td>
<td>T1CKIPPS</td>
<td>•</td>
<td>•</td>
<td></td>
</tr>
<tr>
<td>Timer1 gate</td>
<td>T1GPPS</td>
<td>•</td>
<td>•</td>
<td></td>
</tr>
<tr>
<td>CCP1</td>
<td>CCP1PPS</td>
<td>•</td>
<td>•</td>
<td></td>
</tr>
<tr>
<td>CCP2</td>
<td>CCP2PPS</td>
<td>•</td>
<td>•</td>
<td></td>
</tr>
<tr>
<td>COG</td>
<td>COGINPPS</td>
<td>•</td>
<td>•</td>
<td></td>
</tr>
<tr>
<td>MSSP</td>
<td>SSPCLKPPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>MSSP</td>
<td>SSPDATPPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>MSSP</td>
<td>SSPSSPPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>EUSART</td>
<td>RXPPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>EUSART</td>
<td>CKPPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>All CLCs</td>
<td>CLCIN0PPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>All CLCs</td>
<td>CLCIN1PPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>All CLCs</td>
<td>CLCIN2PPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
<tr>
<td>All CLCs</td>
<td>CLCIN3PPS</td>
<td>•</td>
<td>•</td>
<td>•</td>
</tr>
</tbody>
</table>

Example: CCP1PPS = 0x0B selects RB3 as the input to CCP1.

Note: Inputs are not available on all ports. A check in a port column of a peripheral row indicates that the port selection is valid for that peripheral. Unsupported ports will input a ‘0’. 
REGISTER 12-2: RxPPS: PIN Rx OUTPUT SOURCE SELECTION REGISTER

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/u</th>
<th>R/W-0/u</th>
<th>R/W-0/u</th>
<th>R/W-0/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>RxyPPS&lt;4:0&gt;</td>
</tr>
<tr>
<td>bit 0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:

- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

bit 7-5: Unimplemented: Read as '0'

bit 4-0:

<table>
<thead>
<tr>
<th>RxyPPS&lt;4:0&gt;</th>
<th>Output Signal</th>
</tr>
</thead>
<tbody>
<tr>
<td>11xxx</td>
<td>Reserved</td>
</tr>
<tr>
<td>10111</td>
<td>C2OUT</td>
</tr>
<tr>
<td>10110</td>
<td>C1OUT</td>
</tr>
<tr>
<td>10101</td>
<td>DT&lt;2&gt;</td>
</tr>
<tr>
<td>10100</td>
<td>TX/CK&lt;2&gt;</td>
</tr>
<tr>
<td>10011</td>
<td>Reserved</td>
</tr>
<tr>
<td>10010</td>
<td>Reserved</td>
</tr>
<tr>
<td>10001</td>
<td>SDO/SDA&lt;2&gt;</td>
</tr>
<tr>
<td>10000</td>
<td>SCK/SCL&lt;2&gt;</td>
</tr>
<tr>
<td>01111</td>
<td>PWM4OUT</td>
</tr>
<tr>
<td>01110</td>
<td>PWM3OUT</td>
</tr>
<tr>
<td>01101</td>
<td>CCP2</td>
</tr>
<tr>
<td>01100</td>
<td>CCP1</td>
</tr>
<tr>
<td>01011</td>
<td>COG1D&lt;2&gt;</td>
</tr>
<tr>
<td>01010</td>
<td>COG1C&lt;2&gt;</td>
</tr>
<tr>
<td>01001</td>
<td>COG1B&lt;2&gt;</td>
</tr>
<tr>
<td>01000</td>
<td>COG1A&lt;2&gt;</td>
</tr>
<tr>
<td>00111</td>
<td>CLC4OUT</td>
</tr>
<tr>
<td>00110</td>
<td>CLC3OUT</td>
</tr>
<tr>
<td>00101</td>
<td>CLC2OUT</td>
</tr>
<tr>
<td>00100</td>
<td>CLC1OUT</td>
</tr>
<tr>
<td>00011</td>
<td>NCO1OUT</td>
</tr>
<tr>
<td>00010</td>
<td>Reserved</td>
</tr>
<tr>
<td>00001</td>
<td>Reserved</td>
</tr>
<tr>
<td>00000</td>
<td>LATxy</td>
</tr>
</tbody>
</table>

Table 12-2: AVAILABLE PORTS FOR OUTPUT BY PERIPHERAL

Example: RB3PPS = 0x16 selects RB3 as the comparator 1 output.

Note 1: Outputs are not available on all ports. A check in a port column of a peripheral row indicates that the peripheral selection is valid for that port. Reserved output signals will output a '0'.

2: TRIS control is overridden by the peripheral as required.
REGISTER 12-3: PPSLOCK: PPS LOCK REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>PPSLOCKED</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as '0'
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set  '0' = Bit is cleared

bit 7-1  Unimplemented: Read as '0'
bit 0  PPSLOCKED: PPS Locked bit

1 = PPS is locked. PPS selections can not be changed.
0 = PPS is not locked. PPS selections can be changed.

TABLE 12-3: SUMMARY OF REGISTERS ASSOCIATED WITH THE PPS MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>PPSLOCK</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>155</td>
</tr>
<tr>
<td>INTPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>T0CKIPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>T1CKIPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>T1GPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>CCP1PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>CCP2PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>COGINPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>SSPCLKPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>SSPDATPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>SSPSSPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>RXPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>CKPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>CLCIN0PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>CLCIN1PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>CLCIN2PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>CLCIN3PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153</td>
</tr>
<tr>
<td>RA0PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>RA1PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>RA2PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>RA4PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>RA5PPS</td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>RA6PPS</td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>RA7PPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented, read as '0'. Shaded cells are unused by the PPS module.

Note 1: PIC16(L)F1717/9 only.
## TABLE 12-3: SUMMARY OF REGISTERS ASSOCIATED WITH THE PPS MODULE (CONTINUED)

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>RB0PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RB1PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RB2PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RB3PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RB4PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RB5PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
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<td>154</td>
</tr>
<tr>
<td>RB6PPS</td>
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<td>—</td>
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<td>—</td>
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<td>154</td>
</tr>
<tr>
<td>RB7PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RC0PPS</td>
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<td>—</td>
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<td>—</td>
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<td>154</td>
</tr>
<tr>
<td>RC1PPS</td>
<td>—</td>
<td>—</td>
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<td>—</td>
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<td>154</td>
</tr>
<tr>
<td>RC2PPS</td>
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<tr>
<td>RC3PPS</td>
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<tr>
<td>RC4PPS</td>
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<td>154</td>
</tr>
<tr>
<td>RC5PPS</td>
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<td>154</td>
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<tr>
<td>RC6PPS</td>
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<tr>
<td>RC7PPS</td>
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<td>154</td>
</tr>
<tr>
<td>RD0PPS(f)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RD1PPS(f)</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
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<td>154</td>
</tr>
<tr>
<td>RD2PPS(f)</td>
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<td>154</td>
</tr>
<tr>
<td>RD3PPS(f)</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>154</td>
</tr>
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<td>RD4PPS(f)</td>
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<tr>
<td>RD5PPS(f)</td>
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<tr>
<td>RD6PPS(f)</td>
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<tr>
<td>RD7PPS(f)</td>
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<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RE0PPS(f)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>RE1PPS(f)</td>
<td>—</td>
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<td>—</td>
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<td>RE2PPS(f)</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented, read as ‘0’. Shaded cells are unused by the PPS module.

**Note 1:** PIC16(L)F1717/9 only.
13.0 INTERRUPT-ON-CHANGE

All pins on all ports can be configured to operate as Interrupt-on-Change (IOC) pins. An interrupt can be generated by detecting a signal that has either a rising edge or a falling edge. Any individual pin, or combination of pins, can be configured to generate an interrupt. The interrupt-on-change module has the following features:

- Interrupt-on-Change enable (Master Switch)
- Individual pin configuration
- Rising and falling edge detection
- Individual pin interrupt flags

Figure 13-1 is a block diagram of the IOC module.

13.1 Enabling the Module

To allow individual pins to generate an interrupt, the IOCIE bit of the INTCON register must be set. If the IOCIE bit is disabled, the edge detection on the pin will still occur, but an interrupt will not be generated.

13.2 Individual Pin Configuration

For each pin, a rising edge detector and a falling edge detector are present. To enable a pin to detect a rising edge, the associated bit of the IOCxP register is set. To enable a pin to detect a falling edge, the associated bit of the IOCxN register is set.

A pin can be configured to detect rising and falling edges simultaneously by setting the associated bits in both of the IOCxP and IOCxN registers.

13.3 Interrupt Flags

The bits located in the IOCxF registers are status flags that correspond to the interrupt-on-change pins of each port. If an expected edge is detected on an appropriately enabled pin, then the status flag for that pin will be set, and an interrupt will be generated if the IOCIE bit is set. The IOCIF bit of the INTCON register reflects the status of all IOCxF bits.

13.4 Clearing Interrupt Flags

The individual status flags, (IOCxF register bits), can be cleared by resetting them to zero. If another edge is detected during this clearing operation, the associated status flag will be set at the end of the sequence, regardless of the value actually being written.

In order to ensure that no detected edge is lost while clearing flags, only AND operations masking out known changed bits should be performed. The following sequence is an example of what should be performed.

EXAMPLE 13-1: CLEARING INTERRUPT FLAGS
(PORTA EXAMPLE)

```
MOVLW 0xff
XORWF IOCAF, W
ANDWF IOCAF, F
```

13.5 Operation in Sleep

The interrupt-on-change interrupt sequence will wake the device from Sleep mode, if the IOCIE bit is set.

If an edge is detected while in Sleep mode, the affected IOCxF register will be updated prior to the first instruction executed out of Sleep.
FIGURE 13-1: INTERRUPT-ON-CHANGE BLOCK DIAGRAM (PORTA EXAMPLE)

- IOCANx
- IOCAPx
- Q2
  - data bus = 0 or 1
  - write IOCAFx
- Q4
  - edge detect
- IOC interrupt to CPU core
- from all other IOCnFx individual pin detectors

Diagram showing the block diagram of the interrupt-on-change feature for PORTA example.
### 13.6 Register Definitions: Interrupt-on-Change Control

**REGISTER 13-1: ** IOCAP: INTERRUPT-ON-CHANGE PORTA POSITIVE EDGE REGISTER

<p>| | | | | | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
</tr>
<tr>
<td>IOCAP7</td>
<td>IOCAP6</td>
<td>IOCAP5</td>
<td>IOCAP4</td>
<td>IOCAP3</td>
<td>IOCAP2</td>
<td>IOCAP1</td>
<td>IOCAP0</td>
</tr>
</tbody>
</table>

| bit 7 | bit 0 |

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **‘0’** = Bit is cleared
- **’n/n’** = Value at POR and BOR/Value at all other Resets

**bit 7-0 IOCAP<7:0>:** Interrupt-on-Change PORTA Positive Edge Enable bits

- **1** = Interrupt-on-Change enabled on the pin for a positive going edge. IOCAFx bit and IOCIF flag will be set upon detecting an edge.
- **0** = Interrupt-on-Change disabled for the associated pin.

---

**REGISTER 13-2: ** IOCAN: INTERRUPT-ON-CHANGE PORTA NEGATIVE EDGE REGISTER

<p>| | | | | | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
</tr>
<tr>
<td>IOCAN7</td>
<td>IOCAN6</td>
<td>IOCAN5</td>
<td>IOCAN4</td>
<td>IOCAN3</td>
<td>IOCAN2</td>
<td>IOCAN1</td>
<td>IOCAN0</td>
</tr>
</tbody>
</table>

| bit 7 | bit 0 |

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **‘0’** = Bit is cleared
- **’n/n’** = Value at POR and BOR/Value at all other Resets

**bit 7-0 IOCAN<7:0>:** Interrupt-on-Change PORTA Negative Edge Enable bits

- **1** = Interrupt-on-Change enabled on the pin for a negative going edge. IOCAFx bit and IOCIF flag will be set upon detecting an edge.
- **0** = Interrupt-on-Change disabled for the associated pin.

---

**REGISTER 13-3: ** IOCAF: INTERRUPT-ON-CHANGE PORTA FLAG REGISTER

<p>| | | | | | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>R/W/HS-0/0</td>
<td>R/W/HS-0/0</td>
<td>R/W/HS-0/0</td>
<td>R/W/HS-0/0</td>
<td>R/W/HS-0/0</td>
<td>R/W/HS-0/0</td>
<td>R/W/HS-0/0</td>
<td></td>
</tr>
<tr>
<td>IOCAF7</td>
<td>IOCAF6</td>
<td>IOCAF5</td>
<td>IOCAF4</td>
<td>IOCAF3</td>
<td>IOCAF2</td>
<td>IOCAF1</td>
<td>IOCAF0</td>
</tr>
</tbody>
</table>

| bit 7 | bit 0 |

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **‘0’** = Bit is cleared
- **’n/n’** = Value at POR and BOR/Value at all other Resets
- **HS** = Bit is set in hardware

**bit 7-0 IOCAF<7:0>:** Interrupt-on-Change PORTA Flag bits

- **1** = An enabled change was detected on the associated pin.
  - Set when IOCAPx = 1 and a rising edge was detected on RAx, or when IOCANx = 1 and a falling edge was detected on RAx.
- **0** = No change was detected, or the user cleared the detected change.
REGISTER 13-4: IOCBP: INTERRUPT-ON-CHANGE PORTB POSITIVE EDGE REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>IOCBP7</td>
<td>IOCBP6</td>
<td>IOCBP5</td>
<td>IOCBP4</td>
<td>IOCBP3</td>
<td>IOCBP2</td>
<td>IOCBP1</td>
<td>IOCBP0</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as ‘0’
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set  ‘0’ = Bit is cleared

bit 7-0  IOCBP<7:0>: Interrupt-on-Change PORTB Positive Edge Enable bits
1 = Interrupt-on-Change enabled on the pin for a positive going edge. IOCBFx bit and IOCIF flag will be set upon detecting an edge.
0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-5: IOCBN: INTERRUPT-ON-CHANGE PORTB NEGATIVE EDGE REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>IOCBN7</td>
<td>IOCBN6</td>
<td>IOCBN5</td>
<td>IOCBN4</td>
<td>IOCBN3</td>
<td>IOCBN2</td>
<td>IOCBN1</td>
<td>IOCBN0</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as ‘0’
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set  ‘0’ = Bit is cleared

bit 7-0  IOCBN<7:0>: Interrupt-on-Change PORTB Negative Edge Enable bits
1 = Interrupt-on-Change enabled on the pin for a negative going edge. IOCBFx bit and IOCIF flag will be set upon detecting an edge.
0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-6: IOCBF: INTERRUPT-ON-CHANGE PORTB FLAG REGISTER

<table>
<thead>
<tr>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>IOCBF7</td>
<td>IOCBF6</td>
<td>IOCBF5</td>
<td>IOCBF4</td>
<td>IOCBF3</td>
<td>IOCBF2</td>
<td>IOCBF1</td>
<td>IOCBF0</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as ‘0’
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set  ‘0’ = Bit is cleared  HS - Bit is set in hardware

bit 7-0  IOCBF<7:0>: Interrupt-on-Change PORTB Flag bits
1 = An enabled change was detected on the associated pin.
Set when IOCBPx = 1 and a rising edge was detected on RBx, or when IOCBNx = 1 and a falling edge was detected on RBx.
0 = No change was detected, or the user cleared the detected change.
REGISTER 13-7: IOCCP: INTERRUPT-ON-CHANGE PORTC POSITIVE EDGE REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
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<tbody>
<tr>
<td>IOCCP7</td>
<td>IOCCP6</td>
<td>IOCCP5</td>
<td>IOCCP4</td>
<td>IOCCP3</td>
<td>IOCCP2</td>
<td>IOCCP1</td>
<td>IOCCP0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0 IOCCP<7:0>: Interrupt-on-Change PORTC Positive Edge Enable bits
- 1 = Interrupt-on-Change enabled on the pin for a positive going edge. IOCCFx bit and IOCIF flag will be set upon detecting an edge.
- 0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-8: IOCCN: INTERRUPT-ON-CHANGE PORTC NEGATIVE EDGE REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>IOCCN7</td>
<td>IOCCN6</td>
<td>IOCCN5</td>
<td>IOCCN4</td>
<td>IOCCN3</td>
<td>IOCCN2</td>
<td>IOCCN1</td>
<td>IOCCN0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0 IOCCN<7:0>: Interrupt-on-Change PORTC Negative Edge Enable bits
- 1 = Interrupt-on-Change enabled on the pin for a negative going edge. IOCCFx bit and IOCIF flag will be set upon detecting an edge.
- 0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-9: IOCCF: INTERRUPT-ON-CHANGE PORTC FLAG REGISTER

<table>
<thead>
<tr>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>IOCCF7</td>
<td>IOCCF6</td>
<td>IOCCF5</td>
<td>IOCCF4</td>
<td>IOCCF3</td>
<td>IOCCF2</td>
<td>IOCCF1</td>
<td>IOCCF0</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- HS - Bit is set in hardware

bit 7-0 IOCCF<7:0>: Interrupt-on-Change PORTC Flag bits
- 1 = An enabled change was detected on the associated pin.
  - Set when IOCCPx = 1 and a rising edge was detected on RCx, or when IOCCNx = 1 and a falling edge was detected on RCx.
- 0 = No change was detected, or the user cleared the detected change.
**REGISTER 13-10: IOCEP: INTERRUPT-ON-CHANGE PORTE POSITIVE EDGE REGISTER**

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7-4</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-4 **Unimplemented:** Read as ‘0’

bit 3 **IOCEP3:** Interrupt-on-Change PORTE Positive Edge Enable bits
- 1 = Interrupt-on-Change enabled on the pin for a positive going edge. IOCEFx bit and IOCIF flag will be set upon detecting an edge.
- 0 = Interrupt-on-Change disabled for the associated pin.

**REGISTER 13-11: IOCEN: INTERRUPT-ON-CHANGE PORTE NEGATIVE EDGE REGISTER**

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7-4</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-4 **Unimplemented:** Read as ‘0’

bit 3 **IOCEN3:** Interrupt-on-Change PORTE Negative Edge Enable bits
- 1 = Interrupt-on-Change enabled on the pin for a negative going edge. IOCEFx bit and IOCIF flag will be set upon detecting an edge.
- 0 = Interrupt-on-Change disabled for the associated pin.

bit 2-0 **Unimplemented:** Read as ‘0’
REGISTER 13-12: IOCEF: INTERRUPT-ON-CHANGE PORTE FLAG REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W/HS-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- HS - Bit is set in hardware

bit 7-4 Unimplemented: Read as ‘0’

bit 3 IOCEF3: Interrupt-on-Change PORTE Flag bits
1 = An enabled change was detected on the associated pin.
   Set when IOCEPx = 1 and a rising edge was detected on REx, or when IOCENx = 1 and a falling edge was detected on REx.
0 = No change was detected, or the user cleared the detected change.

bit 2-0 Unimplemented: Read as ‘0’

TABLE 13-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPT-ON-CHANGE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>126</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
</tr>
<tr>
<td>INTCN</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCF</td>
<td>91</td>
</tr>
<tr>
<td>IOCAF</td>
<td>IOCAF7</td>
<td>IOCAF6</td>
<td>IOCAF5</td>
<td>IOCAF4</td>
<td>IOCAF3</td>
<td>IOCAF2</td>
<td>IOCAF1</td>
<td>IOCAF0</td>
<td>159</td>
</tr>
<tr>
<td>IOCAN</td>
<td>IOCAN7</td>
<td>IOCAN6</td>
<td>IOCAN5</td>
<td>IOCAN4</td>
<td>IOCAN3</td>
<td>IOCAN2</td>
<td>IOCAN1</td>
<td>IOCAN0</td>
<td>159</td>
</tr>
<tr>
<td>IOCAP</td>
<td>IOCAP7</td>
<td>IOCAP6</td>
<td>IOCAP5</td>
<td>IOCAP4</td>
<td>IOCAP3</td>
<td>IOCAP2</td>
<td>IOCAP1</td>
<td>IOCAP0</td>
<td>159</td>
</tr>
<tr>
<td>IOCBF</td>
<td>IOCBF7</td>
<td>IOCBF6</td>
<td>IOCBF5</td>
<td>IOCBF4</td>
<td>IOCBF3</td>
<td>IOCBF2</td>
<td>IOCBF1</td>
<td>IOCBF0</td>
<td>160</td>
</tr>
<tr>
<td>IOCBN</td>
<td>IOCBN7</td>
<td>IOCBN6</td>
<td>IOCBN5</td>
<td>IOCBN4</td>
<td>IOCBN3</td>
<td>IOCBN2</td>
<td>IOCBN1</td>
<td>IOCBN0</td>
<td>160</td>
</tr>
<tr>
<td>IOCBP</td>
<td>IOCBP7</td>
<td>IOCBP6</td>
<td>IOCBP5</td>
<td>IOCBP4</td>
<td>IOCBP3</td>
<td>IOCBP2</td>
<td>IOCBP1</td>
<td>IOCBP0</td>
<td>160</td>
</tr>
<tr>
<td>IOCCF</td>
<td>IOCCF7</td>
<td>IOCCF6</td>
<td>IOCCF5</td>
<td>IOCCF4</td>
<td>IOCCF3</td>
<td>IOCCF2</td>
<td>IOCCF1</td>
<td>IOCCF0</td>
<td>161</td>
</tr>
<tr>
<td>IOCCN</td>
<td>IOCCN7</td>
<td>IOCCN6</td>
<td>IOCCN5</td>
<td>IOCCN4</td>
<td>IOCCN3</td>
<td>IOCCN2</td>
<td>IOCCN1</td>
<td>IOCCN0</td>
<td>161</td>
</tr>
<tr>
<td>IOCCP</td>
<td>IOCCP7</td>
<td>IOCCP6</td>
<td>IOCCP5</td>
<td>IOCCP4</td>
<td>IOCCP3</td>
<td>IOCCP2</td>
<td>IOCCP1</td>
<td>IOCCP0</td>
<td>161</td>
</tr>
<tr>
<td>IOCEF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>163</td>
</tr>
<tr>
<td>IOCEN</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>162</td>
</tr>
<tr>
<td>IOCEP</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>162</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used by interrupt-on-change.
14.0 FIXED VOLTAGE REFERENCE (FVR)

The Fixed Voltage Reference, or FVR, is a stable voltage reference, independent of VDD, with 1.024V, 2.048V or 4.096V selectable output levels. The output of the FVR can be configured to supply a reference voltage to the following:

- ADC input channel
- ADC positive reference
- Comparator positive input
- Digital-to-Analog Converter (DAC)

The FVR can be enabled by setting the FVREN bit of the FVRCON register.

14.1 Independent Gain Amplifiers

The output of the FVR supplied to the ADC, comparators, and DAC is routed through two independent programmable gain amplifiers. Each amplifier can be programmed for a gain of 1x, 2x or 4x, to produce the three possible voltage levels.

The ADFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the ADC module. Reference Section 21.0 “Analog-to-Digital Converter (ADC) Module” for additional information.

The CDAFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the DAC and comparator module. Reference Section 23.0 “8-Bit Digital-to-Analog Converter (DAC1) Module” and Section 16.0 “Comparator Module” for additional information.

14.2 FVR Stabilization Period

When the Fixed Voltage Reference module is enabled, it requires time for the reference and amplifier circuits to stabilize. Once the circuits stabilize and are ready for use, the FVRRDY bit of the FVRCON register will be set. See Figure 35-76: FVR Stabilization Period, PIC16LF1717/8/9 Only.
FIGURE 14-1: VOLTAGE REFERENCE BLOCK DIAGRAM

TABLE 14-1: PERIPHERALS REQUIRING THE FIXED VOLTAGE REFERENCE (FVR)

<table>
<thead>
<tr>
<th>Peripheral</th>
<th>Conditions</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>HFINTOSC</td>
<td>FOSC&lt;2:0&gt; = 100 and IRCF&lt;3:0&gt; ≠ 000x</td>
<td>INTOSC is active and device is not in Sleep</td>
</tr>
<tr>
<td>BOR</td>
<td>BOREN&lt;1:0&gt; = 11</td>
<td>BOR always enabled</td>
</tr>
<tr>
<td></td>
<td>BOREN&lt;1:0&gt; = 10 and BORFS = 1</td>
<td>BOR disabled in Sleep mode, BOR Fast Start enabled</td>
</tr>
<tr>
<td></td>
<td>BOREN&lt;1:0&gt; = 01 and BORFS = 1</td>
<td>BOR under software control, BOR Fast Start enabled</td>
</tr>
<tr>
<td>LDO</td>
<td>All PIC16F1717/8/9 devices, when VREGPM = 1 and not in Sleep</td>
<td>The device runs off of the ULP regulator when in Sleep mode</td>
</tr>
</tbody>
</table>
14.3 Register Definitions: FVR Control

REGISTER 14-1: FVRCON: FIXED VOLTAGE REFERENCE CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>R/W</th>
<th>q/l</th>
<th>R/W</th>
<th>q/l</th>
<th>R/W</th>
<th>q/l</th>
<th>R/W</th>
<th>q/l</th>
<th>R/W</th>
<th>q/l</th>
<th>R/W</th>
<th>q/l</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>FVREN: Fixed Voltage Reference Enable bit</td>
<td>R</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>FVRRDY: Fixed Voltage Reference Ready Flag bit</td>
<td>R</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>5</td>
<td>TSEN: Temperature Indicator Enable bit</td>
<td>R</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>TSRNG: Temperature Indicator Range Selection bit</td>
<td>R</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>3-2</td>
<td>CDAFVR&lt;1:0&gt;: Comparator FVR Buffer Gain Selection bits</td>
<td>R</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>1-0</td>
<td>ADFVR&lt;1:0&gt;: ADC FVR Buffer Gain Selection bit</td>
<td>R</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td>q</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
W = Writable bit
U = Unimplemented bit, read as ‘0’
u = Bit is unchanged
x = Bit is unknown
-q/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set
‘0’ = Bit is cleared
q = Value depends on condition

Note 1: FVRRDY is always ‘1’ on PIC16(L)F1717/8/9 only.
2: Fixed Voltage Reference output cannot exceed VDD.
3: See Section 15.0 “Temperature Indicator Module” for additional information.

TABLE 14-2: SUMMARY OF REGISTERS ASSOCIATED WITH FIXED VOLTAGE REFERENCE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>FVRCON</td>
<td>FVREN</td>
<td>FVRRDY</td>
<td>TSEN</td>
<td>TSRNG</td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>ADFVR&lt;1:0&gt;</td>
<td></td>
<td></td>
<td>166</td>
</tr>
</tbody>
</table>

Legend: Shaded cells are not used with the Fixed Voltage Reference.
15.0 TEMPERATURE INDICATOR MODULE

This family of devices is equipped with a temperature circuit designed to measure the operating temperature of the silicon die. The circuit’s range of operating temperature falls between -40°C and +85°C. The output is a voltage that is proportional to the device temperature. The output of the temperature indicator is internally connected to the device ADC.

The circuit may be used as a temperature threshold detector or a more accurate temperature indicator, depending on the level of calibration performed. A one-point calibration allows the circuit to indicate a temperature closely surrounding that point. A two-point calibration allows the circuit to sense the entire range of temperature more accurately. Reference Application Note AN1333, “Use and Calibration of the Internal Temperature Indicator” (DS01333) for more details regarding the calibration process.

15.1 Circuit Operation

Figure 15-1 shows a simplified block diagram of the temperature circuit. The proportional voltage output is achieved by measuring the forward voltage drop across multiple silicon junctions. Equation 15-1 describes the output characteristics of the temperature indicator.

\[
\text{EQUATION 15-1: } V_{\text{OUT}} \text{ RANGES}
\]

<table>
<thead>
<tr>
<th>Range</th>
<th>Equation</th>
</tr>
</thead>
<tbody>
<tr>
<td>High Range</td>
<td>( V_{\text{OUT}} = V_{DD} - 4V_T )</td>
</tr>
<tr>
<td>Low Range</td>
<td>( V_{\text{OUT}} = V_{DD} - 2V_T )</td>
</tr>
</tbody>
</table>

The temperature sense circuit is integrated with the Fixed Voltage Reference (FVR) module. See Section 14.0 “Fixed Voltage Reference (FVR)” for more information.

The circuit is enabled by setting the TSEN bit of the FVRCON register. When disabled, the circuit draws no current.

The circuit operates in either high or low range. The high range, selected by setting the TSRNG bit of the FVRCON register, provides a wider output voltage. This provides more resolution over the temperature range, but may be less consistent from part to part. This range requires a higher bias voltage to operate and thus, a higher VDD is needed.

The low range is selected by clearing the TSRNG bit of the FVRCON register. The low range generates a lower voltage drop and thus, a lower bias voltage is needed to operate the circuit. The low range is provided for low voltage operation.

15.2 Minimum Operating VDD

When the temperature circuit is operated in low range, the device may be operated at any operating voltage that is within specifications.

When the temperature circuit is operated in high range, the device operating voltage, VDD, must be high enough to ensure that the temperature circuit is correctly biased.

Table 15-1 shows the recommended minimum VDD vs. range setting.

<table>
<thead>
<tr>
<th>Min. VDD, TSRNG = 1</th>
<th>Min. VDD, TSRNG = 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>3.6V</td>
<td>1.8V</td>
</tr>
</tbody>
</table>

15.3 Temperature Output

The output of the circuit is measured using the internal Analog-to-Digital Converter. A channel is reserved for the temperature circuit output. Refer to Section 21.0 “Analog-to-Digital Converter (ADC) Module” for detailed information.
15.4 ADC Acquisition Time

To ensure accurate temperature measurements, the user must wait at least 200 $\mu$s after the ADC input multiplexer is connected to the temperature indicator output before the conversion is performed. In addition, the user must wait 200 $\mu$s between sequential conversions of the temperature indicator output.

**TABLE 15-2: SUMMARY OF REGISTERS ASSOCIATED WITH THE TEMPERATURE INDICATOR**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>FVRCON</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>166</td>
<td></td>
<td></td>
<td>166</td>
</tr>
</tbody>
</table>

*Legend:* Shaded cells are unused by the temperature indicator module.
16.0 COMPARATOR MODULE

Comparators are used to interface analog circuits to a digital circuit by comparing two analog voltages and providing a digital indication of their relative magnitudes. Comparators are very useful mixed signal building blocks because they provide analog functionality independent of program execution. The analog comparator module includes the following features:

- Independent comparator control
- Programmable input selection
- Comparator output is available internally/externally
- Programmable output polarity
- Interrupt-on-Change
- Wake-up from Sleep
- Programmable Speed/Power optimization
- PWM shutdown
- Programmable and Fixed Voltage Reference

16.1 Comparator Overview

A single comparator is shown in Figure 16-1 along with the relationship between the analog input levels and the digital output. When the analog voltage at \( V_{IN+} \) is less than the analog voltage at \( V_{IN-} \), the output of the comparator is a digital low level. When the analog voltage at \( V_{IN+} \) is greater than the analog voltage at \( V_{IN-} \), the output of the comparator is a digital high level.

The comparators available for this device are located in Table 16-1.

<table>
<thead>
<tr>
<th>TABLE 16-1: AVAILABLE COMPARATORS</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Device</strong></td>
</tr>
<tr>
<td>PIC16(L)F1717/8/9</td>
</tr>
</tbody>
</table>

**FIGURE 16-1: SINGLE COMPARATOR**

Note: The black areas of the output of the comparator represents the uncertainty due to input offsets and response time.
FIGURE 16-2: COMPARATOR MODULE SIMPLIFIED BLOCK DIAGRAM

Note 1: When CxON = 0, the comparator will produce a '0' at the output.
2: When CxON = 0, all multiplexer inputs are disconnected.
16.2 Comparator Control

Each comparator has two control registers: CMxCON0 and CMxCON1.

The CMxCON0 register (see Register 16-1) contains Control and Status bits for the following:
- Enable
- Output
- Output Polarity
- Zero Latency Filter
- Speed/Power Selection
- Hysteresis Enable
- Output Synchronization

The CMxCON1 register (see Register 16-2) contains Control bits for the following:
- Interrupt Enable
- Interrupt Edge Polarity
- Positive Input Channel Selection
- Negative Input Channel Selection

16.2.1 COMPARATOR ENABLE

Setting the CxON bit of the CMxCON0 register enables the comparator for operation. Clearing the CxON bit disables the comparator resulting in minimum current consumption.

16.2.2 COMPARATOR OUTPUT SELECTION

The output of the comparator can be monitored by reading either the CxOUT bit of the CMxCON0 register or the MCxOUT bit of the CMOUT register. In order to make the output available for an external connection, the following conditions must be true:
- Desired pin PPS control
- Corresponding TRIS bit must be cleared
- CxON bit of the CMxCON0 register must be set

Note 1: The internal output of the comparator is latched with each instruction cycle. Unless otherwise specified, external outputs are not latched.

16.2.3 COMPARATOR OUTPUT POLARITY

Inverting the output of the comparator is functionally equivalent to swapping the comparator inputs. The polarity of the comparator output can be inverted by setting the CxPOL bit of the CMxCON0 register. Clearing the CxPOL bit results in a non-inverted output. Table 16-2 shows the output state versus input conditions, including polarity control.

<table>
<thead>
<tr>
<th>Input Condition</th>
<th>CxPOL</th>
<th>CxOUT</th>
</tr>
</thead>
<tbody>
<tr>
<td>CxVN &gt; CxVP</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>CxVN &lt; CxVP</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>CxVN &gt; CxVP</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>CxVN &lt; CxVP</td>
<td>1</td>
<td>0</td>
</tr>
</tbody>
</table>

16.2.4 COMPARATOR SPEED/POWER SELECTION

The trade-off between speed or power can be optimized during program execution with the CxSP control bit. The default state for this bit is ‘1’, which selects the Normal-Speed mode. Device power consumption can be optimized at the cost of slower comparator propagation delay by clearing the CxSP bit to ‘0’.
16.3 Comparator Hysteresis

A selectable amount of separation voltage can be added to the input pins of each comparator to provide a hysteresis function to the overall operation. Hysteresis is enabled by setting the CxHYS bit of the CMxCON0 register.

See Comparator Specifications in Table 34-18: Comparator Specifications for more information.

16.4 Timer1 Gate Operation

The output resulting from a comparator operation can be used as a source for gate control of Timer1. See Section 27.6 “Timer1 Gate” for more information. This feature is useful for timing the duration or interval of an analog event.

It is recommended that the comparator output be synchronized to Timer1. This ensures that Timer1 does not increment while a change in the comparator is occurring.

16.4.1 COMPARATOR OUTPUT SYNCHRONIZATION

The output from a comparator can be synchronized with Timer1 by setting the CxSYNC bit of the CMxCON0 register.

Once enabled, the comparator output is latched on the falling edge of the Timer1 source clock. If a prescaler is used with Timer1, the comparator output is latched after the prescaling function. To prevent a race condition, the comparator output is latched on the falling edge of the Timer1 clock source and Timer1 increments on the rising edge of its clock source. See the Comparator Block Diagram (Figure 16-2) and the Timer1 Block Diagram (Figure 27-1: Timer1 Block Diagram) for more information.

16.5 Comparator Interrupt

An interrupt can be generated upon a change in the output value of the comparator for each comparator, a rising edge detector and a falling edge detector are present.

When either edge detector is triggered and its associated enable bit is set (CxINTP and/or CxINTN bits of the CMxCON1 register), the Corresponding Interrupt Flag bit (CxIF bit of the PIR2 register) will be set.

To enable the interrupt, you must set the following bits:

- CxON, CxPOL and CxSP bits of the CMxCON0 register
- CxIE bit of the PIE2 register
- CxINTP bit of the CMxCON1 register (for a rising edge detection)
- CxINTN bit of the CMxCON1 register (for a falling edge detection)
- PEIE and GIE bits of the INTCON register

The associated interrupt flag bit, CxIF bit of the PIR2 register, must be cleared in software. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

Note: Although a comparator is disabled, an interrupt can be generated by changing the output polarity with the CxPOL bit of the CMxCON0 register, or by switching the comparator on or off with the CxON bit of the CMxCON0 register.
16.6 Comparator Positive Input Selection
Configuring the CxPCH<2:0> bits of the CMxCON1 register directs an internal voltage reference or an analog pin to the non-inverting input of the comparator:
- CxIN+ analog pin
- DAC output
- FVR (Fixed Voltage Reference)
- VSS (Ground)

See Section 14.0 “Fixed Voltage Reference (FVR)” for more information on the Fixed Voltage Reference module.

See Section 23.0 “8-Bit Digital-to-Analog Converter (DAC1) Module” for more information on the DAC input signal.

Any time the comparator is disabled (CxON = 0), all comparator inputs are disabled.

16.7 Comparator Negative Input Selection
The CxNCH<2:0> bits of the CMxCON1 register direct an analog input pin and internal reference voltage or analog ground to the inverting input of the comparator:
- CxIN- pin
- FVR (Fixed Voltage Reference)
- Analog Ground

Some inverting input selections share a pin with the operational amplifier output function. Enabling both functions at the same time will direct the operational amplifier output to the comparator inverting input.

Note: To use CxINy+ and CxINy- pins as analog input, the appropriate bits must be set in the ANSEL register and the corresponding TRIS bits must also be set to disable the output drivers.

16.8 Comparator Response Time
The comparator output is indeterminate for a period of time after the change of an input source or the selection of a new reference voltage. This period is referred to as the response time. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response time to a comparator input change. See the Comparator and Voltage Reference Specifications in Table 34-18: Comparator Specifications for more details.

16.9 Zero Latency Filter
In high-speed operation, and under proper circuit conditions, it is possible for the comparator output to oscillate. This oscillation can have adverse effects on the hardware and software relying on this signal. Therefore, a digital filter has been added to the comparator output to suppress the comparator output oscillation. Once the comparator output changes, the output is prevented from reversing the change for a nominal time of 20 ns. This allows the comparator output to stabilize without affecting other dependent devices. Refer to Figure 16-3.

FIGURE 16-3: COMPARATOR ZERO LATENCY FILTER OPERATION

CxOUT From Comparator

CxOUT From ZLF

Output waiting for TZLF to expire before an output change is allowed.

TZLF has expired so output change of ZLF is immediate based on comparator output change.
16.10 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 16-4. Since the analog input pins share their connection with a digital input, they have reverse biased ESD protection diodes to VDD and VSS. The analog input, therefore, must be between VSS and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur.

A maximum source impedance of 10 kΩ is recommended for the analog sources. Also, any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current to minimize inaccuracies introduced.

Note 1: When reading a PORT register, all pins configured as analog inputs will read as a ‘0’. Pins configured as digital inputs will convert as an analog input, according to the input specification.

2: Analog levels on any pin defined as a digital input, may cause the input buffer to consume more current than is specified.

FIGURE 16-4: ANALOG INPUT MODEL

Legend: 
- CPIN = Input Capacitance
- ILEAKAGE = Leakage Current at the pin due to various junctions
- RIC = Interconnect Resistance
- RS = Source Impedance
- VA = Analog Voltage
- VT = Threshold Voltage

Note 1: See I/O Ports in Table 34-4: I/O Ports.
### 16.11 Register Definitions: Comparator Control

**REGISTER 16-1: CMxCON0: COMPARATOR Cx CONTROL REGISTER 0**

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CxON</td>
<td>CxOUT</td>
<td>—</td>
<td>CxPOL</td>
<td>CxZLF</td>
<td>CxSP</td>
<td>CxHYS</td>
<td>CxSYNC</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7**
- **CxON:** Comparator Enable bit
  - 1 = Comparator is enabled
  - 0 = Comparator is disabled and consumes no active power

**bit 6**
- **CxOUT:** Comparator Output bit
  - If **CxPOL** = 1 (inverted polarity):
    - 1 = CxVP < CxVN
    - 0 = CxVP > CxVN
  - If **CxPOL** = 0 (non-inverted polarity):
    - 1 = CxVP > CxVN
    - 0 = CxVP < CxVN

**bit 5**
- **Unimplemented:** Read as ‘0’

**bit 4**
- **CxPOL:** Comparator Output Polarity Select bit
  - 1 = Comparator output is inverted
  - 0 = Comparator output is not inverted

**bit 3**
- **CxZLF:** Comparator Zero Latency Filter Enable bit
  - 1 = Comparator output is filtered
  - 0 = Comparator output is unfiltered

**bit 2**
- **CxSP:** Comparator Speed/Power Select bit
  - 1 = Comparator operates in normal power, higher speed mode
  - 0 = Comparator operates in low-power, low-speed mode

**bit 1**
- **CxHYS:** Comparator Hysteresis Enable bit
  - 1 = Comparator hysteresis enabled
  - 0 = Comparator hysteresis disabled

**bit 0**
- **CxSYNC:** Comparator Output Synchronous Mode bit
  - 1 = Comparator output to Timer1 and I/O pin is synchronous to changes on Timer1 clock source. Output updated on the falling edge of Timer1 clock source.
  - 0 = Comparator output to Timer1 and I/O pin is asynchronous.
REGISTER 16-2: CMxCON1: COMPARATOR Cx CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CxINTP</td>
<td>CxINTN</td>
<td>CxPCH&lt;2:0&gt;</td>
<td>CxNCH&lt;2:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

bit 7
CxINTP: Comparator Interrupt on Positive Going Edge Enable bits
1 = The CxIF interrupt flag will be set upon a positive going edge of the CxOUT bit
0 = No interrupt flag will be set on a positive going edge of the CxOUT bit

bit 6
CxINTN: Comparator Interrupt on Negative Going Edge Enable bits
1 = The CxIF interrupt flag will be set upon a negative going edge of the CxOUT bit
0 = No interrupt flag will be set on a negative going edge of the CxOUT bit

bit 5-3
CxPCH<2:0>: Comparator Positive Input Channel Select bits
111 = CxVP connects to AGND
110 = CxVP connects to FVR Buffer 2
101 = CxVP connects to DAC1_output
100 = CxVP connects to DAC2_output
011 = CxVP unconnected, input floating
010 = CxVP unconnected, input floating
001 = CxVP connects to CxIN1+ pin
000 = CxVP connects to CxIN0+ pin

bit 2-0
CxNCH<2:0>: Comparator Negative Input Channel Select bits
111 = CxVN connects to AGND
110 = CxVN connects to FVR Buffer 2
101 = CxVN unconnected, input floating
100 = CxVN unconnected, input floating
011 = CxVN connects to CxIN3- pin
010 = CxVN connects to CxIN2- pin
001 = CxVN connects to CxIN1- pin
000 = CxVN connects to CxIN0- pin
REGISTER 16-3: CMOUT: COMPARATOR OUTPUT REGISTER

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R-0/0</th>
<th>R-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MC2OUT</td>
<td>MC1OUT</td>
</tr>
<tr>
<td>bit 6</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>bit 5</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>bit 4</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>bit 3</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>bit 2</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>bit 1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>bit 0</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
W = Writable bit
U = Unimplemented bit, read as ‘0’
u = Bit is unchanged
x = Bit is unknown
-\textit{n/n} = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set
‘0’ = Bit is cleared

bit 7-2 Unimplemented: Read as ‘0’
bit 1 MC2OUT: Mirror Copy of C2OUT bit
bit 0 MC1OUT: Mirror Copy of C1OUT bit

TABLE 16-3: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARATOR MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>126</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>CM1CON0</td>
<td>C1ON</td>
<td>C1OUT</td>
<td>—</td>
<td>C1POL</td>
<td>C1ZLF</td>
<td>C1SP</td>
<td>C1HYS</td>
<td>C1SYNC</td>
<td>175</td>
</tr>
<tr>
<td>CM2CON0</td>
<td>C2ON</td>
<td>C2OUT</td>
<td>—</td>
<td>C2POL</td>
<td>C2ZLF</td>
<td>C2SP</td>
<td>C2HYS</td>
<td>C2SYNC</td>
<td>175</td>
</tr>
<tr>
<td>CM1CON1</td>
<td>C1NTP</td>
<td>C1INTN</td>
<td>C1PCH&lt;2:0&gt;</td>
<td>—</td>
<td>C1NCH&lt;2:0&gt;</td>
<td>—</td>
<td>C1NCH&lt;2:0&gt;</td>
<td>—</td>
<td>176</td>
</tr>
<tr>
<td>CM2CON1</td>
<td>C2NTP</td>
<td>C2INTN</td>
<td>C2PCH&lt;2:0&gt;</td>
<td>—</td>
<td>C2NCH&lt;2:0&gt;</td>
<td>—</td>
<td>C2NCH&lt;2:0&gt;</td>
<td>—</td>
<td>176</td>
</tr>
<tr>
<td>CMOUT</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MC2OUT 177</td>
</tr>
<tr>
<td>FVRCON</td>
<td>FVREN</td>
<td>FVRRDY</td>
<td>TSEN</td>
<td>TSRNG</td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>ADFVR&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>166</td>
</tr>
<tr>
<td>DAC1CON0</td>
<td>DAC1EN</td>
<td>—</td>
<td>DAC1OE1</td>
<td>DAC1OE2</td>
<td>DAC1PSS&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>DAC1NSS 260</td>
<td></td>
</tr>
<tr>
<td>DAC1CON1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>260</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IO CIF</td>
<td>91</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
<td>93</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td>96</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TRISD(^{(1)})</td>
<td>TRISD7</td>
<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
<td>141</td>
</tr>
<tr>
<td>TRISE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TRISE3 146</td>
</tr>
<tr>
<td>RxYPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RxYPPS&lt;4:0&gt; 154</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are unused by the comparator module.

Note 1: PIC16(L)F1717/9 only.
17.0 **PULSE WIDTH MODULATION (PWM)**

The PWM module generates a Pulse-Width Modulated signal determined by the duty cycle, period, and resolution that are configured by the following registers:

- PR2
- T2CON
- PWMxDCH
- PWMxDCL
- PWMxCON

Figure 17-1 shows a simplified block diagram of PWM operation.

Figure 17-2 shows a typical waveform of the PWM signal.

**FIGURE 17-1: SIMPLIFIED PWM BLOCK DIAGRAM**

For a step-by-step procedure on how to set up this module for PWM operation, refer to Section 17.1.9 “Setup for PWM Operation Using PWMx Pins”.

---

**FIGURE 17-2: PWM OUTPUT**

---

Note 1: 8-bit timer is concatenated with the two Least Significant bits of 1/Fosc adjusted by the Timer2 prescaler to create a 10-bit time base.
17.1 PWMx Pin Configuration

All PWM outputs are multiplexed with the PORT data latch. The user must configure the pins as outputs by clearing the associated TRIS bits.

17.1.1 FUNDAMENTAL OPERATION

The PWM module produces a 10-bit resolution output. Timer2 and PR2 set the period of the PWM. The PWMxDCH and PWMxDCL registers configure the duty cycle. The period is common to all PWM modules, whereas the duty cycle is independently controlled.

**Note:** The Timer2 postscaler is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

All PWM outputs associated with Timer2 are set when TMR2 is cleared. Each PWMx is cleared when TMR2 is equal to the value specified in the corresponding PWMxDCH (8 MSb) and PWMxDCL<7:6> (2 LSb) registers. When the value is greater than or equal to PR2, the PWM output is never cleared (100% duty cycle).

**Note:** The PWMxDCH and PWMxDCL registers are double buffered. The buffers are updated when Timer2 matches PR2. Care should be taken to update both registers before the timer match occurs.

17.1.2 PWM OUTPUT POLARITY

The output polarity is inverted by setting the PWMxPOL bit of the PWMxCON register.

17.1.3 PWM PERIOD

The PWM period is specified by the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 17-1.

**EQUATION 17-1: PWM PERIOD**

\[ PWM \text{ Period} = ((PR2) + 1) \times 4 \times T_{\text{osc}} \times (\text{TMR2 Prescale Value}) \]

**Note:** \( T_{\text{osc}} = 1/F_{\text{osc}} \)

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:
- TMR2 is cleared
- The PWM output is active. (Exception: When the PWM duty cycle = 0%, the PWM output will remain inactive.)
- The PWMxDCH and PWMxDCL register values are latched into the buffers.

**Note:** The Timer2 postscaler has no effect on the PWM operation.

17.1.4 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to the PWMxDCH and PWMxDCL register pair. The PWMxDCH register contains the eight MSbs and the PWMxDCL<7:6>, the two LSbs. The PWMxDCH and PWMxDCL registers can be written to at any time.

Equation 17-2 is used to calculate the PWM pulse width.

Equation 17-3 is used to calculate the PWM duty cycle ratio.

**EQUATION 17-2: PULSE WIDTH**

\[ Pulse \text{ Width} = (\text{PWMxDCH}:\text{PWMxDCL}<7:6>) \times T_{\text{osc}} \times (\text{TMR2 Prescale Value}) \]

**Note:** \( T_{\text{osc}} = 1/F_{\text{osc}} \)

**EQUATION 17-3: DUTY CYCLE RATIO**

\[ Duty \text{ Cycle Ratio} = \frac{(\text{PWMxDCH}:\text{PWMxDCL}<7:6>)}{4(\text{PR2} + 1)} \]

The 8-bit timer TMR2 register is concatenated with the two Least Significant bits of 1/Fosc, adjusted by the Timer2 prescaler to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.
17.1.5 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is ten bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 17-4.

EQUATION 17-4: PWM RESOLUTION

\[
\text{Resolution} = \frac{\log(4(PR2 + 1))}{\log(2)} \text{ bits}
\]

Note: If the pulse width value is greater than the period the assigned PWM pin(s) will remain unchanged.

| TABLE 17-1: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz) |
|----------------|----------------|----------------|----------------|----------------|----------------|----------------|
| PWM Frequency | 0.31 kHz | 4.88 kHz | 19.53 kHz | 78.12 kHz | 156.3 kHz | 208.3 kHz |
| Timer Prescale | 64 | 4 | 1 | 1 | 1 | 1 |
| PR2 Value | 0xFF | 0xFF | 0xFF | 0x3F | 0x1F | 0x17 |
| Maximum Resolution (bits) | 10 | 10 | 10 | 8 | 7 | 6.6 |

| TABLE 17-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz) |
|----------------|----------------|----------------|----------------|----------------|----------------|----------------|
| PWM Frequency | 0.31 kHz | 4.90 kHz | 19.61 kHz | 76.92 kHz | 153.85 kHz | 200.0 kHz |
| Timer Prescale | 64 | 4 | 1 | 1 | 1 | 1 |
| PR2 Value | 0x65 | 0x65 | 0x65 | 0x19 | 0x0C | 0x09 |
| Maximum Resolution (bits) | 8 | 8 | 8 | 6 | 5 | 5 |

17.1.6 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the PWMx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

17.1.7 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency (Fosc). Any changes in the system clock frequency will result in changes to the PWM frequency. Refer to Section 6.0 “Oscillator Module (with Fail-Safe Clock Monitor)” for additional details.

17.1.8 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the PWM registers to their Reset states.
17.1.9 SETUP FOR PWM OPERATION USING PWMx PINS

The following steps should be taken when configuring the module for PWM operation using the PWMx pins:

1. Disable the PWMx pin output driver(s) by setting the associated TRIS bit(s).
2. Clear the PWMxCON register.
3. Load the PR2 register with the PWM period value.
4. Load the PWMxDCH register and bits <7:6> of the PWMxDCL register with the PWM duty cycle value.
5. Configure and start Timer2:
   - Clear the TMR2IF interrupt flag bit of the PIR1 register. See Note below.
   - Configure the T2CKPS bits of the T2CON register with the Timer2 prescale value.
   - Enable Timer2 by setting the TMR2ON bit of the T2CON register.
6. Enable PWM output pin and wait until Timer2 overflows, TMR2IF bit of the PIR1 register is set. See Note below.
7. Enable the PWMx pin output driver(s) by clearing the associated TRIS bit(s) and setting the desired pin PPS control bits.
8. Configure the PWM module by loading the PWMxCON register with the appropriate values.

Note 1: In order to send a complete duty cycle and period on the first PWM output, the above steps must be followed in the order given. If it is not critical to start with a complete PWM signal, then move Step 8 to replace Step 4.

2: For operation with other peripherals only, disable PWMx pin outputs.

17.1.10 SETUP FOR PWM OPERATION TO OTHER DEVICE PERIPHERALS

The following steps should be taken when configuring the module for PWM operation to be used by other device peripherals:

1. Disable the PWMx pin output driver(s) by setting the associated TRIS bit(s).
2. Clear the PWMxCON register.
3. Load the PR2 register with the PWM period value.
4. Load the PWMxDCH register and bits <7:6> of the PWMxDCL register with the PWM duty cycle value.
5. Configure and start Timer2:
   - Clear the TMR2IF interrupt flag bit of the PIR1 register. See Note below.
   - Configure the T2CKPS bits of the T2CON register with the Timer2 prescale value.
   - Enable Timer2 by setting the TMR2ON bit of the T2CON register.
6. Enable PWM output pin:
   - Wait until Timer2 overflows, TMR2IF bit of the PIR1 register is set. See Note below.
7. Configure the PWM module by loading the PWMxCON register with the appropriate values.

Note: In order to send a complete duty cycle and period on the first PWM output, the above steps must be included in the setup sequence. If it is not critical to start with a complete PWM signal on the first output, then step 6 may be ignored.
### 17.2 Register Definitions: PWM Control

#### REGISTER 17-1: PWMxCON: PWM CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value at POR</th>
<th>Value at all other Resets</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>PWMxEN: PWM Module Enable bit</td>
<td>1 = PWM module is enabled</td>
<td>0 = PWM module is disabled</td>
</tr>
<tr>
<td>6</td>
<td>Unimplemented: Read as ‘0’</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>5-0</td>
<td>PWMxOUT: PWM module output level when bit is read.</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>4</td>
<td>PWMxPOL: PWMx Output Polarity Select bit</td>
<td>1 = PWM output is active low.</td>
<td>0 = PWM output is active high.</td>
</tr>
<tr>
<td>3-0</td>
<td>Unimplemented: Read as ‘0’</td>
<td>-</td>
<td>-</td>
</tr>
</tbody>
</table>

#### REGISTER 17-2: PWMxDCH: PWM DUTY CYCLE HIGH BITS

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value at POR</th>
<th>Value at all other Resets</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>PWMxDCH&lt;7:0&gt;: PWM Duty Cycle Most Significant bits</td>
<td>-</td>
<td>-</td>
</tr>
</tbody>
</table>

#### REGISTER 17-3: PWMxDCL: PWM DUTY CYCLE LOW BITS

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value at POR</th>
<th>Value at all other Resets</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-6</td>
<td>PWMxDCL&lt;7:6&gt;: PWM Duty Cycle Least Significant bits</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>5-0</td>
<td>Unimplemented: Read as ‘0’</td>
<td>-</td>
<td>-</td>
</tr>
</tbody>
</table>

---

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

---

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

---

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets
### TABLE 17-3: SUMMARY OF REGISTERS ASSOCIATED WITH PWM

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CCPTMRS</td>
<td></td>
<td></td>
<td>P3TSEL&lt;1:0&gt;</td>
<td></td>
<td>C2TSEL&lt;1:0&gt;</td>
<td></td>
<td>C1TSEL&lt;1:0&gt;</td>
<td></td>
<td>289</td>
</tr>
<tr>
<td>PR2</td>
<td></td>
<td></td>
<td>Timer2 module Period Register</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>285</td>
</tr>
<tr>
<td>PWM3CON</td>
<td>PWM3EN</td>
<td>—</td>
<td>PWM3OUT</td>
<td>PWM3POL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>182</td>
</tr>
<tr>
<td>PWM3DCH</td>
<td></td>
<td></td>
<td>PWM3DCH&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>182</td>
</tr>
<tr>
<td>PWM3DCL</td>
<td>PWM3DCL&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>182</td>
</tr>
<tr>
<td>PWM4CON</td>
<td>PWM4EN</td>
<td>—</td>
<td>PWM4OUT</td>
<td>PWM4POL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>182</td>
</tr>
<tr>
<td>PWM4DCH</td>
<td></td>
<td></td>
<td>PWM4DCH&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>182</td>
</tr>
<tr>
<td>PWM4DCL</td>
<td>PWM4DCL&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>182</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TRISD&lt;1&gt;</td>
<td>TRISD7</td>
<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
<td>141</td>
</tr>
<tr>
<td>TRISE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TRISE3</td>
<td>TRISE2&lt;1&gt;</td>
<td>TRISE1&lt;1&gt;</td>
<td>TRISE0&lt;1&gt;</td>
<td>146</td>
</tr>
<tr>
<td>RxyPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RxyPPS&lt;4:0&gt;</td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>T2CON</td>
<td>—</td>
<td>—</td>
<td>T2OUTPS&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
<td>TMR2ON</td>
<td>T2CKPS&lt;1:0&gt;</td>
<td></td>
<td>287</td>
</tr>
<tr>
<td>TMR2</td>
<td></td>
<td></td>
<td>Timer2 module Register</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>285</td>
</tr>
</tbody>
</table>

**Legend:** — = Unimplemented locations, read as '0'; u = unchanged, x = unknown. Shaded cells are not used by the PWM.

**Note 1:** PIC16(L)F1717/9 only.
18.0 COMPLEMENTARY OUTPUT GENERATOR (COG) MODULE

The primary purpose of the Complementary Output Generator (COG) is to convert a single output PWM signal into a two output complementary PWM signal. The COG can also convert two separate input events into a single or complementary PWM output.

The COG PWM frequency and duty cycle are determined by a rising event input and a falling event input. The rising event and falling event may be the same source. Sources may be synchronous or asynchronous to the COG_clock.

The rate at which the rising event occurs determines the PWM frequency. The time from the rising event input to the falling event input determines the duty cycle.

A selectable clock input is used to generate the phase delay, blanking, and dead-band times. Dead-band time can also be generated with a programmable time delay, which is independent from all clock sources.

Simplified block diagrams of the various COG modes are shown in Figure 18-2 through Figure 18-6.

The COG module has the following features:

- Six modes of operation:
  - Steered PWM mode
  - Synchronous Steered PWM mode
  - Forward Full-Bridge mode
  - Reverse Full-Bridge mode
  - Half-Bridge mode
  - Push-Pull mode
- Selectable COG_clock clock source
- Independently selectable rising event sources
- Independently selectable falling event sources
- Independently selectable edge or level event sensitivity
- Independent output polarity selection
- Phase delay with independent rising and falling delay times
- Dead-band control with:
  - independent rising and falling event dead-band times
  - Synchronous and asynchronous timing
- Blanking control with independent rising and falling event blanking times
- Auto-shutdown control with:
  - Independently selectable shutdown sources
  - Auto-restart enable
  - Auto-shutdown pin override control (high, low, off, and High-Z)

18.1 Fundamental Operation

18.1.1 STEERED PWM MODES

In steered PWM mode, the PWM signal derived from the input event sources is output as a single phase PWM which can be steered to any combination of the four COG outputs. Outputs are selected by setting the GxSTRA through GxSTRD bits of the COGxSTR register (Register 18-9). When the steering bits are cleared, then the output data is the static level determined by the GxSDATA through GxSDATD bits of the COGxSTR register. Output steering takes effect on the instruction cycle following the write to the COGxSTR register.

Synchronous steered PWM mode is identical to the steered PWM mode except that changes to the output steering take effect on the first rising event after the COGxSTR register write. Static output data is not synchronized.

Steering mode configurations are shown in Figure 18-2 and Figure 18-3.

Steered PWM and synchronous steered PWM modes are selected by setting the GxMD bits of the COGxCON0 register (Register 18-1) to '000' and '001' respectively.

18.1.2 FULL-BRIDGE MODES

In both Forward and Reverse Full-Bridge modes, two of the four COG outputs are active and the other two are inactive. Of the two active outputs, one is modulated by the PWM input signal and the other is on at 100% duty cycle. When the direction is changed, the dead-band time is inserted to delay the modulated output. This gives the unmodulated driver time to shut down, thereby, preventing shoot-through current in the series connected power devices.

In Forward Full-Bridge mode, the PWM input modulates the COGxD output and drives the COGA output at 100%.

In Reverse Full-Bridge mode, the PWM input modulates the COGxB output and drives the COGxC output at 100%.

The full-bridge configuration is shown in Figure 18-4. Typical full-bridge waveforms are shown in Figure 18-12 and Figure 18-13.

Full-Bridge Forward and Full-Bridge Reverse modes are selected by setting the GxMD bits of the COGxCON0 register to '010' and '011', respectively.
FIGURE 18-1: EXAMPLE OF FULL-BRIDGE APPLICATION

[Diagram of a full-bridge application with labeled components: COGxA, COGxB, COGxC, COGxD, FET drivers, QA, QB, QC, QD, V+, V-, and Load.]
18.1.3 HALF-BRIDGE MODE

In half-bridge mode, the COG generates a two output complementary PWM waveform from rising and falling event sources. In the simplest configuration, the rising and falling event sources are the same signal, which is a PWM signal with the desired period and duty cycle. The COG converts this single PWM input into a dual complementary PWM output. The frequency and duty cycle of the dual PWM output match those of the single input PWM signal. The off-to-on transition of each output can be delayed from the on-to-off transition of the other output, thereby, creating a time immediately after the PWM transition where neither output is driven. This is referred to as dead time and is covered in Section 18.5 “Dead-Band Control”.

A typical operating waveform, with dead band, generated from a single CCP1 input is shown in Figure 18-9.

The primary output can be steered to either or both COGxA and COGxC. The complementary output can be steered to either or both COGxB and COGxD.

Half-Bridge mode is selected by setting the GxMD bits of the COGxCON0 register to ‘100’.

18.1.4 PUSH-PULL MODE

In Push-Pull mode, the COG generates a single PWM output that alternates, every PWM period, between the two pairs of the COG outputs. COGxA has the same signal as COGxC. COGxB has the same signal as COGxD. The output drive activates with the rising input event and terminates with the falling event input. Each rising event starts a new period and causes the output to switch to the COG pair not used in the previous period.

The push-pull configuration is shown in Figure 18-6. A typical push-pull waveform generated from a single CCP1 input is shown in Figure 18-11.

Push-Pull mode is selected by setting the GxMD bits of the COGxCON0 register to ‘101’.

18.1.5 EVENT DRIVEN PWM (ALL MODES)

Besides generating PWM and complementary outputs from a single PWM input, the COG can also generate PWM waveforms from a periodic rising event and a separate falling event. In this case, the falling event is usually derived from analog feedback within the external PWM driver circuit. In this configuration, high power switching transients may trigger a false falling event that needs to be blanked out. The COG can be configured to blank falling (and rising) event inputs for a period of time immediately following the rising (and falling) event drive output. This is referred to as input blanking and is covered in Section 18.6 “Blanking Control”.

It may be necessary to guard against the possibility of circuit faults. In this case, the active drive must be terminated before the Fault condition causes damage. This is referred to as auto-shutdown and is covered in Section 18.8 “Auto-shutdown Control”.

The COG can be configured to operate in phase delayed conjunction with another PWM. The active drive cycle is delayed from the rising event by a phase delay timer. Phase delay is covered in more detail in Section 18.7 “Phase Delay”.

A typical operating waveform, with phase delay and dead band, generated from a single CCP1 input is shown in Figure 18-10.
FIGURE 18-2: SIMPLIFIED COG BLOCK DIAGRAM (STEERED PWM MODE, GXMD = 0)

COG_clock

GxCS<1:0>

Rising Input Block

NCO1_out
PWM3OUT
CCP2
CCP1
LC1_out
C2OUT
C1OUT
COGINPPS

src7
src6
src5
src4
src3
src2
src1
src0

COG_x<1:0>

GxASE
Auto-shutdown source

Write GxASE High
Write GxASE Low

Set Dominates

Reset Dominates

Rising Input Block

NCO1_out
PWM3OUT
CCP2
CCP1
LC1_out
C2OUT
C1OUT
COGINPPS

src7
src6
src5
src4
src3
src2
src1
src0

GxASE

COGxA

GxPOLA
GxSDATA
GxSTRA
GxASDAC<1:0>

GxASDBD<1:0>

High-Z

GxPOLB
GxSDATB
GxSTRB

GxPOLC
GxSDATC
GxSTRC

GxPOLD
GxSDATD
GxSTRD

GxASE

COGxB

COGxC

COGxD
FIGURE 18-3: SIMPLIFIED COG BLOCK DIAGRAM (SYNCHRONOUS STEERED PWM MODE, GXMD = 1)
FIGURE 18-4: SIMPLIFIED COG BLOCK DIAGRAM (FULL-BRIDGE MODES, FORWARD: GXMD = 2, REVERSE: GXMD = 3)
FIGURE 18-5: SIMPLIFIED COG BLOCK DIAGRAM (HALF-BRIDGE MODE, GXMD = 4)
FIGURE 18-6: SIMPLIFIED COG BLOCK DIAGRAM (PUSH-PULL MODE, GXMD = 5)
FIGURE 18-7: COG (RISING/FALLING) INPUT BLOCK

Phase
Delay

GxPH(R/F)<3:0>

clock

count_en

GxBLK(F/R)<3:0>

Blanking
Cnt/Cir

D Q

LE

D Q

LE

D Q

LE

D Q

LE

D Q

LE

D Q

LE

D Q

LE

D Q

LE

D Q

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D Q

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FIGURE 18-8: COG (RISING/FALLING) DEAD-BAND BLOCK

![Diagram of COG (RISING/FALLING) DEAD-BAND BLOCK]
FIGURE 18-9: TYPICAL HALF-BRIDGE MODE COG OPERATION WITH CCP1

FIGURE 18-10: HALF-BRIDGE MODE COG OPERATION WITH CCP1 AND PHASE DELAY

FIGURE 18-11: PUSH-PULL MODE COG OPERATION WITH CCP1
FIGURE 18-12: FULL-BRIDGE FORWARD MODE COG OPERATION WITH CCP1

FIGURE 18-13: FULL-BRIDGE MODE COG OPERATION WITH CCP1 AND DIRECTION CHANGE
18.2 Clock Sources

The COG_clock is used as the reference clock to the various timers in the peripheral. Timers that use the COG_clock include:

- Rising and falling dead-band time
- Rising and falling blanking time
- Rising and falling event phase delay

Clock sources available for selection include:

- 8 MHz HFINTOSC (active during Sleep)
- Instruction clock (Fosc/4)
- System clock (Fosc)

The clock source is selected with the GxCS<1:0> bits of the COGxCON0 register (Register 18-1).

18.3 Selectable Event Sources

The COG uses any combination of independently selectable event sources to generate the complementary waveform. Sources fall into two categories:

- Rising event sources
- Falling event sources

The rising event sources are selected by setting bits in the COGxRIS register (Register 18-3). The falling event sources are selected by setting bits in the COGxFIS register (Register 18-5). All selected sources are ‘OR’d together to generate the corresponding event signal. Refer to Figure 18-7.

18.3.1 EDGE VS. LEVEL SENSING

Event input detection may be selected as level or edge sensitive. The detection mode is individually selectable for every source. Rising source detection modes are selected with the COGxRSIM register (Register 18-4). Falling source detection modes are selected with the COGxFSIM register (Register 18-6). A set bit enables edge detection for the corresponding event source. A cleared bit enables level detection.

In general, events that are driven from a periodic source should be edge detected and events that are derived from voltage thresholds at the target circuit should be level sensitive. Consider the following two examples:

1. The first example is an application in which the period is determined by a 50% duty cycle clock and the COG output duty cycle is determined by a voltage level fed back through a comparator. If the clock input is level sensitive, duty cycles less than 50% will exhibit erratic operation.

2. The second example is similar to the first except that the duty cycle is close to 100%. The feedback comparator high-to-low transition trips the COG drive off, but almost immediately the period source turns the drive back on. If the off cycle is short enough, the comparator input may not reach the low side of the hysteresis band precluding an output change. The comparator output stays low and without a high-to-low transition to trigger the edge sense, the drive of the COG output will be stuck in a constant drive-on condition. See Figure 18-14.

18.3.2 RISING EVENT

The rising event starts the PWM output active duty cycle period. The rising event is the low-to-high transition of the rising_event output. When the rising event phase delay and dead-band time values are zero, the primary output starts immediately. Otherwise, the primary output is delayed. The rising event source causes all the following actions:

- Start rising event phase delay counter (if enabled).
- Clear complementary output after phase delay.
- Start falling event input blanking (if enabled).
- Start dead-band delay (if enabled).
- Set primary output after dead-band delay expires.

18.3.3 FALLING EVENT

The falling event terminates the PWM output active duty cycle period. The falling event is the high-to-low transition of the falling_event output. When the falling event phase delay and dead-band time values are zero, the complementary output starts immediately. Otherwise, the complementary output is delayed. The falling event source causes all the following actions:

- Start falling event phase delay counter (if enabled).
- Clear primary output.
- Start rising event input blanking (if enabled).
- Start falling event dead-band delay (if enabled).
- Set complementary output after dead-band delay expires.
18.4 Output Control

Upon disabling, or immediately after enabling the COG module, the primary COG outputs are inactive and complementary COG outputs are active.

18.4.1 OUTPUT ENABLES

There are no output enable controls in the COG module. Instead, each device pin has an individual output selection control called the PPS register. All four COG outputs are available for selection in the PPS register of every pin.

When a COG output is enabled by PPS selection, the output on the pin has several possibilities, which depend on the steering control, GxEN bit, and shutdown state as shown in Table 18-1.

<table>
<thead>
<tr>
<th>GxEN COGxSTR bit</th>
<th>Shutdown</th>
<th>Output</th>
</tr>
</thead>
<tbody>
<tr>
<td>x 0</td>
<td>Inactive</td>
<td>Static steering data</td>
</tr>
<tr>
<td>x 1</td>
<td>Active</td>
<td>Shutdown override</td>
</tr>
<tr>
<td>0 1</td>
<td>Inactive</td>
<td>Inactive state</td>
</tr>
<tr>
<td>1 1</td>
<td>Inactive</td>
<td>Active PWM signal</td>
</tr>
</tbody>
</table>

18.4.2 POLARITY CONTROL

The polarity of each COG output can be selected independently. When the output polarity bit is set, the corresponding output is active low. Clearing the output polarity bit configures the output as active high. However, polarity affects the outputs in only one of the four shutdown override modes. See Section 18.8, Auto-shutdown Control for more details.

Output polarity is selected with the GxPOLA through GxPOLD bits of the COGxCON1 register (Register 18-2).

18.5 Dead-Band Control

The dead-band control provides for non-overlapping PWM output signals to prevent shoot-through current in the external power switches. Dead time affects the output only in the Half-Bridge mode and when changing direction in the Full-Bridge mode.

The COG contains two dead-band timers. One dead-band timer is used for rising event dead-band control. The other is used for falling event dead-band control. Timer modes are selectable as either:

• Asynchronous delay chain
• Synchronous counter

The dead-band timer mode is selected for the rising_event and falling_event dead-band times with the respective GxRDBS and GxFDBS bits of the COGxCON1 register (Register 18-2).

In Half-Bridge mode, the rising_event dead-band time delays all selected primary outputs from going active for the selected dead time after the rising event. COGxA and COGxC are the primary outputs in Half-Bridge mode.

In Half-Bridge mode, the falling_event dead-band time delays all selected complementary outputs from going active for the selected dead time after the falling event. COGxB and COGxD are the complementary outputs in Half-Bridge mode.

In Full-Bridge mode, the dead-time delay occurs only during direction changes. The modulated output is delayed for the falling_event dead time after a direction change from forward to reverse. The modulated output is delayed for the rising_event dead time after a direction change from reverse to forward.

18.5.1 ASYNCHRONOUS DELAY CHAIN DEAD-BAND DELAY

Asynchronous dead-band delay is determined by the time it takes the input to propagate through a series of delay elements. Each delay element is a nominal five nanoseconds.

Set the COGxDBR register (Register 18-10) value to the desired number of delay elements in the rising_event dead-band time. Set the COGxDBF register (Register 18-11) value to the desired number of delay elements in the falling_event dead-band time. When the value is zero, dead-band delay is disabled.

18.5.2 SYNCHRONOUS COUNTER DEAD-BAND DELAY

Synchronous counter dead band is timed by counting COG_clock periods from zero up to the value in the dead-band count register. Use Equation 18-1 to calculate dead-band times.

Set the COGxDBR count register value to obtain the desired rising_event dead-band time. Set the COGxDBF count register value to obtain the desired falling_event dead-band time. When the value is zero, dead-band delay is disabled.

18.5.3 SYNCHRONOUS COUNTER DEAD-BAND TIME UNCERTAINTY

When the rising and falling events that trigger the dead-band counters come from asynchronous inputs, it creates uncertainty in the synchronous counter dead-band time. The maximum uncertainty is equal to one COG_clock period. Refer to Example 18-1 for more detail.

When event input sources are asynchronous with no phase delay, use the asynchronous delay chain dead-band mode to avoid the dead-band time uncertainty.
18.5.4 RISING EVENT DEAD BAND
Rising event dead band delays the turn-on of the primary outputs from when complementary outputs are turned off. The rising event dead-band time starts when the rising_event output goes true.
See Section 18.5.1, Asynchronous Delay Chain Dead-Band Delay and Section 18.5.2, Synchronous Counter Dead-Band Delay for more information on setting the rising edge dead-band time.

18.5.5 FALLING EVENT DEAD BAND
Falling event dead band delays the turn-on of complementary outputs from when the primary outputs are turned off. The falling event dead-band time starts when the falling_event output goes true.
See Section 18.5.1, Asynchronous Delay Chain Dead-Band Delay and Section 18.5.2, Synchronous Counter Dead-Band Delay for more information on setting the rising edge dead-band time.

18.5.6 DEAD-BAND OVERLAP
There are two cases of dead-band overlap:
• Rising-to-falling
• Falling-to-rising

18.5.6.1 Rising-to-Falling Overlap
In this case, the falling event occurs while the rising event dead-band counter is still counting. When this happens, the primary drives are suppressed and the dead band extends by the falling event dead-band time. At the termination of the extended dead-band time, the complementary drive goes true.

18.5.6.2 Falling-to-Rising Overlap
In this case, the rising event occurs while the falling event dead-band counter is still counting. When this happens, the complementary drive is suppressed and the dead band extends by the rising event dead-band time. At the termination of the extended dead-band time, the primary drive goes true.

18.6 Blanking Control
Input blanking is a function, whereby, the event inputs can be masked or blanked for a short period of time. This is to prevent electrical transients caused by the turn-on/off of power components from generating a false input event.

The COG contains two blanking counters: one triggered by the rising event and the other triggered by the falling event. The counters are cross coupled with the events they are blanking. The falling event blanking counter is used to blank rising input events and the rising event blanking counter is used to blank falling input events. Once started, blanking extends for the time specified by the corresponding blanking counter.
Blanking is timed by counting COG_clock periods from zero up to the value in the blanking count register. Use Equation 18-1 to calculate blanking times.

18.6.1 FALLING EVENT BLANKING OF RISING EVENT INPUTS
The falling event blanking counter inhibits rising event inputs from triggering a rising event. The falling event blanking time starts when the rising event output drive goes false.
The falling event blanking time is set by the value contained in the COGxBLKF register (Register 18-13). Blanking times are calculated using the formula shown in Equation 18-1.
When the COGxBLKF value is zero, falling event blanking is disabled and the blanking counter output is true, thereby, allowing the event signal to pass straight through to the event trigger circuit.

18.6.2 RISING EVENT BLANKING OF FALLING EVENT INPUTS
The rising event blanking counter inhibits falling event inputs from triggering a falling event. The rising event blanking time starts when the falling event output drive goes false.
The rising event blanking time is set by the value contained in the COGxBLKR register (Register 18-12). When the COGxBLKR value is zero, rising event blanking is disabled and the blanking counter output is true, thereby, allowing the event signal to pass straight through to the event trigger circuit.

18.6.3 BLANKING TIME UNCERTAINTY
When the rising and falling sources that trigger the blanking counters are asynchronous to the COG_clock, it creates uncertainty in the blanking time. The maximum uncertainty is equal to one COG_clock period. Refer to Equation 18-1 and Example 18-1 for more detail.

18.7 Phase Delay
It is possible to delay the assertion of either or both the rising event and falling events. This is accomplished by placing a non-zero value in COGxPHR or COGxPHF phase-delay count register, respectively (Register 18-14 and Register 18-15). Refer to Figure 18-10 for COG operation with CCP1 and phase delay. The delay from the input rising event signal switching to the actual assertion of the events is calculated the same as the dead-band and blanking delays. Refer to Equation 18-1.
When the phase-delay count value is zero, phase delay is disabled and the phase-delay counter output is true, thereby, allowing the event signal to pass straight through to the complementary output driver flop.
18.7.1 CUMULATIVE UNCERTAINTY

It is not possible to create more than one COG_clock of uncertainty by successive stages. Consider that the phase-delay stage comes after the blanking stage, the dead-band stage comes after either the blanking or phase-delay stages, and the blanking stage comes after the dead-band stage. When the preceding stage is enabled, the output of that stage is necessarily synchronous with the COG_clock, which removes any possibility of uncertainty in the succeeding stage.

EQUATION 18-1: PHASE, DEAD-BAND, AND BLANKING TIME CALCULATION

\[
T_{\min} = \frac{\text{Count}}{F_{\text{COG\_clock}}}
\]

\[
T_{\max} = \frac{\text{Count} + 1}{F_{\text{COG\_clock}}}
\]

\[
T_{\text{uncertainty}} = T_{\max} - T_{\min}
\]

Also:

\[
T_{\text{uncertainty}} = \frac{1}{F_{\text{COG\_clock}}}
\]

Where:

<table>
<thead>
<tr>
<th>T</th>
<th>Count</th>
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<tbody>
<tr>
<td>Rising Phase Delay</td>
<td>COGxPHR</td>
</tr>
<tr>
<td>Falling Phase Delay</td>
<td>COGxPHF</td>
</tr>
<tr>
<td>Rising Dead Band</td>
<td>COGxDBR</td>
</tr>
<tr>
<td>Falling Dead Band</td>
<td>COGxDBF</td>
</tr>
<tr>
<td>Rising Event Blanking</td>
<td>COGxBLKR</td>
</tr>
<tr>
<td>Falling Event Blanking</td>
<td>COGxBLKF</td>
</tr>
</tbody>
</table>

EXAMPLE 18-1: TIMER UNCERTAINTY

Given:

\[
\text{Count} = Ah = 10d
\]

\[
F_{\text{COG\_Clock}} = 8\text{MHz}
\]

Therefore:

\[
T_{\text{uncertainty}} = \frac{1}{F_{\text{COG\_clock}}}
\]

\[
= \frac{1}{8\text{MHz}} = 125\text{ns}
\]

Proof:

\[
T_{\min} = \frac{\text{Count}}{F_{\text{COG\_clock}}}
\]

\[
= 125\text{ns} \cdot 10d = 1.25\mu\text{s}
\]

\[
T_{\max} = \frac{\text{Count} + 1}{F_{\text{COG\_clock}}}
\]

\[
= 125\text{ns} \cdot (10d + 1)
\]

\[
= 1.375\mu\text{s}
\]

Therefore:

\[
T_{\text{uncertainty}} = T_{\max} - T_{\min}
\]

\[
= 1.375\mu\text{s} - 1.25\mu\text{s}
\]

\[
= 125\text{ns}
\]
18.8 Auto-shutdown Control

Auto-shutdown is a method to immediately override the COG output levels with specific overrides that allow for safe shutdown of the circuit.

The shutdown state can be either cleared automatically or held until cleared by software. In either case, the shutdown overrides remain in effect until the first rising event after the shutdown is cleared.

18.8.1 SHUTDOWN

The shutdown state can be entered by either of the following two mechanisms:

- Software generated
- External Input

18.8.1.1 Software Generated Shutdown

Setting the GxASE bit of the COGxASD0 register (Register 18-7) will force the COG into the shutdown state.

When auto-restart is disabled, the shutdown state will persist until the first rising event after the GxASE bit is cleared by software.

When auto-restart is enabled, the GxASE bit will clear automatically and resume operation on the first rising event after the shutdown input clears. See Figure 18-15 and Section 18.8.3.2 “Auto-Restart”.

18.8.1.2 External Shutdown Source

External shutdown inputs provide the fastest way to safely suspend COG operation in the event of a Fault condition. When any of the selected shutdown inputs goes true, the output drive latches are reset and the COG outputs immediately go to the selected override levels without software delay.

Any combination of the input sources can be selected to cause a shutdown condition. Shutdown occurs when the selected source is low. Shutdown input sources include:

- Any input pin selected with the COGxPPS control
- C2OUT
- C1OUT
- CLC2OUT

Shutdown inputs are selected independently with bits of the COGxASD1 register (Register 18-8).

Note: Shutdown inputs are level sensitive, not edge sensitive. The shutdown state cannot be cleared as long as the shutdown input level persists, except by disabling auto-shutdown.

18.8.2 PIN OVERRIDE LEVELS

The levels driven to the output pins, while the shutdown is active, are controlled by the GxASDAC<1:0> and GxASDBC<1:0> bits of the COGxASD0 register (Register 18-7). GxASDAC<1:0> controls the COGxA and COGxC override levels and GxASDBC<1:0> controls the COGxB and COGxD override levels. There are four override options for each output pair:

- Forced low
- Forced high
- Tri-state
- PWM inactive state (same state as that caused by a falling event)

Note: The polarity control does not apply to the forced low and high override levels but does apply to the PWM inactive state.

18.8.3 AUTO-SHUTDOWN RESTART

After an auto-shutdown event has occurred, there are two ways to resume operation:

- Software controlled
- Auto-restart

The restart method is selected with the GxARSEN bit of the COGxASD0 register. Waveforms of a software controlled automatic restart are shown in Figure 18-15.

18.8.3.1 Software Controlled Restart

When the GxARSEN bit of the COGxASD0 register is cleared, software must clear the GxASE bit to restart COG operation after an auto-shutdown event.

The COG will resume operation on the first rising event after the GxASE bit is cleared. Clearing the shutdown state requires all selected shutdown inputs to be false, otherwise, the GxASE bit will remain set.

18.8.3.2 Auto-Restart

When the GxARSEN bit of the COGxASD0 register is set, the COG will restart from the auto-shutdown state automatically.

The GxASE bit will clear automatically and the COG will resume operation on the first rising event after all selected shutdown inputs go false.
FIGURE 18-15: AUTO-SHUTDOWN WAVEFORM – CCP1 AS RISING AND FALLING EVENT INPUT SOURCE

- 1
- 2
- 3
- 4
- 5

CCP1
GxARSEN
Shutdown input
GxASE
Cleared in software
Cleared in hardware
GxASDAC
GxASDBD
2b00
2b00
COGxA
COGxB
Operating State
NORMAL OUTPUT
SHUTDOWN
NORMAL OUTPUT
SHUTDOWN
NORMAL OUTPUT
SOFTWARE CONTROLLED RESTART
AUTO-RESTART
18.9 Buffer Updates
Changes to the phase, dead band, and blanking count registers need to occur simultaneously during COG operation to avoid unintended operation that may occur as a result of delays between each register write. This is accomplished with the GxLD bit of the COGxCON0 register and double buffering of the phase, blanking, and dead-band count registers.

Before the COG module is enabled, writing the count registers loads the count buffers without need of the GxLD bit. However, when the COG is enabled, the count buffer updates are suspended after writing the count registers until after the GxLD bit is set. When the GxLD bit is set, the phase, dead-band, and blanking register values are transferred to the corresponding buffers synchronous with COG operation. The GxLD bit is cleared by hardware when the transfer is complete.

18.10 Input and Output Pin Selection
The COG has one selection for an input from a device pin. That one input can be used as rising and falling event source or a fault source. The COGxPPS register is used to select the pin. Refer to Register 12-1 and Register 12-2.

The pin PPS control registers are used to enable the COG outputs. Any combination of outputs to pins is possible including multiple pins for the same output. See the RxypPS control register and Section 12.2 “PPS Outputs” for more details.

18.11 Operation During Sleep
The COG continues to operate in Sleep provided that the COG_clock, rising event, and falling event sources remain active.

The HFINTSOC remains active during Sleep when the COG is enabled and the HFINTOSC is selected as the COG_clock source.

18.12 Configuring the COG
The following steps illustrate how to properly configure the COG to ensure a synchronous start with the rising event input:

1. If a pin is to be used for the COG fault or event input, use the COGxPPS register to configure the desired pin.
2. Clear all ANSEL register bits associated with pins that are used for COG functions.
3. Ensure that the TRIS control bits corresponding to the COG outputs to be used are cleared so that all are configured as inputs. The COG module will disable the output drivers as needed for shutdown.
4. Clear the GxEN bit, if not already cleared.
5. Set desired dead-band times with the COGxDBR and COGxDBF registers and select the source with the COGxRDBS and COGxFDBS bits of the COGxCON1 register.
6. Set desired blanking times with the COGxBLKR and COGxBLKF registers.
7. Set desired phase delay with the COGxPHR and COGxPHF registers.
8. Select the desired shutdown sources with the COGxASD1 register.
9. Setup the following controls in COGxASD0 auto-shutdown register:
   a. Select both output override controls to the desired levels (this is necessary, even if not using auto-shutdown because start-up will be from a shutdown state).
   b. Set the GxASE bit and clear the GxARSEN bit.
10. Select the desired rising and falling event sources with the COGxRIS and COGxFIS registers.
11. Select the desired rising and falling event modes with the COGxRSIM and COGxFSIM registers.
12. Configure the following controls in the COGxCON1 register:
   a. Select the desired clock source
   b. Select the desired dead-band timing sources
13. Configure the following controls in the COGxSTR register:
   a. Set the steering bits of the outputs to be used.
   b. Set the static levels.
14. Set the polarity controls in the COGxCON1 register.
15. Set the GxEN bit.
16. Set the pin PPS controls to direct the COG outputs to the desired pins.
17. If auto-restart is to be used, set the GxARSEN bit and the GxASE will be cleared automatically. Otherwise, clear the GxASE bit to start the COG.
18.13 Register Definitions: COG Control

REGISTER 18-1: COGxCON0: COG CONTROL REGISTER 0

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value</th>
<th>Value</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>GxEN: COGx Enable bit</td>
<td>1</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>GxLD: COGx Load Buffers bit</td>
<td>1</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>5</td>
<td>Unimplemented</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>4-3</td>
<td>GxCS&lt;1:0&gt;: COGx Clock Selection bits</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2-0</td>
<td>GxMD&lt;2:0&gt;: COGx Mode Selection bits</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **q** = Value depends on condition

- **1** = Module is enabled
- **0** = Module is disabled

- **1** = Phase, blanking, and dead-band buffers to be loaded with register values on next input events
- **0** = Register to buffer transfer is complete

- **1** = COG_clock is HFINTOSC (stays active during Sleep)
- **0** = COG_clock is Fosc
- **00** = COG_clock is Fosc/4

- **11** = Reserved. Do not use.
- **101** = COG outputs operate in Push-Pull mode
- **100** = COG outputs operate in Half-Bridge mode
- **011** = COG outputs operate in Reverse Full-Bridge mode
- **010** = COG outputs operate in Forward Full-Bridge mode
- **001** = COG outputs operate in synchronous steered PWM mode
- **000** = COG outputs operate in steered PWM mode
### REGISTER 18-2: COGxCON1: COG CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GxRDBS</td>
<td>GxFDBS</td>
</tr>
<tr>
<td></td>
<td>GxPOLD</td>
</tr>
<tr>
<td></td>
<td>GxPOLC</td>
</tr>
<tr>
<td></td>
<td>GxPOLB</td>
</tr>
<tr>
<td></td>
<td>GxPOLA</td>
</tr>
</tbody>
</table>

#### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **q** = Value depends on condition

- **bit 7**  
  - **GxRDBS**: COGx Rising Event Dead-band Timing Source Select bit
    - **1**: Delay chain and COGxDBR are used for dead-band timing generation
    - **0**: COGx_clock and COGxDBR are used for dead-band timing generation

- **bit 6**  
  - **GxFDBS**: COGx Falling Event Dead-band Timing Source select bit
    - **1**: Delay chain and COGxDF are used for dead-band timing generation
    - **0**: COGx_clock and COGxDBF are used for dead-band timing generation

- **bit 5-4**  
  - **Unimplemented**: Read as '0'.

- **bit 3**  
  - **GxPOLD**: COGxD Output Polarity Control bit
    - **1**: Active level of COGxD output is low
    - **0**: Active level of COGxD output is high

- **bit 2**  
  - **GxPOLC**: COGxC Output Polarity Control bit
    - **1**: Active level of COGxC output is low
    - **0**: Active level of COGxC output is high

- **bit 1**  
  - **GxPOLB**: COGxB Output Polarity Control bit
    - **1**: Active level of COGxB output is low
    - **0**: Active level of COGxB output is high

- **bit 0**  
  - **GxPOLA**: COGxA Output Polarity Control bit
    - **1**: Active level of COGxA output is low
    - **0**: Active level of COGxA output is high
## REGISTER 18-3: COGxRIS: COG RISING EVENT INPUT SELECTION REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GxRIS7</td>
<td>GxRIS6</td>
<td>GxRIS5</td>
<td>GxRIS4</td>
<td>GxRIS3</td>
<td>GxRIS2</td>
<td>GxRIS1</td>
<td>GxRIS0</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
</table>
| GxRIS7: COGx Rising Event Input Source 7 Enable bit  
1 = NCO1_out is enabled as a rising event input  
0 = NCO1_out has no effect on the rising event  
| GxRIS6: COGx Rising Event Input Source 6 Enable bit  
1 = PWM3 output is enabled as a rising event input  
0 = PWM3 has no effect on the rising event  
| GxRIS5: COGx Rising Event Input Source 5 Enable bit  
1 = CCP2 output is enabled as a rising event input  
0 = CCP2 output has no effect on the rising event  
| GxRIS4: COGx Rising Event Input Source 4 Enable bit  
1 = CCP1 is enabled as a rising event input  
0 = CCP1 has no effect on the rising event  
| GxRIS3: COGx Rising Event Input Source 3 Enable bit  
1 = CLC1 output is enabled as a rising event input  
0 = CLC1 output has no effect on the rising event  
| GxRIS2: COGx Rising Event Input Source 2 Enable bit  
1 = Comparator 2 output is enabled as a rising event input  
0 = Comparator 2 output has no effect on the rising event  
| GxRIS1: COGx Rising Event Input Source 1 Enable bit  
1 = Comparator 1 output is enabled as a rising event input  
0 = Comparator 1 output has no effect on the rising event  
| GxRIS0: COGx Rising Event Input Source 0 Enable bit  
1 = Pin selected with COGxPPS control register is enabled as rising event input  
0 = Pin selected with COGxPPS control has no effect on the rising event  

### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **q** = Value depends on condition

---

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### REGISTER 18-4: COGxRSIM: COG RISING EVENT SOURCE INPUT MODE REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GxRSIM7</td>
<td>GxRSIM6</td>
<td>GxRSIM5</td>
<td>GxRSIM4</td>
<td>GxRSIM3</td>
<td>GxRSIM2</td>
<td>GxRSIM1</td>
<td>GxRSIM0</td>
</tr>
</tbody>
</table>

#### bit 7
- **GxRSIM7**: COGx Rising Event Input Source 7 Mode bit
  - **GxRIS7 = 1**: NCO1_out low-to-high transition will cause a rising event after rising event phase delay
  - **GxRIS7 = 0**: NCO1_out has no effect on rising event

#### bit 6
- **GxRSIM6**: COGx Rising Event Input Source 6 Mode bit
  - **GxRIS6 = 1**: PWM3 output low-to-high transition will cause a rising event after rising event phase delay
  - **GxRIS6 = 0**: PWM3 output has no effect on rising event

#### bit 5
- **GxRSIM5**: COGx Rising Event Input Source 5 Mode bit
  - **GxRIS5 = 1**: CCP2 output low-to-high transition will cause a rising event after rising event phase delay
  - **GxRIS5 = 0**: CCP2 output has no effect on rising event

#### bit 4
- **GxRSIM4**: COGx Rising Event Input Source 4 Mode bit
  - **GxRIS4 = 1**: CCP1 low-to-high transition will cause a rising event after rising event phase delay
  - **GxRIS4 = 0**: CCP1 has no effect on rising event

#### bit 3
- **GxRSIM3**: COGx Rising Event Input Source 3 Mode bit
  - **GxRIS3 = 1**: CLC1 output low-to-high transition will cause a rising event after rising event phase delay
  - **GxRIS3 = 0**: CLC1 output has no effect on rising event

#### bit 2
- **GxRSIM2**: COGx Rising Event Input Source 2 Mode bit
  - **GxRIS2 = 1**: Comparator 2 low-to-high transition will cause a rising event after rising event phase delay
  - **GxRIS2 = 0**: Comparator 2 has no effect on rising event

#### bit 1
- **GxRSIM1**: COGx Rising Event Input Source 1 Mode bit
  - **GxRIS1 = 1**: Comparator 1 low-to-high transition will cause a rising event after rising event phase delay
  - **GxRIS1 = 0**: Comparator 1 has no effect on rising event

---

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **q** = Value depends on condition

---

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REGISTER 18-4: COGxRSIM: COG RISING EVENT SOURCE INPUT MODE REGISTER

bit 0   GxRSIM0: COGx Rising Event Input Source 0 Mode bit
   GxRSIM0 = 1:
   1 = Pin selected with COGxPPS control low-to-high transition will cause a rising event after rising event
       phase delay
   0 = Pin selected with COGxPPS control high level will cause an immediate rising event
   GxRSIM0 = 0:
   Pin selected with COGxPPS control has no effect on rising event
REGISTER 18-5: COGxFIS: COG FALLING EVENT INPUT SELECTION REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>GxFIS7</th>
<th>GxFIS6</th>
<th>GxFIS5</th>
<th>GxFIS4</th>
<th>GxFIS3</th>
<th>GxFIS2</th>
<th>GxFIS1</th>
<th>GxFIS0</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>6</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>5</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>4</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>3</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **q** = Value depends on condition

**bit 7**
- **GxFIS7**: COGx Falling Event Input Source 7 Enable bit
  - 1 = NCO1_out is enabled as a falling event input
  - 0 = NCO1_out has no effect on the falling event

**bit 6**
- **GxFIS6**: COGx Falling Event Input Source 6 Enable bit
  - 1 = PWM3 output is enabled as a falling event input
  - 0 = PWM3 has no effect on the falling event

**bit 5**
- **GxFIS5**: COGx Falling Event Input Source 5 Enable bit
  - 1 = CCP2 output is enabled as a falling event input
  - 0 = CCP2 output has no effect on the falling event

**bit 4**
- **GxFIS4**: COGx Falling Event Input Source 4 Enable bit
  - 1 = CCP1 is enabled as a falling event input
  - 0 = CCP1 has no effect on the falling event

**bit 3**
- **GxFIS3**: COGx Falling Event Input Source 3 Enable bit
  - 1 = CLC1 output is enabled as a falling event input
  - 0 = CLC1 output has no effect on the falling event

**bit 2**
- **GxFIS2**: COGx Falling Event Input Source 2 Enable bit
  - 1 = Comparator 2 output is enabled as a falling event input
  - 0 = Comparator 2 output has no effect on the falling event

**bit 1**
- **GxFIS1**: COGx Falling Event Input Source 1 Enable bit
  - 1 = Comparator 1 output is enabled as a falling event input
  - 0 = Comparator 1 output has no effect on the falling event

**bit 0**
- **GxFIS0**: COGx Falling Event Input Source 0 Enable bit
  - 1 = Pin selected with COGxPPS control register is enabled as falling event input
  - 0 = Pin selected with COGxPPS control has no effect on the falling event
**REGISTER 18-6: COGxFSIM: COG FALLING EVENT SOURCE INPUT MODE REGISTER**

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GxFSIM7</td>
<td>GxFSIM6</td>
<td>GxFSIM5</td>
<td>GxFSIM4</td>
<td>GxFSIM3</td>
<td>GxFSIM2</td>
<td>GxFSIM1</td>
<td>GxFSIM0</td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **q** = Value depends on condition

**bit 7**

**GxFSIM7:** COGx Falling Event Input Source 7 Mode bit

- **GxFSIM7 = 1:**
  - 1 = NCO1_out high-to-low transition will cause a falling event after falling event phase delay
  - 0 = NCO1_out low level will cause an immediate falling event

- **GxFSIM7 = 0:**
  - NCO1_out has no effect on falling event

**bit 6**

**GxFSIM6:** COGx Falling Event Input Source 6 Mode bit

- **GxFSIM6 = 1:**
  - 1 = PWM3 output high-to-low transition will cause a falling event after falling event phase delay
  - 0 = PWM3 output low level will cause an immediate falling event

- **GxFSIM6 = 0:**
  - PWM3 output has no effect on falling event

**bit 5**

**GxFSIM5:** COGx Falling Event Input Source 5 Mode bit

- **GxFSIM5 = 1:**
  - 1 = CCP2 output high-to-low transition will cause a falling event after falling event phase delay
  - 0 = CCP2 output low level will cause an immediate falling event

- **GxFSIM5 = 0:**
  - CCP2 output has no effect on falling event

**bit 4**

**GxFSIM4:** COGx Falling Event Input Source 4 Mode bit

- **GxFSIM4 = 1:**
  - 1 = CCP1 high-to-low transition will cause a falling event after falling event phase delay
  - 0 = CCP1 low level will cause an immediate falling event

- **GxFSIM4 = 0:**
  - CCP1 has no effect on falling event

**bit 3**

**GxFSIM3:** COGx Falling Event Input Source 3 Mode bit

- **GxFSIM3 = 1:**
  - 1 = CLC1 output high-to-low transition will cause a falling event after falling event phase delay
  - 0 = CLC1 output low level will cause an immediate falling event

- **GxFSIM3 = 0:**
  - CLC1 output has no effect on falling event

**bit 2**

**GxFSIM2:** COGx Falling Event Input Source 2 Mode bit

- **GxFSIM2 = 1:**
  - 1 = Comparator 2 high-to-low transition will cause a falling event after falling event phase delay
  - 0 = Comparator 2 low level will cause an immediate falling event

- **GxFSIM2 = 0:**
  - Comparator 2 has no effect on falling event

**bit 1**

**GxFSIM1:** COGx Falling Event Input Source 1 Mode bit

- **GxFSIM1 = 1:**
  - 1 = Comparator 1 high-to-low transition will cause a falling event after falling event phase delay
  - 0 = Comparator 1 low level will cause an immediate falling event

- **GxFSIM1 = 0:**
  - Comparator 1 has no effect on falling event
REGISTER 18-6: COGxFSIM: COG FALLING EVENT SOURCE INPUT MODE REGISTER

bit 0  GxFSIM0: COGx Falling Event Input Source 0 Mode bit

GxFSIM0 = 1:
1 = Pin selected with COGxPPS control high-to-low transition will cause a falling event after falling event phase delay
0 = Pin selected with COGxPPS control low level will cause an immediate falling event

GxFSIM0 = 0:
Pin selected with COGxPPS control has no effect on falling event
<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
</table>
| 7   | GxASE: Auto-Shutdown Event Status bit           | 1 = COG is in the shutdown state  
|     |                                                 | 0 = COG is either not in the shutdown state or will exit the shutdown state on the next rising event |
| 6   | GxARSEN: Auto-Restart Enable bit                | 1 = Auto-restart is enabled  
|     |                                                 | 0 = Auto-restart is disabled |
| 5-4 | GxASDBD<1:0>: COGxB and COGxD Auto-shutdown Override Level Select bits  
|     |                                                 | 11 = A logic '1' is placed on COGxB and COGxD when shutdown is active  
|     |                                                 | 10 = A logic '0' is placed on COGxB and COGxD when shutdown is active  
|     |                                                 | 01 = COGxB and COGxD are tri-stated when shutdown is active  
|     |                                                 | 00 = The inactive state of the pin, including polarity, is placed on COGxB and COGxD when shutdown is active |
| 3-2 | GxASDAC<1:0>: COGxA and COGxC Auto-shutdown Override Level Select bits  
|     |                                                 | 11 = A logic '1' is placed on COGxA and COGxC when shutdown is active  
|     |                                                 | 10 = A logic '0' is placed on COGxA and COGxC when shutdown is active  
|     |                                                 | 01 = COGxA and COGxC are tri-stated when shutdown is active  
|     |                                                 | 00 = The inactive state of the pin, including polarity, is placed on COGxA and COGxC when shutdown is active |
| 1-0 | Unimplemented: Read as '0'                      |                            |
### REGISTER 18-8: COGxASD1: COG AUTO-SHUTDOWN CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>bit</th>
<th>Legend</th>
<th>Description</th>
<th>Output Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7-4</td>
<td>Unimplemented</td>
<td>Read as '0'</td>
<td></td>
</tr>
<tr>
<td>bit 3</td>
<td>GxAS3E</td>
<td>COGx Auto-shutdown Source Enable bit 3</td>
<td>1 = COGx is shutdown when CLC2 output is low, 0 = CLC2 output has no effect on shutdown</td>
</tr>
<tr>
<td>bit 2</td>
<td>GxAS2E</td>
<td>COGx Auto-shutdown Source Enable bit 2</td>
<td>1 = COGx is shutdown when Comparator 2 output is low, 0 = Comparator 2 output has no effect on shutdown</td>
</tr>
<tr>
<td>bit 1</td>
<td>GxAS1E</td>
<td>COGx Auto-shutdown Source Enable bit 1</td>
<td>1 = COGx is shutdown when Comparator 1 output is low, 0 = Comparator 1 output has no effect on shutdown</td>
</tr>
<tr>
<td>bit 0</td>
<td>GxAS0E</td>
<td>COGx Auto-shutdown Source Enable bit 0</td>
<td>1 = COGx is shutdown when Pin selected with COGxPPS control is low, 0 = Pin selected with COGxPPS control has no effect on shutdown</td>
</tr>
</tbody>
</table>
**REGISTER 18-9: COGxSTR: COG STEERING CONTROL REGISTER 1**

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GxDATD</td>
<td>GxDATC</td>
<td>GxDATB</td>
<td>GxDA</td>
<td>GxSTRD</td>
<td>GxSTRC</td>
<td>GxSTRB</td>
<td>GxSTRA</td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **q** = Value depends on condition

**bit 7**  
**GxDATD**: COGxD Static Output Data bit  
1 = COGxD static data is high  
0 = COGxD static data is low

**bit 6**  
**GxDATC**: COGxC Static Output Data bit  
1 = COGxC static data is high  
0 = COGxC static data is low

**bit 5**  
**GxDATB**: COGxB Static Output Data bit  
1 = COGxB static data is high  
0 = COGxB static data is low

**bit 4**  
**GxDA**: COGxA Static Output Data bit  
1 = COGxA static data is high  
0 = COGxA static data is low

**bit 3**  
**GxSTRD**: COGxD Steering Control bit  
1 = COGxD output has the COGxD waveform with polarity control from GxPOLD bit  
0 = COGxD output is the static data level determined by the GxDATD bit

**bit 2**  
**GxSTRC**: COGxC Steering Control bit  
1 = COGxC output has the COGxC waveform with polarity control from GxPOLC bit  
0 = COGxC output is the static data level determined by the GxDATC bit

**bit 1**  
**GxSTRB**: COGxB Steering Control bit  
1 = COGxB output has the COGxB waveform with polarity control from GxPOLB bit  
0 = COGxB output is the static data level determined by the GxDATB bit

**bit 0**  
**GxSTRA**: COGxA Steering Control bit  
1 = COGxA output has the COGxA waveform with polarity control from GxPOLA bit  
0 = COGxA output is the static data level determined by the GxDA bit
REGISTER 18-10: COGxDBR: COG RISING EVENT DEAD-BAND COUNT REGISTER

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th>bit 5-0</th>
<th>GxDBR&lt;5:0&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>U-0</td>
<td>U-0</td>
<td>R/W-x/u</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **q** = Value depends on condition

bit 7-6 **Unimplemented**: Read as ‘0’
bit 5-0 **GxDBR<5:0>**: Rising Event Dead-band Count Value bits
- **GxRDBS = 1**: Number of delay chain element periods to delay primary output after rising event
- **GxRDBS = 0**: Number of COGx clock periods to delay primary output after rising event

REGISTER 18-11: COGxDBF: COG FALLING EVENT DEAD-BAND COUNT REGISTER

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th>bit 5-0</th>
<th>GxDBF&lt;5:0&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>U-0</td>
<td>U-0</td>
<td>R/W-x/u</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **q** = Value depends on condition

bit 7-6 **Unimplemented**: Read as ‘0’
bit 5-0 **GxDBF<5:0>**: Falling Event Dead-band Count Value bits
- **GxFDBS = 1**: Number of delay chain element periods to delay complementary output after falling event input
- **GxFDBS = 0**: Number of COGx clock periods to delay complementary output after falling event input
### REGISTER 18-12: COGxBLKR: COG RISING EVENT BLANKING COUNT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>GxBLKR&lt;5:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**

- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- q = Value depends on condition

- **bit 7-6 Unimplemented:** Read as ‘0’
- **bit 5-0 GxBLKR<5:0>:** Rising Event Blanking Count Value bits
  = Number of COGx clock periods to inhibit falling event inputs

### REGISTER 18-13: COGxBLKF: COG FALLING EVENT BLANKING COUNT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**

- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- q = Value depends on condition

- **bit 7-6 Unimplemented:** Read as ‘0’
- **bit 5-0 GxBLKF<5:0>:** Falling Event Blanking Count Value bits
  = Number of COGx clock periods to inhibit rising event inputs
### REGISTER 18-14: COGxPHR: COG RISING EDGE PHASE DELAY COUNT REGISTER

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th>bit 5-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>U-0</td>
<td>U-0</td>
</tr>
<tr>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
</tr>
<tr>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>GxPHR&lt;5:0&gt;</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **q** = Value depends on condition

- **bit 7-6**: Unimplemented: Read as ‘0’
- **bit 5-0**: GxPHR<5:0>: Rising Edge Phase Delay Count Value bits
  = Number of COGx clock periods to delay rising edge event

### REGISTER 18-15: COGxPHF: COG FALLING EDGE PHASE DELAY COUNT REGISTER

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th>bit 5-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>U-0</td>
<td>U-0</td>
</tr>
<tr>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
</tr>
<tr>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>GxPHF&lt;5:0&gt;</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **q** = Value depends on condition

- **bit 7-6**: Unimplemented: Read as ‘0’
- **bit 5-0**: GxPHF<5:0>: Falling Edge Phase Delay Count Value bits
  = Number of COGx clock periods to delay falling edge event
## TABLE 18-2: SUMMARY OF REGISTERS ASSOCIATED WITH COG

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>126</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
</tr>
<tr>
<td>COG1PHR</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANPHR&lt;5:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>216</td>
</tr>
<tr>
<td>COG1PHF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANPHF&lt;5:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>216</td>
</tr>
<tr>
<td>COG1BLKR</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANBLKR&lt;5:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>215</td>
</tr>
<tr>
<td>COG1BLKF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANBLKF&lt;5:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>215</td>
</tr>
<tr>
<td>COG1DBR</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>G1DBR&lt;5:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>214</td>
</tr>
<tr>
<td>COG1DBF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>G1DBF&lt;5:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>214</td>
</tr>
<tr>
<td>COG1RIS</td>
<td>G1RIS7</td>
<td>G1RIS6</td>
<td>G1RIS5</td>
<td>G1RIS4</td>
<td>G1RIS3</td>
<td>G1RIS2</td>
<td>G1RIS1</td>
<td>G1RIS0</td>
<td>205</td>
</tr>
<tr>
<td>COG1RISIM</td>
<td>G1RIS7</td>
<td>G1RIS6</td>
<td>G1RIS5</td>
<td>G1RIS4</td>
<td>G1RIS3</td>
<td>G1RIS2</td>
<td>G1RIS1</td>
<td>G1RIS0</td>
<td>206</td>
</tr>
<tr>
<td>COG1FIS</td>
<td>G1FIS7</td>
<td>G1FIS6</td>
<td>G1FIS5</td>
<td>G1FIS4</td>
<td>G1FIS3</td>
<td>G1FIS2</td>
<td>G1FIS1</td>
<td>G1FIS0</td>
<td>208</td>
</tr>
<tr>
<td>COG1FSIM</td>
<td>G1FSIM7</td>
<td>G1FSIM6</td>
<td>G1FSIM5</td>
<td>G1FSIM4</td>
<td>G1FSIM3</td>
<td>G1FSIM2</td>
<td>G1FSIM1</td>
<td>G1FSIM0</td>
<td>209</td>
</tr>
<tr>
<td>COG1CON0</td>
<td>G1EN</td>
<td>G1LD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>G1CS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td>G1MD&lt;2:0&gt;</td>
</tr>
<tr>
<td>COG1CON1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>COG1ASD0</td>
<td>G1ASE</td>
<td>G1ARSEN</td>
<td>G1ASDBD&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>COG1ASD1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>G1AS3E</td>
<td>G1AS2E</td>
</tr>
<tr>
<td>COG1STR</td>
<td>G1SDATD</td>
<td>G1SDATC</td>
<td>G1SDATB</td>
<td>G1SDATA</td>
<td>G1STRD</td>
<td>G1STRC</td>
<td>G1STRB</td>
<td>G1STRA</td>
<td>213</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>T0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>T0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
<td>COG1PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
<td>93</td>
</tr>
<tr>
<td>PIIR2</td>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td>96</td>
</tr>
<tr>
<td>RxyPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RxyPPS&lt;4:0&gt;</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**  

- `x` = unknown,  
- `@` = unchanged,  
- `–` = unimplemented locations read as '0'.  

Shaded cells are not used by COG.
The Configurable Logic Cell (CLCx) provides programmable logic that operates outside the speed limitations of software execution. The logic cell takes up to 32 input signals and, through the use of configurable gates, reduces the 32 inputs to four logic lines that drive one of eight selectable single-output logic functions.

Input sources are a combination of the following:

- I/O pins
- Internal clocks
- Peripherals
- Register bits

The output can be directed internally to peripherals and to an output pin.

Possible configurations include:

- Combinatorial Logic
  - AND
  - NAND
  - AND-OR
  - AND-OR-INVERT
  - OR-XOR
  - OR-XNOR
- Latches
  - S-R
  - Clocked D with Set and Reset
  - Transparent D with Set and Reset
  - Clocked J-K with Reset

Refer to Figure 19-1 for a simplified diagram showing signal flow through the CLCx.

Note 1: See Figure 19-2.
Note 2: See Figure 19-3.
19.1 CLCx Setup

Programming the CLCx module is performed by configuring the four stages in the logic signal flow. The four stages are:

- Data selection
- Data gating
- Logic function selection
- Output polarity

Each stage is set up at runtime by writing to the corresponding CLCx Special Function Registers. This has the added advantage of permitting logic reconfiguration on-the-fly during program execution.

19.1.1 DATA SELECTION

There are 32 signals available as inputs to the configurable logic. Four 32-input multiplexers are used to select the inputs to pass on to the next stage.

Data selection is through four multiplexers as indicated on the left side of Figure 19-2. Data inputs in the figure are identified by a generic numbered input name.

Table 19-1 correlates the generic input name to the actual signal for each CLC module. The column labeled Icx.dy indicates the MUX selection code for the selected data input. Dxy is an abbreviation for the MUX select input codes: LCxD1S<4:0> through LCxD4S<4:0>.

Data inputs are selected with CLCxSEL0 through CLCxSEL3 registers (Register 19-3 through Register 19-6).

<table>
<thead>
<tr>
<th>Data Input</th>
<th>Icx.dy Dxy</th>
<th>CLCx</th>
</tr>
</thead>
<tbody>
<tr>
<td>LCx_in[31]</td>
<td>11111</td>
<td>Fosc</td>
</tr>
<tr>
<td>LCx_in[30]</td>
<td>11110</td>
<td>HFINTOSC</td>
</tr>
<tr>
<td>LCx_in[29]</td>
<td>11101</td>
<td>LFINTOSC</td>
</tr>
<tr>
<td>LCx_in[28]</td>
<td>11100</td>
<td>ADCRC</td>
</tr>
<tr>
<td>LCx_in[27]</td>
<td>11101</td>
<td>IOCIF set signal</td>
</tr>
<tr>
<td>LCx_in[26]</td>
<td>11100</td>
<td>T2_match</td>
</tr>
<tr>
<td>LCx_in[25]</td>
<td>11101</td>
<td>T1_overflow</td>
</tr>
<tr>
<td>LCx_in[24]</td>
<td>11000</td>
<td>T0_overflow</td>
</tr>
<tr>
<td>LCx_in[23]</td>
<td>10111</td>
<td>T6_match</td>
</tr>
<tr>
<td>LCx_in[22]</td>
<td>10110</td>
<td>T4_match</td>
</tr>
<tr>
<td>LCx_in[21]</td>
<td>10101</td>
<td>DT from EUSART</td>
</tr>
<tr>
<td>LCx_in[20]</td>
<td>10100</td>
<td>TX/CK from EUSART</td>
</tr>
<tr>
<td>LCx_in[19]</td>
<td>10111</td>
<td>ZCDx_out from Zero-Cross Detect</td>
</tr>
<tr>
<td>LCx_in[18]</td>
<td>10010</td>
<td>NCO1_out</td>
</tr>
<tr>
<td>LCx_in[17]</td>
<td>10001</td>
<td>SDO/SDA from MSSP</td>
</tr>
<tr>
<td>LCx_in[16]</td>
<td>10000</td>
<td>SCK from MSSP</td>
</tr>
<tr>
<td>LCx_in[15]</td>
<td>01111</td>
<td>PWM4_out</td>
</tr>
<tr>
<td>LCx_in[14]</td>
<td>01110</td>
<td>PWM3_out</td>
</tr>
<tr>
<td>LCx_in[13]</td>
<td>01101</td>
<td>CCP2 output</td>
</tr>
<tr>
<td>LCx_in[12]</td>
<td>01100</td>
<td>CCP1 output</td>
</tr>
<tr>
<td>LCx_in[11]</td>
<td>01011</td>
<td>COG1B</td>
</tr>
<tr>
<td>LCx_in[10]</td>
<td>01010</td>
<td>COG1A</td>
</tr>
<tr>
<td>LCx_in[9]</td>
<td>01001</td>
<td>sync_C2OUT</td>
</tr>
<tr>
<td>LCx_in[8]</td>
<td>01000</td>
<td>sync_C1OUT</td>
</tr>
<tr>
<td>LCx_in[7]</td>
<td>00111</td>
<td>LC4_out from the CLC4</td>
</tr>
<tr>
<td>LCx_in[6]</td>
<td>00110</td>
<td>LC3_out from the CLC3</td>
</tr>
<tr>
<td>LCx_in[5]</td>
<td>00101</td>
<td>LC2_out from the CLC2</td>
</tr>
<tr>
<td>LCx_in[4]</td>
<td>00100</td>
<td>LC1_out from the CLC1</td>
</tr>
<tr>
<td>LCx_in[3]</td>
<td>00011</td>
<td>CLCIN3 pin input selected in CLCIN3PPS register</td>
</tr>
<tr>
<td>LCx_in[2]</td>
<td>00010</td>
<td>CLCIN2 pin input selected in CLCIN2PPS register</td>
</tr>
<tr>
<td>LCx_in[1]</td>
<td>00001</td>
<td>CLCIN1 pin input selected in CLCIN1PPS register</td>
</tr>
<tr>
<td>LCx_in[0]</td>
<td>00000</td>
<td>CLCIN0 pin input selected in CLCIN0PPS register</td>
</tr>
</tbody>
</table>

Note: Data selections are undefined at power-up.
19.1.2 DATA GATING

Outputs from the input multiplexers are directed to the desired logic function input through the data gating stage. Each data gate can direct any combination of the four selected inputs.

**Note:** Data gating is undefined at power-up.

The gate stage is more than just signal direction. The gate can be configured to direct each input signal as inverted or non-inverted data. Directed signals are ANDed together in each gate. The output of each gate can be inverted before going on to the logic function stage.

The gating is in essence a 1-to-4 input AND/NAND/OR/NOR gate. When every input is inverted and the output is inverted, the gate is an OR of all enabled data inputs. When the inputs and output are not inverted, the gate is an AND or all enabled inputs.

Table 19-2 summarizes the basic logic that can be obtained in gate 1 by using the gate logic select bits. The table shows the logic of four input variables, but each gate can be configured to use less than four. If no inputs are selected, the output will be zero or one, depending on the gate output polarity bit.

**TABLE 19-2: DATA GATING LOGIC**

<table>
<thead>
<tr>
<th>CLCxGLS0</th>
<th>LCxG1POL</th>
<th>Gate Logic</th>
</tr>
</thead>
<tbody>
<tr>
<td>0x55</td>
<td>1</td>
<td>AND</td>
</tr>
<tr>
<td>0x55</td>
<td>0</td>
<td>NAND</td>
</tr>
<tr>
<td>0xAA</td>
<td>1</td>
<td>NOR</td>
</tr>
<tr>
<td>0xAA</td>
<td>0</td>
<td>OR</td>
</tr>
<tr>
<td>0x00</td>
<td>0</td>
<td>Logic 0</td>
</tr>
<tr>
<td>0x00</td>
<td>1</td>
<td>Logic 1</td>
</tr>
</tbody>
</table>

It is possible (but not recommended) to select both the true and negated values of an input. When this is done, the gate output is zero, regardless of the other inputs, but may emit logic glitches (transient-induced pulses). If the output of the channel must be zero or one, the recommended method is to set all gate bits to zero and use the gate polarity bit to set the desired level.

Data gating is configured with the logic gate select registers as follows:

- **Gate 1:** CLCxGLS0 ([Register 19-7])
- **Gate 2:** CLCxGLS1 ([Register 19-8])
- **Gate 3:** CLCxGLS2 ([Register 19-9])
- **Gate 4:** CLCxGLS3 ([Register 19-10])

Register number suffixes are different than the gate numbers because other variations of this module have multiple gate selections in the same register.

Data gating is indicated in the right side of Figure 19-2. Only one gate is shown in detail. The remaining three gates are configured identically with the exception that the data enables correspond to the enables for that gate.

19.1.3 LOGIC FUNCTION

There are eight available logic functions including:

- AND-OR
- OR-XOR
- AND
- S-R Latch
- D Flip-Flop with Set and Reset
- D Flip-Flop with Reset
- J-K Flip-Flop with Reset
- Transparent Latch with Set and Reset

Logic functions are shown in Figure 19-3. Each logic function has four inputs and one output. The four inputs are the four data gate outputs of the previous stage. The output is fed to the inversion stage and from there to other peripherals, an output pin, and back to the CLCx itself.

19.1.4 OUTPUT POLARITY

The last stage in the configurable logic cell is the output polarity. Setting the LCxPOL bit of the CLCxCON register inverts the output signal from the logic stage. Changing the polarity while the interrupts are enabled will cause an interrupt for the resulting output transition.

**Note:** Data gating is undefined at power-up.
19.1.5 CLCx SETUP STEPS
The following steps should be followed when setting up the CLCx:

- Disable CLCx by clearing the LCxEN bit.
- Select desired inputs using CLCxSEL0 through CLCxSEL3 registers (See Table 19-1).
- Clear any associated ANSEL bits.
- Set all TRIS bits associated with inputs.
- Clear all TRIS bits associated with outputs.
- Enable the chosen inputs through the four gates using CLCxGLS0, CLCxGLS1, CLCxGLS2, and CLCxGLS3 registers.
- Select the gate output polarities with the LCxPOLy bits of the CLCxPOL register.
- Select the desired logic function with the LCxMODE<2:0> bits of the CLCxCON register.
- Select the desired polarity of the logic output with the LCxPOL bit of the CLCxCON register. (This step may be combined with the previous gate output polarity step).
- If driving a device pin, set the desired pin PPS control register and also clear the TRIS bit corresponding to that output.
- If interrupts are desired, configure the following bits:
  - Set the LCxINTP bit in the CLCxCON register for rising event.
  - Set the LCxINTN bit in the CLCxCON register for falling event.
  - Set the CLCxIE bit of the associated PIE registers.
  - Set the GIE and PEIE bits of the INTCON register.
- Enable the CLCx by setting the LCxEN bit of the CLCxCON register.

19.2 CLCx Interrupts
An interrupt will be generated upon a change in the output value of the CLCx when the appropriate interrupt enables are set. A rising edge detector and a falling edge detector are present in each CLC for this purpose.

- The CLCxIF bit of the associated PIR registers will be set when either edge detector is triggered and its associated enable bit is set. The LCxINTP enables rising edge interrupts and the LCxINTN bit enables falling edge interrupts. Both are located in the CLCxCON register.

To fully enable the interrupt, set the following bits:

- LCxEN bit of the CLCxCON register
- CLCxIE bit of the associated PIE registers
- LCxINTP bit of the CLCxCON register (for a rising edge detection)
- LCxINTN bit of the CLCxCON register (for a falling edge detection)
- PEIE and GIE bits of the INTCON register

The CLCxIF bit of the associated PIR registers, must be cleared in software as part of the interrupt service. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

19.3 Output Mirror Copies
Mirror copies of all LCxCON output bits are contained in the CLCxDATA register. Reading this register reads the outputs of all CLCs simultaneously. This prevents any reading skew introduced by testing or reading the CLCxOUT bits in the individual CLCxCON registers.

19.4 Effects of a Reset
The CLCxCON register is cleared to zero as the result of a Reset. All other selection and gating values remain unchanged.

19.5 Operation During Sleep
The CLC module operates independently from the system clock and will continue to run during Sleep, provided that the input sources selected remain active.

- The HFINTOSC remains active during Sleep when the CLC module is enabled and the HFINTOSC is selected as an input source, regardless of the system clock source selected.

In other words, if the HFINTOSC is simultaneously selected as the system clock and as a CLC input source, when the CLC is enabled, the CPU will go idle during Sleep, but the CLC will continue to operate and the HFINTOSC will remain active.

This will have a direct effect on the Sleep mode current.
FIGURE 19-2: INPUT DATA SELECTION AND GATING

Data Selection

Note: All controls are undefined at power-up.
FIGURE 19-3: PROGRAMMABLE LOGIC FUNCTIONS

<table>
<thead>
<tr>
<th>AND - OR</th>
<th>OR - XOR</th>
</tr>
</thead>
<tbody>
<tr>
<td><img src="image1" alt="Diagram 1" /></td>
<td><img src="image2" alt="Diagram 2" /></td>
</tr>
<tr>
<td>LCxMODE&lt;2:0&gt;= 000</td>
<td>LCxMODE&lt;2:0&gt;= 001</td>
</tr>
<tr>
<td>4-Input AND</td>
<td>S-R Latch</td>
</tr>
<tr>
<td><img src="image3" alt="Diagram 3" /></td>
<td><img src="image4" alt="Diagram 4" /></td>
</tr>
<tr>
<td>LCxMODE&lt;2:0&gt;= 010</td>
<td>LCxMODE&lt;2:0&gt;= 011</td>
</tr>
<tr>
<td>1-Input D Flip-Flop with S and R</td>
<td>2-Input D Flip-Flop with R</td>
</tr>
<tr>
<td><img src="image5" alt="Diagram 5" /></td>
<td><img src="image6" alt="Diagram 6" /></td>
</tr>
<tr>
<td>LCxMODE&lt;2:0&gt;= 100</td>
<td>LCxMODE&lt;2:0&gt;= 101</td>
</tr>
<tr>
<td>J-K Flip-Flop with R</td>
<td>1-Input Transparent Latch with S and R</td>
</tr>
<tr>
<td><img src="image7" alt="Diagram 7" /></td>
<td><img src="image8" alt="Diagram 8" /></td>
</tr>
<tr>
<td>LCxMODE&lt;2:0&gt;= 110</td>
<td>LCxMODE&lt;2:0&gt;= 111</td>
</tr>
</tbody>
</table>
## 19.6 Register Definitions: CLC Control

**REGISTER 19-1: CLCxCON: CONFIGURABLE LOGIC CELL CONTROL REGISTER**

<table>
<thead>
<tr>
<th>Bit</th>
<th>LCxEN</th>
<th>LCxOUT</th>
<th>LCxINTP</th>
<th>LCxINTN</th>
<th>LCxMODE&lt;2:0&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>U</td>
<td>R</td>
<td>W</td>
<td>U</td>
<td>R</td>
</tr>
<tr>
<td>6</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<tr>
<td>5</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>4</td>
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<tr>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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</tr>
<tr>
<td>0</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- '1' = Bit is set
- '0' = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

**bit 7** LCxEN: Configurable Logic Cell Enable bit
- 1 = Configurable logic cell is enabled and mixing input signals
- 0 = Configurable logic cell is disabled and has logic zero output

**bit 6** Unimplemented: Read as '0'

**bit 5** LCxOUT: Configurable Logic Cell Data Output bit
Read-only: logic cell output data, after LCxPOL; sampled from lcx_out wire.

**bit 4** LCxINTP: Configurable Logic Cell Positive Edge Going Interrupt Enable bit
- 1 = CLCxIF will be set when a rising edge occurs on lcx_out
- 0 = CLCxIF will not be set

**bit 3** LCxINTN: Configurable Logic Cell Negative Edge Going Interrupt Enable bit
- 1 = CLCxIF will be set when a falling edge occurs on lcx_out
- 0 = CLCxIF will not be set

**bit 2-0** LCxMODE<2:0>: Configurable Logic Cell Functional Mode bits
- 111 = Cell is 1-input transparent latch with S and R
- 110 = Cell is J-K flip-flop with R
- 101 = Cell is 2-input D flip-flop with R
- 100 = Cell is 1-input D flip-flop with S and R
- 011 = Cell is S-R latch
- 010 = Cell is 4-input AND
- 001 = Cell is OR-XOR
- 000 = Cell is AND-OR
REGISTER 19-2: CLCxPOL: SIGNAL POLARITY CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>LCxPOL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>LCxG4POL</td>
<td>LCxG3POL</td>
<td>LCxG2POL</td>
<td>LCxG1POL</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7  **LCxPOL**: LCOUT Polarity Control bit
1 = The output of the logic cell is inverted
0 = The output of the logic cell is not inverted

bit 6-4  **Unimplemented**: Read as ‘0’

bit 3  **LCxG4POL**: Gate 4 Output Polarity Control bit
1 = The output of gate 4 is inverted when applied to the logic cell
0 = The output of gate 4 is not inverted

bit 2  **LCxG3POL**: Gate 3 Output Polarity Control bit
1 = The output of gate 3 is inverted when applied to the logic cell
0 = The output of gate 3 is not inverted

bit 1  **LCxG2POL**: Gate 2 Output Polarity Control bit
1 = The output of gate 2 is inverted when applied to the logic cell
0 = The output of gate 2 is not inverted

bit 0  **LCxG1POL**: Gate 1 Output Polarity Control bit
1 = The output of gate 1 is inverted when applied to the logic cell
0 = The output of gate 1 is not inverted

REGISTER 19-3: CLCxSEL0: GENERIC CLCx DATA 1 SELECT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>LCxD1S&lt;4:0&gt;</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-5  **Unimplemented**: Read as ‘0’

bit 4-0  **LCxD1S<4:0>**: CLCx Data1 Input Selection bits
See Table 19-1.
REGISTER 19-4:  CLCxSEL1: GENERIC CLCx DATA 2 SELECT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

LCxD2S<4:0>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

bit 7-5  **Unimplemented**: Read as '0'

bit 4-0  **LCxD2S<4:0>**: CLCx Data 2 Input Selection bits

See Table 19-1.

REGISTER 19-5:  CLCxSEL2: GENERIC CLCx DATA 3 SELECT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

LCxD3S<4:0>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

bit 7-5  **Unimplemented**: Read as '0'

bit 4-0  **LCxD3S<4:0>**: CLCx Data 3 Input Selection bits

See Table 19-1.

REGISTER 19-6:  CLCxSEL3: GENERIC CLCx DATA 4 SELECT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

LCxD4S<4:0>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

bit 7-5  **Unimplemented**: Read as '0'

bit 4-0  **LCxD4S<4:0>**: CLCx Data 4 Input Selection bits

See Table 19-1.
## REGISTER 19-7: CLCxGLS0: GATE 1 LOGIC SELECT REGISTER

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 0</th>
</tr>
</thead>
</table>

<table>
<thead>
<tr>
<th>LCxG1D4T</th>
<th>LCxG1D4N</th>
<th>LCxG1D3T</th>
<th>LCxG1D3N</th>
<th>LCxG1D2T</th>
<th>LCxG1D2N</th>
<th>LCxG1D1T</th>
<th>LCxG1D1N</th>
</tr>
</thead>
<tbody>
<tr>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
</tr>
</tbody>
</table>

### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- ‘-n/n’ = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

### Bit Descriptions:
- **bit 7**  
  **LCxG1D4T**: Gate 1 Data 4 True (non-inverted) bit
  - 1 = LCxD4T is gated into LCxG1
  - 0 = LCxD4T is not gated into LCxG1
- **bit 6**  
  **LCxG1D4N**: Gate 1 Data 4 Negated (inverted) bit
  - 1 = LCxD4N is gated into LCxG1
  - 0 = LCxD4N is not gated into LCxG1
- **bit 5**  
  **LCxG1D3T**: Gate 1 Data 3 True (non-inverted) bit
  - 1 = LCxD3T is gated into LCxG1
  - 0 = LCxD3T is not gated into LCxG1
- **bit 4**  
  **LCxG1D3N**: Gate 1 Data 3 Negated (inverted) bit
  - 1 = LCxD3N is gated into LCxG1
  - 0 = LCxD3N is not gated into LCxG1
- **bit 3**  
  **LCxG1D2T**: Gate 1 Data 2 True (non-inverted) bit
  - 1 = LCxD2T is gated into LCxG1
  - 0 = LCxD2T is not gated into LCxG1
- **bit 2**  
  **LCxG1D2N**: Gate 1 Data 2 Negated (inverted) bit
  - 1 = LCxD2N is gated into LCxG1
  - 0 = LCxD2N is not gated into LCxG1
- **bit 1**  
  **LCxG1D1T**: Gate 1 Data 1 True (non-inverted) bit
  - 1 = LCxD1T is gated into LCxG1
  - 0 = LCxD1T is not gated into LCxG1
- **bit 0**  
  **LCxG1D1N**: Gate 1 Data 1 Negated (inverted) bit
  - 1 = LCxD1N is gated into LCxG1
  - 0 = LCxD1N is not gated into LCxG1
## REGISTER 19-8: CLCxGLS1: GATE 2 LOGIC SELECT REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>LCxG2D4T</th>
<th>LCxG2D4N</th>
<th>LCxG2D3T</th>
<th>LCxG2D3N</th>
<th>LCxG2D2T</th>
<th>LCxG2D2N</th>
<th>LCxG2D1T</th>
<th>LCxG2D1N</th>
</tr>
</thead>
<tbody>
<tr>
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</tr>
<tr>
<td>6</td>
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</tr>
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<td></td>
</tr>
<tr>
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<tr>
<td>0</td>
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<td></td>
<td></td>
</tr>
</tbody>
</table>

### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **‘-n/n’** = Value at POR and BOR/Value at all other Resets

- **bit 7**
  - **LCxG2D4T**: Gate 2 Data 4 True (non-inverted) bit
    - 1 = LCxD4T is gated into LCxG2
    - 0 = LCxD4T is not gated into LCxG2

- **bit 6**
  - **LCxG2D4N**: Gate 2 Data 4 Negated (inverted) bit
    - 1 = LCxD4N is gated into LCxG2
    - 0 = LCxD4N is not gated into LCxG2

- **bit 5**
  - **LCxG2D3T**: Gate 2 Data 3 True (non-inverted) bit
    - 1 = LCxD3T is gated into LCxG2
    - 0 = LCxD3T is not gated into LCxG2

- **bit 4**
  - **LCxG2D3N**: Gate 2 Data 3 Negated (inverted) bit
    - 1 = LCxD3N is gated into LCxG2
    - 0 = LCxD3N is not gated into LCxG2

- **bit 3**
  - **LCxG2D2T**: Gate 2 Data 2 True (non-inverted) bit
    - 1 = LCxD2T is gated into LCxG2
    - 0 = LCxD2T is not gated into LCxG2

- **bit 2**
  - **LCxG2D2N**: Gate 2 Data 2 Negated (inverted) bit
    - 1 = LCxD2N is gated into LCxG2
    - 0 = LCxD2N is not gated into LCxG2

- **bit 1**
  - **LCxG2D1T**: Gate 2 Data 1 True (non-inverted) bit
    - 1 = LCxD1T is gated into LCxG2
    - 0 = LCxD1T is not gated into LCxG2

- **bit 0**
  - **LCxG2D1N**: Gate 2 Data 1 Negated (inverted) bit
    - 1 = LCxD1N is gated into LCxG2
    - 0 = LCxD1N is not gated into LCxG2
### REGISTER 19-9: CLCxGLS2: GATE 3 LOGIC SELECT REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>LCxG3D4T</td>
<td>LCxG3D4N</td>
<td>LCxG3D3T</td>
<td>LCxG3D3N</td>
<td>LCxG3D2T</td>
<td>LCxG3D2N</td>
<td>LCxG3D1T</td>
<td>LCxG3D1N</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7**
- **LCxG3D4T**: Gate 3 Data 4 True (non-inverted) bit
  - 1 = LCxD4T is gated into LCxG3
  - 0 = LCxD4T is not gated into LCxG3

**bit 6**
- **LCxG3D4N**: Gate 3 Data 4 Negated (inverted) bit
  - 1 = LCxD4N is gated into LCxG3
  - 0 = LCxD4N is not gated into LCxG3

**bit 5**
- **LCxG3D3T**: Gate 3 Data 3 True (non-inverted) bit
  - 1 = LCxD3T is gated into LCxG3
  - 0 = LCxD3T is not gated into LCxG3

**bit 4**
- **LCxG3D3N**: Gate 3 Data 3 Negated (inverted) bit
  - 1 = LCxD3N is gated into LCxG3
  - 0 = LCxD3N is not gated into LCxG3

**bit 3**
- **LCxG3D2T**: Gate 3 Data 2 True (non-inverted) bit
  - 1 = LCxD2T is gated into LCxG3
  - 0 = LCxD2T is not gated into LCxG3

**bit 2**
- **LCxG3D2N**: Gate 3 Data 2 Negated (inverted) bit
  - 1 = LCxD2N is gated into LCxG3
  - 0 = LCxD2N is not gated into LCxG3

**bit 1**
- **LCxG3D1T**: Gate 3 Data 1 True (non-inverted) bit
  - 1 = LCxD1T is gated into LCxG3
  - 0 = LCxD1T is not gated into LCxG3

**bit 0**
- **LCxG3D1N**: Gate 3 Data 1 Negated (inverted) bit
  - 1 = LCxD1N is gated into LCxG3
  - 0 = LCxD1N is not gated into LCxG3
### REGISTER 19-10: CLCxGLS3: GATE 4 LOGIC SELECT REGISTER

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<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
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<tbody>
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<td>LCxG4D3N</td>
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<td>LCxG4D2N</td>
<td>LCxG4D1T</td>
<td>LCxG4D1N</td>
<td></td>
</tr>
</tbody>
</table>

#### Legend:
- **R** = Readable bit
- **W** =Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

- **bit 7**
  - **LCxG4D4T**: Gate 4 Data 4 True (non-inverted) bit
    - **1** = LCxD4T is gated into LCxG4
    - **0** = LCxD4T is not gated into LCxG4

- **bit 6**
  - **LCxG4D4N**: Gate 4 Data 4 Negated (inverted) bit
    - **1** = LCxD4N is gated into LCxG4
    - **0** = LCxD4N is not gated into LCxG4

- **bit 5**
  - **LCxG4D3T**: Gate 4 Data 3 True (non-inverted) bit
    - **1** = LCxD3T is gated into LCxG4
    - **0** = LCxD3T is not gated into LCxG4

- **bit 4**
  - **LCxG4D3N**: Gate 4 Data 3 Negated (inverted) bit
    - **1** = LCxD3N is gated into LCxG4
    - **0** = LCxD3N is not gated into LCxG4

- **bit 3**
  - **LCxG4D2T**: Gate 4 Data 2 True (non-inverted) bit
    - **1** = LCxD2T is gated into LCxG4
    - **0** = LCxD2T is not gated into LCxG4

- **bit 2**
  - **LCxG4D2N**: Gate 4 Data 2 Negated (inverted) bit
    - **1** = LCxD2N is gated into LCxG4
    - **0** = LCxD2N is not gated into LCxG4

- **bit 1**
  - **LCxG4D1T**: Gate 4 Data 1 True (non-inverted) bit
    - **1** = LCxD1T is gated into LCxG4
    - **0** = LCxD1T is not gated into LCxG4

- **bit 0**
  - **LCxG4D1N**: Gate 4 Data 1 Negated (inverted) bit
    - **1** = LCxD1N is gated into LCxG4
    - **0** = LCxD1N is not gated into LCxG4
REGISTER 19-11: CLCDATA: CLC DATA OUTPUT

| bit 7-4 | Unimplemented: Read as '0' |
| bit 3  | MCL4OUT: Mirror copy of LC4OUT bit |
| bit 2  | MLC3OUT: Mirror copy of LC3OUT bit |
| bit 1  | MLC2OUT: Mirror copy of LC2OUT bit |
| bit 0  | MLC1OUT: Mirror copy of LC1OUT bit |

Legend:
- \( R \) = Readable bit
- \( W \) = Writable bit
- \( U \) = Unimplemented bit, read as '0'
- \( u \) = Bit is unchanged
- \( x \) = Bit is unknown
- \( -n/n \) = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

TABLE 19-3: SUMMARY OF REGISTERS ASSOCIATED WITH CLCx

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<tr>
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<th>Bit7</th>
<th>Bit6</th>
<th>Bit5</th>
<th>Bit4</th>
<th>Bit3</th>
<th>Bit2</th>
<th>Bit1</th>
<th>Bit0</th>
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</tbody>
</table>

Legend: — = unimplemented read as '0'. Shaded cells are not used for CLC module.

Note 1: PIC16(L)F1717/9 only.
### TABLE 19-3: SUMMARY OF REGISTERS ASSOCIATED WITH CLCx (CONTINUED)

<table>
<thead>
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<th>Name</th>
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<th>Bit6</th>
<th>Bit5</th>
<th>Bit4</th>
<th>Bit3</th>
<th>Bit2</th>
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<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
<td>141</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented read as ‘0’. Shaded cells are not used for CLC module.

**Note 1:** PIC16(L)F1717/9 only.
20.0 NUMERICALLY CONTROLLED OSCILLATOR (NCO) MODULE

The Numerically Controlled Oscillator (NCOx) module is a timer that uses the overflow from the addition of an increment value to divide the input frequency. The advantage of the addition method over simple counter driven timer is that the resolution of division does not vary with the divider value. The NCOx is most useful for applications that require frequency accuracy and fine resolution at a fixed duty cycle.

Features of the NCOx include:
- 20-bit increment function
- Fixed Duty Cycle (FDC) mode
- Pulse Frequency (PF) mode
- Output pulse width control
- Multiple clock input sources
- Output polarity control
- Interrupt capability

Figure 20-1 is a simplified block diagram of the NCOx module.

20.1 NCOx Operation

The NCOx operates by repeatedly adding a fixed value to an accumulator. Additions occur at the input clock rate. The accumulator will overflow with a carry periodically, which is the raw NCOx output (NCO_overflow). This effectively reduces the input clock by the ratio of the addition value to the maximum accumulator value. See Equation 20-1.

The NCOx output can be further modified by stretching the pulse or toggling a flip-flop. The modified NCOx output is then distributed internally to other peripherals and optionally output to a pin. The accumulator overflow also generates an interrupt (NCO_interrupt).

The NCOx period changes in discrete steps to create an average frequency. This output depends on the ability of the receiving circuit (i.e., CWG or external resonant converter circuitry) to average the NCOx output to reduce uncertainty.

20.1.1 NCOx CLOCK SOURCES

Clock sources available to the NCOx include:
- HFINTOSC
- Fosc
- LC3_out

The NCOx clock source is selected by configuring the NxCKS<2:0> bits in the NCOxCLK register.

20.1.2 ACCUMULATOR

The accumulator is a 20-bit register. Read and write access to the accumulator is available through three registers:
- NCOxACCL
- NCOxACCH
- NCOxACCU

20.1.3 ADDER

The NCOx adder is a full adder, which operates independently from the system clock. The addition of the previous result and the increment value replaces the accumulator value on the rising edge of each input clock.

20.1.4 INCREMENT REGISTERS

The increment value is stored in three registers making up a 20-bit increment. In order of LSB to MSB they are:
- NCOxINCL
- NCOxINCH
- NCOxINCU

When the NCO module is enabled, the NCOxINCU and NCOxINCH registers should be written first, then the NCOxINCL register. Writing to the NCOxINCL register initiates the increment buffer registers to be loaded simultaneously on the second rising edge of the NCOx_clk signal.

The registers are readable and writable. The increment registers are double-buffered to allow value changes to be made without first disabling the NCOx module.

When the NCO module is disabled, the increment buffers are loaded immediately after a write to the increment registers.

Note: The increment buffer registers are not user-accessible.

EQUATION 20-1:

\[ \text{Overflow} = \frac{\text{NCO Clock Frequency} \times \text{Increment Value}}{2^n} \]

\[ n = \text{Accumulator width in bits} \]
Note 1: The increment registers are double-buffered to allow for value changes to be made without first disabling the NCO module. The full increment value is loaded into the buffer registers on the second rising edge of the NCOx_clk signal that occurs immediately after a write to NCOxNCL register. The buffers are not user-accessible and are shown here for reference.
20.2 Fixed Duty Cycle (FDC) Mode

In Fixed Duty Cycle (FDC) mode, every time the accumulator overflows (NCO_overflow), the output is toggled. This provides a 50% duty cycle, provided that the increment value remains constant. For more information, see Figure 20-2.

The FDC mode is selected by clearing the NxPFM bit in the NCOxCON register.

20.3 Pulse Frequency (PF) Mode

In Pulse Frequency (PF) mode, every time the accumulator overflows (NCO_overflow), the output becomes active for one or more clock periods. Once the clock period expires, the output returns to an inactive state. This provides a pulsed output.

The output becomes active on the rising clock edge immediately following the overflow event. For more information, see Figure 20-2.

The value of the active and inactive states depends on the polarity bit, NxPOL in the NCOxCON register.

The PF mode is selected by setting the NxPFM bit in the NCOxCON register.

20.3.1 OUTPUT PULSE WIDTH CONTROL

When operating in PF mode, the active state of the output can vary in width by multiple clock periods. Various pulse widths are selected with the NxPWS<2:0> bits in the NCOxCLK register.

When the selected pulse width is greater than the accumulator overflow time frame, the output of the NCOx operation is indeterminate.

20.4 Output Polarity Control

The last stage in the NCOx module is the output polarity. The NxPOL bit in the NCOxCON register selects the output polarity. Changing the polarity while the interrupts are enabled will cause an interrupt for the resulting output transition.

The NCOx output can be used internally by source code or other peripherals. Accomplish this by reading the NxOUT (read-only) bit of the NCOxCON register.

The NCOx output signal is available to the following peripherals:

- CLC
- CWG

20.5 Interrupts

When the accumulator overflows (NCO_overflow), the NCOx Interrupt Flag bit, NCOxF, of the PIRx register is set. To enable the interrupt event (NCO_interrupt), the following bits must be set:

- NxEN bit of the NCOxCON register
- NCOxIE bit of the PIEx register
- PEIE bit of the INTCON register
- GIE bit of the INTCON register

The interrupt must be cleared by software by clearing the NCOxF bit in the Interrupt Service Routine.

20.6 Effects of a Reset

All of the NCOx registers are cleared to zero as the result of a Reset.

20.7 Operation in Sleep

The NCO module operates independently from the system clock and will continue to run during Sleep, provided that the clock source selected remains active.

The HFINTOSC remains active during Sleep when the NCO module is enabled and the HFINTOSC is selected as the clock source, regardless of the system clock source selected.

In other words, if the HFINTOSC is simultaneously selected as the system clock and the NCO clock source, when the NCO is enabled, the CPU will go idle during Sleep, but the NCO will continue to operate and the HFINTOSC will remain active.

This will have a direct effect on the Sleep mode current.
FIGURE 20-2: NCO – FIXED DUTY CYCLE (FDC) AND PULSE FREQUENCY MODE (PFM) OUTPUT OPERATION DIAGRAM

- NCOx Clock Source
- NCOx Increment Value
  - 4000h
- NCOx Accumulator Value
  - 0000h
  - 04000h
  - 08000h
  - FC000h
  - 0000h
  - 04000h
  - 08000h
- NCO_overflow
- NCO_interrupt
- NCOx Output FDC Mode
- NCOx Output PF Mode
  - NCOxPWS = 000
  - NCOxPWS = 001
20.8 Register Definitions: NCOx Control Registers

REGISTER 20-1: NCOxCON: NCOx CONTROL REGISTER

<table>
<thead>
<tr>
<th></th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>NxEN</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>NxOUT</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>NxPOL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>NxPFM</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

Legend:

- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

bit 7  **NxEN**: NCOx Enable bit
- 1 = NCOx module is enabled
- 0 = NCOx module is disabled

bit 6  **Unimplemented**: Read as ‘0’

bit 5  **NxOUT**: NCOx Output bit
- 1 = NCOx output is high
- 0 = NCOx output is low

bit 4  **NxPOL**: NCOx Polarity bit
- 1 = NCOx output signal is active-low (inverted)
- 0 = NCOx output signal is active-high (non-inverted)

bit 3-1  **Unimplemented**: Read as ‘0’

bit 0  **NxPFM**: NCOx Pulse Frequency Mode bit
- 1 = NCOx operates in Pulse Frequency mode
- 0 = NCOx operates in Fixed Duty Cycle mode
REGISTER 20-2: NCOxCLK: NCOx INPUT CLOCK CONTROL REGISTER

<table>
<thead>
<tr>
<th>Legend:</th>
</tr>
</thead>
<tbody>
<tr>
<td>R = Readable bit</td>
</tr>
<tr>
<td>W = Writable bit</td>
</tr>
<tr>
<td>U = Unimplemented bit, read as '0'</td>
</tr>
<tr>
<td>u = Bit is unchanged</td>
</tr>
<tr>
<td>x = Bit is unknown</td>
</tr>
<tr>
<td>-n/n = Value at POR and BOR/Value at all other Resets</td>
</tr>
<tr>
<td>'1' = Bit is set</td>
</tr>
<tr>
<td>'0' = Bit is cleared</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 7-5</th>
<th>NxPWS&lt;2:0&gt;: NCOx Output Pulse Width Select bits**(1, 2)**</th>
</tr>
</thead>
<tbody>
<tr>
<td>111</td>
<td>128 NCOx clock periods</td>
</tr>
<tr>
<td>110</td>
<td>64 NCOx clock periods</td>
</tr>
<tr>
<td>101</td>
<td>32 NCOx clock periods</td>
</tr>
<tr>
<td>100</td>
<td>16 NCOx clock periods</td>
</tr>
<tr>
<td>011</td>
<td>8 NCOx clock periods</td>
</tr>
<tr>
<td>010</td>
<td>4 NCOx clock periods</td>
</tr>
<tr>
<td>001</td>
<td>2 NCOx clock periods</td>
</tr>
<tr>
<td>000</td>
<td>1 NCOx clock periods</td>
</tr>
</tbody>
</table>

bit 4-2 | Unimplemented: Read as '0' |

bit 1-0 | NxCKS<1:0>: NCOx Clock Source Select bits |
<table>
<thead>
<tr>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>11</td>
<td>Reserved</td>
</tr>
<tr>
<td>10</td>
<td>LC3_out</td>
</tr>
<tr>
<td>01</td>
<td>Fosc</td>
</tr>
<tr>
<td>00</td>
<td>HFINTOSC (16 MHz)</td>
</tr>
</tbody>
</table>

Note 1: NxPWS applies only when operating in Pulse Frequency mode.

2: If NCOx pulse width is greater than NCO_overflow period, operation is undeterminate.

REGISTER 20-3: NCOxACCL: NCOx ACCUMULATOR REGISTER – LOW BYTE

<table>
<thead>
<tr>
<th>Legend:</th>
</tr>
</thead>
<tbody>
<tr>
<td>R = Readable bit</td>
</tr>
<tr>
<td>W = Writable bit</td>
</tr>
<tr>
<td>U = Unimplemented bit, read as '0'</td>
</tr>
<tr>
<td>u = Bit is unchanged</td>
</tr>
<tr>
<td>x = Bit is unknown</td>
</tr>
<tr>
<td>-n/n = Value at POR and BOR/Value at all other Resets</td>
</tr>
<tr>
<td>'1' = Bit is set</td>
</tr>
<tr>
<td>'0' = Bit is cleared</td>
</tr>
</tbody>
</table>

| bit 7-0 | NCOxACCL<7:0>: NCOx Accumulator, Low Byte |
REGISTER 20-4: NCOxACCH: NCOx ACCUMULATOR REGISTER – HIGH BYTE

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>NCOxAC&lt;15:8&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0  NCOxAC<15:8>: NCOx Accumulator, High Byte

REGISTER 20-5: NCOxACCU: NCOx ACCUMULATOR REGISTER – UPPER BYTE

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>NCOxAC&lt;19:16&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-4  Unimplemented: Read as ‘0’

bit 3-0  NCOxAC<19:16>: NCOx Accumulator, Upper Byte

REGISTER 20-6: NCOxINCL: NCOx INCREMENT REGISTER – LOW BYTE

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>NCOxINC&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-0  NCOxINC<7:0>: NCOx Increment, Low Byte

Note 1: Write the NCOxINCH register first, then the NCOxINCL register. See Section 20.1.4 “Increment Registers” for more information.
REGISTER 20-7:  NCOxINCH: NCOx INCREMENT REGISTER – HIGH BYTE\(^{(1)}\)

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>NCOxINC&lt;15:8&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- \( -n/n \) = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

bit 7-0  \( \text{NCOx INC}<15:8> \): NCOx Increment, High Byte

**Note 1:** Write the NCOxINCH register first, then the NCOxINCL register. See Section 20.1.4 “Increment Registers” for more information.

REGISTER 20-8:  NCOxINCU: NCOx INCREMENT REGISTER – UPPER BYTE\(^{(1)}\)

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>( \text{NCOx INC}&lt;19:16&gt; )</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- \( -n/n \) = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

bit 7-4  \( \text{Unimplemented} \): Read as ‘0’

bit 3-0  \( \text{NCOx INC}<19:16> \): NCOx Increment, Upper Byte

**Note 1:** Write the NCOxINCU register first, then the NCOxINCL register. See Section 20.1.4 “Increment Registers” for more information.
## TABLE 20-1: SUMMARY OF REGISTERS ASSOCIATED WITH NCOx

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCF</td>
<td>91</td>
</tr>
<tr>
<td>NCO1ACCU</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCO1ACC&lt;19:16&gt;</td>
<td></td>
<td></td>
<td>239</td>
</tr>
<tr>
<td>NCO1ACCH</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCO1ACC&lt;15:8&gt;</td>
<td></td>
<td></td>
<td>239</td>
</tr>
<tr>
<td>NCO1ACCL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCO1ACC&lt;7:0&gt;</td>
<td></td>
<td></td>
<td>238</td>
</tr>
<tr>
<td>NCO1CLK</td>
<td>N1PWS&lt;2:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>N1CKS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td>238</td>
</tr>
<tr>
<td>NCO1CON</td>
<td>N1EN</td>
<td>—</td>
<td>N1OUT</td>
<td>N1POL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>N1PFM</td>
<td>237</td>
</tr>
<tr>
<td>NCO1INCU</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCO1INC&lt;19:16&gt;</td>
<td></td>
<td></td>
<td>240</td>
</tr>
<tr>
<td>NCO1INCH</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCO1INC&lt;15:8&gt;</td>
<td></td>
<td></td>
<td>240</td>
</tr>
<tr>
<td>NCO1INCL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCO1INC&lt;7:0&gt;</td>
<td></td>
<td></td>
<td>239</td>
</tr>
<tr>
<td>PIE3</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCOIE</td>
<td>COGIE</td>
<td>ZCDIE</td>
<td>94</td>
</tr>
<tr>
<td>PIR3</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>NCOIF</td>
<td>COGIF</td>
<td>ZCDIF</td>
<td>97</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TRISD(1)</td>
<td>TRISD7</td>
<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
<td>141</td>
</tr>
<tr>
<td>RxyPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RxyPPS&lt;4:0&gt;</td>
<td></td>
<td></td>
<td>154</td>
</tr>
</tbody>
</table>

**Legend:**  
- x = unknown,  
- u = unchanged,  
- – = unimplemented read as ‘0’,  
- q = value depends on condition.  
- Shaded cells are not used for NCOx module.

**Note 1:** PIC16(L)F1717/9 only.
21.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE

The Analog-to-Digital Converter (ADC) allows conversion of an analog input signal to a 10-bit binary representation of that signal. This device uses analog inputs, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates a 10-bit binary result via successive approximation and stores the conversion result into the ADC result registers (ADRESH:ADRESL register pair). Figure 21-1 shows the block diagram of the ADC.

The ADC voltage reference is software selectable to be either internally generated or externally supplied.

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.

FIGURE 21-1: ADC BLOCK DIAGRAM

Note 1: When ADON = 0, all multiplexer inputs are disconnected.
2: PIC16(L)F1717/9 only.
21.1 ADC Configuration

When configuring and using the ADC the following functions must be considered:

• Port configuration
• Channel selection
• ADC voltage reference selection
• ADC conversion clock source
• Interrupt control
• Result formatting

21.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin should be configured for analog by setting the associated TRIS and ANSEL bits. Refer to Section 11.0 “I/O Ports” for more information.

21.1.2 CHANNEL SELECTION

There are up to 21 channel selections available on the PIC16(L)F1718 and 32 channel selections on the PIC16(L)F1717/9:

• AN<19:8, 4:0> pins (PIC16(L)F1718)
• AN<27:0> pins (PIC16(L)F1717/9)
• Temperature Indicator
• DAC_output
• FVR_buffer1
• FVR_buffer2

The CHS bits of the ADCON0 register (Register 21-1) determine which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion. Refer to Section 21.2 “ADC Operation” for more information.

21.1.3 ADC VOLTAGE REFERENCE

The ADPREF bits of the ADCON1 register provides control of the positive voltage reference. The positive voltage reference can be:

• VREF+ pin
• VDD
• FVR 2.048V
• FVR 4.096V (Not available on LF devices)

The ADNREF bit of the ADCON1 register provides control of the negative voltage reference. The negative voltage reference can be:

• VREF- pin
• VSS


21.1.4 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCS bits of the ADCON1 register. There are seven possible clock options:

• FOSC/2
• FOSC/4
• FOSC/8
• FOSC/16
• FOSC/32
• FOSC/64
• FRC (internal RC oscillator)

The time to complete one bit conversion is defined as TAD. One full 10-bit conversion requires 11.5 TAD periods as shown in Figure 21-2.

For correct conversion, the appropriate TAD specification must be met. Refer to Table 34-16: ADC Conversion Requirements for more information. Table 21-1 gives examples of appropriate ADC clock selections.

Note: Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.
### TABLE 21-1: ADC CLOCK PERIOD (TAD) VS. DEVICE OPERATING FREQUENCIES

<table>
<thead>
<tr>
<th>ADC Clock Period (TAD)</th>
<th>32 MHz</th>
<th>20 MHz</th>
<th>16 MHz</th>
<th>8 MHz</th>
<th>4 MHz</th>
<th>1 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADC Clock Source</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Fosc/2</td>
<td>000</td>
<td>62.5 ns(2)</td>
<td>100 ns(2)</td>
<td>125 ns(2)</td>
<td>250 ns(2)</td>
<td>500 ns(2)</td>
</tr>
<tr>
<td>Fosc/4</td>
<td>100</td>
<td>125 ns(2)</td>
<td>200 ns(2)</td>
<td>250 ns(2)</td>
<td>500 ns(2)</td>
<td>1.0 µs</td>
</tr>
<tr>
<td>Fosc/8</td>
<td>001</td>
<td>0.5 µs(2)</td>
<td>400 ns(2)</td>
<td>0.5 µs(2)</td>
<td>1.0 µs</td>
<td>2.0 µs</td>
</tr>
<tr>
<td>Fosc/16</td>
<td>101</td>
<td>800 ns</td>
<td>800 ns</td>
<td>1.0 µs</td>
<td>2.0 µs</td>
<td>4.0 µs</td>
</tr>
<tr>
<td>Fosc/32</td>
<td>010</td>
<td>1.0 µs</td>
<td>1.6 µs</td>
<td>2.0 µs</td>
<td>4.0 µs</td>
<td>8.0 µs(3)</td>
</tr>
<tr>
<td>Fosc/64</td>
<td>110</td>
<td>2.0 µs</td>
<td>3.2 µs</td>
<td>4.0 µs</td>
<td>8.0 µs(3)</td>
<td>16.0 µs(2)</td>
</tr>
<tr>
<td>FRC</td>
<td>x11</td>
<td>1.0-6.0 µs(1,4)</td>
<td>1.0-6.0 µs(1,4)</td>
<td>1.0-6.0 µs(1,4)</td>
<td>1.0-6.0 µs(1,4)</td>
<td>1.0-6.0 µs(1,4)</td>
</tr>
</tbody>
</table>

**Legend:** Shaded cells are outside of recommended range.

**Note 1:** See TAD parameter for FRC source typical TAD value.

**Note 2:** These values violate the required TAD time.

**Note 3:** Outside the recommended TAD time.

**Note 4:** The ADC clock period (TAD) and total ADC conversion time can be minimized when the ADC clock is derived from the system clock Fosc. However, the FRC oscillator source must be used when conversions are to be performed with the device in Sleep mode.

### FIGURE 21-2: ANALOG-TO-DIGITAL CONVERSION TAD CYCLES

- **TAD1**
- **TAD2**
- **TAD3**
- **TAD4**
- **TAD5**
- **TAD6**
- **TAD7**
- **TAD8**
- **TAD9**
- **TAD10**
- **TAD11**

- **TAD1**
- **TAD2**
- **TAD3**
- **TAD4**
- **TAD5**
- **TAD6**
- **TAD7**
- **TAD8**
- **TAD9**
- **TAD10**
- **TAD11**

- **THCD**
- **Conversion Starts**
- **Holding capacitor disconnected from analog input (THCD).**
- **Set GO bit**
- **Enable ADC (ADON bit) and Select channel (ACS bits)**
- **On the following cycle:**
  - **ADRESH:ADRESL** is loaded,
  - **GO bit** is cleared,
  - **ADIF** bit is set,
  - **holding capacitor** is reconnected to analog input
21.1.5 INTERRUPTS

The ADC module allows for the ability to generate an interrupt upon completion of an Analog-to-Digital conversion. The ADC Interrupt Flag is the ADIF bit in the PIR1 register. The ADC Interrupt Enable is the ADIE bit in the PIE1 register. The ADIF bit must be cleared in software.

**Note 1:** The ADIF bit is set at the completion of every conversion, regardless of whether or not the ADC interrupt is enabled.

**Note 2:** The ADC operates during Sleep only when the FRC oscillator is selected.

This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the ADIE bit of the PIE1 register and the PEIE bit of the INTCON register must both be set and the GIE bit of the INTCON register must be cleared. If all three of these bits are set, the execution will switch to the Interrupt Service Routine.

21.1.6 RESULT FORMATTING

The 10-bit ADC conversion result can be supplied in two formats, left justified or right justified. The ADFM bit of the ADCON1 register controls the output format. Figure 21-3 shows the two output formats.

**FIGURE 21-3: 10-BIT ADC CONVERSION RESULT FORMAT**

<table>
<thead>
<tr>
<th>ADFM = 0</th>
<th>ADRESH</th>
<th>ADRESL</th>
</tr>
</thead>
<tbody>
<tr>
<td>MSB</td>
<td>bit 7</td>
<td>LSB</td>
</tr>
</tbody>
</table>

10-bit ADC Result

<table>
<thead>
<tr>
<th>ADFM = 1</th>
</tr>
</thead>
<tbody>
<tr>
<td>MSB</td>
</tr>
</tbody>
</table>

Unimplemented: Read as '0'

| ADRESH |
|---|---|
| MSB | bit 7 |
21.2  ADC Operation

21.2.1  STARTING A CONVERSION
To enable the ADC module, the ADON bit of the ADCON0 register must be set to a ‘1’. Setting the GO/DONE bit of the ADCON0 register to a ‘1’ will start the Analog-to-Digital conversion.

Note: The GO/DONE bit should not be set in the same instruction that turns on the ADC. Refer to Section 21.2.6 “ADC Conversion Procedure”.

21.2.2  COMPLETION OF A CONVERSION
When the conversion is complete, the ADC module will:

• Clear the GO/DONE bit
• Set the ADIF Interrupt Flag bit
• Update the ADRESH and ADRESL registers with new conversion result

21.2.3  TERMINATING A CONVERSION
If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRESH and ADRESL registers will be updated with the partially complete Analog-to-Digital conversion sample. Incomplete bits will match the last bit converted.

Note: A device Reset forces all registers to their Reset state. Thus, the ADC module is turned off and any pending conversion is terminated.

21.2.4  ADC OPERATION DURING SLEEP
The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC oscillator source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

21.2.5  AUTO-CONVERSION TRIGGER
The auto-conversion trigger allows periodic ADC measurements without software intervention. When a rising edge of the selected source occurs, the GO/DONE bit is set by hardware. The auto-conversion trigger source is selected with the TRIGSEL<3:0> bits of the ADCON2 register.

Using the auto-conversion trigger does not assure proper ADC timing. It is the user’s responsibility to ensure that the ADC timing requirements are met. See Table 21-2 for auto-conversion sources.

**TABLE 21-2: AUTO-CONVERSION SOURCES**

<table>
<thead>
<tr>
<th>Source Peripheral</th>
<th>Signal Name</th>
</tr>
</thead>
<tbody>
<tr>
<td>CCP1</td>
<td>—</td>
</tr>
<tr>
<td>CCP2</td>
<td>—</td>
</tr>
<tr>
<td>Timer0</td>
<td>T0_overflow</td>
</tr>
<tr>
<td>Timer1</td>
<td>T1_overflow</td>
</tr>
<tr>
<td>Timer2</td>
<td>T2_match</td>
</tr>
<tr>
<td>Timer4</td>
<td>T4_match</td>
</tr>
<tr>
<td>Timer6</td>
<td>T6_match</td>
</tr>
<tr>
<td>Comparator C1</td>
<td>sync_C1OUT</td>
</tr>
<tr>
<td>Comparator C2</td>
<td>sync_C2OUT</td>
</tr>
<tr>
<td>CLC1</td>
<td>LC1_out</td>
</tr>
<tr>
<td>CLC2</td>
<td>LC2_out</td>
</tr>
<tr>
<td>CLC3</td>
<td>LC3_out</td>
</tr>
<tr>
<td>CLC4</td>
<td>LC4_out</td>
</tr>
</tbody>
</table>
21.2.6  ADC CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an Analog-to-Digital conversion:

1. Configure Port:
   • Disable pin output driver (Refer to the TRIS register)
   • Configure pin as analog (Refer to the ANSEL register)
   • Disable weak pull-ups either globally (Refer to the OPTION_REG register) or individually (Refer to the appropriate WPUx register)

2. Configure the ADC module:
   • Select ADC conversion clock
   • Configure voltage reference
   • Select ADC input channel
   • Turn on ADC module

3. Configure ADC interrupt (optional):
   • Clear ADC interrupt flag
   • Enable ADC interrupt
   • Enable peripheral interrupt
   • Enable global interrupt

4. Wait the required acquisition time(2).

5. Start conversion by setting the GO/DONE bit.

6. Wait for ADC conversion to complete by one of the following:
   • Polling the GO/DONE bit
   • Waiting for the ADC interrupt (interrupts enabled)

7. Read ADC Result.

8. Clear the ADC interrupt flag (required if interrupt is enabled).

---

EXAMPLE 21-1:  ADC CONVERSION

This code block configures the ADC for polling, Vdd and Vss references, FRC oscillator and AN0 input.

Conversion start & polling for completion are included.

```
; This code block configures the ADC
; for polling, Vdd and Vss references, FRC oscillator and AN0 input.
; Conversion start & polling for completion are included.
; BANKSEL ADCON1 ;
MOVLW B'11110000' ;Right justify, FRC oscillator
MOVWF ADCON1 ;Vdd and Vss Vref
BANKSEL TRISA ;
BSF TRISA,0 ;Set RA0 to input
BANKSEL ANSEL ;
BSF ANSEL,0 ;Set RA0 to analog
BANKSEL WPUA
BCF WPUA,0 ;Disable weak pull-up on RA0
BANKSEL ADCON0 ;
MOVLW B'00000001' ;Select channel AN0
MOVF ADRESH,W ;Read upper 2 bits
MOVF RESULTHI ;Store in GPR space
BANKSEL ADRESL ;
MOVF ADRESL,W ;Read lower 8 bits
MOVF RESULTLO ;Store in GPR space
```

---

**Note 1:** The global interrupt can be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.

**2:** Refer to **Section 21.4 “ADC Acquisition Requirements”**.
### 21.3 Register Definitions: ADC Control

**REGISTER 21-1: ADCON0: ADC CONTROL REGISTER 0**

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 6-2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CHS&lt;4:0&gt;</td>
<td>GO/DONE</td>
<td>ADON</td>
<td>Unimplemented</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **\(-n/n\)** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

**bit 7**  
Unimplemented: Read as ‘0’

**bit 6-2**  
CHS<4:0>: Analog Channel Select bits
- **11111** = FVR (Fixed Voltage Reference) Buffer 1 Output(2)
- **11110** = DAC1_output(1)
- **11101** = Temperature Indicator(3)
- **11100** = DAC2_output(5)
- **11011** = AN27(4)
- **11010** = AN26(4)
- **11001** = AN25(4)
- **11000** = AN24(4)
- **10111** = AN23(4)
- **10110** = AN22(4)
- **10101** = AN21(4)
- **10100** = AN20(4)
- **10011** = AN19
- **10010** = AN18
- **10001** = AN17
- **10000** = AN16
- **01111** = AN15
- **01110** = AN14
- **01101** = AN13
- **01100** = AN12
- **01011** = AN11
- **01010** = AN10
- **01001** = AN9
- **01000** = AN8
- **00111** = AN7(4)
- **00110** = AN6(4)
- **00101** = AN5(4)
- **00100** = AN4
- **00011** = AN3
- **00010** = AN2
- **00001** = AN1
- **00000** = AN0

**bit 1**  
GO/DONE: ADC Conversion Status bit
- **1** = ADC conversion cycle in progress. Setting this bit starts an ADC conversion cycle.
- This bit is automatically cleared by hardware when the ADC conversion has completed.
- **0** = ADC conversion completed/not in progress

**bit 0**  
ADON: ADC Enable bit
- **1** = ADC is enabled
- **0** = ADC is disabled and consumes no operating current

**Note 1:**  
See Section 23.0 “8-Bit Digital-to-Analog Converter (DAC1) Module” for more information.

**Note 2:**  
See Section 14.0 “Fixed Voltage Reference (FVR)” for more information.

**Note 3:**  
See Section 15.0 “Temperature Indicator Module” for more information.

**Note 4:**  
PIC16(L)F1717/9 only.

**Note 5:**  
See Section 24.0 “5-Bit Digital-to-Analog Converter (DAC2) Module” for more information.
### REGISTER 21-2: ADCON1: ADC CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>bit</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>ADFM: ADC Result Format Select bit</td>
<td>R = Readable bit</td>
</tr>
<tr>
<td>6-4</td>
<td>ADCS&lt;2:0&gt;: ADC Conversion Clock Select bits</td>
<td>W = Writable bit</td>
</tr>
<tr>
<td>3</td>
<td>Unimplemented: Read as ‘0’</td>
<td>U = Unimplemented bit, read as ‘0’</td>
</tr>
<tr>
<td>2</td>
<td>ADNREF: A/D Negative Voltage Reference Configuration bit</td>
<td>x = Bit is unknown</td>
</tr>
<tr>
<td>1-0</td>
<td>ADPREF&lt;1:0&gt;: ADC Positive Voltage Reference Configuration bits</td>
<td>-n/n = Value at POR and BOR/Value at all other Resets</td>
</tr>
</tbody>
</table>

#### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

#### bit 7
- **ADFM**: ADC Result Format Select bit
  - 1 = Right justified. Six Most Significant bits of ADRESH are set to ‘0’ when the conversion result is loaded.
  - 0 = Left justified. Six Least Significant bits of ADRESL are set to ‘0’ when the conversion result is loaded.

#### bit 6-4
- **ADCS<2:0>: ADC Conversion Clock Select bits**
  - 111 = FRC (clock supplied from an internal RC oscillator)
  - 110 = FOSC/64
  - 101 = FOSC/16
  - 100 = FOSC/4
  - 011 = FRC (clock supplied from an internal RC oscillator)
  - 010 = FOSC/32
  - 001 = FOSC/8
  - 000 = FOSC/2

#### bit 3
- **Unimplemented**: Read as ‘0’

#### bit 2
- **ADNREF**: A/D Negative Voltage Reference Configuration bit
  - 1 = VREF- is connected to VREF- pin
  - 0 = VREF- is connected to VSS

#### bit 1-0
- **ADPREF<1:0>: ADC Positive Voltage Reference Configuration bits**
  - 11 = VREF+ is connected to internal Fixed Voltage Reference (FVR) module(1)
  - 10 = VREF+ is connected to external VREF+ pin(1)
  - 01 = Reserved
  - 00 = VREF+ is connected to VDD

**Note 1:** When selecting the VREF+ pin as the source of the positive reference, be aware that a minimum voltage specification exists. See Table 34-16: ADC Conversion Requirements for details.
### REGISTER 21-3: ADCON2: ADC CONTROL REGISTER 2

<table>
<thead>
<tr>
<th>Bit</th>
<th>TRIGSEL&lt;3:0&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>—</td>
</tr>
<tr>
<td>6</td>
<td>—</td>
</tr>
<tr>
<td>5</td>
<td>—</td>
</tr>
<tr>
<td>4</td>
<td>—</td>
</tr>
<tr>
<td>3</td>
<td>—</td>
</tr>
<tr>
<td>2</td>
<td>—</td>
</tr>
<tr>
<td>1</td>
<td>—</td>
</tr>
<tr>
<td>0</td>
<td>—</td>
</tr>
</tbody>
</table>

#### Legend:
- **R**: Readable bit
- **W**: Writable bit
- **U**: Unimplemented bit, read as '0'
- **u**: Bit is unchanged
- **x**: Bit is unknown
- **-n/n**: Value at POR and BOR/Value at all other Resets
- **'1'**: Bit is set
- **'0'**: Bit is cleared

#### bit 7-4

- **TRIGSEL<3:0>: Auto-Conversion Trigger Selection bits**
  - **0000**: No auto-conversion trigger selected
  - **0001**: CCP1
  - **0010**: CCP2
  - **0011**: Timer0 – T0_overflow
  - **0100**: Timer1 – T1_overflow
  - **0101**: Timer2 – T2_match
  - **0110**: Comparator C1 – sync_C1OUT
  - **0111**: Comparator C2 – sync_C2OUT
  - **1000**: CLC1 – LC1_out
  - **1001**: CLC2 – LC2_out
  - **1010**: CLC3 – LC3_out
  - **1011**: CLC4 – LC4_out
  - **1100**: Timer4 – T4_match
  - **1101**: Timer6 – T6_match
  - **1110**: Reserved
  - **1111**: Reserved

#### Note 1:
This is a rising edge sensitive input for all sources.

#### Note 2:
Signal also sets its corresponding interrupt flag.

### REGISTER 21-4: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 0

<table>
<thead>
<tr>
<th>Bit</th>
<th>ADRES&lt;9:2&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>—</td>
</tr>
<tr>
<td>6</td>
<td>—</td>
</tr>
<tr>
<td>5</td>
<td>—</td>
</tr>
<tr>
<td>4</td>
<td>—</td>
</tr>
<tr>
<td>3</td>
<td>—</td>
</tr>
<tr>
<td>2</td>
<td>—</td>
</tr>
<tr>
<td>1</td>
<td>—</td>
</tr>
<tr>
<td>0</td>
<td>—</td>
</tr>
</tbody>
</table>

#### Legend:
- **R**: Readable bit
- **W**: Writable bit
- **U**: Unimplemented bit, read as '0'
- **u**: Bit is unchanged
- **x**: Bit is unknown
- **-n/n**: Value at POR and BOR/Value at all other Resets
- **'1'**: Bit is set
- **'0'**: Bit is cleared

#### bit 7-0

- **ADRES<9:2>: ADC Result Register bits**
  - Upper eight bits of 10-bit conversion result
### REGISTER 21-5: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 0

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Read/Write/Unimplemented</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR / Value at all other Resets

**bit 7-6**
- **ADRES<1:0>: ADC Result Register bits**
  - Lower two bits of 10-bit conversion result

**bit 5-0**
- **Reserved: Do not use.**

### REGISTER 21-6: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 1

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Read/Write/Unimplemented</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR / Value at all other Resets

**bit 7-2**
- **Reserved: Do not use.**

**bit 1-0**
- **ADRES<9:8>: ADC Result Register bits**
  - Upper two bits of 10-bit conversion result

### REGISTER 21-7: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 1

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Read/Write/Unimplemented</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
<td>R/W-x/u</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR / Value at all other Resets

**bit 7-0**
- **ADRES<7:0>: ADC Result Register bits**
  - Lower eight bits of 10-bit conversion result
21.4 ADC Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 21-4. The source impedance (RS) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), refer to Figure 21-4. The maximum recommended impedance for analog sources is 10 kΩ. As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an ADC acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 21-1 may be used. This equation assumes that 1/2 LSB error is used (1,024 steps for the ADC). The 1/2 LSB error is the maximum error allowed for the ADC to meet its specified resolution.

**EQUATION 21-1: ACQUISITION TIME EXAMPLE**

Assumptions:
- Temperature = 50°C and external impedance of 10kΩ 5.0V VDD
- \( T_{ACQ} = \text{Amplifier Settling Time} + \text{Hold Capacitor Charging Time} + \text{Temperature Coefficient} \)
- \( T_{ACQ} = 2\mu s + T_C + T_{COFF} = 2\mu s + T_C + [(\text{Temperature} - 25°C)(0.05\mu s/°C)] \)

The value for \( T_C \) can be approximated with the following equations:

\[
V_{APPLIED}\left(1 - \frac{l}{2^n + 1}\right) = V_{CHOLD} \hspace{1cm} ;[1] \ V_{CHOLD} \text{ charged to within 1/2 lsb}
\]

\[
V_{APPLIED}\left(1 - e^{-\frac{T_C}{RC}}\right) = V_{CHOLD} \hspace{1cm} ;[2] \ V_{CHOLD} \text{ charge response to } V_{APPLIED}
\]

\[
V_{APPLIED}\left(1 - e^{-\frac{T_C}{RC}}\right) = V_{APPLIED}\left(1 - \frac{l}{2^n + 1}\right) \hspace{1cm} \text{combining [1] and [2]}
\]

Note: Where \( n \) = number of bits of the ADC.

Solving for \( T_C \):

\[
T_C = -CHOLD(R_{IC} + RSS + RS) \ln(1/2047)
\]
\[
= -10pF(1kΩ + 7kΩ + 10kΩ) \ln(0.0004885)
\]
\[
= 1.37\mu s
\]

Therefore:

\[
T_{ACQ} = 2\mu s + 892\text{ns} + [(50°C- 25°C)(0.05\mu s/°C)]
\]
\[
= 4.62\mu s
\]

**Note 1:** The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.
FIGURE 21-4: ANALOG INPUT MODEL

Legend:
- **CHOLD** = Sample/Hold Capacitance
- **CPIN** = Input Capacitance
- **I LEAKAGE** = Leakage current at the pin due to various junctions
- **Ric** = Interconnect Resistance
- **RSS** = Resistance of Sampling Switch
- **SS** = Sampling Switch
- **VT** = Threshold Voltage

Note 1: Refer to Table 34-4: I/O Ports (parameter D060).

FIGURE 21-5: ADC TRANSFER FUNCTION

Legend:
- **ADC Output Code**
- **Ref-** = Reference Input Voltage
- **Ref+** = Reference Input Voltage
- **0.5 LSB** = Half of the Least Significant Bit
- **1.5 LSB** = Three-quarters of the Least Significant Bit
- **Full-Scale Range**
- **Zero-Scale Transition**
- **Full-Scale Transition**

Note 1: Refer to Table 34-4: I/O Ports (parameter D060).
**TABLE 21-3: SUMMARY OF REGISTERS ASSOCIATED WITH ADC**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADCON0</td>
<td>—</td>
<td>—</td>
<td>CHS&lt;4:0&gt;</td>
<td>—</td>
<td>—</td>
<td>GO/DONE</td>
<td>ADON</td>
<td>248</td>
<td></td>
</tr>
<tr>
<td>ADCON1</td>
<td>ADFM</td>
<td>—</td>
<td>ADCS&lt;2:0&gt;</td>
<td>—</td>
<td>—</td>
<td>ADNREF</td>
<td>ADPREF&lt;1:0&gt;</td>
<td>249</td>
<td></td>
</tr>
<tr>
<td>ADCON2</td>
<td>—</td>
<td>TRIGSEL&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>250</td>
</tr>
<tr>
<td>ADRESH</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>251</td>
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<tr>
<td>ADRESL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>251</td>
</tr>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSI5</td>
<td>ANSI4</td>
<td>ANSI3</td>
<td>ANSI2</td>
<td>ANSI1</td>
<td>ANSI0</td>
</tr>
<tr>
<td>ANSELC</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
</tr>
<tr>
<td>ANSELD(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSD7</td>
<td>ANSD6</td>
<td>ANSD5</td>
<td>ANSD4</td>
<td>ANSD3</td>
<td>ANSD2</td>
</tr>
<tr>
<td>ANSELE(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSE2</td>
<td>ANSE1</td>
<td>ANSE0</td>
<td>—</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IO CIF</td>
<td>91</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>95</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TRISD(1)</td>
<td>TRISD7</td>
<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
<td>141</td>
</tr>
<tr>
<td>TRISE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TRISE3</td>
<td>TRISE2(1)</td>
<td>TRISE1(1)</td>
<td>TRISE0(1)</td>
</tr>
<tr>
<td>FVRCON</td>
<td>FVREN</td>
<td>FVRDY</td>
<td>TSEN</td>
<td>TSRNG</td>
<td>—</td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>ADFVR&lt;1:0&gt;</td>
<td>—</td>
<td>166</td>
</tr>
<tr>
<td>DAC1CON0</td>
<td>DAC1EN</td>
<td>—</td>
<td>DAC10E1</td>
<td>DAC10E2</td>
<td>DAC1PSS&lt;1:0&gt;</td>
<td>—</td>
<td>DAC1NSS</td>
<td>260</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**  
- x = unknown, u = unchanged, — = unimplemented read as ‘0’, q = value depends on condition. Shaded cells are not used for the ADC module.

**Note 1:**  
PIC16(L)F1717/9 only.
22.0 OPERATIONAL AMPLIFIER (OPA) MODULES

The Operational Amplifier (OPA) is a standard three-terminal device requiring external feedback to operate. The OPA module has the following features:

- External connections to I/O ports
- Low leakage inputs
- Factory Calibrated Input Offset Voltage

Note 1: The OPAxSP bit must be set. Low-Power mode is not supported.
22.1 OPA Module Performance

Common AC and DC performance specifications for the OPA module:

- Common Mode Voltage Range
- Leakage Current
- Input Offset Voltage
- Open Loop Gain
- Gain Bandwidth Product

**Common mode voltage range** is the specified voltage range for the OPA+ and OPA- inputs, for which the OPA module will perform to within its specifications. The OPA module is designed to operate with input voltages between VSS and VDD. Behavior for Common mode voltages greater than VDD, or below VSS, are not guaranteed.

**Leakage current** is a measure of the small source or sink currents on the OPA+ and OPA- inputs. To minimize the effect of leakage currents, the effective impedances connected to the OPA+ and OPA- inputs should be kept as small as possible and equal.

**Input offset voltage** is a measure of the voltage difference between the OPA+ and OPA- inputs in a closed loop circuit with the OPA in its linear region. The offset voltage will appear as a DC offset in the output equal to the input offset voltage, multiplied by the gain of the circuit. The input offset voltage is also affected by the Common mode voltage. The OPA is factory calibrated to minimize the input offset voltage of the module.

**Open loop gain** is the ratio of the output voltage to the differential input voltage, (OPA+) - (OPA-). The gain is greatest at DC and falls off with frequency.

**Gain Bandwidth Product** or GBWP is the frequency at which the open loop gain falls off to 0 dB.

22.1.1 OPA Module Control

The OPA module is enabled by setting the OPAxEN bit of the OPAxCON register. When enabled, the OPA forces the output driver of OPAxOUT pin into tri-state to prevent contention between the driver and the OPA output.

**Note:** When the OPA module is enabled, the OPAxOUT pin is driven by the op amp output, not by the PORT digital driver. Refer to **Table 34-17: Operational Amplifier (OPA)** for the op amp output drive capability.

22.1.2 UNITY GAIN MODE

The OPAxUG bit of the OPAxCON register selects the Unity Gain mode. When unity gain is selected, the OPA output is connected to the inverting input and the OPAxIN pin is relinquished, releasing the pin for general purpose input and output.

22.2 Effects of Reset

A device Reset forces all registers to their Reset state. This disables the OPA module.
22.3 Register Definitions: Op Amp Control

REGISTER 22-1: OPAxCON: OPERATIONAL AMPLIFIERS (OPAx) CONTROL REGISTERS

<table>
<thead>
<tr>
<th>Bit</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>U-0</td>
<td>R/W-0/0</td>
<td>U-0</td>
<td>U-0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td></td>
</tr>
<tr>
<td>OPAxEN</td>
<td>OPAxSP</td>
<td>—</td>
<td>OPAxUG</td>
<td>—</td>
<td>—</td>
<td>OPAxCH&lt;1:0&gt;</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- q = Value depends on condition

bit 7  OPAxEN: Op Amp Enable bit
- 1 = Op amp is enabled
- 0 = Op amp is disabled and consumes no active power

bit 6  OPAxSP: Op Amp Speed/Power Select bit
- 1 = Op amp operates in high GBWP mode
- 0 = Reserved. Do not use.

bit 5  Unimplemented: Read as ‘0’

bit 4  OPAxUG: Op Amp Unity Gain Select bit
- 1 = OPA output is connected to inverting input. OPAxIN- pin is available for general purpose I/O.
- 0 = Inverting input is connected to the OPAxIN- pin

bit 3-2  Unimplemented: Read as ‘0’

bit 1-0  OPAxCH<1:0>: Non-inverting Channel Selection bits
- 11 = Non-inverting input connects to FVR Butter 2 output
- 10 = Non-inverting input connects to DAC1_output
- 01 = Non-inverting input connects to DAC2_output
- 00 = Non-inverting input connects to OPAxIN+ pin

TABLE 22-1: SUMMARY OF REGISTERS ASSOCIATED WITH OP AMPS

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>126</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>DAC1CON0</td>
<td>DAC1EN</td>
<td>—</td>
<td>DAC1OE1</td>
<td>DAC1OE2</td>
<td>DAC1PSS&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>DAC1NSS</td>
<td>260</td>
</tr>
<tr>
<td>DAC1CON1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>260</td>
</tr>
<tr>
<td>FVRCON</td>
<td>FVREN</td>
<td>FVRRDY</td>
<td>TSEN</td>
<td>TSRNG</td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>ADFVR&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>166</td>
</tr>
<tr>
<td>OPA1CON</td>
<td>OPA1EN</td>
<td>OPA1SP</td>
<td>—</td>
<td>OPA1UG</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>257</td>
</tr>
<tr>
<td>OPA2CON</td>
<td>OPA2EN</td>
<td>OPA2SP</td>
<td>—</td>
<td>OPA2UG</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>257</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>125</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>131</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used by op amps.
23.0 8-BIT DIGITAL-TO-ANALOG CONVERTER (DAC1) MODULE

The Digital-to-Analog Converter supplies a variable voltage reference, ratiometric with the input source, with 256 selectable output levels.

The input of the DAC can be connected to:
- External VREF pins
- VDD supply voltage
- FVR (Fixed Voltage Reference)

The output of the DAC can be configured to supply a reference voltage to the following:
- Comparator positive input
- ADC input channel
- DAC1OUT1 pin
- DAC1OUT2 pin

The Digital-to-Analog Converter (DAC) is enabled by setting the DAC1EN bit of the DAC1CON0 register.

23.1 Output Voltage Selection

The DAC has 256 voltage level ranges. The 256 levels are set with the DAC1R<7:0> bits of the DAC1CON1 register.

The DAC output voltage is determined by Equation 23-1:

\[
\text{IF DAC1EN = 1} \\
V_{OUT} = \left( V_{\text{SOURCE}+} - V_{\text{SOURCE}-} \right) \times \frac{\text{DAC1R}[7:0]}{2^8} + V_{\text{SOURCE}-} \\
V_{\text{SOURCE}+} = VDD, VREF, \text{or FVR BUFFER 2} \\
V_{\text{SOURCE}-} = VSS
\]

23.2 Ratiometric Output Level

The DAC output value is derived using a resistor ladder with each end of the ladder tied to a positive and negative voltage reference input source. If the voltage of either input source fluctuates, a similar fluctuation will result in the DAC output value.

The value of the individual resistors within the ladder can be found in Table 34-19: 8-Bit Digital-to-Analog Converter (DAC) Specifications.

23.3 DAC Voltage Reference Output

The DAC voltage can be output to the DAC1OUT1 and DAC1OUT2 pins by setting the respective DAC1OE1 and DAC1OE2 pins of the DAC1CON0 register. Selecting the DAC reference voltage for output on either DAC1OUTx pin automatically overrides the digital output buffer and digital input threshold detector functions of that pin. Reading the DAC1OUTx pin when it has been configured for DAC reference voltage output will always return a '0'.

Due to the limited current drive capability, a buffer must be used on the DAC voltage reference output for external connections to either DAC1OUTx pin. Figure 23-2 shows an example buffering technique.
23.4 Operation During Sleep

The DAC continues to function during Sleep. When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the DAC1CON0 register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

23.5 Effects of a Reset

A device Reset affects the following:
- DAC is disabled
- DAC output voltage is removed from the DAC1OUT pin
- The DAC1R<7:0> range select bits are cleared
23.6 Register Definitions: DAC Control

**REGISTER 23-1: DAC1CON0: DAC1 CONTROL REGISTER 0**

<table>
<thead>
<tr>
<th></th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC1EN</td>
<td>—</td>
<td>—</td>
<td>DAC1OE1</td>
<td>DAC1OE2</td>
<td>DAC1PSS&lt;1:0&gt;</td>
<td>—</td>
<td>DAC1NSS</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- n/n = Value at POR and BOR/Value at all other Resets

<table>
<thead>
<tr>
<th>bit 7</th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC1EN: DAC1 Enable bit</td>
<td></td>
<td></td>
</tr>
<tr>
<td>1 = DAC is enabled</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0 = DAC is disabled</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 6</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Unimplemented: Read as ‘0’</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 5</td>
<td></td>
<td></td>
</tr>
<tr>
<td>DAC1OE1: DAC1 Voltage Output 1 Enable bit</td>
<td></td>
<td></td>
</tr>
<tr>
<td>1 = DAC voltage level is also an output on the DAC1OUT1 pin</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0 = DAC voltage level is disconnected from the DAC1OUT1 pin</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 4</td>
<td></td>
<td></td>
</tr>
<tr>
<td>DAC1OE2: DAC1 Voltage Output 2 Enable bit</td>
<td></td>
<td></td>
</tr>
<tr>
<td>1 = DAC voltage level is also an output on the DAC1OUT2 pin</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0 = DAC voltage level is disconnected from the DAC1OUT2 pin</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 3-2</td>
<td></td>
<td></td>
</tr>
<tr>
<td>DAC1PSS&lt;1:0&gt;: DAC1 Positive Source Select bits</td>
<td></td>
<td></td>
</tr>
<tr>
<td>11 = Reserved, do not use</td>
<td></td>
<td></td>
</tr>
<tr>
<td>10 = FVR Buffer2 output</td>
<td></td>
<td></td>
</tr>
<tr>
<td>01 = VREF+ pin</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00 = VDD</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 1</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Unimplemented: Read as ‘0’</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 0</td>
<td></td>
<td></td>
</tr>
<tr>
<td>DAC1NSS: DAC1 Negative Source Select bits</td>
<td></td>
<td></td>
</tr>
<tr>
<td>1 = VREF- pin</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0 = VSS</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**REGISTER 23-2: DAC1CON1: DAC1 CONTROL REGISTER 1**

<table>
<thead>
<tr>
<th></th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC1R&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

<table>
<thead>
<tr>
<th>bit 7-0</th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC1R&lt;7:0&gt;: DAC1 Voltage Output Select bits</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**TABLE 23-1: SUMMARY OF REGISTERS ASSOCIATED WITH THE DAC1 MODULE**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC1CON0</td>
<td>DAC1EN</td>
<td>—</td>
<td>DAC1OE1</td>
<td>DAC1OE2</td>
<td>DAC1PSS&lt;1:0&gt;</td>
<td>—</td>
<td>DAC1NSS</td>
<td>260</td>
<td></td>
</tr>
<tr>
<td>DAC1CON1</td>
<td>—</td>
<td>DAC1R&lt;7:0&gt;</td>
<td>—</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>260</td>
</tr>
</tbody>
</table>

Legend: — = Unimplemented location, read as ‘0’. Shaded cells are not used with the DAC module.
24.0 5-BIT DIGITAL-TO-ANALOG CONVERTER (DAC2) MODULE

The Digital-to-Analog Converter supplies a variable voltage reference, ratiometric with the input source, with 32 selectable output levels.

The input of the DAC can be connected to:
- External VREF pins
- VDD supply voltage
- FVR (Fixed Voltage Reference)

The output of the DAC can be configured to supply a reference voltage to the following:
- Comparator positive input
- ADC input channel
- DAC2OUT1/DAC2OUT2 pin
- Comparators
- Op Amps

The Digital-to-Analog Converter (DAC) can be enabled by setting the DAC2EN bit of the DAC2CON0 register.

EQUATION 24-1: DAC OUTPUT VOLTAGE

\[
V_{OUT} = \begin{cases} 
(V_{SOURCE}^+ - V_{SOURCE}^-) \times \frac{DACR[4:0]}{2^5} + V_{SOURCE}^- & \text{IF DACEN} = 1 \\
V_{SOURCE}^+ & \text{IF DACEN} = 0 \text{ and DACLPS} = 1 \text{ and DACR[4:0]} = 11111 \\
V_{SOURCE}^- & \text{IF DACEN} = 0 \text{ and DACLPS} = 0 \text{ and DACR[4:0]} = 00000 \\
V_{SOURCE}^+ = VDD, VREF, or FVR BUFFER 2 \\
V_{SOURCE}^- = VSS 
\end{cases}
\]

24.1 Output Voltage Selection

The DAC has 32 voltage level ranges. The 32 levels are set with the DAC2R<4:0> bits of the DAC2CON1 register.

The DAC output voltage is determined by the following equations:

24.2 Ratiometric Output Level

The DAC output value is derived using a resistor ladder with each end of the ladder tied to a positive and negative voltage reference input source. If the voltage of either input source fluctuates, a similar fluctuation will result in the DAC output value.

The value of the individual resistors within the ladder can be found in Table 34-20: 5-Bit Digital-to-Analog Converter (DAC) Specifications.

24.3 DAC Voltage Reference Output

The DAC can be output to the DAC2OUT1 and DAC2OUT2 pins by setting the DAC2OE1 and DAC2OE2 bits of the DAC2CON0 register to '1'.

Selecting the DAC reference voltage for output on the DACOUTx pin automatically overrides the digital output buffer and digital input threshold detector functions of that pin. Reading the DACOUTx pin when it has been configured for DAC reference voltage output will always return a '0'.

Due to the limited current drive capability, a buffer must be used on the DAC voltage reference output for external connections to either DAC2OUTx pin.

Figure 24-2 shows an example buffering technique.
24.4 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the DAC2CON0 register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

24.5 Effects of a Reset

A device Reset affects the following:

- DAC is disabled
- DAC output voltage is removed from the DAC2OUTx pin
- The DAC2R<4:0> range select bits are cleared
24.6 Register Definitions: DAC Control

REGISTER 24-1: DAC2CON0: VOLTAGE REFERENCE CONTROL REGISTER 0

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC2EN</td>
<td>—</td>
<td>DAC2OE1</td>
<td>DAC2OE2</td>
<td>DAC2PSS&lt;1:0&gt;</td>
<td>—</td>
<td>DAC2NSS</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
W = Writable bit
U = Unimplemented bit, read as ‘0’
u = Bit is unchanged
x = Bit is unknown
‘1’ = Bit is set
‘0’ = Bit is cleared
-n/n = Value at POR and BOR/Value at all other Resets

bit 7    
DAC2EN: DAC2 Enable bit
1 = DAC is enabled
0 = DAC is disabled

bit 6    
Unimplemented: Read as ‘0’

bit 5    
DAC2OE1: DAC2 Voltage Output Enable bit
1 = DAC voltage level is also an output on the DAC2OUT1 pin
0 = DAC voltage level is disconnected from the DAC2OUT1 pin

bit 4    
DAC2OE2: DAC2 Voltage Output Enable bit
1 = DAC voltage level is also an output on the DAC2OUT2 pin
0 = DAC voltage level is disconnected from the DAC2OUT2 pin

bit 3-2  
DAC2PSS<1:0>: DAC2 Positive Source Select bits
11 = Reserved, do not use
10 = FVR Buffer2 output
01 = VREF+ pin
00 = VDD

bit 1    
Unimplemented: Read as ‘0’

bit 0    
DAC2NSS: DAC2 Negative Source Select bits
1 = VREF-
0 = VSS

REGISTER 24-2: DAC2CON1: VOLTAGE REFERENCE CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>DAC2R&lt;4:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
W = Writable bit
U = Unimplemented bit, read as ‘0’
u = Bit is unchanged
x = Bit is unknown
‘1’ = Bit is set
‘0’ = Bit is cleared

bit 7-5  
Unimplemented: Read as ‘0’

bit 4-0  
DAC2R<4:0>: DAC Voltage Output Select bits

TABLE 24-1: SUMMARY OF REGISTERS ASSOCIATED WITH THE DAC2 MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC2CON0</td>
<td>DAC2EN</td>
<td>—</td>
<td>DAC2OE1</td>
<td>DAC2OE2</td>
<td>DAC2PSS&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>DAC2NSS</td>
<td>263</td>
</tr>
<tr>
<td>DAC2CON1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>263</td>
</tr>
</tbody>
</table>

Legend: — = Unimplemented location, read as ‘0’. Shaded cells are not used with the DAC module.
25.0 ZERO-CROSS DETECTION (ZCD) MODULE

The ZCD module detects when an A/C signal crosses through the ground potential. The actual zero-crossing threshold is the zero-crossing reference voltage, ZCPINV, which is typically 0.75V above ground.

The connection to the signal to be detected is through a series current limiting resistor. The module applies a current source or sink to the ZCD pin to maintain a constant voltage on the pin, thereby preventing the pin voltage from forward biasing the ESD protection diodes. When the applied voltage is greater than the reference voltage, the module sinks current. When the applied voltage is less than the reference voltage, the module sources current. The current source and sink action keeps the pin voltage constant over the full range of the applied voltage. The ZCD module is shown in the simplified block diagram Figure 25-2.

The ZCD module is useful when monitoring an AC waveform for, but not limited to, the following purposes:

- A/C period measurement
- Accurate long term time measurement
- Dimmer phase delayed drive
- Low EMI cycle switching

25.1 External Resistor Selection

The ZCD module requires a current limiting resistor in series with the external voltage source. The impedance and rating of this resistor depends on the external source peak voltage. Select a resistor value that will drop all of the peak voltage when the current through the resistor is nominally $300 \mu A$. Refer to Equation 25-1 and Figure 25-1. Make sure that the ZCD I/O pin internal weak pull-up is disabled so it does not interfere with the current source and sink.

\[
R_{series} = \frac{V_{peak}}{3 \times 10^{-4}}
\]

Figure 25-1: External Voltage

Figure 25-2: Simplified ZCD Block Diagram
25.2 ZCD Logic Output

The ZCD module includes a Status bit, which can be read to determine whether the current source or sink is active. The OUT bit of the ZCDCON register is set when the current sink is active, and cleared when the current source is active. The OUT bit is affected by the polarity bit.

25.3 ZCD Logic Polarity

The POL bit of the ZCDxCON register inverts the OUT bit relative to the current source and sink output. When the POL bit is set, a OUT high indicates that the current source is active, and a low output indicates that the current sink is active.

The POL bit affects the ZCD interrupts. See Section 25.4 “ZCD Interrupts”.

25.4 ZCD Interrupts

An interrupt will be generated upon a change in the ZCD logic output when the appropriate interrupt enables are set. A rising edge detector and a falling edge detector are present in the ZCD for this purpose. The ZCDIF bit of the PIR3 register will be set when either edge detector is triggered and its associated enable bit is set. The INTP enables rising edge interrupts and the INTN bit enables falling edge interrupts. Both are located in the ZCDxCON register.

To fully enable the interrupt, the following bits must be set:

- ZCDIE bit of the PIE3 register
- INTP bit of the ZCDxCON register (for a rising edge detection)
- INTN bit of the ZCDxCON register (for a falling edge detection)
- PEIE and GIE bits of the INTCON register

Changing the POL bit will cause an interrupt, regardless of the level of the EN bit.

The ZCDIF bit of the PIR3 register must be cleared in software as part of the interrupt service. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

25.5 Correcting for ZCPINV Offset

The actual voltage at which the ZCD switches is the reference voltage at the noninverting input of the ZCD op amp. For external voltage source waveforms other than square waves this voltage offset from zero causes the zero-cross event to occur either too early or too late.

25.5.1 CORRECTION BY AC COUPLING

When the external voltage source is sinusoidal, the effects of the ZCPINV offset can be eliminated by isolating the external voltage source from the ZCD pin with a capacitor in addition to the voltage reducing resistor. The capacitor will cause a phase shift resulting in the ZCD output switch in advance of the actual zero-crossing event. The phase shift will be the same for both rising and falling zero crossings, which can be compensated for by either delaying the CPU response to the ZCD switch by a timer or other means, or selecting a capacitor value large enough that the phase shift is negligible.

To determine the series resistor and capacitor values for this configuration, start by computing the impedance, $Z$, to obtain a peak current of 300 μA. Next, arbitrarily select a suitably large non-polar capacitor and compute its reactance, $X_C$, at the external voltage source frequency. Finally, compute the series resistor, capacitor peak voltage, and phase shift by the formulas shown in Equation 25-2.
**EQUATION 25-2: R-C CALCULATIONS**

\[ V_{\text{peak}} = \text{external voltage source peak voltage} \]
\[ f = \text{external voltage source frequency} \]
\[ C = \text{series capacitor} \]
\[ R = \text{series resistor} \]
\[ V_C = \text{Peak capacitor voltage} \]
\[ \phi = \text{Capacitor induced zero crossing phase advance in radians} \]
\[ T_\phi = \text{Time ZC event occurs before actual zero crossing} \]

\[ Z = \frac{V_{\text{peak}}}{3 \times 10^4} \]
\[ X_C = \frac{1}{(2\pi f C)} \]
\[ R = \sqrt{Z^2 - X_C^2} \]
\[ V_C = X_C (3 \times 10^{-4}) \]
\[ \phi = Tan^{-1} \left( \frac{X_C}{R} \right) \]
\[ T_\phi = \frac{\phi}{(2\pi f)} \]
\[ V_{\text{rms}} = 120 \]

**EQUATION 25-3: R-C CALCULATIONS EXAMPLE**

\[ V_{\text{peak}} = V_{\text{rms}} \cdot \sqrt{2} = 169.7 \]
\[ f = 60 \text{ Hz} \]
\[ C = 0.1 \mu\text{f} \]

\[ Z = \frac{V_{\text{peak}}}{3 \times 10^{-4}} = \frac{169.7}{3 \times 10^{-4}} = 565.7 \text{ kOhms} \]
\[ X_C = \frac{1}{(2\pi f C)} = \frac{1}{(2\pi \cdot 60 \cdot 1 \cdot 10^{-7})} = 26.53 \text{ kOhms} \]
\[ R = 560 \text{ kOhms} \]

\[ Z_R = \sqrt{(R^2 + X_C^2)} = 560.6 \text{ kOhm (using actual resistor)} \]

\[ I_{\text{peak}} = \frac{V_{\text{peak}}}{Z_R} = 302.7 \times 10^{-6} \]
\[ V_C = X_C \cdot I_{\text{peak}} = 8.0 \text{ V} \]
\[ \phi = Tan^{-1} \left( \frac{X_C}{R} \right) = 0.047 \text{ radians} \]
\[ T_\phi = \frac{\phi}{(2\pi f)} = 125.6 \mu\text{s} \]
25.5.2  CORRECTION BY OFFSET CURRENT

When the waveform is varying relative to Vss then the zero cross is detected too early as the waveform falls and too late as the waveform rises. When the waveform is varying relative to VDD then the zero cross is detected too late as the waveform rises and too early as the waveform falls. The actual offset time can be determined for sinusoidal waveforms with the corresponding equations shown in Equation 25-4.

EQUATION 25-4:  ZCD EVENT OFFSET

When External Voltage Source is relative to Vss:

\[ T_{offset} = \frac{\text{asin} \left( \frac{Z_{cpinv}}{V_{peak}} \right)}{2\pi \cdot \text{Freq}} \]

When External Voltage Source is relative to VDD:

\[ T_{offset} = \frac{\text{asin} \left( \frac{V_{DD} - Z_{cpinv}}{V_{peak}} \right)}{2\pi \cdot \text{Freq}} \]

This offset time can be compensated for by adding a pull-up or pull-down biasing resistor to the ZCD pin. A pull-up resistor is used when the external voltage source is varying relative to Vss. A pull-down resistor is used when the voltage is varying relative to VDD. The resistor adds a bias to the ZCD pin so that the target external voltage source must go to zero to pull the pin voltage to the ZCPINV switching voltage. The pull-up or pull-down value can be determined with the equations shown in Equation 25-5.

EQUATION 25-5:  ZCD PULL-UP/DOWN

When External Signal is relative to Vss:

\[ R_{pullup} = \frac{R_{series} \cdot (V_{pullup} - Z_{cpinv})}{Z_{cpinv}} \]

When External Signal is relative to VDD:

\[ R_{pulldown} = \frac{R_{series} \cdot (Z_{cpinv})}{(V_{DD} - Z_{cpinv})} \]

The pull-up and pull-down resistor values are significantly affected by small variations of ZCPINV. Measuring ZCPINV can be difficult, especially when the waveform is relative to VDD. However, by combining Equation 25-4 and Equation 25-5 the resistor value can be determined from the time difference between the ZCDOUT high and low periods. Note that the time difference, \( \Delta T \), is \( 4 \cdot T_{offset} \). The equation for determining the pull-up and pull-down resistor values from the high and low ZCDOUT periods is shown in Equation 25-6. The ZCDOUT signal can be directly observed on a pin by routing the ZCDOUT signal through one of the CLCs.

EQUATION 25-6:

\[ R = R_{series} \left( \frac{V_{bias}}{V_{peak} \left( \text{sin} \left( \pi \text{Freq} \left( \frac{\Delta T}{2} \right) \right) \right)} - 1 \right) \]

R is pull-up or pull-down resistor

\( V_{bias} \) is \( V_{pullup} \) when R is pull-up or VDD when R is pull-down

\( \Delta T \) is the ZCDOUT high and low period difference

25.6  Handling \( V_{peak} \) variations

If the peak amplitude of the external voltage is expected to vary then the series resistor must be selected to keep the ZCD current source and sink below the design maximum range of \( \pm 600 \mu A \) for the maximum expected voltage and high enough to be detected accurately at the minimum peak voltage. A general rule of thumb is that the maximum peak voltage can be no more than six times the minimum peak voltage. To ensure that the maximum current does not exceed \( \pm 600 \mu A \) and the minimum is at least \( \pm 100 \mu A \), compute the series resistance as shown in Equation 25-7. The compensating pull-up for this series resistance can be determined with Equation 25-5 because the pull-up value is independent from the peak voltage.

EQUATION 25-7:  SERIES R FOR V RANGE

\[ R_{series} = \frac{V_{maxpeak} + V_{minpeak}}{7 \times 10^{-4}} \]
25.7 Operation During Sleep
The ZCD current sources and interrupts are unaffected by Sleep.

25.8 Effects of a Reset
The ZCD circuit can be configured to default to the active or inactive state on Power-on-Reset (POR). When the ZCD Configuration bit is cleared, the ZCD circuit will be active at POR. When the ZCD Configuration bit is set, the ZCDEN bit of the ZCDCON register must be set to enable the ZCD module.
25.9 Register Definitions: ZCD Control

Long bit name prefixes for the zero-cross detect peripheral is shown in Table 25-1. Refer to Section 1.1.2.2 “Long Bit Names” for more information.

**TABLE 25-1:**

<table>
<thead>
<tr>
<th>Peripheral</th>
<th>Bit Name Prefix</th>
</tr>
</thead>
<tbody>
<tr>
<td>ZCD1</td>
<td>ZCD1</td>
</tr>
</tbody>
</table>

**REGISTER 25-1:** ZCDxCON: ZERO-CROSS DETECTION CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value at POR and BOR/Value at all other Resets</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>EN: Zero-Cross Detection Enable bit(1)</td>
<td>1 = Zero-cross detect is enabled. ZCD pin is forced to output to source and sink current. 0 = Zero-cross detect is disabled. ZCD pin operates according to PPS and TRIS controls.</td>
</tr>
<tr>
<td>bit 6</td>
<td>Unimplemented: Read as '0'</td>
<td></td>
</tr>
<tr>
<td>bit 5</td>
<td>OUT: Zero-Cross Detection Logic Level bit</td>
<td></td>
</tr>
<tr>
<td></td>
<td>POL bit = 0:</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = ZCD pin is sinking current</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = ZCD pin is sourcing current</td>
<td></td>
</tr>
<tr>
<td></td>
<td>POL bit = 1:</td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = ZCD pin is sourcing current</td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = ZCD pin is sinking current</td>
<td></td>
</tr>
<tr>
<td>bit 4</td>
<td>POL: Zero-Cross Detection Logic Output Polarity bit</td>
<td>1 = ZCD logic output is inverted 0 = ZCD logic output is not inverted</td>
</tr>
<tr>
<td>bit 3-2</td>
<td>Unimplemented: Read as '0'</td>
<td></td>
</tr>
<tr>
<td>bit 1</td>
<td>INTP: Zero-Cross Positive Edge Interrupt Enable bit</td>
<td>1 = ZCDIF bit is set on low-to-high OUT transition 0 = ZCDIF bit is unaffected by low-to-high OUT transition</td>
</tr>
<tr>
<td>bit 0</td>
<td>INTN: Zero-Cross Negative Edge Interrupt Enable bit</td>
<td>1 = ZCDIF bit is set on high-to-low OUT transition 0 = ZCDIF bit is unaffected by high-to-low OUT transition</td>
</tr>
</tbody>
</table>

**Legend:**

- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared
- q = value depends on configuration bits

**Note 1:** The EN bit has no effect when the ZCD Configuration bit is cleared.
### TABLE 25-2: SUMMARY OF REGISTERS ASSOCIATED WITH THE ZCD MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIE3</td>
<td>—</td>
<td>—</td>
<td>COG2IE</td>
<td>ZCDIE</td>
<td>CLC4IE</td>
<td>CLC3IE</td>
<td>CLC2IE</td>
<td>CLC1IE</td>
<td>94</td>
</tr>
<tr>
<td>PIR3</td>
<td>—</td>
<td>—</td>
<td>COG2IF</td>
<td>ZCDIF</td>
<td>CLC4IF</td>
<td>CLC3IF</td>
<td>CLC2IF</td>
<td>CLC1IF</td>
<td>97</td>
</tr>
<tr>
<td>ZCD1CON</td>
<td>EN</td>
<td>—</td>
<td>OUT</td>
<td>—</td>
<td>POL</td>
<td>—</td>
<td>—</td>
<td>INTP</td>
<td>INTN</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented, read as ‘0’. Shaded cells are unused by the ZCD module.

### TABLE 25-3: SUMMARY OF CONFIGURATION WORD WITH THE ZCD MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG2</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>LVP</td>
<td>DEBUG</td>
<td>LPBOR</td>
<td>BORV</td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>ZCD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>PPS1WAY</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used by the ZCD module.
26.0 TIMER0 MODULE

The Timer0 module is an 8-bit timer/counter with the following features:

- 8-bit timer/counter register (TMR0)
- 8-bit prescaler (independent of Watchdog Timer)
- Programmable internal or external clock source
- Programmable external clock edge selection
- Interrupt on overflow
- TMR0 can be used to gate Timer1

Figure 26-1 is a block diagram of the Timer0 module.

26.1 Timer0 Operation

The Timer0 module can be used as either an 8-bit timer or an 8-bit counter.

26.1.1 8-BIT TIMER MODE

The Timer0 module will increment every instruction cycle, if used without a prescaler. 8-bit Timer mode is selected by clearing the TMR0CS bit of the OPTION_REG register.

When TMR0 is written, the increment is inhibited for two instruction cycles immediately following the write.

Note: The value written to the TMR0 register can be adjusted, in order to account for the two instruction cycle delay when TMR0 is written.

26.1.2 8-BIT COUNTER MODE

In 8-Bit Counter mode, the Timer0 module will increment on every rising or falling edge of the T0CKI pin.

8-Bit Counter mode using the T0CKI pin is selected by setting the TMR0CS bit in the OPTION_REG register to ‘1’.

The rising or falling transition of the incrementing edge for either input source is determined by the TMR0SE bit in the OPTION_REG register.

FIGURE 26-1: BLOCK DIAGRAM OF THE TIMER0

![Block diagram of the Timer0 module]
26.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A software programmable prescaler is available for exclusive use with Timer0. The prescaler is enabled by clearing the PSA bit of the OPTION_REG register.

**Note:** The Watchdog Timer (WDT) uses its own independent prescaler.

There are eight prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION_REG register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be disabled by setting the PSA bit of the OPTION_REG register.

The prescaler is not readable or writable. All instructions writing to the TMR0 register will clear the prescaler.

26.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The TMR0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The TMR0IF bit can only be cleared in software. The Timer0 interrupt enable is the TMR0IE bit of the INTCON register.

**Note:** The Timer0 interrupt cannot wake the processor from Sleep since the timer is frozen during Sleep.

26.1.5 8-BIT COUNTER MODE SYNCHRONIZATION

When in 8-Bit Counter mode, the incrementing edge on the T0CKI pin must be synchronized to the instruction clock. Synchronization can be accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the instruction clock. The high and low periods of the external clocking source must meet the timing requirements as shown in Table 34-12: Timer0 and Timer1 External Clock Requirements.

26.1.6 OPERATION DURING SLEEP

Timer0 cannot operate while the processor is in Sleep mode. The contents of the TMR0 register will remain unchanged while the processor is in Sleep mode.
26.2 Register Definitions: Option Register

REGISTER 26-1: OPTION_REG: OPTION REGISTER

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>WPUEN</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS&lt;2:0&gt;</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- n/n = Value at POR and BOR/Value at all other Resets

bit 7  **WPUEN**: Weak Pull-Up Enable bit
1 = All weak pull-ups are disabled (except MCLR, if it is enabled)
0 = Weak pull-ups are enabled by individual WPUx latch values

bit 6  **INTEDG**: Interrupt Edge Select bit
1 = Interrupt on rising edge of INT pin
0 = Interrupt on falling edge of INT pin

bit 5  **TMR0CS**: Timer0 Clock Source Select bit
1 = Transition on T0CKI pin
0 = Internal instruction cycle clock (Fosc/4)

bit 4  **TMR0SE**: Timer0 Source Edge Select bit
1 = Increment on high-to-low transition on T0CKI pin
0 = Increment on low-to-high transition on T0CKI pin

bit 3  **PSA**: Prescaler Assignment bit
1 = Prescaler is not assigned to the Timer0 module
0 = Prescaler is assigned to the Timer0 module

bit 2-0  **PS<2:0>**: Prescaler Rate Select bits

<table>
<thead>
<tr>
<th>Bit Value</th>
<th>Timer0 Rate</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>1 : 2</td>
</tr>
<tr>
<td>001</td>
<td>1 : 4</td>
</tr>
<tr>
<td>010</td>
<td>1 : 8</td>
</tr>
<tr>
<td>011</td>
<td>1 : 16</td>
</tr>
<tr>
<td>100</td>
<td>1 : 32</td>
</tr>
<tr>
<td>101</td>
<td>1 : 64</td>
</tr>
<tr>
<td>110</td>
<td>1 : 128</td>
</tr>
<tr>
<td>111</td>
<td>1 : 256</td>
</tr>
</tbody>
</table>

TABLE 26-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
<td>OPTION_REG</td>
<td>WPUEN</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS&lt;2:0&gt;</td>
<td></td>
<td></td>
<td>273</td>
</tr>
<tr>
<td>TMR0</td>
<td>Timer0 Module Register</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>271*</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
</tbody>
</table>

Legend:
- — = Unimplemented location, read as ‘0’.
- Shaded cells are not used by the Timer0 module.
- * Page provides register information.
27.0 TIMER1 MODULE WITH GATE CONTROL

The Timer1 module is a 16-bit timer/counter with the following features:

- 16-bit timer/counter register pair (TMR1H:TMR1L)
- Programmable internal or external clock source
- 2-bit prescaler
- Dedicated 32 kHz oscillator circuit
- Optionally synchronized comparator out
- Multiple Timer1 gate (count enable) sources
- Interrupt on overflow
- Wake-up on overflow (external clock, Asynchronous mode only)
- Time base for the Capture/Compare function
- Auto-conversion Trigger (with CCP)
- Selectable Gate Source Polarity
- Gate Toggle mode
- Gate Single-pulse mode
- Gate Value Status
- Gate Event Interrupt

Figure 27-1 is a block diagram of the Timer1 module.

FIGURE 27-1: TIMER1 BLOCK DIAGRAM

Note 1: ST Buffer is high speed type when using T1CKI.
2: Timer1 register increments on rising edge.
3: Synchronize does not operate while in Sleep.
27.1 Timer1 Operation

The Timer1 module is a 16-bit incrementing counter which is accessed through the TMR1H:TMR1L register pair. Writes to TMR1H or TMR1L directly update the counter.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or counter and increments on every selected edge of the external source.

Timer1 is enabled by configuring the TMR1ON and TMR1GE bits in the T1CON and T1GCON registers, respectively. Table 27-1 displays the Timer1 enable selections.

<table>
<thead>
<tr>
<th>TMR1ON</th>
<th>TMR1GE</th>
<th>Timer1 Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>Off</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>Off</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Always On</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>Count Enabled</td>
</tr>
</tbody>
</table>

27.2 Clock Source Selection

The TMR1CS<1:0> and T1OSCEN bits of the T1CON register are used to select the clock source for Timer1. Table 27-2 displays the clock source selections.

27.2.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected, the TMR1H:TMR1L register pair will increment on multiples of Fosc as determined by the Timer1 prescaler.

When the Fosc internal clock source is selected, the Timer1 register value will increment by four counts every instruction clock cycle. Due to this condition, a 2 LSB error in resolution will occur when reading the Timer1 value. To utilize the full resolution of Timer1, an asynchronous input signal must be used to gate the Timer1 clock input.

The following asynchronous sources may be used:
- Asynchronous event on the T1G pin to Timer1 gate
- C1 or C2 comparator input to Timer1 gate

27.2.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When enabled to count, Timer1 is incremented on the rising edge of the external clock input T1CKI, which can be synchronized to the microcontroller system clock or can run asynchronously.

When used as a timer with a clock oscillator, an external 32.768 kHz crystal can be used in conjunction with the dedicated internal oscillator circuit.

**Note:** In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge after any one or more of the following conditions:
- Timer1 enabled after POR
- Write to TMR1H or TMR1L
- Timer1 is disabled
- Timer1 is disabled (TMR1ON = 0) when T1CKI is high then Timer1 is enabled (TMR1ON=1) when T1CKI is low.

<table>
<thead>
<tr>
<th>TMR1CS&lt;1:0&gt;</th>
<th>T1OSCEN</th>
<th>Clock Source</th>
</tr>
</thead>
<tbody>
<tr>
<td>11</td>
<td>x</td>
<td>LFINTOSC</td>
</tr>
<tr>
<td>10</td>
<td>0</td>
<td>External Clocking on T1CKI Pin</td>
</tr>
<tr>
<td>01</td>
<td>x</td>
<td>System Clock (Fosc)</td>
</tr>
<tr>
<td>00</td>
<td>x</td>
<td>Instruction Clock (Fosc/4)</td>
</tr>
</tbody>
</table>
27.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The T1CKPS bits of the T1CON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

27.4 Timer1 (Secondary) Oscillator

A dedicated low-power 32.768 kHz oscillator circuit is built-in between pins SOSCI (input) and SOSCO (amplifier output). This internal circuit is to be used in conjunction with an external 32.768 kHz crystal.

The oscillator circuit is enabled by setting the T1OSCEN bit of the T1CON register. The oscillator will continue to run during Sleep.

Note: The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to using Timer1. A suitable delay similar to the OST delay can be implemented in software by clearing the TMR1IF bit then presetting the TMR1H:TMR1L register pair to FC00h. The TMR1IF flag will be set when 1024 clock cycles have elapsed, thereby indicating that the oscillator is running and reasonably stable.

27.5 Timer1 Operation in Asynchronous Counter Mode

If the control bit T1SYNC of the T1CON register is set, the external clock input is not synchronized. The timer increments asynchronously to the internal phase clocks. If the external clock source is selected then the timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see Section 27.5.1 “Reading and Writing Timer1 in Asynchronous Counter Mode”).

Note: When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.

27.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TMR1L register pair.

27.6 Timer1 Gate

Timer1 can be configured to count freely or the count can be enabled and disabled using Timer1 gate circuitry. This is also referred to as Timer1 Gate Enable. Timer1 gate can also be driven by multiple selectable sources.

27.6.1 TIMER1 GATE ENABLE

The Timer1 Gate Enable mode is enabled by setting the TMR1GE bit of the T1GCON register. The polarity of the Timer1 Gate Enable mode is configured using the T1GPOL bit of the T1GCON register.

When Timer1 Gate Enable mode is enabled, Timer1 will increment on the rising edge of the Timer1 clock source. When Timer1 Gate Enable mode is disabled, no incrementing will occur and Timer1 will hold the current count. See Figure 27-3 for timing details.

### TABLE 27-3: TIMER1 GATE ENABLE SELECTIONS

<table>
<thead>
<tr>
<th>T1CLK</th>
<th>T1GPOL</th>
<th>T1G</th>
<th>Timer1 Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>↑</td>
<td>0</td>
<td>0</td>
<td>Counts</td>
</tr>
<tr>
<td>↑</td>
<td>0</td>
<td>1</td>
<td>Holds Count</td>
</tr>
<tr>
<td>↑</td>
<td>1</td>
<td>0</td>
<td>Holds Count</td>
</tr>
<tr>
<td>↑</td>
<td>1</td>
<td>1</td>
<td>Counts</td>
</tr>
</tbody>
</table>

Note: When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.
27.6.2 TIMER1 GATE SOURCE SELECTION

Timer1 gate source selections are shown in Table 27-4. Source selection is controlled by the T1GSS bits of the T1GCON register. The polarity for each available source is also selectable. Polarity selection is controlled by the T1GPOL bit of the T1GCON register.

<table>
<thead>
<tr>
<th>T1GSS</th>
<th>Timer1 Gate Source</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Timer1 Gate Pin</td>
</tr>
<tr>
<td>01</td>
<td>Overflow of Timer0 (TMR0 increments from FFh to 00h)</td>
</tr>
<tr>
<td>10</td>
<td>Comparator 1 Output sync_C1OUT (optionally Timer1 synchronized output)</td>
</tr>
<tr>
<td>11</td>
<td>Comparator 2 Output sync_C2OUT (optionally Timer1 synchronized output)</td>
</tr>
</tbody>
</table>

Note: Enabling Toggle mode at the same time as changing the gate polarity may result in indeterminate operation.

27.6.3 TIMER1 GATE TOGGLE MODE

When Timer1 Gate Toggle mode is enabled, it is possible to measure the full-cycle length of a Timer1 gate signal, as opposed to the duration of a single level pulse.

The Timer1 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. See Figure 27-4 for timing details.

Timer1 Gate Toggle mode is enabled by setting the T1GTM bit of the T1GCON register. When the T1GTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

27.6.4 TIMER1 GATE SINGLE-PULSE MODE

When Timer1 Gate Single-Pulse mode is enabled, it is possible to capture a single-pulse gate event. Timer1 Gate Single-Pulse mode is first enabled by setting the T1GSPM bit in the T1GCON register. Next, the T1GGO/DONE bit in the T1GCON register must be set. The Timer1 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the T1GGO/DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1 until the T1GGO/DONE bit is once again set in software. See Figure 27-5 for timing details.

If the Single-Pulse Gate mode is disabled by clearing the T1GSPM bit in the T1GCON register, the T1GGO/DONE bit should also be cleared.

Enabling the Toggle mode and the Single-Pulse mode simultaneously will permit both sections to work together. This allows the cycle times on the Timer1 gate source to be measured. See Figure 27-6 for timing details.

27.6.5 TIMER1 GATE VALUE STATUS

When Timer1 Gate Value Status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the T1GV bit in the T1GCON register. The T1GV bit is valid even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

27.6.6 TIMER1 GATE EVENT INTERRUPT

When Timer1 Gate Event Interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of T1GV occurs, the TMR1GIF flag bit in the PIR1 register will be set. If the TMR1GIE bit in the PIE1 register is set, then an interrupt will be recognized.

The TMR1GIF flag bit operates even when the Timer1 gate is not enabled (TMR1GE bit is cleared).
27.7 Timer1 Interrupt

The Timer1 register pair (TMR1H:TMR1L) increments to FFFFh and rolls over to 0000h. When Timer1 rolls over, the Timer1 interrupt flag bit of the PIR1 register is set. To enable the interrupt on rollover, you must set these bits:

- TMR1ON bit of the T1CON register
- TMR1IE bit of the PIE1 register
- PEIE bit of the INTCON register
- GIE bit of the INTCON register

The interrupt is cleared by clearing the TMR1IF bit in the Interrupt Service Routine.

Note: The TMR1H:TMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.

27.8 Timer1 Operation During Sleep

Timer1 can only operate during Sleep when setup in Asynchronous Counter mode. In this mode, an external crystal or clock source can be used to increment the counter. To set up the timer to wake the device:

- TMR1ON bit of the T1CON register must be set
- TMR1IE bit of the PIE1 register must be set
- PEIE bit of the INTCON register must be set
- T1SYNC bit of the T1CON register must be set
- TMR1CS bits of the T1CON register must be configured
- T1OSCEN bit of the T1CON register must be configured

The device will wake-up on an overflow and execute the next instructions. If the GIE bit of the INTCON register is set, the device will call the Interrupt Service Routine.

Secondary oscillator will continue to operate in Sleep regardless of the T1SYNC bit setting.

FIGURE 27-2: TIMER1 INCREMENTING EDGE

T1CKI = 1  
when TMR1  
Enabled

T1CKI = 0  
when TMR1  
Enabled

Note 1: Arrows indicate counter increments.

2: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge of the clock.

27.9 CCP Capture/Compare Time Base

The CCP modules use the TMR1H:TMR1L register pair as the time base when operating in Capture or Compare mode.

In Capture mode, the value in the TMR1H:TMR1L register pair is copied into the CCPR1H:CCPR1L register pair on a configured event.

In Compare mode, an event is triggered when the value CCPR1H:CCPR1L register pair matches the value in the TMR1H:TMR1L register pair. This event can be an auto-conversion trigger.

For more information, see Section 29.0 “Capture/Compare/PWM Modules”.

27.10 CCP Auto-Conversion Trigger

When any of the CCP’s are configured to trigger an auto-conversion, the trigger will clear the TMR1H:TMR1L register pair. This auto-conversion does not cause a Timer1 interrupt. The CCP module may still be configured to generate a CCP interrupt.

In this mode of operation, the CCPR1H:CCPR1L register pair becomes the period register for Timer1. Timer1 should be synchronized and Fosc/4 should be selected as the clock source in order to utilize the auto-conversion trigger. Asynchronous operation of Timer1 can cause an auto-conversion trigger to be missed.

In the event that a write to TMR1H or TMR1L coincides with an auto-conversion trigger from the CCP, the write will take precedence.

For more information, see Section 29.2.4 “Auto-Conversion Trigger”.

Note: The TMR1H:TMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.
FIGURE 27-3: TIMER1 GATE ENABLE MODE

FIGURE 27-4: TIMER1 GATE TOGGLE MODE
FIGURE 27-5: TIMER1 GATE SINGLE-PULSE MODE

TMR1GE

T1GPOL

T1GSPM

T1GGO/
DONE

T1g_in

T1CKI

T1GVAL

Timer1

N

N + 1

N + 2

TMR1GIF

Counting enabled on rising edge of T1G

Set by software

Cleared by hardware on falling edge of T1GVAL

Cleared by software

Set by hardware on falling edge of T1GVAL

Cleared by software
FIGURE 27-6: TIMER1 GATE SINGLE-PULSE AND TOGGLE COMBINED MODE

- TMR1GE
- T1GVAL
- T1GSPM
- T1GTM
- T1GGO/
- T1GPO
- T1G VAL
- Timer1
- TMR1GIF

- Set by software
- Cleared by hardware on falling edge of T1GVAL
- Counting enabled on rising edge of T1G
- Cleared by software
- Set by hardware on falling edge of T1GVAL
- Cleared by software
### 27.11 Register Definitions: Timer1 Control

#### REGISTER 27-1: T1CON: TIMER1 CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5-4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>TMR1CS&lt;1:0&gt;</td>
<td>T1CKPS&lt;1:0&gt;</td>
<td>T1OSCEN</td>
<td>T1SYNC</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

#### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**bit 7-6**  
**TMR1CS<1:0>:** Timer1 Clock Source Select bits
- 11 = LFINTOSC
- 10 = Timer1 clock source is pin or oscillator:
  - If **T1OSCEN** = 0:
    - External clock from T1CKI pin (on the rising edge)
  - If **T1OSCEN** = 1:
    - Crystal oscillator on SOSCI/SOSCO pins
- 01 = Timer1 clock source is system clock (FOSC)
- 00 = Timer1 clock source is instruction clock (FOSC/4)

**bit 5-4**  
**T1CKPS<1:0>:** Timer1 Input Clock Prescale Select bits
- 11 = 1:8 Prescale value
- 10 = 1:4 Prescale value
- 01 = 1:2 Prescale value
- 00 = 1:1 Prescale value

**bit 3**  
**T1OSCEN:** LP Oscillator Enable Control bit
- 1 = Dedicated secondary oscillator circuit enabled
- 0 = Dedicated secondary oscillator circuit disabled

**bit 2**  
**T1SYNC:** Timer1 Synchronization Control bit
- 1 = Do not synchronize asynchronous clock input
- 0 = Synchronize asynchronous clock input with system clock (FOSC)

**bit 1**  
**Unimplemented:** Read as ‘0’

**bit 0**  
**TMR1ON:** Timer1 On bit
- 1 = Enables Timer1
- 0 = Stops Timer1 and clears Timer1 gate flip-flop
## REGISTER 27-2: T1GCON: TIMER1 GATE CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value at POR and BOR/Value at all other Resets</th>
</tr>
</thead>
</table>
| 7   | TMR1GE: Timer1 Gate Enable bit | If TMR1ON = 0: Bit is ignored  
If TMR1ON = 1:  
1 = Timer1 counting is controlled by the Timer1 gate function  
0 = Timer1 counts regardless of Timer1 gate function |
| 6   | T1GPOL: Timer1 Gate Polarity bit | 1 = Timer1 gate is active-high (Timer1 counts when gate is high)  
0 = Timer1 gate is active-low (Timer1 counts when gate is low) |
| 5   | T1GTM: Timer1 Gate Toggle Mode bit | 1 = Timer1 Gate Toggle mode is enabled  
0 = Timer1 Gate Toggle mode is disabled and toggle flip-flop is cleared  
Timer1 gate flip-flop toggles on every rising edge. |
| 4   | T1GSPM: Timer1 Gate Single-Pulse Mode bit | 1 = Timer1 Gate Single-Pulse mode is enabled and is controlling Timer1 gate  
0 = Timer1 Gate Single-Pulse mode is disabled |
| 3   | T1GGO/DONE: Timer1 Gate Single-Pulse Acquisition Status bit | 1 = Timer1 gate single-pulse acquisition is ready, waiting for an edge  
0 = Timer1 gate single-pulse acquisition has completed or has not been started |
| 2   | T1GVAL: Timer1 Gate Value Status bit | Indicates the current state of the Timer1 gate that could be provided to TMR1H:TMR1L  
Unaffected by Timer1 Gate Enable (TMR1GE) |
| 1-0 | T1GSS<1:0>: Timer1 Gate Source Select bits | 11 = Comparator 2 optionally synchronized output (sync_C2OUT)  
10 = Comparator 1 optionally synchronized output (sync_C1OUT)  
01 = Timer0 overflow output  
00 = Timer1 gate pin |

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **HC** = Bit is cleared by hardware
### TABLE 27-5: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER1

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>CCP1CON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>DC1B&lt;1:0&gt;</td>
<td>CCP1M&lt;3:0&gt;</td>
<td>297</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>CCP2CON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>DC2B&lt;1:0&gt;</td>
<td>CCP2M&lt;3:0&gt;</td>
<td>297</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td></td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td></td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td></td>
</tr>
<tr>
<td>TMR1H</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TMR1L</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td></td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td></td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td></td>
</tr>
<tr>
<td>T1CON</td>
<td>TMR1CS&lt;1:0&gt;</td>
<td>T1CKPS&lt;1:0&gt;</td>
<td>T1OSCEN</td>
<td>T1SYNC</td>
<td>—</td>
<td>TMR1ON</td>
<td></td>
<td></td>
<td>282</td>
</tr>
<tr>
<td>T1GCON</td>
<td>TMR1GE</td>
<td>T1GPO</td>
<td>T1GTM</td>
<td>T1GSPM</td>
<td>T1GGO/DONE</td>
<td>T1GVAL</td>
<td>T1GSS&lt;1:0&gt;</td>
<td></td>
<td>283</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used by the Timer1 module.

* Page provides register information.
28.0  TIMER2/4/6 MODULE

The Timer2/4/6 modules are 8-bit timers that incorporate the following features:

- 8-bit Timer and Period registers (TMR2 and PR2, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4, 1:16, and 1:64)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMR2 match with PR2, respectively
- Optional use as the shift clock for the MSSP module

See Figure 28-1 for a block diagram of Timer2.

Three identical Timer2 modules are implemented on this device. To maintain consistency with earlier devices, the timers are named Timer2, Timer4, and Timer6. All references to Timer2 apply as well to Timer4 and Timer6.

FIGURE 28-1:  TIMER2 BLOCK DIAGRAM
28.1 Timer2 Operation

The clock input to the Timer2 modules is the system instruction clock (Fosc/4).
TMR2 increments from 00h on each clock edge.
A 4-bit counter/prescaler on the clock input allows direct input, divide-by-4 and divide-by-16 prescale options. These options are selected by the prescaler control bits, T2CKPS<1:0> of the T2CON register. The value of TMR2 is compared to that of the Period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/postscaler (see Section 28.2 “Timer2 Interrupt”).

The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, whereas the PR2 register initializes to FFh. Both the prescaler and postscaler counters are cleared on the following events:
- A write to the TMR2 register
- A write to the T2CON register
- Power-on Reset (POR)
- Brown-out Reset (BOR)
- MCLR Reset
- Watchdog Timer (WDT) Reset
- Stack Overflow Reset
- Stack Underflow Reset
- RESET Instruction

Note: TMR2 is not cleared when T2CON is written.

28.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2-to-PR2 match) provides the input for the 4-bit counter/postscaler. This counter generates the TMR2 match interrupt flag which is latched in TMR2IF of the PIR1 register. The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE, of the PIE1 register.

A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS<3:0>, of the T2CON register.

28.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the CCP modules, where it is used as a time base for operations in PWM mode.

Timer2 can be optionally used as the shift clock source for the MSSP module operating in SPI mode. Additional information is provided in Section 30.0 “Master Synchronous Serial Port (MSSP) Module”.

28.4 Timer2 Operation During Sleep

The Timer2 timers cannot be operated while the processor is in Sleep mode. The contents of the TMR2 and PR2 registers will remain unchanged while the processor is in Sleep mode.
28.5 Register Definitions: Timer2 Control

REGISTER 28-1: T2CON: TIMER2 CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit</th>
<th>Symbol</th>
<th>Function</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>T2OUTPS&lt;3:0&gt;</td>
<td>Timer2 Output Postscaler Select bits</td>
<td>1111 = 1:16 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1110 = 1:15 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1101 = 1:14 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1100 = 1:13 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1011 = 1:12 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1010 = 1:11 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1001 = 1:10 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1000 = 1:9 Postscaler</td>
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<tr>
<td></td>
<td></td>
<td></td>
<td>0111 = 1:8 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0110 = 1:7 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0101 = 1:6 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0100 = 1:5 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0011 = 1:4 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0010 = 1:3 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0001 = 1:2 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0000 = 1:1 Postscaler</td>
</tr>
<tr>
<td>6-3</td>
<td></td>
<td>Timer2 Output Postscaler Select bits</td>
<td>0111 = 1:8 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0110 = 1:7 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0101 = 1:6 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0100 = 1:5 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0011 = 1:4 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0010 = 1:3 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0001 = 1:2 Postscaler</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0000 = 1:1 Postscaler</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

bit 2  **TMR2ON:** Timer2 On bit
1 = Timer2 is on
0 = Timer2 is off

bit 1-0  **T2CKPS<1:0>:** Timer2 Clock Prescale Select bits
11 = Prescaler is 64
10 = Prescaler is 16
01 = Prescaler is 4
00 = Prescaler is 1
### TABLE 28-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER2

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CCP2CON</td>
<td>—</td>
<td>—</td>
<td>DC2B&lt;1:0&gt;</td>
<td>CCP2M&lt;3:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>297</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td></td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSR1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td></td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSR1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td></td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
<td></td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td></td>
</tr>
<tr>
<td>PR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Timer2 Module Period Register</td>
</tr>
<tr>
<td>T2CON</td>
<td>—</td>
<td></td>
<td>T2OUTPS&lt;3:0&gt;</td>
<td>TMR2ON</td>
<td>T2CKPS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>287</td>
</tr>
<tr>
<td>TMR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Holding Register for the 8-bit TMR2 Register</td>
</tr>
<tr>
<td>PR4</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td>Timer4 Module Period Register</td>
</tr>
<tr>
<td>T4CON</td>
<td>—</td>
<td></td>
<td>T4OUTPS&lt;3:0&gt;</td>
<td>TMR4ON</td>
<td>T4CKPS&lt;1:0&gt;</td>
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<td></td>
<td></td>
<td>287</td>
</tr>
<tr>
<td>TMR4</td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Holding Register for the 8-bit TMR4 Register</td>
</tr>
<tr>
<td>PR6</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td>Timer6 Module Period Register</td>
</tr>
<tr>
<td>T6CON</td>
<td>—</td>
<td></td>
<td>T6OUTPS&lt;3:0&gt;</td>
<td>TMR6ON</td>
<td>T6CKPS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>287</td>
</tr>
<tr>
<td>TMR6</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Holding Register for the 8-bit TMR6 Register</td>
</tr>
</tbody>
</table>

**Legend:**
- — = unimplemented location, read as ‘0’.
- Shaded cells are not used for Timer2 module.
- * Page provides register information.
### 28.6 CCP/PWM Clock Selection

The PIC16(L)F1717/8/9 allows each individual CCP and PWM module to select the timer source that controls the module. Each module has an independent selection.

As there are up to three 8-bit timers with auto-reload (Timer2, Timer4, and Timer6), PWM mode on the CCP and PWM modules can use any of these timers.

The CCPTMRS register is used to select which timer is used.

### 28.7 Register Definitions: CCP/PWM Timers Control

**REGISTER 28-2: CCPTMRS: PWM TIMER SELECTION CONTROL REGISTER 0**

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th>bit 5-4</th>
<th>bit 3-2</th>
<th>bit 1-0</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>P4TSEL&lt;1:0&gt;: PWM4 Timer Selection</strong></td>
<td><strong>P3TSEL&lt;1:0&gt;: PWM3 Timer Selection</strong></td>
<td><strong>C2TSEL&lt;1:0&gt;: CCP2 (PWM2) Timer Selection</strong></td>
<td><strong>C1TSEL&lt;1:0&gt;: CCP1 (PWM1) Timer Selection</strong></td>
</tr>
<tr>
<td>11 = Reserved</td>
<td>11 = Reserved</td>
<td>11 = Reserved</td>
<td>11 = Reserved</td>
</tr>
<tr>
<td>10 = PWM4 is based off Timer6</td>
<td>10 = PWM3 is based off Timer6</td>
<td>10 = CCP2 is based off Timer 6 in PWM mode</td>
<td>10 = CCP1 is based off Timer6 in PWM mode</td>
</tr>
<tr>
<td>01 = PWM4 is based off Timer4</td>
<td>01 = PWM3 is based off Timer4</td>
<td>01 = CCP2 is based off Timer 4 in PWM mode</td>
<td>01 = CCP1 is based off Timer4 in PWM mode</td>
</tr>
<tr>
<td>00 = PWM4 is based off Timer2</td>
<td>00 = PWM3 is based off Timer2</td>
<td>00 = CCP2 is based off Timer 2 in PWM mode</td>
<td>00 = CCP1 is based off Timer2 in PWM mode</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets
29.0 CAPTURE/COMPARE/PWM MODULES

The Capture/Compare/PWM module is a peripheral which allows the user to time and control different events, and to generate Pulse-Width Modulation (PWM) signals. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate Pulse-Width Modulated signals of varying frequency and duty cycle.

This family of devices contains two standard Capture/Compare/PWM modules (CCP1 and CCP2). The Capture and Compare functions are identical for all CCP modules.

Note 1: In devices with more than one CCP module, it is very important to pay close attention to the register names used. A number placed after the module acronym is used to distinguish between separate modules. For example, the CCP1CON and CCP2CON control the same operational aspects of two completely different CCP modules.

2: Throughout this section, generic references to a CCP module in any of its operating modes may be interpreted as being equally applicable to CCPx module. Register names, module signals, I/O pins, and bit names may use the generic designator ‘x’ to indicate the use of a numeral to distinguish a particular module, when required.

29.1 Capture Mode

The Capture mode function described in this section is available and identical for all CCP modules.

Capture mode makes use of the 16-bit Timer1 resource. When an event occurs on the CCPx pin, the 16-bit CCPRxH:CCPRxL register pair captures and stores the 16-bit value of the TMR1H:TMR1L register pair, respectively. An event is defined as one of the following and is configured by the CCPxM<3:0> bits of the CCPxCON register:

- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

When a capture is made, the Interrupt Request Flag bit CCPxIF of the PIRx register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPRxH, CCPRxL register pair is read, the old captured value is overwritten by the new captured value.

Figure 29-1 shows a simplified diagram of the capture operation.

29.1.1 CCP PIN CONFIGURATION

In Capture mode, the CCPx pin should be configured as an input by setting the associated TRIS control bit.

Note: If the CCPx pin is configured as an output, a write to the port can cause a capture condition.
29.1.2 TIMER1 MODE RESOURCE

Timer1 must be running in Timer mode or Synchronized Counter mode for the CCP module to use the capture feature. In Asynchronous Counter mode, the capture operation may not work.

See Section 27.0 “Timer1 Module with Gate Control” for more information on configuring Timer1.

29.1.3 SOFTWARE INTERRUPT MODE

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit of the PIEx register clear to avoid false interrupts. Additionally, the user should clear the CCPxIF interrupt flag bit of the PIRx register following any change in Operating mode.

Note: Clocking Timer1 from the system clock (Fosc) should not be used in Capture mode. In order for Capture mode to recognize the trigger event on the CCPx pin, Timer1 must be clocked from the instruction clock (Fosc/4) or from an external clock source.

29.1.4 CCP PRESCALER

There are four prescaler settings specified by the CCPxM<3:0> bits of the CCPxCON register. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another does not clear the prescaler and may generate a false interrupt. To avoid this unexpected operation, turn the module off by clearing the CCPxCON register before changing the prescaler. Example 29-1 demonstrates the code to perform this function.

EXAMPLE 29-1: CHANGING BETWEEN CAPTURE PRESCALERS

BANKSEL CCPxCON ;Set Bank bits to point to CCPxCON
CLRF CCPxCON ;Turn CCP module off
MOVLW NEW_CAPT_PS ;Load the W reg with the new prescaler
MOVWF CCPxCON ;move value and CCP ON
MOVWF CCPxCON ;Load CCPxCON with this value

29.1.5 CAPTURE DURING SLEEP

Capture mode depends upon the Timer1 module for proper operation. There are two options for driving the Timer1 module in Capture mode. It can be driven by the instruction clock (Fosc/4), or by an external clock source.

When Timer1 is clocked by Fosc/4, Timer1 will not increment during Sleep. When the device wakes from Sleep, Timer1 will continue from its previous state.

Capture mode will operate during Sleep when Timer1 is clocked by an external clock source.

Note: Clocking Timer1 from the system clock (Fosc) should not be used in Capture mode. In order for Capture mode to recognize the trigger event on the CCPx pin, Timer1 must be clocked from the instruction clock (Fosc/4) or from an external clock source.
29.2 Compare Mode

The Compare mode function described in this section is available and identical for all CCP modules.

Compare mode makes use of the 16-bit Timer1 resource. The 16-bit value of the CCPRxH:CCPRxL register pair is constantly compared against the 16-bit value of the TMR1H:TMR1L register pair. When a match occurs, one of the following events can occur:

- Toggle the CCPx output
- Set the CCPx output
- Clear the CCPx output
- Generate an Auto-conversion Trigger
- Generate a Software Interrupt

The action on the pin is based on the value of the CCPxM<3:0> control bits of the CCPxCON register. At the same time, the interrupt flag CCPxIF bit is set.

All Compare modes can generate an interrupt.

Figure 29-2 shows a simplified diagram of the compare operation.

FIGURE 29-2: COMPARE MODE OPERATION BLOCK DIAGRAM

29.2.1 CCPX PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the associated TRIS bit.

Note: Clearing the CCPxCON register will force the CCPx compare output latch to the default low level. This is not the PORT I/O data latch.

29.2.2 TIMER1 MODE RESOURCE

In Compare mode, Timer1 must be running in either Timer mode or Synchronized Counter mode. The compare operation may not work in Asynchronous Counter mode.

See Section 27.0 “Timer1 Module with Gate Control” for more information on configuring Timer1.

Note: Clocking Timer1 from the system clock (FOSC) should not be used in Compare mode. In order for Compare mode to recognize the trigger event on the CCPx pin, Timer1 must be clocked from the instruction clock (Fosc/4) or from an external clock source.

29.2.3 SOFTWARE INTERRUPT MODE

When Generate Software Interrupt mode is chosen (CCPxM<3:0> = 1010), the CCPx module does not assert control of the CCPx pin (see the CCPxCON register).

29.2.4 AUTO-CONVERSION TRIGGER

When Auto-Conversion Trigger mode is chosen (CCPxM<3:0> = 1011), the CCPx module does the following:

- Resets Timer1
- Starts an ADC conversion if ADC is enabled

The CCPx module does not assert control of the CCPx pin in this mode.

The auto-conversion trigger output of the CCP occurs immediately upon a match between the TMR1H, TMR1L register pair and the CCPRxH, CCPRxL register pair. The TMR1H, TMR1L register pair is not reset until the next rising edge of the Timer1 clock. The auto-conversion trigger output starts an ADC conversion (if the ADC module is enabled). This allows the CCPRxH, CCPRxL register pair to effectively provide a 16-bit programmable period register for Timer1.

Refer to Section 29.2.4 “Auto-Conversion Trigger” for more information.

Note 1: The auto-conversion trigger from the CCP module does not set interrupt flag bit TMR1IF of the PIR1 register.

2: Removing the match condition by changing the contents of the CCPRxH and CCPRxL register pair, between the clock edge that generates the auto-conversion trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.

29.2.5 COMPARE DURING SLEEP

The Compare mode is dependent upon the system clock (Fosc) for proper operation. Since Fosc is shut down during Sleep mode, the Compare mode will not function properly during Sleep.

Note 1: The auto-conversion trigger from the CCP module does not set interrupt flag bit TMR1IF of the PIR1 register.

2: Removing the match condition by changing the contents of the CCPRxH and CCPRxL register pair, between the clock edge that generates the auto-conversion trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.
29.3 PWM Overview

Pulse-Width Modulation (PWM) is a scheme that provides power to a load by switching quickly between fully on and fully off states. The PWM signal resembles a square wave where the high portion of the signal is considered the on state and the low portion of the signal is considered the off state. The high portion, also known as the pulse width, can vary in time and is defined in steps. A larger number of steps applied, which lengthens the pulse width, also supplies more power to the load. Lowering the number of steps applied, which shortens the pulse width, supplies less power. The PWM period is defined as the duration of one complete cycle or the total amount of on and off time combined.

PWM resolution defines the maximum number of steps that can be present in a single PWM period. A higher resolution allows for more precise control of the pulse width time and in turn the power that is applied to the load.

The term duty cycle describes the proportion of the on time to the off time and is expressed in percentages, where 0% is fully off and 100% is fully on. A lower duty cycle corresponds to less power applied and a higher duty cycle corresponds to more power applied.

Figure 29-3 shows a typical waveform of the PWM signal.

29.3.1 STANDARD PWM OPERATION

The standard PWM function described in this section is available and identical for all CCP modules.

The standard PWM mode generates a Pulse-Width Modulation (PWM) signal on the CCPx pin with up to ten bits of resolution. The period, duty cycle, and resolution are controlled by the following registers:

- PR2 registers
- T2CON registers
- CCPRxL registers
- CCPxCON registers

Figure 29-4 shows a simplified block diagram of PWM operation.

**Note:** The corresponding TRIS bit must be cleared to enable the PWM output on the CCPx pin.

**FIGURE 29-4:** SIMPLIFIED PWM BLOCK DIAGRAM

**Note 1:** The 8-bit timer TMR2 register is concatenated with the 2-bit internal system clock (Fosc), or two bits of the prescaler, to create the 10-bit time base.

**Note 2:** In PWM mode, CCPR1H is a read-only register.

**FIGURE 29-3:** CCP PWM OUTPUT SIGNAL
29.3.2 SETUP FOR PWM OPERATION
The following steps should be taken when configuring the CCP module for standard PWM operation:

1. Use the desired output pin RxPPP control to select CCPx as the source and disable the CCPx pin output driver by setting the associated TRIS bit.
2. Load the PR2 register with the PWM period value.
3. Configure the CCP module for the PWM mode by loading the CCPxCON register with the appropriate values.
4. Load the CCPRxL register and the DCxB bits of the CCPxCON register, with the PWM duty cycle value.
5. Configure and start Timer2:
   - Clear the TMR2IF interrupt flag bit of the PIRx register. See Note below.
   - Configure the T2CKPS bits of the T2CON register with the Timer prescale value.
   - Enable the Timer by setting the TMR2ON bit of the T2CON register.
6. Enable PWM output pin:
   - Wait until the Timer overflows and the TMR2IF bit of the PIR1 register is set. See Note below.
   - Enable the CCPx pin output driver by clearing the associated TRIS bit.

Note: In order to send a complete duty cycle and period on the first PWM output, the above steps must be included in the setup sequence. If it is not critical to start with a complete PWM signal on the first output, then step 6 may be ignored.

29.3.3 TIMER2 TIMER RESOURCE
The PWM standard mode makes use of the 8-bit Timer2 timer resources to specify the PWM period.

29.3.4 PWM PERIOD
The PWM period is specified by the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 29-1.

\[
\text{PWM Period} = \left( (\text{PR2} + 1) \cdot 4 \cdot T_{\text{OSC}} \right) \cdot (\text{TMR2 Prescale Value})
\]

Note 1: \( T_{\text{OSC}} = 1/F_{\text{OSC}} \)

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:
- TMR2 is cleared
- The CCPx pin is set. (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM duty cycle is latched from CCPRxL into CCPRxH

Note: The Timer postscaler (see Section 28.1 “Timer2 Operation”) is not used in the determination of the PWM frequency.

29.3.5 PWM DUTY CYCLE
The PWM duty cycle is specified by writing a 10-bit value to multiple registers: CCPRxL register and DCxB<1:0> bits of the CCPxCON register. The CCPRxL contains the eight MSbs and the DCxB<1:0> bits of the CCPxCON register contain the two LSbs. CCPRxL and DCxB<1:0> bits of the CCPxCON register can be written to at any time. The duty cycle value is not latched into CCPRxH until after the period completes (i.e., a match between PR2 and TMR2 registers occurs). While using the PWM, the CCPRxH register is read-only.

Equation 29-2 is used to calculate the PWM pulse width.
Equation 29-3 is used to calculate the PWM duty cycle ratio.

\[
\text{Pulse Width} = \left( \text{CCPRxL:CCPxCON}<5:4> \right) \cdot T_{\text{OSC}} \cdot (\text{TMR2 Prescale Value})
\]

\[
\text{Duty Cycle Ratio} = \left( \frac{\text{CCPRxL:CCPxCON}<5:4>}{4(\text{PR2} + 1)} \right)
\]

The CCPRxH register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.
The 8-bit timer TMR2 register is concatenated with either the 2-bit internal system clock (Fosc), or two bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

When the 10-bit time base matches the CCPRxH and 2-bit latch, then the CCPx pin is cleared (see Figure 29-4).

29.3.6 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is 10 bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 29-4.

EQUATION 29-4: PWM RESOLUTION

\[
\text{Resolution} = \frac{\log(4(\text{PR2} + 1))}{\log(2)} \text{ bits}
\]

Note: If the pulse width value is greater than the period, the assigned PWM pin(s) will remain unchanged.

<table>
<thead>
<tr>
<th>TABLE 29-1: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PWM Frequency</td>
</tr>
<tr>
<td>Timer Prescale</td>
</tr>
<tr>
<td>PR2 Value</td>
</tr>
<tr>
<td>Maximum Resolution (bits)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>TABLE 29-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PWM Frequency</td>
</tr>
<tr>
<td>Timer Prescale</td>
</tr>
<tr>
<td>PR2 Value</td>
</tr>
<tr>
<td>Maximum Resolution (bits)</td>
</tr>
</tbody>
</table>

29.3.7 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the CCPx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

29.3.8 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See Section 6.0 “Oscillator Module (with Fail-Safe Clock Monitor)” for additional details.
29.3.9 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

**TABLE 29-3: SUMMARY OF REGISTERS ASSOCIATED WITH CCP**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CCP1CON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>DC1B&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>297</td>
</tr>
<tr>
<td>CCP1L</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>294*</td>
</tr>
<tr>
<td>CCPTMRS</td>
<td>P4TSEL&lt;1:0&gt;</td>
<td>—</td>
<td>P3TSEL&lt;1:0&gt;</td>
<td>C2TSEL&lt;1:0&gt;</td>
<td>—</td>
<td>C1TSEL&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>289</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IO CIF</td>
<td>91</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
<td>93</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>95</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td>96</td>
</tr>
<tr>
<td>PR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>285*</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>RxyPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RxyPPS&lt;4:0&gt;</td>
</tr>
<tr>
<td>CCP1PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>CCP1PPS&lt;4:0&gt;</td>
</tr>
<tr>
<td>CCP2PPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>CCP2PPS&lt;4:0&gt;</td>
</tr>
<tr>
<td>T2CON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>T2OUTPS&lt;3:0&gt;</td>
</tr>
<tr>
<td>TMR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>TMR2ON</td>
</tr>
</tbody>
</table>

Legend: — = Unimplemented location, read as ‘0’. Shaded cells are not used by the CCP.

* Page provides register information.
29.4 Register Definitions: CCP Control

REGISTER 29-1: CCPxCON: CCPx CONTROL REGISTER

| bit 7-6 | | | | | | | |
|---|---|---|---|---|---|---|
| U-0 | U-0 | R/W-0/0 | R/W-0/0 | R/W-0/0 | R/W-0/0 | R/W-0/0 |
| — | — | DCxB<1:0> | | | | |

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Reset

bit 7-6 **Unimplemented**: Read as ‘0’

bit 5-4 **DCxB<1:0>**: PWM Duty Cycle Least Significant bits

- **Capture mode**: Unused
- **Compare mode**: Unused

- **PWM mode**:
  These bits are the two LSbs of the PWM duty cycle. The eight MSbs are found in CCPRxL.

bit 3-0 **CCPxM<3:0>**: CCPx Mode Select bits

- **11xx** = PWM mode
- **1011** = Compare mode: Auto-conversion Trigger (sets CCPxIF bit), starts ADC conversion if TRIGSEL = CCPx (see Register 21-3)
- **1010** = Compare mode: generate software interrupt only
- **1001** = Compare mode: clear output on compare match (set CCPxIF)
- **1000** = Compare mode: set output on compare match (set CCPxIF)
- **0111** = Capture mode: every 16th rising edge
- **0110** = Capture mode: every 4th rising edge
- **0101** = Capture mode: every rising edge
- **0100** = Capture mode: every falling edge
- **0011** = Reserved
- **0010** = Compare mode: toggle output on match
- **0001** = Reserved
- **0000** = Capture/Compare/PWM off (resets CCPx module)
30.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

30.1 MSSP Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I²C)

The SPI interface supports the following modes and features:

- Master mode
- Slave mode
- Clock Parity
- Slave Select Synchronization (Slave mode only)
- Daisy-chain connection of slave devices

Figure 30-1 is a block diagram of the SPI interface module.

FIGURE 30-1: MSSP BLOCK DIAGRAM (SPI MODE)
The I²C interface supports the following modes and features:

- Master mode
- Slave mode
- Byte NACKing (Slave mode)
- Limited multi-master support
- 7-bit and 10-bit addressing
- Start and Stop interrupts
- Interrupt masking
- Clock stretching
- Bus collision detection
- General call address matching
- Address masking
- Address Hold and Data Hold modes
- Selectable SDA hold times

Figure 30-2 is a block diagram of the I²C interface module in Master mode. Figure 30-3 is a diagram of the I²C interface module in Slave mode.

**FIGURE 30-2: MSSP BLOCK DIAGRAM (I²C MASTER MODE)**

**Note 1:** SDA pin selections must be the same for input and output.
**Note 2:** SCL pin selections must be the same for input and output.
FIGURE 30-3: MSSP BLOCK DIAGRAM (I^2C SLAVE MODE)

Note 1: SDA pin selections must be the same for input and output.
2: SCL pin selections must be the same for input and output.
30.2 SPI Mode Overview

The Serial Peripheral Interface (SPI) bus is a synchronous serial data communication bus that operates in Full-Duplex mode. Devices communicate in a master/slave environment where the master device initiates the communication. A slave device is controlled through a Chip Select known as Slave Select.

The SPI bus specifies four signal connections:
- Serial Clock (SCK)
- Serial Data Out (SDO)
- Serial Data In (SDI)
- Slave Select (SS)

Figure 30-1 shows the block diagram of the MSSP module when operating in SPI mode.

The SPI bus operates with a single master device and one or more slave devices. When multiple slave devices are used, an independent Slave Select connection is required from the master device to each slave device.

Figure 30-4 shows a typical connection between a master device and multiple slave devices.

The master selects only one slave at a time. Most slave devices have tri-state outputs so their output signal appears disconnected from the bus when they are not selected.

Transmissions involve two shift registers, eight bits in size, one in the master and one in the slave. With either the master or the slave device, data is always shifted out one bit at a time, with the Most Significant bit (MSb) shifted out first. At the same time, a new Least Significant bit (LSb) is shifted into the same register.

Figure 30-5 shows a typical connection between two processors configured as master and slave devices.

Data is shifted out of both shift registers on the programmed clock edge and latched on the opposite edge of the clock.

The master device transmits information out on its SDO output pin which is connected to, and received by, the slave’s SDI input pin. The slave device transmits information out on its SDO output pin, which is connected to, and received by, the master’s SDI input pin.

To begin communication, the master device first sends out the clock signal. Both the master and the slave devices should be configured for the same clock polarity.

The master device starts a transmission by sending out the MSb from its shift register. The slave device reads this bit from that same line and saves it into the LSb position of its shift register.

During each SPI clock cycle, a full-duplex data transmission occurs. This means that while the master device is sending out the MSb from its shift register (on its SDO pin) and the slave device is reading this bit and saving it as the LSb of its shift register, that the slave device is also sending out the MSb from its shift register (on its SDO pin) and the master device is reading this bit and saving it as the LSb of its shift register.

After eight bits have been shifted out, the master and slave have exchanged register values.

If there is more data to exchange, the shift registers are loaded with new data and the process repeats itself.

Whether the data is meaningful or not (dummy data), depends on the application software. This leads to three scenarios for data transmission:
- Master sends useful data and slave sends dummy data.
- Master sends useful data and slave sends useful data.
- Master sends dummy data and slave sends useful data.

Transmissions may involve any number of clock cycles. When there is no more data to be transmitted, the master stops sending the clock signal and it deselects the slave.

Every slave device connected to the bus that has not been selected through its slave select line must disregard the clock and transmission signals and must not transmit any data of its own.
30.2.1 SPI MODE REGISTERS

The MSSP module has five registers for SPI mode operation. These are:

- MSSP STATUS register (SSP1STAT)
- MSSP Control register 1 (SSP1CON1)
- MSSP Control register 3 (SSP1CON3)
- MSSP Data Buffer register (SSP1BUF)
- MSSP Address register (SSP1ADD)
- MSSP Shift register (SSP1SR)

(SPP1CON1 and SSP1STAT are the control and STATUS registers in SPI mode operation. The SSP1CON1 register is readable and writable. The lower six bits of the SSP1STAT are read-only. The upper two bits of the SSP1STAT are read/write.

In one SPI master mode, SSP1ADD can be loaded with a value used in the Baud Rate Generator. More information on the Baud Rate Generator is available in Section 30.7 “Baud Rate Generator”.

SSP1SR is the shift register used for shifting data in and out. SSP1BUF provides indirect access to the SSP1SR register. SSP1BUF is the buffer register to which data bytes are written, and from which data bytes are read.

In receive operations, SSP1SR and SSP1BUF together create a buffered receiver. When SSP1SR receives a complete byte, it is transferred to SSP1BUF and the SSP1IF interrupt is set.

During transmission, the SSP1BUF is not buffered. A write to SSP1BUF will write to both SSP1BUF and SSP1SR.

30.2.2 SPI MODE OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSP1CON1<5:0> and SSP1STAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- Clock Polarity (Idle state of SCK)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- Slave Select mode (Slave mode only)

To enable the serial port, SSP Enable bit, SSPEN of the SSP1CON1 register, must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSP1CONx registers and then set the SSPEN bit. This configures the SDI, SDO, SCK and SS pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- SDI must have corresponding TRIS bit set
- SDO must have corresponding TRIS bit cleared
- SCK (Master mode) must have corresponding TRIS bit cleared
- SCK (Slave mode) must have corresponding TRIS bit set
- SS must have corresponding TRIS bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.
The MSSP consists of a transmit/receive shift register (SSP1SR) and a buffer register (SSP1BUF). The SSP1SR shifts the data in and out of the device, MSb first. The SSP1BUF holds the data that was written to the SSP1SR until the received data is ready. Once the eight bits of data have been received, that byte is moved to the SSP1BUF register. Then, the Buffer Full detect bit, BF of the SSP1STAT register, and the interrupt flag bit, SSP1IF, are set. This double-buffering of the received data (SSP1BUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSP1BUF register during transmission/reception of data will be ignored and the Write Collision Detect bit WCOL of the SSP1CON1 register, will be set. User software must clear the WCOL bit to allow the following write(s) to the SSP1BUF register to complete successfully.

When the application software is expecting to receive valid data, the SSP1BUF should be read before the next byte of data to transfer is written to the SSP1BUF. The Buffer Full bit, BF of the SSP1STAT register, indicates when SSP1BUF has been loaded with the received data (transmission is complete). When the SSP1BUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur.

The SSP1SR is not directly readable or writable and can only be accessed by addressing the SSP1BUF register. Additionally, the SSP1STAT register indicates the various status conditions.

**FIGURE 30-5: SPI MASTER/SLAVE CONNECTION**

![SPI Master/Slave Connection Diagram](image-url)
30.2.3 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK line. The master determines when the slave (Processor 2, Figure 30-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSP1BUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSP1SR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSP1BUF register as if a normal received byte (interrupts and Status bits appropriately set).

The clock polarity is selected by appropriately programming the CKP bit of the SSP1CON1 register and the CKE bit of the SSP1STAT register. This then, would give waveforms for SPI communication as shown in Figure 30-6, Figure 30-8, Figure 30-9 and Figure 30-10, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or TCY)
- Fosc/16 (or 4 * TCY)
- Fosc/64 (or 16 * TCY)
- Timer2 output/2
- Fosc/(4 * (SSP1ADD + 1))

Figure 30-6 shows the waveforms for Master mode.

When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSP1BUF is loaded with the received data is shown.

**Note:** In Master mode the clock signal output to the SCK pin is also the clock signal input to the peripheral. The pin selected for output with the RxyPPS register must also be selected as the peripheral input with the SSPCLKPPS register.
30.2.4 SPI SLAVE MODE

In Slave mode, the data is transmitted and received as external clock pulses appear on SCK. When the last bit is latched, the SSP1IF interrupt flag bit is set.

Before enabling the module in SPI Slave mode, the clock line must match the proper Idle state. The clock line can be observed by reading the SCK pin. The Idle state is determined by the CKP bit of the SSP1CON1 register.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. The shift register is clocked from the SCK pin input and when a byte is received, the device will generate an interrupt. If enabled, the device will wake-up from Sleep.

30.2.4.1 Daisy-Chain Configuration

The SPI bus can sometimes be connected in a daisy-chain configuration. The first slave output is connected to the second slave input, the second slave output is connected to the third slave input, and so on. The final slave output is connected to the master input. Each slave sends out, during a second group of clock pulses, an exact copy of what was received during the first group of clock pulses. The whole chain acts as one large communication shift register. The daisy-chain feature only requires a single Slave Select line from the master device.

Figure 30-7 shows the block diagram of a typical daisy-chain connection when operating in SPI mode.

In a daisy-chain configuration, only the most recent byte on the bus is required by the slave. Setting the BOEN bit of the SSP1CON3 register will enable writes to the SSP1BUF register, even if the previous byte has not been read. This allows the software to ignore data that may not apply to it.
30.2.5 SLAVE SELECT SYNCHRONIZATION

The Slave Select can also be used to synchronize communication. The Slave Select line is held high until the master device is ready to communicate. When the Slave Select line is pulled low, the slave knows that a new transmission is starting.

If the slave fails to receive the communication properly, it will be reset at the end of the transmission, when the Slave Select line returns to a high state. The slave is then ready to receive a new transmission when the Slave Select line is pulled low again. If the Slave Select line is not used, there is a risk that the slave will eventually become out of sync with the master. If the slave misses a bit, it will always be one bit off in future transmissions. Use of the Slave Select line allows the slave and master to align themselves at the beginning of each transmission.

The SS pin allows a Synchronous Slave mode. The SPI must be in Slave mode with SS pin control enabled (SSP1CON1<3:0> = 0100).

When the SS pin is low, transmission and reception are enabled and the SDO pin is driven.

When the SS pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

Note 1: When the SPI is in Slave mode with SS pin control enabled (SSP1CON1<3:0> = 0100), the SPI module will reset if the SS pin is set to VDD.

2: When the SPI is used in Slave mode with CKE set; the user must enable SS pin control.

3: While operated in SPI Slave mode the SMP bit of the SSPSTAT register must remain clear.

When the SPI module resets, the bit counter is forced to ‘0’. This can be done by either forcing the SS pin to a high level or clearing the SSPEN bit.

FIGURE 30-7: SPI DAISY-CHAIN CONNECTION
FIGURE 30-8: SLAVE SELECT SYNCHRONOUS WAVEFORM

- SS
- SCK (CKP = 0, CKE = 0)
- SCK (CKP = 1, CKE = 0)
- Write to SSP1BUF
- SSP1BUF to SSP1SR
- SSP1BUF to SSP1SR, Shift register SSP1SR and bit count are reset
- SDO
- SDI
- Input Sample
- SSP1IF Interrupt Flag
- SSP1SR to SSP1BUF
FIGURE 30-9: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)

FIGURE 30-10: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)
30.2.6  SPI OPERATION IN SLEEP MODE

In SPI Master mode, module clocks may be operating at a different speed than when in Full-Power mode; in the case of the Sleep mode, all clocks are halted.

Special care must be taken by the user when the MSSP clock is much faster than the system clock.

In Slave mode, when MSSP interrupts are enabled, after the master completes sending data, an MSSP interrupt will wake the controller from Sleep.

If an exit from Sleep mode is not desired, MSSP interrupts should be disabled.

In SPI Master mode, when the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the device wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in Sleep mode and data to be shifted into the SPI Transmit/Receive Shift register. When all eight bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device.

### TABLE 30-1: SUMMARY OF REGISTERS ASSOCIATED WITH SPI OPERATION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>ANSA5</td>
<td>ANSA4</td>
<td>ANSA3</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>126</td>
</tr>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
</tr>
<tr>
<td>ANSELD(1)</td>
<td>ANSD7</td>
<td>ANSD6</td>
<td>ANSD5</td>
<td>ANSD4</td>
<td>ANSD3</td>
<td>ANSD2</td>
<td>ANSD1</td>
<td>ANSD0</td>
<td>142</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
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<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>RxyPPS&lt;4:0&gt; 154</td>
</tr>
<tr>
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<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>SSPCLKPPS&lt;4:0&gt; 153</td>
</tr>
<tr>
<td>SSPDATPPS</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>SSPDATPPS&lt;4:0&gt; 153</td>
</tr>
<tr>
<td>SSPSSPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>SSPSSPPS&lt;4:0&gt; 153</td>
</tr>
<tr>
<td>SSP1BUF</td>
<td>Synchronous Serial Port Receive Buffer/Transmit Register</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>302*</td>
</tr>
<tr>
<td>SSP1CON1</td>
<td>WCOL</td>
<td>SSPOV</td>
<td>SSPEN</td>
<td>CKP</td>
<td>SSPM&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>349</td>
</tr>
<tr>
<td>SSP1CON3</td>
<td>ACKTIM</td>
<td>PCIE</td>
<td>SCIE</td>
<td>BOEN</td>
<td>SDAHT</td>
<td>SBCDE</td>
<td>AHEN</td>
<td>DHEN</td>
<td>352</td>
</tr>
<tr>
<td>SSP1STAT</td>
<td>SMP</td>
<td>CKE</td>
<td>D/A</td>
<td>P</td>
<td>S</td>
<td>R/W</td>
<td>UA</td>
<td>BF</td>
<td>348</td>
</tr>
<tr>
<td>TRISA</td>
<td>TRISA7</td>
<td>TRISA6</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>125</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
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<tr>
<td>TRISC</td>
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<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TRISD(1)</td>
<td>TRISD7</td>
<td>TRISD6</td>
<td>TRISD5</td>
<td>TRISD4</td>
<td>TRISD3</td>
<td>TRISD2</td>
<td>TRISD1</td>
<td>TRISD0</td>
<td>141</td>
</tr>
</tbody>
</table>

**Legend:** — = Unimplemented location, read as ‘0’. Shaded cells are not used by the MSSP in SPI mode.

*Page provides register information.

**Note 1:** PIC16(L)F1717/9 only.
30.3 I<sub>2</sub>C MODE OVERVIEW

The Inter-Integrated Circuit (I<sup>2</sup>C) bus is a multi-master serial data communication bus. Devices communicate in a master/slave environment where the master devices initiate the communication. A slave device is controlled through addressing.

The I<sup>2</sup>C bus specifies two signal connections:
- Serial Clock (SCL)
- Serial Data (SDA)

Figure 30-11 shows the block diagram of the MSSP module when operating in I<sup>2</sup>C mode.

Both the SCL and SDA connections are bidirectional open-drain lines, each requiring pull-up resistors for the supply voltage. Pulling the line to ground is considered a logical zero and letting the line float is considered a logical one.

Figure 30-11 shows a typical connection between two processors configured as master and slave devices.

The I<sup>2</sup>C bus can operate with one or more master devices and one or more slave devices.

There are four potential modes of operation for a given device:
- Master Transmit mode
  (master is transmitting data to a slave)
- Master Receive mode
  (master is receiving data from a slave)
- Slave Transmit mode
  (slave is transmitting data to a master)
- Slave Receive mode
  (slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device.

If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an ACK. The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

The Acknowledge bit (ACK) is an active-low signal, which holds the SDA line low to indicate to the transmitter that the slave device has received the transmitted data and is ready to receive more.

The transition of a data bit is always performed while the SCL line is held low. Transitions that occur while the SCL line is held high are used to indicate Start and Stop bits.

If the master intends to write to the slave, then it repeatedly sends out a byte of data, with the slave responding after each byte with an ACK bit. In this example, the master device is in Master Transmit mode and the slave is in Slave Receive mode.

If the master intends to read from the slave, then it repeatedly receives a byte of data from the slave, and responds after each byte with an ACK bit. In this example, the master device is in Master Receive mode and the slave is Slave Transmit mode.

On the last byte of data communicated, the master device may end the transmission by sending a Stop bit. If the master device is in Receive mode, it sends the Stop bit in place of the last ACK bit. A Stop bit is indicated by a low-to-high transition of the SDA line while the SCL line is held high.

In some cases, the master may want to maintain control of the bus and re-initiate another transmission. If so, the master device may send another Start bit in place of the Stop bit or last ACK bit when it is in receive mode.

The I<sup>2</sup>C bus specifies three message protocols:
- Single message where a master writes data to a slave.
- Single message where a master reads data from a slave.
- Combined message where a master initiates a minimum of two writes, or two reads, or a combination of writes and reads, to one or more slaves.
When one device is transmitting a logical one, or letting the line float, and a second device is transmitting a logical zero, or holding the line low, the first device can detect that the line is not a logical one. This detection, when used on the SCL line, is called clock stretching. Clock stretching gives slave devices a mechanism to control the flow of data. When this detection is used on the SDA line, it is called arbitration. Arbitration ensures that there is only one master device communicating at any single time.

30.3.1 CLOCK STRETCHING

When a slave device has not completed processing data, it can delay the transfer of more data through the process of clock stretching. An addressed slave device may hold the SCL clock line low after receiving or sending a bit, indicating that it is not yet ready to continue. The master that is communicating with the slave will attempt to raise the SCL line in order to transfer the next bit, but will detect that the clock line has not yet been released. Because the SCL connection is open-drain, the slave has the ability to hold that line low until it is ready to continue communicating.

Clock stretching allows receivers that cannot keep up with a transmitter to control the flow of incoming data.

30.3.2 ARBITRATION

Each master device must monitor the bus for Start and Stop bits. If the device detects that the bus is busy, it cannot begin a new message until the bus returns to an Idle state.

However, two master devices may try to initiate a transmission on or about the same time. When this occurs, the process of arbitration begins. Each transmitter checks the level of the SDA data line and compares it to the level that it expects to find. The first transmitter to observe that the two levels do not match, loses arbitration, and must stop transmitting on the SDA line.

For example, if one transmitter holds the SDA line to a logical one (lets it float) and a second transmitter holds it to a logical zero (pulls it low), the result is that the SDA line will be low. The first transmitter then observes that the level of the line is different than expected and concludes that another transmitter is communicating.

The first transmitter to notice this difference is the one that loses arbitration and must stop driving the SDA line. If this transmitter is also a master device, it also must stop driving the SCL line. It then can monitor the lines for a Stop condition before trying to reissue its transmission. In the meantime, the other device that has not noticed any difference between the expected and actual levels on the SDA line continues with its original transmission. It can do so without any complications, because so far, the transmission appears exactly as expected with no other transmitter disturbing the message.

Slave Transmit mode can also be arbitrated, when a master addresses multiple slaves, but this is less common.

If two master devices are sending a message to two different slave devices at the address stage, the master sending the lower slave address always wins arbitration. When two master devices send messages to the same slave address, and addresses can sometimes refer to multiple slaves, the arbitration process must continue into the data stage.

Arbitration usually occurs very rarely, but it is a necessary process for proper multi-master support.
30.4 **I²C MODE OPERATION**

All MSSP I²C communication is byte oriented and shifted out MSb first. Six SFR registers and two interrupt flags interface the module with the PIC® microcontroller and user software. Two pins, SDA and SCL, are exercised by the module to communicate with other external I²C devices.

30.4.1 **BYTE FORMAT**

All communication in I²C is done in 9-bit segments. A byte is sent from a master to a slave or vice-versa, followed by an Acknowledge bit sent back. After the eighth falling edge of the SCL line, the device outputting data on the SDA changes that pin to an input and reads in an acknowledge value on the next clock pulse.

The clock signal, SCL, is provided by the master. Data is valid to change while the SCL signal is low, and sampled on the rising edge of the clock. Changes on the SDA line while the SCL line is high define special conditions on the bus, explained below.

30.4.2 **DEFINITION OF I²C TERMINOLOGY**

There is language and terminology in the description of I²C communication that have definitions specific to I²C. That word usage is defined below and may be used in the rest of this document without explanation. Table 30-2 was adapted from the Philips I²C specification.

30.4.3 **SDA AND SCL PINS**

Selection of any I²C mode with the SSPEN bit set, forces the SCL and SDA pins to be open-drain. These pins should be set by the user to inputs by setting the appropriate TRIS bits.

| Note 1: | Data is tied to output zero when an I²C mode is enabled. |
| 2: | Any device pin can be selected for SDA and SCL functions with the PPS peripheral. These functions are bidirectional. The SDA input is selected with the SSPDATPPS registers. The SCL input is selected with the SSPCLKPPS registers. Outputs are selected with the RxyPPS registers. It is the user’s responsibility to make the selections so that both the input and the output for each function is on the same pin. |

30.4.4 **SDA HOLD TIME**

The hold time of the SDA pin is selected by the SDAHT bit of the SSP1CON3 register. Hold time is the time SDA is held valid after the falling edge of SCL. Setting the SDAHT bit selects a longer 300 ns minimum hold time and may help on buses with large capacitance.

**TABLE 30-2: I²C BUS TERMS**

<table>
<thead>
<tr>
<th>TERM</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>Transmitter</td>
<td>The device which shifts data out onto the bus.</td>
</tr>
<tr>
<td>Receiver</td>
<td>The device which shifts data in from the bus.</td>
</tr>
<tr>
<td>Master</td>
<td>The device that initiates a transfer, generates clock signals and terminates a transfer.</td>
</tr>
<tr>
<td>Slave</td>
<td>The device addressed by the master.</td>
</tr>
<tr>
<td>Multi-master</td>
<td>A bus with more than one device that can initiate data transfers.</td>
</tr>
<tr>
<td>Arbitration</td>
<td>Procedure to ensure that only one master at a time controls the bus. Winning arbitration ensures that the message is not corrupted.</td>
</tr>
<tr>
<td>Synchronization</td>
<td>Procedure to synchronize the clocks of two or more devices on the bus.</td>
</tr>
<tr>
<td>Idle</td>
<td>No master is controlling the bus, and both SDA and SCL lines are high.</td>
</tr>
<tr>
<td>Active</td>
<td>Any time one or more master devices are controlling the bus.</td>
</tr>
<tr>
<td>Addressed Slave</td>
<td>Slave device that has received a matching address and is actively being clocked by a master.</td>
</tr>
<tr>
<td>Matching Address</td>
<td>Address byte that is clocked into a slave that matches the value stored in SSP1ADD.</td>
</tr>
<tr>
<td>Write Request</td>
<td>Slave receives a matching address with R/W bit clear, and is ready to clock in data.</td>
</tr>
<tr>
<td>Read Request</td>
<td>Master sends an address byte with the R/W bit set, indicating that it wishes to clock data out of the Slave. This data is the next and all following bytes until a Restart or Stop.</td>
</tr>
<tr>
<td>Clock Stretching</td>
<td>When a device on the bus hold SCL low to stall communication.</td>
</tr>
<tr>
<td>Bus Collision</td>
<td>Any time the SDA line is sampled low by the module while it is outputting and expected high state.</td>
</tr>
</tbody>
</table>
30.4.5 START CONDITION

The I²C specification defines a Start condition as a transition of SDA from a high to a low state while SCL line is high. A Start condition is always generated by the master and signifies the transition of the bus from an Idle to an Active state. Figure 30-12 shows waveforms for Start and Stop conditions.

A bus collision can occur on a Start condition if the module samples the SDA line low before asserting it low. This does not conform to the I²C Specification that states no bus collision can occur on a Start.

30.4.6 STOP CONDITION

A Stop condition is a transition of the SDA line from low-to-high state while the SCL line is high.

Note: At least one SCL low time must appear before a Stop is valid, therefore, if the SDA line goes low then high again while the SCL line stays high, only the Start condition is detected.

30.4.7 RESTART CONDITION

A Restart is valid any time that a Stop would be valid. A master can issue a Restart if it wishes to hold the bus after terminating the current transfer. A Restart has the same effect on the slave that a Start would, resetting all slave logic and preparing it to clock in an address. The master may want to address the same or another slave. Figure 30-13 shows the waveform for a Restart condition.

In 10-bit Addressing Slave mode a Restart is required for the master to clock data out of the addressed slave. Once a slave has been fully addressed, matching both high and low address bytes, the master can issue a Restart and the high address byte with the R/W bit set. The slave logic will then hold the clock and prepare to clock out data.

After a full match with R/W clear in 10-bit mode, a prior match flag is set and maintained until a Stop condition, a high address with R/W clear, or high address match fails.

30.4.8 START/STOP CONDITION

INTERRUPT MASKING

The SCIE and PCIE bits of the SSP1CON3 register can enable the generation of an interrupt in Slave modes that do not typically support this function. Slave modes where interrupt on Start and Stop detect are already enabled, these bits will have no effect.
30.4.9 ACKNOWLEDGE SEQUENCE

The 9th SCL pulse for any transferred byte in I²C is dedicated as an Acknowledge. It allows receiving devices to respond back to the transmitter by pulling the SDA line low. The transmitter must release control of the line during this time to shift in the response. The Acknowledge (ACK) is an active-low signal, pulling the SDA line low indicates to the transmitter that the device has received the transmitted data and is ready to receive more.

The result of an ACK is placed in the ACKSTAT bit of the SSP1CON2 register.

Slave software, when the AHEN and DHEN bits are set, allow the user to set the ACK value sent back to the transmitter. The ACKDT bit of the SSP1CON2 register is set/cleared to determine the response.

Slave hardware will generate an ACK response if the AHEN and DHEN bits of the SSP1CON3 register are clear.

There are certain conditions where an ACK will not be sent by the slave. If the BF bit of the SSP1STAT register or the SSPOV bit of the SSP1CON1 register are set when a byte is received.

When the module is addressed, after the eighth falling edge of SCL on the bus, the ACKTIM bit of the SSP1CON3 register is set. The ACKTIM bit indicates the acknowledge time of the active bus. The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is enabled.

30.5 I²C SLAVE MODE OPERATION

The MSSP Slave mode operates in one of four modes selected by the SSPM bits of SSP1CON1 register. The modes can be divided into 7-bit and 10-bit Addressing mode. 10-bit Addressing modes operate the same as 7-bit with some additional overhead for handling the larger addresses.

Modes with Start and Stop bit interrupts operate the same as the other modes with SSP1IF additionally getting set upon detection of a Start, Restart, or Stop condition.

30.5.1 SLAVE MODE ADDRESSES

The SSP1ADD register (Register 30-6) contains the Slave mode address. The first byte received after a Start or Restart condition is compared against the value stored in this register. If the byte matches, the value is loaded into the SSP1BUF register and an interrupt is generated. If the value does not match, the module goes idle and no indication is given to the software that anything happened.

The SSP Mask register (Register 30-5) affects the address matching process. See Section 30.5.9 “SSP Mask Register” for more information.

30.5.1.1 I²C Slave 7-Bit Addressing Mode

In 7-bit Addressing mode, the LSb of the received data byte is ignored when determining if there is an address match.

30.5.1.2 I²C Slave 10-Bit Addressing Mode

In 10-bit Addressing mode, the first received byte is compared to the binary value of ‘1 1 1 1 0 A9 A8 0’. A9 and A8 are the two MSb’s of the 10-bit address and stored in bits 2 and 1 of the SSP1ADD register.

After the acknowledge of the high byte the UA bit is set and SCL is held low until the user updates SSP1ADD with the low address. The low address byte is clocked in and all eight bits are compared to the low address value in SSP1ADD. Even if there is not an address match; SSP1IF and UA are set, and SCL is held low until SSP1ADD is updated to receive a high byte again. When SSP1ADD is updated the UA bit is cleared. This ensures the module is ready to receive the high address byte on the next communication.

A high and low address match as a write request is required at the start of all 10-bit addressing communication. A transmission can be initiated by issuing a Restart once the slave is addressed, and clocking in the high address with the R/W bit set. The slave hardware will then acknowledge the read request and prepare to clock out data. This is only valid for a slave after it has received a complete high and low address byte match.
30.5.2 SLAVE RECEPTION

When the \( R/W \) bit of a matching received address byte is clear, the \( R/W \) bit of the SSP1STAT register is cleared. The received address is loaded into the SSP1BUF register and acknowledged.

When the overflow condition exists for a received address, then read Acknowledge is given. An overflow condition is defined as either bit BF of the SSP1STAT register is set, or bit SSPOV of the SSP1CON1 register is set. The BOEN bit of the SSP1CON3 register modifies this operation. For more information see Register 30-4.

An MSSP interrupt is generated for each transferred data byte. Flag bit, SSP1IF, must be cleared by software.

When the SEN bit of the SSP1CON2 register is set, SCL will be held low (clock stretch) following each received byte. The clock must be released by setting the CKP bit of the SSP1CON1 register, except sometimes in 10-bit mode. See Section 30.5.6.2 “10-Bit Addressing Mode” for more detail.

30.5.2.1 7-Bit Addressing Reception

This section describes a standard sequence of events for the MSSP module configured as an \( \text{I}^2\text{C} \) slave in 7-bit Addressing mode. Figure 30-14 and Figure 30-15 is used as a visual reference for this description.

This is a step by step process of what typically must be done to accomplish \( \text{I}^2\text{C} \) communication.

1. Start bit detected.
2. S bit of SSP1STAT is set; SSP1IF is set if interrupt on Start detect is enabled.
3. Matching address with \( R/W \) bit clear is received.
4. The slave pulls SDA low sending an ACK to the master, and sets SSP1IF bit.
5. Software clears the SSP1IF bit.
6. Slave can look at the ACKTIM bit of the SSP1CON3 register to determine if the SSP1IF was after or before the ACK.
7. Slave reads the address value from SSP1BUF, clearing the BF flag.
8. Slave sets ACK value clocked out to the master by setting ACKDT.
9. Slave releases the clock by setting CKP.
10. SSP1IF is set after an ACK, not after a NACK.

Note: SSP1IF is still set after the 9th falling edge of SCL even if there is no clock stretching and BF has been cleared. Only if NACK is sent to master is SSP1IF not set

30.5.2.2 7-Bit Reception with AHEN and DHEN

Slave device reception with AHEN and DHEN set operate the same as without these options with extra interrupts and clock stretching added after the eighth falling edge of SCL. These additional interrupts allow the slave software to decide whether it wants to ACK the receive address or data byte, rather than the hardware. This functionality adds support for PMBus™ that was not present on previous versions of this module.

This list describes the steps that need to be taken by slave software to use these options for \( \text{I}^2\text{C} \) communication. Figure 30-16 displays a module using both address and data holding. Figure 30-17 includes the operation with the SEN bit of the SSP1CON2 register set.

1. S bit of SSP1STAT is set; SSP1IF is set if interrupt on Start detect is enabled.
2. Matching address with \( R/W \) bit clear is clocked in. SSP1IF is set and CKP cleared after the eighth falling edge of SCL.
3. Slave clears the SSP1IF.
4. Slave can look at the ACKTIM bit of the SSP1CON3 register to determine if the SSP1IF was after or before the ACK.
5. Slave reads the address value from SSP1BUF, clearing the BF flag.
6. Slave sets ACK value clocked out to the master by setting ACKDT.
7. Slave releases the clock by setting CKP.
8. SSP1IF is set after an ACK, not after a NACK.
9. If SEN = 1, the slave hardware will stretch the clock after the ACK.
10. Slave clears SSP1IF.
11. SSP1IF set and CKP cleared after eighth falling edge of SCL for a received data byte.
12. Slave looks at ACKTIM bit of SSP1CON3 to determine the source of the interrupt.
13. Slave reads the received data from SSP1BUF clearing BF.
14. Steps 7-14 are the same for each received data byte.
15. Communication is ended by either the slave sending an ACK = 1, or the master sending a Stop condition. If a Stop is sent and Interrupt on Stop Detect is disabled, the slave will only know by polling the P bit of the SSTOPSTAT register.

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FIGURE 30-14: \( {\text{i}^2\text{C}} \) SLAVE, 7-BIT ADDRESS, RECEPTION (\( \text{SEN} = 0, \text{AHEN} = 0, \text{DHEN} = 0 \))
FIGURE 30-15: I$^2$C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 0, DHEN = 0)
FIGURE 30-16: \( \text{i}^2\text{C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 1, DHEN = 1)} \)
FIGURE 30-17: \( i^2 C \) SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 1, DHEN = 1)

- Master sends Stop condition.
- Master releases SDA to slave for ACK sequence.
- Master sends ACK.
- Slave software clears ACKDT to ACK.
- Received address is loaded into SSP1BUF.
- ACKTIM is cleared by software.
- SSP1BUF can be read anytime before next byte is loaded.
- Release SCL on 9th rising edge of SCL.
- Received data is available on SSP1BUF.
- Slave sends not ACK.
- ACKTIM is set by hardware on 8th falling edge of SCL.
- When DHEN = 1, on the 8th falling edge of a received data byte, CKP is cleared.
- When AHEN = 1, on the 8th falling edge of an address byte, CKP is cleared.
- When R/W = 0, master releases SDA to slave for ACK sequence.
- No interrupt after if not ACK from slave.
- ACKTIM is not cleared if not ACK.
- Stop condition.
- No interrupt after.
- Not ACK.
- No interrupt after.
30.5.3 SLAVE TRANSMISSION

When the R/W bit of the incoming address byte is set and an address match occurs, the R/W bit of the SSP1STAT register is set. The received address is loaded into the SSP1BUF register, and an ACK pulse is sent by the slave on the ninth bit.

Following the ACK, slave hardware clears the CKP bit and the SCL pin is held low (see Section 30.5.6 “Clock Stretching” for more detail). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data.

The transmit data must be loaded into the SSP1BUF register which also loads the SSP1SR register. Then the SCL pin should be released by setting the CKP bit of the SSP1CON1 register. The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time.

The ACK pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. This ACK value is copied to the ACKSTAT bit of the SSP1CON2 register. If ACKSTAT is set (not ACK), then the data transfer is complete. In this case, when the not ACK is latched by the slave, the slave goes idle and waits for another occurrence of the Start bit. If the SDA line was low (ACK), the next transmit data must be loaded into the SSP1BUF register. Again, the SCL pin must be released by setting bit CKP.

An MSSP interrupt is generated for each data transfer byte. The SSP1IF bit must be cleared by software and the SSP1STAT register is used to determine the status of the byte. The SSP1IF bit is set on the falling edge of the ninth clock pulse.

30.5.3.1 Slave Mode Bus Collision

A slave receives a Read request and begins shifting data out on the SDA line. If a bus collision is detected and the SBCDE bit of the SSP1CON3 register is set, the BCL1IF bit of the PIR2 register is set. Once a bus collision is detected, the slave goes idle and waits to be addressed again. User software can use the BCL1IF bit to handle a slave bus collision.
30.5.3.2 7-Bit Transmission

A master device can transmit a read request to a slave, and then clock data out of the slave. The list below outlines what software for a slave will need to do to accomplish a standard transmission. Figure 30-18 can be used as a reference to this list.

1. Master sends a Start condition on SDA and SCL.
2. S bit of SSP1STAT is set; SSP1IF is set if interrupt on Start detect is enabled.
3. Matching address with R/W bit set is received by the Slave setting SSP1IF bit.
4. Slave hardware generates an \( \overline{\text{ACK}} \) and sets SSP1IF.
5. SSP1IF bit is cleared by user.
6. Software reads the received address from SSP1BUF, clearing BF.
7. R/W is set so CKP was automatically cleared after the \( \overline{\text{ACK}} \).
8. The slave software loads the transmit data into SSP1BUF.
9. CKP bit is set releasing SCL, allowing the master to clock the data out of the slave.
10. SSP1IF is set after the \( \overline{\text{ACK}} \) response from the master is loaded into the ACKSTAT register.
11. SSP1IF bit is cleared.
12. The slave software checks the ACKSTAT bit to see if the master wants to clock out more data.

**Note 1:** If the master \( \overline{\text{ACK}} \)s the clock will be stretched.

**Note 2:** ACKSTAT is the only bit updated on the rising edge of SCL (9th) rather than the falling.

13. Steps 9-13 are repeated for each transmitted byte.
14. If the master sends a not \( \overline{\text{ACK}} \); the clock is not held, but SSP1IF is still set.
15. The master sends a Restart condition or a Stop.
16. The slave is no longer addressed.
FIGURE 30-18: I²C SLAVE, 7-BIT ADDRESS, TRANSMISSION (AHEN = 0)

- Receiving Address
- Automatic
- Transmitting Data
- Automatic

Master sends Stop condition

- SF
- SP
- ACK
- BF
- CKP
- ACKSTAT
- R/W
- D/A

Received address is read from SSP1BUF when R/W = 1

When R/W is set to 0, address byte is compared to 7th SCL falling edge

Indicates an address has been received

Masters not ACK

ACK is copied into ACKSTAT

CKP is not held for not ACK

BF is automatically cleared after 8th falling edge of SCL

Data to transmit is loaded into SSP1BUF

Set by software

Cleared by software

Masters send Stop condition
30.5.3.3 7-Bit Transmission with Address Hold Enabled

Setting the AHEN bit of the SSP1CON3 register enables additional clock stretching and interrupt generation after the eighth falling edge of a received matching address. Once a matching address has been clocked in, CKP is cleared and the SSP1IF interrupt is set.

Figure 30-19 displays a standard waveform of a 7-bit address slave transmission with AHEN enabled.

1. Bus starts idle.
2. Master sends Start condition; the S bit of SSP1STAT is set; SSP1IF is set if interrupt on Start detect is enabled.
3. Master sends matching address with R/W bit set. After the eighth falling edge of the SCL line the CKP bit is cleared and SSP1IF interrupt is generated.
4. Slave software clears SSP1IF.
5. Slave software reads ACKTIM bit of SSP1CON3 register, and R/W and D/A of the SSP1STAT register to determine the source of the interrupt.
6. Slave reads the address value from the SSP1BUF register clearing the BF bit.
7. Slave software decides from this information if it wishes to ACK or not ACK and sets the ACKDT bit of the SSP1CON2 register accordingly.
8. Slave sets the CKP bit releasing SCL.
9. Master clocks in the ACK value from the slave.
10. Slave hardware automatically clears the CKP bit and sets SSP1IF after the ACK if the R/W bit is set.
11. Slave software clears SSP1IF.
12. Slave loads value to transmit to the master into SSP1BUF setting the BF bit.

**Note:** SSP1BUF cannot be loaded until after the ACK.

13. Slave sets the CKP bit releasing the clock.
14. Master clocks out the data from the slave and sends an ACK value on the 9th SCL pulse.
15. Slave hardware copies the ACK value into the ACKSTAT bit of the SSP1CON2 register.
16. Steps 10-15 are repeated for each byte transmitted to the master from the slave.
17. If the master sends a not ACK the slave releases the bus allowing the master to send a Stop and end the communication.

**Note:** Master must send a not ACK on the last byte to ensure that the slave releases the SCL line to receive a Stop.
FIGURE 30-19: I2C SLAVE, 7-BIT ADDRESS, TRANSMISSION (AHEN = 1)

- Receiving Address
  - Address is read from SSP1BUF
  - BF is automatically cleared after 8th falling edge of SCL
  - Data to transmit is loaded into SSP1BUF

- Transmitting Data
  - ACKDT is cleared by software
  - ACKDT is automatically set at 8th falling edge of SCL
  - BF is automatically cleared after 8th falling edge of SCL
  - Master's ACK response is copied to SSP1STAT

- ACKSTAT
  - Master releases SDA to slave for ACK sequence

- CKP
  - When AHEN = 1:
    - CKP is cleared by hardware after receiving matching address
      - When R/W = 1:
        - OKP is always cleared after ACK
  - Set by software, releases SCL
  - CKP not cleared after not ACK

- ACKTIM
  - ACKTIM is set on 8th falling edge of SCL
  - ACKTIM is cleared on 9th rising edge of SCL

- Master sends Stop condition

Notes:
- Master sends Stop condition
- R/W = 1
- D/A

Legend:
- SDA
- SCL
- SSP1IF
- BF
- ACKDT
- ACKSTAT
- CKP
- R/W
- D/A

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30.5.4 SLAVE MODE 10-BIT ADDRESS RECEPTION

This section describes a standard sequence of events for the MSSP module configured as an I²C slave in 10-bit Addressing mode. Figure 30-20 is used as a visual reference for this description.

This is a step by step process of what must be done by slave software to accomplish I²C communication.

1. Bus starts idle.
2. Master sends Start condition; S bit of SSP1STAT is set; SSP1IF is set if interrupt on Start detect is enabled.
3. Master sends matching high address with R/W bit clear; UA bit of the SSP1STAT register is set.
4. Slave sends ACK and SSP1IF is set.
5. Software clears the SSP1IF bit.
6. Software reads received address from SSP1BUF clearing the BF flag.
7. Slave loads low address into SSP1ADD, releasing SCL.
8. Master sends matching low address byte to the slave; UA bit is set.

**Note:** Updates to the SSP1ADD register are not allowed until after the ACK sequence.

9. Slave sends ACK and SSP1IF is set.

**Note:** If the low address does not match, SSP1IF and UA are still set so that the slave software can set SSP1ADD back to the high address. BF is not set because there is no match. CKP is unaffected.

10. Slave clears SSP1IF.
11. Slave reads the received matching address from SSP1BUF clearing BF.
12. Slave loads high address into SSP1ADD.
13. Master clocks a data byte to the slave and clocks out the slaves ACK on the 9th SCL pulse; SSP1IF is set.
14. If SEN bit of SSP1CON2 is set, CKP is cleared by hardware and the clock is stretched.
15. Slave clears SSP1IF.
16. Slave reads the received byte from SSP1BUF clearing BF.
17. If SEN is set the slave sets CKP to release the SCL.
18. Steps 13-17 repeat for each received byte.
19. Master sends Stop to end the transmission.

30.5.5 10-BIT ADDRESSING WITH ADDRESS OR DATA HOLD

Reception using 10-bit addressing with AHEN or DHEN set is the same as with 7-bit modes. The only difference is the need to update the SSP1ADD register using the UA bit. All functionality, specifically when the CKP bit is cleared and SCL line is held low are the same. Figure 30-21 can be used as a reference of a slave in 10-bit addressing with AHEN set.

Figure 30-22 shows a standard waveform for a slave transmitter in 10-bit Addressing mode.

Note: Updates to the SSP1ADD register are not
Note: If the low address does not match, SSP1IF and UA are still set so that the slave software can set SSP1ADD back to the high address. BF is not set because there is no match. CKP is unaffected.
FIGURE 30-20:  \( \text{I}^2\text{C} \) SLAVE, 10-BIT ADDRESS, RECEPTION (\( \text{SEN} = 1 \), \( \text{AHEN} = 0 \), \( \text{DHEN} = 0 \))

- Master sends Stop condition
- Receive First Address Byte
- Receive Second Address Byte
- ACK
- Receive Data
- ACK
- Receive Data
- ACK

- When \( \text{SEN} = 1 \);
- \( \text{CKP} \) is cleared after 9th falling edge of received byte
- \( \text{SCL} \) is held low
- Software updates SSP1ADD
- SSP1BUF
- \( \text{BF} \)
- If address matches
- SSP1ADD is loaded into SSP1BUF
- \( \text{SSP1IF} \)
- \( \text{UA} \)
- \( \text{OKP} \)
- \( \text{SCL} \) is held low
- Data is read from SSP1BUF
- \( \text{SDA} \)
- \( \text{SCL} \)
- \( \text{UA} \)
- \( \text{CKP} \)

- Set by hardware, released by \( \text{CKP} \)
- "When \( \text{OKP} = 1 \), \( \text{CKP} \) is cleared after 9th falling edge of received byte"
FIGURE 30-21:  I²C SLAVE, 10-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 1, DHEN = 0)
FIGURE 30-22:  \( \text{i}^2\text{C} \) SLAVE, 10-BIT ADDRESS, TRANSMISSION (\( \text{SEN} = 0 \), \( \text{AHEN} = 0 \), \( \text{DHEN} = 0 \))
30.5.6 CLOCK STRETCHING

Clock stretching occurs when a device on the bus holds the SCL line low, effectively pausing communication. The slave may stretch the clock to allow more time to handle data or prepare a response for the master device. A master device is not concerned with stretching as anytime it is active on the bus and not transferring data it is stretching. Any stretching done by a slave is invisible to the master software and handled by the hardware that generates SCL.

The CKP bit of the SSP1CON1 register is used to control stretching in software. Any time the CKP bit is cleared, the module will wait for the SCL line to go low and then hold it. Setting CKP will release SCL and allow more communication.

30.5.6.1 Normal Clock Stretching

Following an ACK if the R/W bit of SSP1STAT is set, a read request, the slave hardware will clear CKP. This allows the slave time to update SSP1BUF with data to transfer to the master. If the SEN bit of SSP1CON2 is set, the slave hardware will always stretch the clock after the ACK sequence. Once the slave is ready; CKP is set by software and communication resumes.

**Note 1:** The BF bit has no effect on if the clock will be stretched or not. This is different than previous versions of the module that would not stretch the clock, clear CKP, if SSP1BUF was read before the 9th falling edge of SCL.

**2:** Previous versions of the module did not stretch the clock for a transmission if SSP1BUF was loaded before the 9th falling edge of SCL. It is now always cleared for read requests.

30.5.6.2 10-Bit Addressing Mode

In 10-bit Addressing mode, when the UA bit is set the clock is always stretched. This is the only time the SCL is stretched without CKP being cleared. SCL is released immediately after a write to SSP1ADD.

**Note:** Previous versions of the module did not stretch the clock if the second address byte did not match.

30.5.6.3 Byte NACKing

When AHEN bit of SSP1CON3 is set; CKP is cleared by hardware after the eighth falling edge of SCL for a received matching address byte. When DHEN bit of SSP1CON3 is set; CKP is cleared after the eighth falling edge of SCL for received data.

Stretching after the eighth falling edge of SCL allows the slave to look at the received address or data and decide if it wants to ACK the received data.
30.5.7 CLOCK SYNCHRONIZATION AND THE CKP BIT

Any time the CKP bit is cleared, the module will wait for the SCL line to go low and then hold it. However, clearing the CKP bit will not assert the SCL output low until the SCL output is already sampled low. Therefore, the CKP bit will not assert the SCL line until an external I^2C master device has already asserted the SCL line. The SCL output will remain low until the CKP bit is set and all other devices on the I^2C bus have released SCL. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCL (see Figure 30-23).

FIGURE 30-23: CLOCK SYNCHRONIZATION TIMING
30.5.8 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I2C bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master device. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an acknowledge.

The general call address is a reserved address in the I2C protocol, defined as address 0x00. When the GCEN bit of the SSP1CON2 register is set, the slave module will automatically ACK the reception of this address regardless of the value stored in SSP1ADD. After the slave clocks in an address of all zeros with the R/W bit clear, an interrupt is generated and slave software can read SSP1BUF and respond. Figure 30-24 shows a general call reception sequence.

In 10-bit Address mode, the UA bit will not be set on the reception of the general call address. The slave will prepare to receive the second byte as data, just as it would in 7-bit mode.

If the AHEN bit of the SSP1CON3 register is set, just as with any other address reception, the slave hardware will stretch the clock after the eighth falling edge of SCL. The slave must then set its ACKDT value and release the clock with communication progressing as it would normally.

30.5.9 SSP MASK REGISTER

An SSP Mask (SSP1MSK) register (Register 30-5) is available in I2C Slave mode as a mask for the value held in the SSP1SR register during an address comparison operation. A zero (‘0’) bit in the SSP1MSK register has the effect of making the corresponding bit of the received address a “don’t care”.

This register is reset to all ‘1’s upon any Reset condition and, therefore, has no effect on standard SSP operation until written with a mask value.

The SSP Mask register is active during:

- 7-bit Address mode: address compare of SSP1ADD<7:1>.
- 10-bit Address mode: address compare of SSP1ADD<7:0> only. The SSP mask has no effect during the reception of the first (high) byte of the address.
30.6 I²C Master Mode

Master mode is enabled by setting and clearing the appropriate SSPM bits in the SSP1CON1 register and by setting the SSPEN bit. In Master mode, the SDA and SCK pins must be configured as inputs. The MSSP peripheral hardware will override the output driver TRIS controls when necessary to drive the pins low.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I²C bus may be taken when the P bit is set, or the bus is idle.

In Firmware Controlled Master mode, user code conducts all I²C bus operations based on Start and Stop bit condition detection. Start and Stop condition detection is the only active circuitry in this mode. All other communication is done by the user software directly manipulating the SDA and SCL lines.

The following events will cause the SSP Interrupt Flag bit, SSP1IF, to be set (SSP interrupt, if enabled):

- Start condition detected
- Stop condition detected
- Data transfer byte transmitted/received
- Acknowledge transmitted/received
- Repeated Start generated

Note 1: The MSSP module, when configured in I²C Master mode, does not allow queuing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSP1BUF register to initiate transmission before the Start condition is complete. In this case, the SSP1BUF will not be written to and the WCOL bit will be set, indicating that a write to the SSP1BUF did not occur.

Note 2: Master mode suspends Start/Stop detection when sending the Start/Stop condition by means of the SEN/PEN control bits. The SSPxIF bit is set at the end of the Start/Stop generation when hardware clears the control bit.

30.6.1 I²C MASTER MODE OPERATION

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic ‘0’. Serial data is transmitted eight bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic ‘1’. Thus, the first byte transmitted is a 7-bit slave address followed by a ‘1’ to indicate the receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received eight bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

A Baud Rate Generator is used to set the clock frequency output on SCL. See Section 30.7 “Baud Rate Generator” for more detail.
30.6.2 CLOCK ARBITRATION

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, releases the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 30-25).

FIGURE 30-25: BAUD RATE GENERATOR TIMING WITH CLOCK ARBITRATION

30.6.3 WCOL STATUS FLAG

If the user writes the SSP1BUF when a Start, Restart, Stop, Receive or Transmit sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write does not occur). Any time the WCOL bit is set it indicates that an action on SSP1BUF was attempted while the module was not idle.

Note: Because queuing of events is not allowed, writing to the lower five bits of SSP1CON2 is disabled until the Start condition is complete.
30.6.4 \(^2\)C MASTER MODE START

CONDITION TIMING

To initiate a Start condition (Figure 30-26), the user sets the Start Enable bit, SEN bit of the SSP1CON2 register. If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (TBRG), the SDA pin is driven low. The action of the SDA being driven low while SCL is high is the Start condition and causes the S bit of the SSP1STAT register to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit of the SSP1CON2 register will be automatically cleared by hardware; the Baud Rate Generator is suspended, leaving the SDA line held low and the Start condition is complete.

**Note 1:** If at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition, the SCL line is sampled low before the SDA line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCL1IF, is set, the Start condition is aborted and the \(^2\)C module is reset into its Idle state.

2: The Philips \(^2\)C specification states that a bus collision cannot occur on a Start.

**FIGURE 30-26:** FIRST START BIT TIMING

![First Start Bit Timing Diagram](image)
30.6.5 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition (Figure 30-27) occurs when the RSEN bit of the SSP1CON2 register is programmed high and the master state machine is no longer active. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be deasserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA = 0) for one TBRG while SCL is high. SCL is asserted low. Following this, the RSEN bit of the SSP1CON2 register will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit of the SSP1STAT register will be set. The SSP1IF bit will not be set until the Baud Rate Generator has timed out.

Note 1: If RSEN is programmed while any other event is in progress, it will not take effect.

2: A bus collision during the Repeated Start condition occurs if:
   • SDA is sampled low when SCL goes from low-to-high.
   • SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data ‘1’.

FIGURE 30-27: REPEATED START CONDITION WAVEFORM
30.6.6 I²C MASTER MODE

TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSP1BUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted. SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high. When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an ACK bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKSTAT bit on the rising edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared. If not, the bit is set. After the ninth clock, the SSP1IF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSP1BUF, leaving SCL low and SDA unchanged (Figure 30-28).

After the write to the SSP1BUF, each bit of the address will be shifted out on the falling edge of SCL until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will release the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT Status bit of the SSP1CON2 register. Following the falling edge of the ninth clock transmission of the address, the SSP1IF is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSP1BUF takes place, holding SCL low and allowing SDA to float.

30.6.6.1 BF Status Flag

In Transmit mode, the BF bit of the SSP1STAT register is set when the CPU writes to SSP1BUF and is cleared when all eight bits are shifted out.

30.6.6.2 WCOL Status Flag

If the user writes the SSP1BUF when a transmit is already in progress (i.e., SSP1SR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

WCOL must be cleared by software before the next transmission.

30.6.6.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit of the SSP1CON2 register is cleared when the slave has sent an Acknowledge (ACK = 0) and is set when the slave does not Acknowledge (ACK = 1). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

30.6.6.4 Typical Transmit Sequence

1. The user generates a Start condition by setting the SEN bit of the SSP1CON2 register.
2. SSP1IF is set by hardware on completion of the Start.
3. SSP1IF is cleared by software.
4. The MSSP module will wait the required start time before any other operation takes place.
5. The user loads the SSP1BUF with the slave address to transmit.
6. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSP1BUF is written to.
7. The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
8. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSP1IF bit.
9. The user loads the SSP1BUF with eight bits of data.
10. Data is shifted out the SDA pin until all eight bits are transmitted.
11. The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
12. Steps 8-11 are repeated for all transmitted data bytes.
13. The user generates a Stop or Restart condition by setting the PEN or RSEN bits of the SSP1CON2 register. Interrupt is generated once the Stop/Restart condition is complete.
FIGURE 30-28: I²C MASTER MODE WAVEFORM (TRANSMISSION, 7 OR 10-BIT ADDRESS)
30.6.7  I²C MASTER MODE RECEPTION

Master mode reception (Figure 30-29) is enabled by programming the Receive Enable bit, RCEN bit of the SSP1CON2 register.

Note: The MSSP module must be in an Idle state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting and on each rollover, the state of the SCL pin changes (high-to-low/low-to-high) and data is shifted into the SSP1SR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSP1SR are loaded into the SSP1BUF, the BF flag bit is set, the SSP1IF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable, ACKEN bit of the SSP1CON2 register.

30.6.7.1 BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSP1BUF from SSP1SR. It is cleared when the SSP1BUF register is read.

30.6.7.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when eight bits are received into the SSP1SR and the BF flag bit is already set from a previous reception.

30.6.7.3 WCOL Status Flag

If the user writes the SSP1BUF when a receive is already in progress (i.e., SSP1SR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

30.6.7.4 Typical Receive Sequence

1. The user generates a Start condition by setting the SEN bit of the SSP1CON2 register.
2. SSP1IF is set by hardware on completion of the Start.
3. SSP1IF is cleared by software.
4. User writes SSP1BUF with the slave address to transmit and the R/W bit set.
5. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSP1BUF is written to.
6. The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
7. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSP1IF bit.
8. User sets the RCEN bit of the SSP1CON2 register and the master clocks in a byte from the slave.
9. After the eighth falling edge of SCL, SSP1IF and BF are set.
10. Master clears SSP1IF and reads the received byte from SSPUF, clears BF.
11. Master sets ACK value sent to slave in ACKDT bit of the SSP1CON2 register and initiates the ACK by setting the ACKEN bit.
12. Master’s ACK is clocked out to the slave and SSP1IF is set.
13. User clears SSP1IF.
14. Steps 8-13 are repeated for each received byte from the slave.
15. Master sends a not ACK or Stop to end communication.
FIGURE 30-29: \(^2\)C MASTER MODE WAVEFORM (RECEPTION, 7-BIT ADDRESS)
30.6.8 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit, ACKEN bit of the SSP1CON2 register. When this bit is set, the SCL pin is pulled low and the contents of the Acknowledge data bit are presented on the SDA pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCL pin is deasserted (pulled high). When the SCL pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG. The SCL pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into Idle mode (Figure 30-30).

30.6.9 STOP CONDITION TIMING

A Stop bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN bit of the SSP1CON2 register. At the end of a receive/transmit, the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the Baud Rate Generator is reloaded and counts down to '0'. When the Baud Rate Generator times out, the SCL pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDA pin will be deasserted. When the SDA pin is sampled high while SCL is high, the P bit of the SSP1STAT register is set. A TBRG later, the PEN bit is cleared and the SSP1IF bit is set (Figure 30-31).

30.6.8.1 WCOL Status Flag

If the user writes the SSP1BUF when an Acknowledge sequence is in progress, then WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

FIGURE 30-30: ACKNOWLEDGE SEQUENCE WAVEFORM

![Acknowledge Sequence Waveform Diagram]

Note: TBRG = one Baud Rate Generator period.

FIGURE 30-31: STOP CONDITION RECEIVE OR TRANSMIT MODE

![Stop Condition Receive or Transmit Mode Diagram]

Note: TBRG = one Baud Rate Generator period.
30.6.10 SLEEP OPERATION
While in Sleep mode, the \( \text{I}^{\text{2}}\text{C} \) slave module can receive addresses or data and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

30.6.11 EFFECTS OF A RESET
A Reset disables the MSSP module and terminates the current transfer.

30.6.12 MULTI-MASTER MODE
In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the \( \text{I}^{\text{2}}\text{C} \) bus may be taken when the P bit of the SSP1STAT register is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the SSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDA line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed by hardware with the result placed in the BCL1IF bit.

The states where arbitration can be lost are:
- Address Transfer
- Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

30.6.13 MULTI-MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION
Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a ‘1’ on SDA, by letting SDA float high and another master asserts a ‘0’. When the SCL pin floats high, data should be stable. If the expected data on SDA is a ‘1’ and the data sampled on the SDA pin is ‘0’, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCL1IF and reset the \( \text{I}^{\text{2}}\text{C} \) port to its Idle state (Figure 30-32).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDA and SCL lines are deasserted and the SSP1BUF can be written to. When the user services the bus collision Interrupt Service Routine and if the \( \text{I}^{\text{2}}\text{C} \) bus is free, the user can resume communication by asserting a Start condition.

If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are deasserted and the respective control bits in the SSP1CON2 register are cleared. When the user services the bus collision Interrupt Service Routine and if the \( \text{I}^{\text{2}}\text{C} \) bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDA and SCL pins. If a Stop condition occurs, the SSP1IF bit will be set. A write to the SSP1BUF will start the transmission of data at the first data bit, regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the \( \text{I}^{\text{2}}\text{C} \) bus can be taken when the P bit is set in the SSP1STAT register, or the bus is Idle and the S and P bits are cleared.

FIGURE 30-32: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE

![Bus Collision Timing Diagram](Image)

- Data changes while SCL = 0
- SDA line pulled low by another source
- Sample SDA. While SCL is high, data does not match what is driven by the master.
- Bus collision has occurred.
- SDA released by master
- Set bus collision interrupt (BCL1IF)
30.6.13.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 30-33).
b) SCL is sampled low before SDA is asserted low (Figure 30-34).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:
• the Start condition is aborted,
• the BCL1IF flag is set and
• the MSSP module is reset to its Idle state (Figure 30-33).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded and counts down. If the SCL pin is sampled low while SDA is high, a bus collision occurs because it is assumed that another master is attempting to drive a data ‘1’ during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 30-35). If, however, a ‘1’ is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to zero; if the SCL pin is sampled as ‘0’ during this time, a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

FIGURE 30-33: BUS COLLISION DURING START CONDITION (SDA ONLY)
FIGURE 30-34: BUS COLLISION DURING START CONDITION (SCL = 0)

SDA = 0, SCL = 1

SCL = 0 before SDA = 0, SCL = 1

SEN

SCL = 0 before SDA = 0, SCL = 1

BCL1IF

Interrupt cleared by software

S

‘0’

SSP1IF

‘0’

FIGURE 30-35: BRG RESET DUE TO SDA ARBITRATION DURING START CONDITION

SDA = 0, SCL = 1

Set S

Set S

Set S

Set S

SSP1IF

Interrupts cleared by software

SDA

SDA pulled low by other master

Reset BRG and assert SDA.

SCL

S

SEN

Set SEN, enable Start sequence if SDA = 1, SCL = 1

‘0’

BCL1IF

‘0’

SDA = 0, SCL = 1, set SSP1IF
30.6.13.2  Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

a) A low level is sampled on SDA when SCL goes from low level to high level (Case 1).

b) SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data ‘1’ (Case 2).

When the user releases SDA and the pin is allowed to float high, the BRG is loaded with SSP1ADD and counts down to zero. The SCL pin is then deasserted and when sampled high, the SDA pin is sampled.

If SDA is low, a bus collision has occurred (i.e., another master is attempting to transmit a data ‘0’, Figure 30-36). If SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high-to-low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data ‘1’ during the Repeated Start condition, see Figure 30-37.

If, at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.

FIGURE 30-36:  BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)

FIGURE 30-37:  BUS COLLISION DURING REPEATED START CONDITION (CASE 2)
30.6.13.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

a) After the SDA pin has been deasserted and allowed to float high, SDA is sampled low after the BRG has timed out (Case 1).

b) After the SCL pin is deasserted, SCL is sampled low before SDA goes high (Case 2).

The Stop condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSP1ADD and counts down to zero. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 30-38). If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 30-39).

**FIGURE 30-38: BUS COLLISION DURING A STOP CONDITION (CASE 1)**

<table>
<thead>
<tr>
<th>TBRG</th>
<th>TBRG</th>
<th>TBRG</th>
<th>SDA sampled low after TBRG, set BCL1IF</th>
</tr>
</thead>
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<tr>
<td>SDA</td>
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<td></td>
<td></td>
</tr>
<tr>
<td>SCL</td>
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</tr>
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<td>BCL1IF</td>
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</tr>
<tr>
<td>P</td>
<td></td>
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<tr>
<td>SSP1IF</td>
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</tr>
</tbody>
</table>

**FIGURE 30-39: BUS COLLISION DURING A STOP CONDITION (CASE 2)**

<table>
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<th>TBRG</th>
<th>TBRG</th>
</tr>
</thead>
<tbody>
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</tr>
<tr>
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</tr>
<tr>
<td>PEN</td>
<td></td>
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</tr>
<tr>
<td>BCL1IF</td>
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<tr>
<td>P</td>
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<td></td>
</tr>
<tr>
<td>SSP1IF</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

SDA asserted low.
### TABLE 30-3: SUMMARY OF REGISTERS ASSOCIATED WITH I\(^2\)C OPERATION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Reset Values on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>C2IE</td>
<td>C1IE</td>
<td>—</td>
<td>BCL1IE</td>
<td>TMR6IE</td>
<td>TMR4IE</td>
<td>CCP2IE</td>
<td>93</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>95</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>C2IF</td>
<td>C1IF</td>
<td>—</td>
<td>BCL1IF</td>
<td>TMR6IF</td>
<td>TMR4IF</td>
<td>CCP2IF</td>
<td>96</td>
</tr>
<tr>
<td>RxyPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
<tr>
<td>SSPCLKPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>153</td>
</tr>
<tr>
<td>SSPDATPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>153</td>
</tr>
<tr>
<td>SSP1ADD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>353</td>
</tr>
<tr>
<td>SSP1BUF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>302*</td>
</tr>
<tr>
<td>SSP1CON1</td>
<td>WCOL</td>
<td>SSPOV</td>
<td>SSPEN</td>
<td>CKP</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>349</td>
</tr>
<tr>
<td>SSP1CON2</td>
<td>GCEN</td>
<td>ACKSTAT</td>
<td>ACKDT</td>
<td>ACKEN</td>
<td>RCEN</td>
<td>PEN</td>
<td>RSEN</td>
<td>SEN</td>
<td>351</td>
</tr>
<tr>
<td>SSP1CON3</td>
<td>ACKTIM</td>
<td>PCIE</td>
<td>SCIE</td>
<td>BOEN</td>
<td>SDAHT</td>
<td>SBCDE</td>
<td>AHEN</td>
<td>DHEN</td>
<td>352</td>
</tr>
<tr>
<td>SSP1MSK</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>353</td>
</tr>
<tr>
<td>SSP1STAT</td>
<td>SMP</td>
<td>CKE</td>
<td>D/Ä</td>
<td>P</td>
<td>S</td>
<td>R/W</td>
<td>UA</td>
<td>BF</td>
<td>348</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used by the MSSP module in I\(^2\)C mode.

* Page provides register information.
30.7 BAUD RATE GENERATOR

The MSSP module has a Baud Rate Generator available for clock generation in both I²C and SPI Master modes. The Baud Rate Generator (BRG) reload value is placed in the SSP1ADD register (Register 30-6). When a write occurs to SSP1BUF, the Baud Rate Generator will automatically begin counting down.

Once the given operation is complete, the internal clock will automatically stop counting and the clock pin will remain in its last state.

An internal signal “Reload” in Figure 30-40 triggers the value from SSP1ADD to be loaded into the BRG counter. This occurs twice for each oscillation of the module clock line. The logic dictating when the reload signal is asserted depends on the mode the MSSP is being operated in.

Table 30-4 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSP1ADD.

**EQUATION 30-1:**

\[
F_{\text{CLOCK}} = \frac{F_{\text{OSC}}}{(SSP_{\text{XADD}} + 1)(4)}
\]

**FIGURE 30-40: BAUD RATE GENERATOR BLOCK DIAGRAM**

Note: Values of 0x00, 0x01 and 0x02 are not valid for SSP1ADD when used as a Baud Rate Generator for I²C. This is an implementation limitation.

<table>
<thead>
<tr>
<th>Fosc</th>
<th>Fcy</th>
<th>BRG Value</th>
<th>FCLOCK (2 Rollovers of BRG)</th>
</tr>
</thead>
<tbody>
<tr>
<td>32 MHz</td>
<td>8 MHz</td>
<td>13h</td>
<td>400 kHz</td>
</tr>
<tr>
<td>32 MHz</td>
<td>8 MHz</td>
<td>19h</td>
<td>308 kHz</td>
</tr>
<tr>
<td>32 MHz</td>
<td>8 MHz</td>
<td>4Fh</td>
<td>100 kHz</td>
</tr>
<tr>
<td>16 MHz</td>
<td>4 MHz</td>
<td>09h</td>
<td>400 kHz</td>
</tr>
<tr>
<td>16 MHz</td>
<td>4 MHz</td>
<td>0Ch</td>
<td>308 kHz</td>
</tr>
<tr>
<td>16 MHz</td>
<td>4 MHz</td>
<td>27h</td>
<td>100 kHz</td>
</tr>
<tr>
<td>4 MHz</td>
<td>1 MHz</td>
<td>09h</td>
<td>100 kHz</td>
</tr>
</tbody>
</table>

Note: Refer to the I/O port electrical and timing specifications in Table 34-10 and Figure 34-7 to ensure the system is designed to support IOL requirements.
30.8 Register Definitions: MSSP Control

REGISTER 30-1: SSP1STAT: SSP STATUS REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>SMP</td>
<td>CKE</td>
<td>D/Å</td>
<td>P</td>
<td>S</td>
<td>R/W</td>
<td>UA</td>
<td>BF</td>
</tr>
</tbody>
</table>

Legend:
- \( R \) = Readable bit
- \( W \) = Writable bit
- \( U \) = Unimplemented bit, read as ‘0’
- \( u \) = Bit is unchanged
- \( x \) = Bit is unknown
- \(-n/n\) = Value at POR and BOR/Value at all other Resets

- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

- \( \overline{R/W} \): Read/Write bit information (I\(^2\)C mode only)
- \( \overline{R/W} \) holds the \( \overline{R/W} \) bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit, or not ACK bit.

<table>
<thead>
<tr>
<th>bit 7</th>
<th>SMP: SPI Data Input Sample bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>SPI Master mode:</td>
</tr>
<tr>
<td></td>
<td>1 = Input data sampled at end of data output time</td>
</tr>
<tr>
<td></td>
<td>0 = Input data sampled at middle of data output time</td>
</tr>
<tr>
<td></td>
<td>SPI Slave mode:</td>
</tr>
<tr>
<td></td>
<td>SMP must be cleared when SPI is used in Slave mode</td>
</tr>
<tr>
<td></td>
<td>In I(^2)C Master or Slave mode:</td>
</tr>
<tr>
<td></td>
<td>1 = Slew rate control disabled for Standard Speed mode (100 kHz and 1 MHz)</td>
</tr>
<tr>
<td></td>
<td>0 = Slew rate control enabled for High-Speed mode (400 kHz)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 6</th>
<th>CKE: SPI Clock Edge Select bit (SPI mode only)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>In SPI Master or Slave mode:</td>
</tr>
<tr>
<td></td>
<td>1 = Transmit occurs on transition from active to Idle clock state</td>
</tr>
<tr>
<td></td>
<td>0 = Transmit occurs on transition from Idle to active clock state</td>
</tr>
<tr>
<td></td>
<td>In I(^2)C™ mode only:</td>
</tr>
<tr>
<td></td>
<td>1 = Enable input logic so that thresholds are compliant with SMBus specification</td>
</tr>
<tr>
<td></td>
<td>0 = Disable SMBus specific inputs</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 5</th>
<th>D/Å: Data/Address bit (I(^2)C mode only)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1 = Indicates that the last byte received or transmitted was data</td>
</tr>
<tr>
<td></td>
<td>0 = Indicates that the last byte received or transmitted was address</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 4</th>
<th>P: Stop bit (I(^2)C mode only. This bit is cleared when the MSSP module is disabled, SSPEN is cleared.)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1 = Indicates that a Stop bit has been detected last (this bit is ‘0’ on Reset)</td>
</tr>
<tr>
<td></td>
<td>0 = Stop bit was not detected last</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 3</th>
<th>S: Start bit (I(^2)C mode only. This bit is cleared when the MSSP module is disabled, SSPEN is cleared.)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1 = Indicates that a Start bit has been detected last (this bit is ‘0’ on Reset)</td>
</tr>
<tr>
<td></td>
<td>0 = Start bit was not detected last</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 2</th>
<th>R/W: Read/Write bit information (I(^2)C mode only)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>This bit holds the R/W bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit, or not ACK bit.</td>
</tr>
</tbody>
</table>

- In I\(^2\)C Slave mode:
  - 1 = Read |
  - 0 = Write |

- In I\(^2\)C Master mode:
  - 1 = Transmit is in progress |
  - 0 = Transmit is not in progress |

- OR-ing this bit with SEN, RSEN, PEN, RCEN or ACKEN will indicate if the MSSP is in Idle mode.

<table>
<thead>
<tr>
<th>bit 1</th>
<th>UA: Update Address bit (10-bit I(^2)C mode only)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1 = Indicates that the user needs to update the address in the SSP1ADD register</td>
</tr>
<tr>
<td></td>
<td>0 = Address does not need to be updated</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 0</th>
<th>BF: Buffer Full Status bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Receive (SPI and I(^2)C modes):</td>
</tr>
<tr>
<td></td>
<td>1 = Receive complete, SSP1BUF is full</td>
</tr>
<tr>
<td></td>
<td>0 = Receive not complete, SSP1BUF is empty</td>
</tr>
</tbody>
</table>

- Transmit (I\(^2\)C mode only):
  - 1 = Data transmit in progress (does not include the ACK and Stop bits), SSP1BUF is full |
  - 0 = Data transmit complete (does not include the ACK and Stop bits), SSP1BUF is empty |
REGISTER 30-2:  SSP1CON1: SSP CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>R/C/HS-0/0</th>
<th>R/C/HS-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>WCOL</td>
<td>SSPOV</td>
<td>SSPEN</td>
<td>CKP</td>
<td>SSPM&lt;3:0&gt;</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- HS = Bit is set by hardware
- C = User cleared

bit 7  WCOL: Write Collision Detect bit
- Master mode:
  - 1 = A write to the SSP1BUF register was attempted while the I2C conditions were not valid for a transmission to be started
  - 0 = No collision
- Slave mode:
  - 1 = The SSP1BUF register is written while it is still transmitting the previous word (must be cleared in software)
  - 0 = No collision

bit 6  SSPOV: Receive Overflow Indicator bit(1)
- In SPI mode:
  - 1 = A new byte is received while the SSP1BUF register is still holding the previous data. In case of overflow, the data in SSP1SR is lost. Overflow can only occur in Slave mode. In Slave mode, the user must read the SSP1BUF, even if only transmitting data, to avoid setting overflow. In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSP1BUF register (must be cleared in software).
  - 0 = No overflow
- In I2C mode:
  - 1 = A byte is received while the SSP1BUF register is still holding the previous byte. SSPOV is a “don’t care” in Transmit mode (must be cleared in software).
  - 0 = No overflow

bit 5  SSPEN: Synchronous Serial Port Enable bit
- In both modes, when enabled, these pins must be properly configured as input or output
- In SPI mode:
  - 1 = Enables serial port and configures SCK, SDIO, SDI and SS as the source of the serial port pins(2)
  - 0 = Disables serial port and configures these pins as I/O port pins
- In I2C mode:
  - 1 = Enables the serial port and configures the SDA and SCL pins as the source of the serial port pins(3)
  - 0 = Disables serial port and configures these pins as I/O port pins

bit 4  CKP: Clock Polarity Select bit
- In SPI mode:
  - 1 = Idle state for clock is a high level
  - 0 = Idle state for clock is a low level
- In I2C Slave mode:
  - SCL release control
    - 1 = Enable clock
    - 0 = Holds clock low (clock stretch). (Used to ensure data setup time.)
- In I2C Master mode:
  - Unused in this mode

bit 3-0  SSPM<3:0>: Synchronous Serial Port Mode Select bits
- 1111 = I2C Slave mode, 10-bit address with Start and Stop bit interrupts enabled
- 1110 = I2C Slave mode, 7-bit address with Start and Stop bit interrupts enabled
- 1101 = Reserved
- 1100 = Reserved
- 1011 = I2C firmware controlled Master mode (slave idle)
- 1010 = SPI Master mode, clock = Fosc/(4 * (SSP1ADD+1))(5)
- 1001 = Reserved
- 1000 = I2C Master mode, clock = Fosc / (4 * (SSP1ADD+1))(4)
- 0111 = I2C Slave mode, 10-bit address
- 0110 = I2C Slave mode, 7-bit address
- 0101 = SPI Slave mode, clock = SCK pin, SS pin control disabled, SS can be used as I/O pin
- 0100 = SPI Slave mode, clock = SCK pin, SS pin control enabled
- 0011 = SPI Master mode, clock = T2_match/2
- 0010 = SPI Master mode, clock = Fosc/64
- 0001 = SPI Master mode, clock = Fosc/16
- 0000 = SPI Master mode, clock = Fosc/4
Note 1: In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSP1BUF register.

2: When enabled, these pins must be properly configured as input or output. Use SSPSSPPS, SSPCLKPPS, SSPDATPPS, and RxyPPS to select the pins.

3: When enabled, the SDA and SCL pins must be configured as inputs. Use SSPCLKPPS, SSPDATPPS, and RxyPPS to select the pins.

4: SSP1ADD values of 0, 1 or 2 are not supported for I2C mode.

5: SSP1ADD value of ‘0’ is not supported. Use SSPM = 0000 instead.
REGISTER 30-3:  SSP1CON2: SSP CONTROL REGISTER 2(1)

<table>
<thead>
<tr>
<th></th>
<th>GCEN</th>
<th>ACKSTAT</th>
<th>ACKDT</th>
<th>ACKEN</th>
<th>RCEN</th>
<th>PEN</th>
<th>RSEN</th>
<th>SEN</th>
</tr>
</thead>
<tbody>
<tr>
<td>R/W-0/0</td>
<td>R-0/0</td>
<td>R/W-0/0</td>
<td>R/S-0/0</td>
<td>R/S-0/0</td>
<td>R/S-0/0</td>
<td>R/S-0/0</td>
<td>R/W-0/0</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
W = Writable bit
U = Unimplemented bit, read as ‘0’
u = Bit is unchanged
x = Bit is unknown
-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set
'0' = Bit is cleared
HC = Cleared by hardware
S = User set

bit 7  GCEN: General Call Enable bit (in \( \mathrm{I}^2 \mathrm{C} \) Slave mode only)
1 = Enable interrupt when a general call address (0x00 or 00h) is received in the SSP1SR
0 = General call address disabled

bit 6  ACKSTAT: Acknowledge Status bit (in \( \mathrm{I}^2 \mathrm{C} \) mode only)
1 = Acknowledge was not received
0 = Acknowledge was received

bit 5  ACKDT: Acknowledge Data bit (in \( \mathrm{I}^2 \mathrm{C} \) mode only)
In Receive mode:
Value transmitted when the user initiates an Acknowledge sequence at the end of a receive
1 = Not Acknowledge
0 = Acknowledge

bit 4  ACKEN: Acknowledge Sequence Enable bit (in \( \mathrm{I}^2 \mathrm{C} \) Master mode only)
In Master Receive mode:
1 = Initiate Acknowledge sequence on SDA and SCL pins, and transmit ACKDT data bit. Automatically cleared by hardware.
0 = Acknowledge sequence idle

bit 3  RCEN: Receive Enable bit (in \( \mathrm{I}^2 \mathrm{C} \) Master mode only)
1 = Enables Receive mode for \( \mathrm{I}^2 \mathrm{C} \)
0 = Receive idle

bit 2  PEN: Stop Condition Enable bit (in \( \mathrm{I}^2 \mathrm{C} \) Master mode only)

SCKMSSP Release Control:
1 = Initiate Stop condition on SDA and SCL pins. Automatically cleared by hardware.
0 = Stop condition Idle

bit 1  RSEN: Repeated Start Condition Enable bit (in \( \mathrm{I}^2 \mathrm{C} \) Master mode only)
1 = Initiate Repeated Start condition on SDA and SCL pins. Automatically cleared by hardware.
0 = Repeated Start condition Idle

bit 0  SEN: Start Condition Enable/Stretch Enable bit

In Master mode:
1 = Initiate Start condition on SDA and SCL pins. Automatically cleared by hardware.
0 = Start condition Idle

In Slave mode:
1 = Clock stretching is enabled for both slave transmit and slave receive (stretch enabled)
0 = Clock stretching is disabled

Note 1: For bits ACKEN, RCEN, PEN, RSEN, SEN: If the \( \mathrm{I}^2 \mathrm{C} \) module is not in the Idle mode, this bit may not be set (no spooling) and the SSP1BUF may not be written (or writes to the SSP1BUF are disabled).
REGISTER 30-4: SSPCON3: SSP CONTROL REGISTER 3

<table>
<thead>
<tr>
<th>Bit</th>
<th>Acknowledge Time Status bit (I2C mode only)</th>
<th>PCIE: Stop Condition Interrupt Enable bit (I2C Slave mode only)</th>
<th>SCIE: Start Condition Interrupt Enable bit (I2C Slave mode only)</th>
<th>BOEN: Buffer Overwrite Enable bit</th>
<th>SDAHT: SDA Hold Time Selection bit (I2C mode only)</th>
<th>SBCDE: Slave Mode Bus Collision Detect Enable bit (I2C Slave mode only)</th>
<th>AHEN: Address Hold Enable bit (I2C Slave mode only)</th>
<th>DHEN: Data Hold Enable bit (I2C Slave mode only)</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>1 = Indicates the I2C bus is in an Acknowledge sequence, set on eighth falling edge of SCL clock</td>
<td>1 = Enable interrupt on detection of Stop condition</td>
<td>1 = Enable interrupt on detection of Start or Restart conditions</td>
<td>1 = SSP1BUF updates every time that a new data byte is shifted in ignoring the BF bit</td>
<td>1 = Minimum of 300 ns hold time on SDA after the falling edge of SCL</td>
<td>If, on the rising edge of SCL, SDA is sampled low when the module is outputting a high state, the BCL1IF bit of the PIR2 register is set, and bus goes idle</td>
<td>1 = Following the eighth falling edge of SCL for a matching received address byte; CKP bit of the SSP1CON1 register will be cleared and the SCL will be held low.</td>
<td>1 = Following the eighth falling edge of SCL for a received data byte; slave hardware clears the CKP bit of the SSP1CON1 register and SCL is held low.</td>
</tr>
<tr>
<td>6</td>
<td>0 = Not an Acknowledge sequence, cleared on 9th rising edge of SCL clock</td>
<td>0 = Stop detection interrupts are disabled</td>
<td>0 = Start detection interrupts are disabled</td>
<td>0 = If new byte is received with BF bit of the SSP1STAT register already set, SSPOV bit of the SSP1CON1 register is set, and the buffer is not updated</td>
<td>0 = Minimum of 100 ns hold time on SDA after the falling edge of SCL</td>
<td>1 = Enable slave bus collision interrupts</td>
<td>0 = Address holding is disabled</td>
<td>0 = Data holding is disabled</td>
</tr>
<tr>
<td>5</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>4</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>3</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **1’** = Bit is set
- **0’** = Bit is cleared

**Note 1:** For daisy-chained SPI operation; allows the user to ignore all but the last received byte. SSPOV is still set when a new byte is received and BF = 1, but hardware continues to write the most recent byte to SSP1BUF.

**Note 2:** This bit has no effect in Slave modes that Start and Stop condition detection is explicitly listed as enabled.

**Note 3:** The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is set.
REGISTER 30-5:  SSP1MSK: SSP MASK REGISTER

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>MSK&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7-1 MSK<7:1>: Mask bits
- 1 = The received address bit n is compared to SSP1ADD<n> to detect I2C address match
- 0 = The received address bit n is not used to detect I2C address match

bit 0 MSK<0>: Mask bit for I2C Slave mode, 10-bit Address
- I2C Slave mode, 10-bit address (SSPM<3:0> = 0111 or 1111):
  - 1 = The received address bit 0 is compared to SSP1ADD<0> to detect I2C address match
  - 0 = The received address bit 0 is not used to detect I2C address match
- I2C Slave mode, 7-bit address, the bit is ignored

REGISTER 30-6:  SSP1ADD: MSSP ADDRESS AND BAUD RATE REGISTER (I2C MODE)

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADD&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

Master mode:

bit 7-0 ADD<7:0>: Baud Rate Clock Divider bits
- SCL pin clock period = ((ADD<7:0> + 1) * 4)/FOSC

10-Bit Slave mode – Most Significant Address Byte:

bit 7-3 Not used: Unused for Most Significant Address Byte. Bit state of this register is a "don't care". Bit pattern sent by master is fixed by I2C specification and must be equal to '11110'. However, those bits are compared by hardware and are not affected by the value in this register.

bit 2-1 ADD<2:1>: Two Most Significant bits of 10-bit address
bit 0 Not used: Unused in this mode. Bit state is a "don't care".

10-Bit Slave mode – Least Significant Address Byte:

bit 7-0 ADD<7:0>: Eight Least Significant bits of 10-bit address

7-Bit Slave mode:

bit 7-1 ADD<7:1>: 7-bit address
bit 0 Not used: Unused in this mode. Bit state is a "don't care".
31.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer independent of device program execution. The EUSART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or half-duplex synchronous system. Full-Duplex mode is useful for communications with peripheral systems, such as CRT terminals and personal computers. Half-Duplex Synchronous mode is intended for communications with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs or other microcontrollers. These devices typically do not have internal clocks for baud rate generation and require the external clock signal provided by a master synchronous device.

The EUSART module includes the following capabilities:
- Full-duplex asynchronous transmit and receive
- Two-character input buffer
- One-character output buffer
- Programmable 8-bit or 9-bit character length
- Address detection in 9-bit mode
- Input buffer overrun error detection
- Received character framing error detection
- Half-duplex synchronous master
- Half-duplex synchronous slave
- Programmable clock polarity in synchronous modes
- Sleep operation

The EUSART module implements the following additional features, making it ideally suited for use in Local Interconnect Network (LIN) bus systems:
- Automatic detection and calibration of the baud rate
- Wake-up on Break reception
- 13-bit Break character transmit

Block diagrams of the EUSART transmitter and receiver are shown in Figure 31-1 and Figure 31-2.

The EUSART transmit output (TX_out) is available to the TX/CK pin and internally to the following peripherals:
- Configurable Logic Cell (CLC)

FIGURE 31-1: EUSART TRANSMIT BLOCK DIAGRAM
The operation of the EUSART module is controlled through three registers:

- Transmit Status and Control (TX1STA)
- Receive Status and Control (RC1STA)
- Baud Rate Control (BAUD1CON)

These registers are detailed in Register 31-1, Register 31-2 and Register 31-3, respectively.

The RX and CK input pins are selected with the RXPPS and CKPPS registers, respectively. TX, CK, and DT output pins are selected with each pin’s RxyPPS register. Since the RX input is coupled with the DT output in Synchronous mode, it is the user’s responsibility to select the same pin for both of these functions when operating in Synchronous mode. The EUSART control logic will control the data direction drivers automatically.
31.1 EUSART Asynchronous Mode

The EUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH Mark state which represents a ‘1’ data bit, and a VOL Space state which represents a ‘0’ data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the Mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is eight bits. Each transmitted bit persists for a period of 1/(Baud Rate). An on-chip dedicated 8-bit/16-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See Table 31-5 for examples of baud rate configurations.

The EUSART transmits and receives the LSB first. The EUSART’s transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

31.1.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 31-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TX1REG register.

31.1.1.1 Enabling the Transmitter

The EUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

• TXEN = 1
• SYNC = 0
• SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TX1STA register enables the transmitter circuitry of the EUSART. Clearing the SYNC bit of the TX1STA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RC1STA register enables the EUSART and automatically configures the TX/CK I/O pin as an output. If the TX/CK pin is shared with an analog peripheral, the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

Note: The TXIF Transmitter Interrupt flag is set when the TXEN enable bit is set.

31.1.1.2 Transmitting Data

A transmission is initiated by writing a character to the TX1REG register. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TX1REG is immediately transferred to the TSR register. If the TSR still contains all or part of a previous character, the new character data is held in the TX1REG until the Stop bit of the previous character has been transmitted. The pending character in the TX1REG is then transferred to the TSR in one TCY immediately following the Stop bit transmission. The transmission of the Start bit, data bits and Stop bit sequence commences immediately following the transfer of the data to the TSR from the TX1REG.

31.1.1.3 Transmit Data Polarity

The polarity of the transmit data can be controlled with the SCKP bit of the BAUD1CON register. The default state of this bit is ‘0’ which selects high true transmit idle and data bits. Setting the SCKP bit to ‘1’ will invert the transmit data resulting in low true idle and data bits. The SCKP bit controls transmit data polarity in Asynchronous mode only. In Synchronous mode, the SCKP bit has a different function. See Section 31.5.1.2 “Clock Polarity”.

31.1.1.4 Transmit Interrupt Flag

The TXIF interrupt flag bit of the PIR1 register is set whenever the EUSART transmitter is enabled and no character is being held for transmission in the TX1REG. In other words, the TXIF bit is only clear when the TSR is busy with a character and a new character has been queued for transmission in the TX1REG. The TXIF flag bit is not cleared immediately upon writing TX1REG. TXIF becomes valid in the second instruction cycle following the write execution. Polling TXIF immediately following the TX1REG write will return invalid results. The TXIF bit is read-only, it cannot be set or cleared by software.

The TXIF interrupt can be enabled by setting the TXIE interrupt enable bit of the PIE1 register. However, the TXIF flag bit will be set whenever the TX1REG is empty, regardless of the state of TXIE enable bit.

To use interrupts when transmitting data, set the TXIE bit only when there is more data to send. Clear the TXIE interrupt enable bit upon writing the last character of the transmission to the TX1REG.
31.1.1.5 TSR Status

The TRMT bit of the TX1STA register indicates the status of the TSR register. This is a read-only bit. The TRMT bit is set when the TSR register is empty and is cleared when a character is transferred to the TSR register from the TX1REG. The TRMT bit remains clear until all bits have been shifted out of the TSR register. No interrupt logic is tied to this bit, so the user has to poll this bit to determine the TSR status.

**Note:** The TSR register is not mapped in data memory, so it is not available to the user.

31.1.1.6 Transmitting 9-Bit Characters

The EUSART supports 9-bit character transmissions. When the TX9 bit of the TX1STA register is set, the EUSART will shift nine bits out for each character transmitted. The TX9D bit of the TX1STA register is the ninth, and Most Significant data bit. When transmitting 9-bit data, the TX9D data bit must be written before writing the eight Least Significant data bits into the TX1REG. All nine bits of data will be transferred to the TSR shift register immediately after the TX1REG is written.

A special 9-bit Address mode is available for use with multiple receivers. See Section 31.1.2.7 “Address Detection” for more information on the Address mode.

31.1.1.7 Asynchronous Transmission Setup

1. Initialize the SP1BRGH, SP1BRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 31.4 “EUSART Baud Rate Generator (BRG)”.
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If 9-bit transmission is desired, set the TX9 control bit. A set ninth data bit will indicate that the eight Least Significant data bits are an address when the receiver is set for address detection.
4. Set SCKP bit if inverted transmit is desired.
5. Enable the transmission by setting the TXEN control bit. This will cause the TXIF interrupt bit to be set.
6. If interrupts are desired, set the TXIE interrupt enable bit of the PIE1 register. An interrupt will occur immediately provided that the GIE and PEIE bits of the INTCON register are also set.
7. If 9-bit transmission is selected, the ninth bit should be loaded into the TX9D data bit.
8. Load 8-bit data into the TX1REG register. This will start the transmission.

**FIGURE 31-3: ASYNCHRONOUS TRANSMISSION**
FIGURE 31-4: ASYNCHRONOUS TRANSMISSION (BACK-TO-BACK)

Note: This timing diagram shows two consecutive transmissions.

TABLE 31-1: SUMMARY OF REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUD1CON</td>
<td></td>
<td>ABDOVF</td>
<td></td>
<td></td>
<td>SCKP</td>
<td>BRG16</td>
<td></td>
<td>WUE</td>
<td>ABDEN 365</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1E</td>
<td>CCP1E</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1F</td>
<td>CCP1F</td>
<td>TMR2F</td>
<td>TMR1F</td>
<td>95</td>
</tr>
<tr>
<td>RC1STA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>364</td>
</tr>
<tr>
<td>RxyPPS</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>154</td>
</tr>
<tr>
<td>SP1BRGL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>RxyPPS&lt;4:0&gt;</td>
<td>366*</td>
</tr>
<tr>
<td>SP1BRGH</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>SP1BRG&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>366*</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TX1REG</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>356*</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used for asynchronous transmission.

* Page provides register information.
31.1.2   EUSART ASYNCHRONOUS RECEIVER

The Asynchronous mode is typically used in RS-232 systems. The receiver block diagram is shown in Figure 31-2. The data is received on the RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at 16 times the baud rate, whereas the serial Receive Shift Register (RSR) operates at the bit rate. When all eight or nine bits of the character have been shifted in, they are immediately transferred to a two character First-In-First-Out (FIFO) memory. The FIFO buffering allows reception of two complete characters and the start of a third character before software must start servicing the EUSART receiver. The FIFO and RSR registers are not directly accessible by software. Access to the received data is via the RC1REG register.

31.1.2.1   Enabling the Receiver

The EUSART receiver is enabled for asynchronous operation by configuring the following three control bits:

- CREN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the CREN bit of the RC1STA register enables the receiver circuitry of the EUSART. Clearing the SYNC bit of the TX1STA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RC1STA register enables the EUSART. The programmer must set the corresponding TRIS bit to configure the RX/DT I/O pin as an input.

Note: If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.

31.1.2.2   Receiving Data

The receiver data recovery circuit initiates character reception on the falling edge of the first bit. The first bit, also known as the Start bit, is always a zero. The data recovery circuit counts one-half bit time to the center of the Start bit and verifies that the bit is still a zero. If it is not a zero then the data recovery circuit aborts character reception, without generating an error, and resumes looking for the falling edge of the Start bit. If the Start bit zero verification succeeds then the data recovery circuit counts a full bit time to the center of the next bit. The bit is then sampled by a majority detect circuit and the resulting '0' or '1' is shifted into the RSR. This repeats until all data bits have been sampled and shifted into the RSR. One final bit time is measured and the level sampled. This is the Stop bit, which is always a '1'. If the data recovery circuit samples a '0' in the Stop bit position then a framing error is set for this character, otherwise the framing error is cleared for this character. See Section 31.1.2.4 “Receive Framing Error” for more information on framing errors.

Immediately after all data bits and the Stop bit have been received, the character in the RSR is transferred to the EUSART receive FIFO and the RCIF interrupt flag bit of the PIR1 register is set. The top character in the FIFO is transferred out of the FIFO by reading the RC1REG register.

Note: If the receive FIFO is overrun, no additional characters will be received until the overrun condition is cleared. See Section 31.1.2.5 “Receive Overrun Error” for more information on overrun errors.

31.1.2.3   Receive Interrupts

The RCIF interrupt flag bit of the PIR1 register is set whenever the EUSART receiver is enabled and there is an unread character in the receive FIFO. The RCIF interrupt flag bit is read-only, it cannot be set or cleared by software.

RCIF interrupts are enabled by setting all of the following bits:

- RCIE, Interrupt Enable bit of the PIE1 register
- PEIE, Peripheral Interrupt Enable bit of the INTCON register
- GIE, Global Interrupt Enable bit of the INTCON register

The RCIF interrupt flag bit will be set when there is an unread character in the FIFO, regardless of the state of interrupt enable bits.
31.1.2.4 Receive Framing Error
Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RC1STA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RC1REG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error \((\text{FERR} = 1)\) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RC1STA register which resets the EUSART. Clearing the CREN bit of the RC1STA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

31.1.2.5 Receive Overrun Error
The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RC1STA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RC1STA register or by resetting the EUSART by clearing the SPEN bit of the RC1STA register.

31.1.2.6 Receiving 9-Bit Characters
The EUSART supports 9-bit character reception. When the RX9 bit of the RC1STA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RC1STA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RC1REG.

31.1.2.7 Address Detection
A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RC1STA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

Note: If all receive characters in the receive FIFO have framing errors, repeated reads of the RC1REG will not clear the FERR bit.
31.1.2.8 Asynchronous Reception Setup:
1. Initialize the SP1BRGH, SP1BRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 31.4 “EUSART Baud Rate Generator (BRG)”).
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
4. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
5. If 9-bit reception is desired, set the RX9 bit.
6. Enable reception by setting the CREN bit.
7. The RCIF interrupt flag bit will be set when a character is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
8. Read the RC1STA register to get the error flags and, if 9-bit data reception is enabled, the ninth data bit.
9. Get the received eight Least Significant data bits from the receive buffer by reading the RC1REG register.
10. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.

31.1.2.9 9-bit Address Detection Mode Setup
This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:
1. Initialize the SP1BRGH, SP1BRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 31.4 “EUSART Baud Rate Generator (BRG)”).
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
4. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
5. Enable 9-bit reception by setting the RX9 bit.
6. Enable address detection by setting the ADDEN bit.
7. Enable reception by setting the CREN bit.
8. The RCIF interrupt flag bit will be set when a character with the ninth bit set is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
9. Read the RC1STA register to get the error flags. The ninth data bit will always be set.
10. Get the received eight Least Significant data bits from the receive buffer by reading the RC1REG register. Software determines if this is the device’s address.
11. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.
12. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and generate interrupts.

FIGURE 31-5: ASYNCHRONOUS RECEPTION

Note: This timing diagram shows three words appearing on the RX input. The RC1REG (receive buffer) is read after the third word, causing the OERR (overrun) bit to be set.
31.2 Clock Accuracy with Asynchronous Operation

The factory calibrates the internal oscillator block output (INTOSC). However, the INTOSC frequency may drift as Vdd or temperature changes, and this directly affects the asynchronous baud rate. Two methods may be used to adjust the baud rate clock, but both require a reference clock source of some kind.

The first (preferred) method uses the OSCTUNE register to adjust the INTOSC output. Adjusting the value in the OSCTUNE register allows for fine resolution changes to the system clock source. See Section 6.2.2.3 “Internal Oscillator Frequency Adjustment” for more information.

The other method adjusts the value in the Baud Rate Generator. This can be done automatically with the Auto-Baud Detect feature (see Section 31.4.1 “Auto-Baud Detect”). There may not be fine enough resolution when adjusting the Baud Rate Generator to compensate for a gradual change in the peripheral clock frequency.
31.3 Register Definitions: EUSART Control

REGISTER 31-1: TX1STA: TRANSMIT STATUS AND CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-1/1</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN$^{(1)}$</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
</tr>
</tbody>
</table>

Legend:

R = Readable bit  W = Writable bit  U = Unimplemented bit, read as ‘0’
u = Bit is unchanged  x = Bit is unknown -n/n = Value at POR and BOR/Value at all other Resets

‘1’ = Bit is set  ‘0’ = Bit is cleared

bit 7  

**CSRC**: Clock Source Select bit

Asynchronous mode:

Don’t care

Synchronous mode:

1 = Master mode (clock generated internally from BRG)
0 = Slave mode (clock from external source)

bit 6  

**TX9**: 9-bit Transmit Enable bit

1 = Selects 9-bit transmission
0 = Selects 8-bit transmission

bit 5  

**TXEN**: Transmit Enable bit$^{(1)}$

1 = Transmit enabled
0 = Transmit disabled

bit 4  

**SYNC**: EUSART Mode Select bit

1 = Synchronous mode
0 = Asynchronous mode

bit 3  

**SENDB**: Send Break Character bit

Asynchronous mode:

1 = Send Sync Break on next transmission (cleared by hardware upon completion)
0 = Sync Break transmission completed

Synchronous mode:

Don’t care

bit 2  

**BRGH**: High Baud Rate Select bit

Asynchronous mode:

1 = High speed
0 = Low speed

Synchronous mode:

Unused in this mode

bit 1  

**TRMT**: Transmit Shift Register Status bit

1 = TSR empty
0 = TSR full

bit 0  

**TX9D**: Ninth bit of Transmit Data

Can be address/data bit or a parity bit.

**Note 1**:  SREN/CREN overrides TXEN in Sync mode.
**REGISTER 31-2: RC1STA: RECEIVE STATUS AND CONTROL REGISTER**

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

**bit 7**
- **SPEN**: Serial Port Enable bit
  - **1** = Serial port enabled
  - **0** = Serial port disabled (held in Reset)

**bit 6**
- **RX9**: 9-Bit Receive Enable bit
  - **1** = Selects 9-bit reception
  - **0** = Selects 8-bit reception

**bit 5**
- **SREN**: Single Receive Enable bit
  - **Asynchronous mode**: Don’t care
  - **Synchronous mode – Master**: 1 = Enables single receive
    0 = Disables single receive
    This bit is cleared after reception is complete.
  - **Synchronous mode – Slave**: Don’t care

**bit 4**
- **CREN**: Continuous Receive Enable bit
  - **Asynchronous mode**: 1 = Enables receiver
    0 = Disables receiver
  - **Synchronous mode**: 1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN)
    0 = Disables continuous receive

**bit 3**
- **ADDEN**: Address Detect Enable bit
  - **Asynchronous mode 9-bit (RX9 = 1)**: 1 = Enables address detection, enable interrupt and load the receive buffer when RSR<8> is set
    0 = Disables address detection, all bytes are received and ninth bit can be used as parity bit
  - **Asynchronous mode 8-bit (RX9 = 0)**: Don’t care

**bit 2**
- **FERR**: Framing Error bit
  - **1** = Framing error (can be updated by reading RC1REG register and receive next valid byte)
  - **0** = No framing error

**bit 1**
- **OERR**: Overrun Error bit
  - **1** = Overrun error (can be cleared by clearing bit CREN)
  - **0** = No overrun error

**bit 0**
- **RX9D**: Ninth bit of Received Data
  - This can be address/data bit or a parity bit and must be calculated by user firmware.
REGISTER 31-3: BAUD1CON: BAUD RATE CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>ABDVF</td>
<td>RCIDL</td>
<td>Unimplemented</td>
<td>SCKP</td>
<td>BRG16</td>
<td>Unimplemented</td>
<td>WUE</td>
<td>ABDEN</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

bit 7  **ABDOVF**: Auto-Baud Detect Overflow bit
   **Asynchronous mode:**
   1 = Auto-baud timer overflowed
   0 = Auto-baud timer did not overflow
   **Synchronous mode:**
   Don’t care

bit 6  **RCIDL**: Receive Idle Flag bit
   **Asynchronous mode:**
   1 = Receiver is Idle
   0 = Start bit has been received and the receiver is receiving
   **Synchronous mode:**
   Don’t care

bit 5  **Unimplemented**: Read as ‘0’

bit 4  **SCKP**: Synchronous Clock Polarity Select bit
   **Asynchronous mode:**
   1 = Transmit inverted data to the TX/CK pin
   0 = Transmit non-inverted data to the TX/CK pin
   **Synchronous mode:**
   1 = Data is clocked on rising edge of the clock
   0 = Data is clocked on falling edge of the clock

bit 3  **BRG16**: 16-bit Baud Rate Generator bit
   1 = 16-bit Baud Rate Generator is used
   0 = 8-bit Baud Rate Generator is used

bit 2  **Unimplemented**: Read as ‘0’

bit 1  **WUE**: Wake-up Enable bit
   **Asynchronous mode:**
   1 = Receiver is waiting for a falling edge. No character will be received, byte RCIF will be set. WUE will automatically clear after RCIF is set.
   0 = Receiver is operating normally
   **Synchronous mode:**
   Don’t care

bit 0  **ABDEN**: Auto-Baud Detect Enable bit
   **Asynchronous mode:**
   1 = Auto-Baud Detect mode is enabled (clears when auto-baud is complete)
   0 = Auto-Baud Detect mode is disabled
   **Synchronous mode:**
   Don’t care
31.4 EUSART Baud Rate Generator (BRG)

The Baud Rate Generator (BRG) is an 8-bit or 16-bit timer that is dedicated to the support of both the asynchronous and synchronous EUSART operation. By default, the BRG operates in 8-bit mode. Setting the BRG16 bit of the BAUD1CON register selects 16-bit mode.

The SP1BRGH, SP1BRGL register pair determines the period of the free running baud rate timer. In Asynchronous mode the multiplier of the baud rate period is determined by both the BRGH bit of the TX1STA register and the BRG16 bit of the BAUD1CON register. In Synchronous mode, the BRGH bit is ignored.

Table 31-3 contains the formulas for determining the baud rate. Example 31-1 provides a sample calculation for determining the baud rate and baud rate error.

Typical baud rates and error values for various Asynchronous modes have been computed for your convenience and are shown in Table 31-5. It may be advantageous to use the high baud rate (BRGH = 1), or the 16-bit BRG (BRG16 = 1) to reduce the baud rate error. The 16-bit BRG mode is used to achieve slow baud rates for fast oscillator frequencies.

Writing a new value to the SP1BRGH, SP1BRGL register pair causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

If the system clock is changed during an active receive operation, a receive error or data loss may result. To avoid this problem, check the status of the RCIDL bit to make sure that the receive operation is idle before changing the system clock.

### EXAMPLE 31-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:

\[
\text{Desired Baud Rate} = \frac{\text{FOSC}}{64(\text{SPBRGH:SPBRGL} + 1)}
\]

Solving for SPBRGH:SPBRGL:

\[
X = \frac{\text{FOSC}}{64 \cdot \text{Desired Baud Rate} - 1}
\]

\[
= \frac{16000000}{9600 - 1}
\]

\[
= [25.042] = 25
\]

Calculated Baud Rate = \(
\frac{16000000}{64(25 + 1)}
\)

\[
= 9615
\]

Error = \(
\frac{\text{Calc. Baud Rate} - \text{Desired Baud Rate}}{\text{Desired Baud Rate}}
\)

\[
= \frac{(9615 - 9600)}{9600} = 0.16\%
\]
### TABLE 31-3: BAUD RATE FORMULAS

<table>
<thead>
<tr>
<th>Configuration Bits</th>
<th>BRG/EUSART Mode</th>
<th>Baud Rate Formula</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Sync</td>
<td>BRG16</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>x</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>x</td>
</tr>
</tbody>
</table>

Legend:  
\( x \) = Don’t care, \( n \) = value of SP1BRGH, SP1BRGL register pair.

### TABLE 31-4: SUMMARY OF REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUD1CON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>ABDEN</td>
<td>365</td>
</tr>
<tr>
<td>RC1STA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>364</td>
</tr>
<tr>
<td>SP1BRGL</td>
<td></td>
<td></td>
<td></td>
<td>SP1BRG&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>366</td>
</tr>
<tr>
<td>SP1BRGH</td>
<td></td>
<td></td>
<td></td>
<td>SP1BRG&lt;15:8&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>366</td>
</tr>
<tr>
<td>TX1STA</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>363</td>
</tr>
</tbody>
</table>

Legend:  
— = unimplemented location, read as ‘0’. Shaded cells are not used for the Baud Rate Generator.  
* Page provides register information.
## TABLE 31-5: BAUD RATES FOR ASYNCHRONOUS MODES

### BAUD RATE

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Fosc = 32.000 MHz</th>
<th>Fosc = 20.000 MHz</th>
<th>Fosc = 18.432 MHz</th>
<th>Fosc = 11.0592 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>SPBRG value (decimal)</td>
<td>Actual Rate</td>
</tr>
<tr>
<td>300</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>1200</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2400</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>9600</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>10417</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>19.2k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>57.6k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>115.2k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

### BAUD RATE

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Fosc = 8.000 MHz</th>
<th>Fosc = 4.000 MHz</th>
<th>Fosc = 3.6864 MHz</th>
<th>Fosc = 1.000 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>SPBRG value (decimal)</td>
<td>Actual Rate</td>
</tr>
<tr>
<td>300</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>1200</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2400</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>9600</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>10417</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>19.2k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>57.6k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>115.2k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

### BAUD RATE

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Fosc = 32.000 MHz</th>
<th>Fosc = 20.000 MHz</th>
<th>Fosc = 18.432 MHz</th>
<th>Fosc = 11.0592 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>SPBRG value (decimal)</td>
<td>Actual Rate</td>
</tr>
<tr>
<td>300</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>1200</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>2400</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>9600</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>10417</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>19.2k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>57.6k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>115.2k</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>
### TABLE 31-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>FOSC = 8.000 MHz</th>
<th>FOSC = 4.000 MHz</th>
<th>FOSC = 3.6864 MHz</th>
<th>FOSC = 1.000 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>SPBRG value (decimal)</td>
<td>Actual Rate</td>
</tr>
<tr>
<td>300</td>
<td>300.0</td>
<td>0.00</td>
<td>6666</td>
<td>300.0</td>
</tr>
<tr>
<td>1200</td>
<td>1200.0</td>
<td>-0.02</td>
<td>3332</td>
<td>1200.0</td>
</tr>
<tr>
<td>2400</td>
<td>2401.0</td>
<td>-0.04</td>
<td>832</td>
<td>2399.0</td>
</tr>
<tr>
<td>9600</td>
<td>9615.0</td>
<td>0.16</td>
<td>51</td>
<td>9615.0</td>
</tr>
<tr>
<td>10417</td>
<td>10417.0</td>
<td>0.00</td>
<td>47</td>
<td>10417.0</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.23k</td>
<td>0.16</td>
<td>103</td>
<td>19.23k</td>
</tr>
<tr>
<td>57.6k</td>
<td>55556.0</td>
<td>-3.55</td>
<td>8</td>
<td>56186.0</td>
</tr>
<tr>
<td>115.2k</td>
<td>117.6k</td>
<td>2.12</td>
<td>16</td>
<td>113.636</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>FOSC = 8.000 MHz</th>
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<th>FOSC = 1.000 MHz</th>
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<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>SPBRG value (decimal)</td>
<td>Actual Rate</td>
</tr>
<tr>
<td>300</td>
<td>299.9</td>
<td>-0.02</td>
<td>1666</td>
<td>300.1</td>
</tr>
<tr>
<td>1200</td>
<td>1199.0</td>
<td>-0.08</td>
<td>416</td>
<td>1202.0</td>
</tr>
<tr>
<td>2400</td>
<td>2404.0</td>
<td>0.16</td>
<td>207</td>
<td>2404.0</td>
</tr>
<tr>
<td>9600</td>
<td>9615.0</td>
<td>0.16</td>
<td>51</td>
<td>9615.0</td>
</tr>
<tr>
<td>10417</td>
<td>10417.0</td>
<td>0.00</td>
<td>47</td>
<td>10417.0</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.23k</td>
<td>0.16</td>
<td>12</td>
<td>19.23k</td>
</tr>
<tr>
<td>57.6k</td>
<td>55556.0</td>
<td>-3.55</td>
<td>8</td>
<td>—</td>
</tr>
<tr>
<td>115.2k</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>
### TABLE 31-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>FOSC = 32.000 MHz</th>
<th>FOSC = 20.000 MHz</th>
<th>FOSC = 18.432 MHz</th>
<th>FOSC = 11.0592 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>SPBRG value (decimal)</td>
<td>Actual Rate</td>
</tr>
<tr>
<td>300</td>
<td>300.0</td>
<td>0.00</td>
<td>26666</td>
<td>300.0</td>
</tr>
<tr>
<td>1200</td>
<td>1200</td>
<td>0.00</td>
<td>6666</td>
<td>1200</td>
</tr>
<tr>
<td>2400</td>
<td>2400</td>
<td>0.01</td>
<td>3332</td>
<td>2400</td>
</tr>
<tr>
<td>9600</td>
<td>9604</td>
<td>0.04</td>
<td>832</td>
<td>9597</td>
</tr>
<tr>
<td>10417</td>
<td>10417</td>
<td>0.00</td>
<td>767</td>
<td>10417</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.18k</td>
<td>-0.08</td>
<td>416</td>
<td>19.23k</td>
</tr>
<tr>
<td>57.6k</td>
<td>57.55k</td>
<td>-0.08</td>
<td>138</td>
<td>57.47k</td>
</tr>
<tr>
<td>115.2k</td>
<td>115.9k</td>
<td>0.64</td>
<td>68</td>
<td>116.3k</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>FOSC = 8.000 MHz</th>
<th>FOSC = 4.000 MHz</th>
<th>FOSC = 3.6864 MHz</th>
<th>FOSC = 1.000 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>SPBRG value (decimal)</td>
<td>Actual Rate</td>
</tr>
<tr>
<td>300</td>
<td>300.0</td>
<td>0.00</td>
<td>6666</td>
<td>300.0</td>
</tr>
<tr>
<td>1200</td>
<td>1200</td>
<td>-0.02</td>
<td>1666</td>
<td>1200</td>
</tr>
<tr>
<td>2400</td>
<td>2401</td>
<td>0.04</td>
<td>832</td>
<td>2398</td>
</tr>
<tr>
<td>9600</td>
<td>9615</td>
<td>0.16</td>
<td>207</td>
<td>9615</td>
</tr>
<tr>
<td>10417</td>
<td>10417</td>
<td>0.19</td>
<td>191</td>
<td>10417</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.23k</td>
<td>0.16</td>
<td>103</td>
<td>19.23k</td>
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<tr>
<td>57.6k</td>
<td>57.14k</td>
<td>-0.79</td>
<td>34</td>
<td>58.82k</td>
</tr>
<tr>
<td>115.2k</td>
<td>117.6k</td>
<td>2.12</td>
<td>16</td>
<td>111.1k</td>
</tr>
</tbody>
</table>
31.4.1 AUTO-BAUD DETECT

The EUSART module supports automatic detection and calibration of the baud rate.

In the Auto-Baud Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RX signal, the RX signal is timing the BRG. The Baud Rate Generator is used to time the period of a received 55h (ASCII "U") which is the Sync character for the LIN bus. The unique feature of this character is that it has five rising edges including the Stop bit edge.

Setting the ABDEN bit of the BAUD1CON register starts the auto-baud calibration sequence. While the ABD sequence takes place, the EUSART state machine is held in Idle. On the first rising edge of the receive line, after the Start bit, the SPBRG begins counting up using the BRG counter clock as shown in Figure 31-6. The fifth rising edge will occur on the RX pin at the end of the eighth bit period. At that time, an accumulated value totaling the proper BRG period is left in the SP1BRGH, SP1BRGL register pair, the ABDEN bit is automatically cleared and the RCIF interrupt flag is set. The value in the RC1REG needs to be read to clear the RCIF interrupt. RC1REG content should be discarded. When calibrating for modes that do not use the SP1BRGH register the user can verify that the SP1BRGL register did not overflow by checking for 00h in the SP1BRGH register.

The BRG auto-baud clock is determined by the BRG16 and BRGH bits as shown in Table 31-6. During ABD, both the SP1BRGH and SP1BRGL registers are used as a 16-bit counter, independent of the BRG16 bit setting. While calibrating the baud rate period, the SP1BRGH and SP1BRGL registers are clocked at 1/8th the BRG base clock rate. The resulting byte measurement is the average bit time when clocked at full speed.

**Note 1:** If the WUE bit is set with the ABDEN bit, auto-baud detection will occur on the byte following the Break character (see Section 31.4.3 “Auto-Wake-up on Break”).

2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible.

3: During the auto-baud process, the auto-baud counter starts counting at one. Upon completion of the auto-baud sequence, to achieve maximum accuracy, subtract 1 from the SP1BRGH:SP1BRGL register pair.

<table>
<thead>
<tr>
<th>TABLE 31-6: BRG COUNTER CLOCK RATES</th>
</tr>
</thead>
<tbody>
<tr>
<td>BRG16</td>
</tr>
<tr>
<td>-------</td>
</tr>
<tr>
<td>00</td>
</tr>
<tr>
<td>01</td>
</tr>
<tr>
<td>10</td>
</tr>
<tr>
<td>11</td>
</tr>
</tbody>
</table>

Note: During the ABD sequence, SP1BRRL and SP1BRGL registers are both used as a 16-bit counter, independent of the BRG16 setting.

**Note 1:** The ABD sequence requires the EUSART module to be configured in Asynchronous mode.

---

**FIGURE 31-6: AUTOMATIC BAUD RATE CALIBRATION**
31.4.2 AUTO-BAUD OVERFLOW

During the course of automatic baud detection, the ABDOVF bit of the BAUD1CON register will be set if the baud rate counter overflows before the fifth rising edge is detected on the RX pin. The ABDOVF bit indicates that the counter has exceeded the maximum count that can fit in the 16 bits of the SPxBRGH:SPxBRGL register pair. The overflow condition will set the RCIF flag. The counter continues to count until the fifth rising edge is detected on the RX pin. The RCIDL bit will remain false (‘0’) until the fifth rising edge at which time the RCIDL bit will be set. If the RC1REG is read after the overflow occurs, but before the fifth rising edge, the fifth rising edge will set the RCIF again.

Terminating the auto-baud process early to clear an overflow condition will prevent proper detection of the sync character’s fifth rising edge. If any falling edges of the sync character have not yet occurred when the ABDEN bit is cleared, those will be falsely detected as start bits. The following steps are recommended to clear the overflow condition:

1. Read RCREG to clear RCIF.
2. If RCIDL is zero then wait for RCIF and repeat step 1.
3. Clear the ABDOVF bit.

31.4.3 AUTO-WAKE-UP ON BREAK

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper character reception cannot be performed. The Auto-Wake-up feature allows the controller to wake-up due to activity on the RX/DT line. This feature is available only in Asynchronous mode.

The Auto-Wake-up feature is enabled by setting the WUE bit of the BAUD1CON register. Once set, the normal receive sequence on RX/DT is disabled, and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX/DT line. (This coincides with the start of a Sync Break or a wake-up signal character for the LIN protocol.)

The EUSART module generates an RCIF interrupt coincident with the wake-up event. The interrupt is generated synchronously to the Q clocks in normal CPU operating modes (Figure 31-7), and asynchronously if the device is in Sleep mode (Figure 31-8). The interrupt condition is cleared by reading the RC1REG register.

The WUE bit is automatically cleared by the low-to-high transition on the RX line at the end of the Break. This signals to the user that the Break event is over. At this point, the EUSART module is in Idle mode waiting to receive the next character.

31.4.3.1 Special Considerations

Break Character

To avoid character errors or character fragments during a wake-up event, the wake-up character must be all zeros.

When the wake-up is enabled the function works independent of the low time on the data stream. If the WUE bit is set and a valid non-zero character is received, the low time from the Start bit to the first rising edge will be interpreted as the wake-up event. The remaining bits in the character will be received as a fragmented character and subsequent characters can result in framing or overrun errors.

Therefore, the initial character in the transmission must be all ‘0’s. This must be ten or more bit times, 13-bit times recommended for LIN bus, or any number of bit times for standard RS-232 devices.

Oscillator Start-up Time

Oscillator start-up time must be considered, especially in applications using oscillators with longer start-up intervals (i.e., LP, XT or HS/PLL mode). The Sync Break (or wake-up signal) character must be of sufficient length, and be followed by a sufficient interval, to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

WUE Bit

The wake-up event causes a receive interrupt by setting the RCIF bit. The WUE bit is cleared in hardware by a rising edge on RX/DT. The interrupt condition is then cleared in software by reading the RC1REG register and discarding its contents.

To ensure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process before setting the WUE bit. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.
FIGURE 31-7: AUTO-WAKE-UP BIT (WUE) TIMING DURING NORMAL OPERATION

Note 1: The EUSART remains in Idle while the WUE bit is set.

FIGURE 31-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP

Note 1: If the wake-up event requires long oscillator warm-up time, the automatic clearing of the WUE bit can occur while the $stposc$ signal is still active. This sequence should not depend on the presence of Q clocks.

2: The EUSART remains in Idle while the WUE bit is set.
31.4.4 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. A Break character consists of a Start bit, followed by 12 '0' bits and a Stop bit.

To send a Break character, set the SENDB and TXEN bits of the TX1STA register. The Break character transmission is then initiated by a write to the TX1REG. The value of data written to TX1REG will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

The TRMT bit of the TX1STA register indicates when the transmit operation is active or idle, just as it does during normal transmission. See Figure 31-9 for the timing of the Break character sequence.

31.4.4.1 Break and Sync Transmit Sequence

The following sequence will start a message frame header made up of a Break, followed by an auto-baud Sync byte. This sequence is typical of a LIN bus master.

1. Configure the EUSART for the desired mode.
2. Set the TXEN and SENDB bits to enable the Break sequence.
3. Load the TX1REG with a dummy character to initiate transmission (the value is ignored).
4. Write '55h' to TX1REG to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TX1REG becomes empty, as indicated by the TXIF, the next data byte can be written to TX1REG.

31.4.5 RECEIVING A BREAK CHARACTER

The Enhanced EUSART module can receive a Break character in two ways.

The first method to detect a Break character uses the FERR bit of the RC1STA register and the received data as indicated by RC1REG. The Baud Rate Generator is assumed to have been initialized to the expected baud rate.

A Break character has been received when;
- RCIF bit is set
- FERR bit is set
- RC1REG = 00h

The second method uses the Auto-Wake-up feature described in Section 31.4.3 “Auto-Wake-up on Break”. By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCIF interrupt, and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Detect feature. For both methods, the user can set the ABDEN bit of the BAUD1CON register before placing the EUSART in Sleep mode.

**FIGURE 31-9: SEND BREAK CHARACTER SEQUENCE**
31.5 EUSART Synchronous Mode

Synchronous serial communications are typically used in systems with a single master and one or more slaves. The master device contains the necessary circuitry for baud rate generation and supplies the clock for all devices in the system. Slave devices can take advantage of the master clock by eliminating the internal clock generation circuitry.

There are two signal lines in Synchronous mode: a bidirectional data line and a clock line. Slaves use the external clock supplied by the master to shift the serial data into and out of their respective receive and transmit shift registers. Since the data line is bidirectional, synchronous operation is half-duplex only. Half-duplex refers to the fact that master and slave devices can receive and transmit data but not both simultaneously. The EUSART can operate as either a master or slave device.

Start and Stop bits are not used in synchronous transmissions.

31.5.1 SYNCHRONOUS MASTER MODE

The following bits are used to configure the EUSART for synchronous master operation:

- **SYNC = 1**
- **CSRC = 1**
- **SREN = 0** (for transmit); **SREN = 1** (for receive)
- **CREN = 0** (for transmit); **CREN = 1** (for receive)
- **SPEN = 1**

Setting the SYNC bit of the TX1STA register configures the device for synchronous operation. Setting the CSRC bit of the TX1STA register configures the device as a master. Clearing the SREN and CREN bits of the RC1STA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RC1STA register enables the EUSART.

31.5.1.1 Master Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a master transmits the clock on the TX/CK line. The TX/CK pin output driver is automatically enabled when the EUSART is configured for synchronous transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One clock cycle is generated for each data bit. Only as many clock cycles are generated as there are data bits.

31.5.1.2 Clock Polarity

A clock polarity option is provided for Microwire compatibility. Clock polarity is selected with the SCKP bit of the BAUD1CON register. Setting the SCKP bit sets the clock Idle state as high. When the SCKP bit is set, the data changes on the falling edge of each clock. Clearing the SCKP bit sets the Idle state as low. When the SCKP bit is cleared, the data changes on the rising edge of each clock.

31.5.1.3 Synchronous Master Transmission

Data is transferred out of the device on the RX/DT pin. The RX/DT and TX/CK pin output drivers are automatically enabled when the EUSART is configured for synchronous master transmit operation.

A transmission is initiated by writing a character to the TX1REG register. If the TSR still contains all or part of a previous character the new character data is held in the TX1REG until the last bit of the previous character has been transmitted. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TX1REG is immediately transferred to the TSR. The transmission of the character commences immediately following the transfer of the data to the TSR from the TX1REG.

Each data bit changes on the leading edge of the master clock and remains valid until the subsequent leading clock edge.

**Note:** The TSR register is not mapped in data memory, so it is not available to the user.

31.5.1.4 Synchronous Master Transmission Setup

1. Initialize the SP1BRGH, SP1BRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 31.4 “EUSART Baud Rate Generator (BRG)”).
2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
3. Disable Receive mode by clearing bits SREN and CREN.
4. Enable Transmit mode by setting the TXEN bit.
5. If 9-bit transmission is desired, set the TX9 bit.
6. If interrupts are desired, set the TXIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
7. If 9-bit transmission is selected, the ninth bit should be loaded in the TX9D bit.
8. Start transmission by loading data to the TX1REG register.
FIGURE 31-10: SYNCHRONOUS TRANSMISSION

Note: Sync Master mode, SP1BRGL = 0, continuous transmission of two 8-bit words.

FIGURE 31-11: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)
## TABLE 31-7: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
</tr>
<tr>
<td>BAUDCON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>ABDEN</td>
<td>365</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GI</td>
<td>ADIE</td>
<td>RClE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GI</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>95</td>
</tr>
<tr>
<td>RC1STA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>364</td>
</tr>
<tr>
<td>RxyPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
</tr>
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<td>SP1BRGL</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<tr>
<td>SP1BRGH</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>366</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TX1REG</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>EUSART Transmit Data Register 356*</td>
</tr>
<tr>
<td>TX1STA</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>363</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used for synchronous master transmission.
* Page provides register information.
31.5.1.5  Synchronous Master Reception

Data is received at the RX/DT pin. The RX/DT pin output driver is automatically disabled when the EUSART is configured for synchronous master receive operation.

In Synchronous mode, reception is enabled by setting either the Single Receive Enable bit (SREN of the RC1STA register) or the Continuous Receive Enable bit (CREN of the RC1STA register).

When SREN is set and CREN is clear, only as many clock cycles are generated as there are data bits in a single character. The SREN bit is automatically cleared at the completion of one character. When CREN is set, clocks are continuously generated until CREN is cleared. If CREN is cleared in the middle of a character the CK clock stops immediately and the partial character is discarded. If SREN and CREN are both set, then SREN is cleared at the completion of the first character and CREN takes precedence.

To initiate reception, set either SREN or CREN. Data is sampled at the RX/DT pin on the trailing edge of the TX/CK clock pin and is shifted into the Receive Shift Register (RSR). When a complete character is received into the RSR, the RCIF bit is set and the character is automatically transferred to the two character receive FIFO. The Least Significant eight bits of the top character in the receive FIFO are available in RC1REG. The RCIF bit remains set as long as there are unread characters in the receive FIFO.

**Note:** If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.

31.5.1.6  Slave Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a slave receives the clock on the TX/CK line. The TX/CK pin output driver is automatically disabled when the device is configured for synchronous slave transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One data bit is transferred for each clock cycle. Only as many clock cycles should be received as there are data bits.

**Note:** If the device is configured as a slave and the TX/CK function is on an analog pin, the corresponding ANSEL bit must be cleared.

31.5.1.7  Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before RC1REG is read to access the FIFO. When this happens the OERR bit of the RC1STA register is set. Previous data in the FIFO will not be overwritten. The two characters in the FIFO buffer can be read, however, no additional characters will be received until the error is cleared. The OERR bit can only be cleared by clearing the overrun condition. If the overrun error occurred when the SREN bit is set and CREN is clear then the error is cleared by reading RC1REG. If the overrun occurred when the CREN bit is set then the error condition is cleared by either clearing the CREN bit of the RC1STA register or by clearing the SPEN bit which resets the EUSART.

31.5.1.8  Receiving 9-Bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RC1STA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RC1STA register is the ninth, and Most Significant, data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RC1REG.

31.5.1.9  Synchronous Master Reception Setup

1. Initialize the SP1BRGH, SP1BRGL register pair for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
4. Ensure bits CREN and SREN are clear.
5. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
6. If 9-bit reception is desired, set bit RX9.
7. Start reception by setting the SREN bit or for continuous reception, set the CREN bit.
8. Interrupt flag bit RCIF will be set when reception of a character is complete. An interrupt will be generated if the enable bit RCIE was set.
9. Read the RC1STA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
10. Read the 8-bit received data by reading the RC1REG register.
11. If an overrun error occurs, clear the error by either clearing the CREN bit of the RC1STA register or by clearing the SPEN bit which resets the EUSART.
FIGURE 31-12: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

<table>
<thead>
<tr>
<th>RX/DT pin</th>
<th>bit 0</th>
<th>bit 1</th>
<th>bit 2</th>
<th>bit 3</th>
<th>bit 4</th>
<th>bit 5</th>
<th>bit 6</th>
<th>bit 7</th>
</tr>
</thead>
<tbody>
<tr>
<td>TX/CK pin (SCKP = 0)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TX/CK pin (SCKP = 1)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Write to bit SREN</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SREN bit</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>CREN bit</td>
<td>’0’</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>’0’</td>
</tr>
<tr>
<td>RCIF bit (Interrupt)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td></td>
</tr>
</tbody>
</table>

Note: Timing diagram demonstrates Sync Master mode with bit SREN = 1 and bit BRGH = 0.

TABLE 31-8: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
</tr>
<tr>
<td>BAUD1CON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>ABDEN</td>
<td>365</td>
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<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR2IE</td>
<td>92</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>95</td>
</tr>
<tr>
<td>RC1REG</td>
<td></td>
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<td></td>
<td></td>
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<tr>
<td>RC1STA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>364</td>
</tr>
<tr>
<td>RXPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>153</td>
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<tr>
<td>RxyPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
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<tr>
<td>SP1BRGL</td>
<td></td>
<td></td>
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<td></td>
<td>366*</td>
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<tr>
<td>SP1BRGH</td>
<td></td>
<td></td>
<td>SP1BRG&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>366*</td>
</tr>
<tr>
<td>TRISB</td>
<td>TRISB7</td>
<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
<td>131</td>
</tr>
<tr>
<td>TRISC</td>
<td>TRISC7</td>
<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
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<td>TRISC0</td>
<td>136</td>
</tr>
<tr>
<td>TX1STA</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>363</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used for synchronous master reception.
* Page provides register information.
31.5.2 SYNCHRONOUS SLAVE MODE

The following bits are used to configure the EUSART for synchronous slave operation:

- **SYNC** = 1
- **CSRC** = 0
- **SREN** = 0 (for transmit); **SREN** = 1 (for receive)
- **CREN** = 0 (for transmit); **CREN** = 1 (for receive)
- **SPEN** = 1

Setting the **SYNC** bit of the TX1STA register configures the device for synchronous operation. Clearing the **CSRC** bit of the TX1STA register configures the device as a slave. Clearing the **SREN** and **CREN** bits of the RC1STA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the **SPEN** bit of the RC1STA register enables the EUSART.

31.5.2.1 EUSART Synchronous Slave Transmit

The operation of the Synchronous Master and Slave modes are identical (see Section 31.5.1.3 “Synchronous Master Transmission”), except in the case of the Sleep mode.

If two words are written to the TX1REG and then the **SLEEP** instruction is executed, the following will occur:

1. The first character will immediately transfer to the TSR register and transmit.
2. The second word will remain in the TX1REG register.
3. The **TXIF** bit will not be set.
4. After the first character has been shifted out of TSR, the TX1REG register will transfer the second character to the TSR and the **TXIF** bit will now be set.
5. If the **PEIE** and **TXIE** bits are set, the interrupt will wake the device from Sleep and execute the next instruction. If the **GIE** bit is also set, the program will call the Interrupt Service Routine.

31.5.2.2 Synchronous Slave Transmission Setup

1. Set the **SYNC** and **SPEN** bits and clear the **CSRC** bit.
2. Clear the **ANSEL** bit for the CK pin (if applicable).
3. Clear the **CREN** and **SREN** bits.
4. If interrupts are desired, set the **TXIE** bit of the PIE1 register and the **GIE** and **PEIE** bits of the **INTCON** register.
5. If 9-bit transmission is desired, set the **TX9** bit.
6. Enable transmission by setting the **TXEN** bit.
7. If 9-bit transmission is selected, insert the Most Significant bit into the TX9D bit.
8. Start transmission by writing the Least Significant eight bits to the TX1REG register.
31.5.2.3 EUSART Synchronous Slave Reception

The operation of the Synchronous Master and Slave modes is identical (Section 31.5.1.5 “Synchronous Master Reception”), with the following exceptions:

- **Sleep**
  - CREN bit is always set, therefore the receiver is never idle
- **SREN bit**, which is a “don’t care” in Slave mode

A character may be received while in Sleep mode by setting the CREN bit prior to entering Sleep. Once the word is received, the RSR register will transfer the data to the RC1REG register. If the RCIE enable bit is set, the interrupt generated will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will branch to the interrupt vector.

31.5.2.4 Synchronous Slave Reception Setup

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Clear the ANSEL bit for both the CK and DT pins (if applicable).
3. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
4. If 9-bit reception is desired, set the RX9 bit.
5. Set the CREN bit to enable reception.
6. The RCIF bit will be set when reception is complete. An interrupt will be generated if the RCIE bit was set.
7. If 9-bit mode is enabled, retrieve the Most Significant bit from the RX9D bit of the RC1STA register.
8. Retrieve the eight Least Significant bits from the receive FIFO by reading the RC1REG register.
9. If an overrun error occurs, clear the error by either clearing the CREN bit of the RC1STA register or by clearing the SPEN bit which resets the EUSART.

### TABLE 31-9: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
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<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
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<tr>
<td>BAUD1CON</td>
<td>ABDVF</td>
<td>RCDL</td>
<td>—</td>
<td>SCKP</td>
<td>—</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>365</td>
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<tr>
<td>CKPPS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
<td>—</td>
<td>153</td>
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<tr>
<td>INTCN</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
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<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>91</td>
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<tr>
<td>PIE1</td>
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<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>92</td>
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<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
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<td>CCP1IF</td>
<td>TMR2IF</td>
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<td>SREN</td>
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<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>364</td>
</tr>
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<td>RxyPPS</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>154</td>
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<td>TRISB6</td>
<td>TRISB5</td>
<td>TRISB4</td>
<td>TRISB3</td>
<td>TRISB2</td>
<td>TRISB1</td>
<td>TRISB0</td>
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<td>TRISC6</td>
<td>TRISC5</td>
<td>TRISC4</td>
<td>TRISC3</td>
<td>TRISC2</td>
<td>TRISC1</td>
<td>TRISC0</td>
<td>136</td>
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<tr>
<td>TX1REG</td>
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<td>356*</td>
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<tr>
<td>TX1STA</td>
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<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>363</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used for synchronous slave transmission.

* Page provides register information.

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### TABLE 31-10: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELB</td>
<td>—</td>
<td>—</td>
<td>ANSB5</td>
<td>ANSB4</td>
<td>ANSB3</td>
<td>ANSB2</td>
<td>ANSB1</td>
<td>ANSB0</td>
<td>132</td>
</tr>
<tr>
<td>ANSELC</td>
<td>ANSC7</td>
<td>ANSC6</td>
<td>ANSC5</td>
<td>ANSC4</td>
<td>ANSC3</td>
<td>ANSC2</td>
<td>—</td>
<td>—</td>
<td>137</td>
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<tr>
<td>BAUD1CON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
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<td>SCKP</td>
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<td>—</td>
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<tr>
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<td>IOCIF</td>
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</tr>
<tr>
<td>PIE1</td>
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<td>RCIE</td>
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<td>CCP1IE</td>
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<tr>
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<td>TMR1GIF</td>
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<td>RCIF</td>
<td>TXIF</td>
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<td>RC1REG</td>
<td>EUSART Receive Data Register</td>
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<td>FERR</td>
<td>OERR</td>
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<td>TRISB</td>
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<td>TRISB5</td>
<td>TRISB4</td>
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<td>TX1STA</td>
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<td>BRGH</td>
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<td>TX9D</td>
<td>363</td>
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</tbody>
</table>

**Legend:** 
- — = unimplemented location, read as ‘0’. Shaded cells are not used for synchronous slave reception.
- * Page provides register information.
31.6 EUSART Operation During Sleep

The EUSART will remain active during Sleep only in the Synchronous Slave mode. All other modes require the system clock and therefore cannot generate the necessary signals to run the Transmit or Receive Shift registers during Sleep.

Synchronous Slave mode uses an externally generated clock to run the Transmit and Receive Shift registers.

31.6.1 SYNCHRONOUS RECEIVE DURING SLEEP

To receive during Sleep, all the following conditions must be met before entering Sleep mode:

- RC1STA and TX1STA Control registers must be configured for Synchronous Slave Reception (see Section 31.5.2.4 “Synchronous Slave Reception Setup”).
- If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- The RCIF interrupt flag must be cleared by reading RC1REG to unload any pending characters in the receive buffer.

Upon entering Sleep mode, the device will be ready to accept data and clocks on the RX/DT and TX/CK pins, respectively. When the data word has been completely clocked in by the external device, the RCIF interrupt flag bit of the PIR1 register will be set. Thereby, waking the processor from Sleep.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit is also set, then the Interrupt Service Routine at address 0004h will be called.

31.6.2 SYNCHRONOUS TRANSMIT DURING SLEEP

To transmit during Sleep, all the following conditions must be met before entering Sleep mode:

- The RC1STA and TX1STA Control registers must be configured for synchronous slave transmission (see Section 31.5.2.2 “Synchronous Slave Transmission Setup”).
- The TXIF interrupt flag must be cleared by writing the output data to the TX1REG, thereby filling the TSR and transmit buffer.
- If interrupts are desired, set the TXIE bit of the PIE1 register and the PEIE bit of the INTCON register.

Upon entering Sleep mode, the device will be ready to accept clocks on TX/CK pin and transmit data on the RX/DT pin. When the data word in the TSR has been completely clocked out by the external device, the pending byte in the TX1REG will transfer to the TSR and the TXIF flag will be set. Thereby, waking the processor from Sleep. At this point, the TX1REG is available to accept another character for transmission, which will clear the TXIF flag.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit is also set then the Interrupt Service Routine at address 0004h will be called.
32.0 IN-CIRCUIT SERIAL PROGRAMMING (ICSP™)

ICSP programming allows customers to manufacture circuit boards with unprogrammed devices. Programming can be done after the assembly process, allowing the device to be programmed with the most recent firmware or a custom firmware. Five pins are needed for ICSP programming:
- ICSPCLK
- ICSPDAT
- MCLR/VPP
- VDD
- VSS

In Program/Verify mode, the program memory, user IDs, and the Configuration Words are programmed through serial communications. The ICSPDAT pin is a bidirectional I/O used for transferring the serial data and the ICSPCLK pin is the clock input. For more information on ICSP refer to the “PIC16(L)F170X Memory Programming Specification” (DS41683).

32.1 High-Voltage Programming Entry Mode

The device is placed into High-Voltage Programming Entry mode by holding the ICSPCLK and ICSPDAT pins low then raising the voltage on MCLR/VPP to VIHH.

32.2 Low-Voltage Programming Entry Mode

The Low-Voltage Programming Entry mode allows the PIC Flash MCUs to be programmed using VDD only, without high voltage. When the LVP bit of Configuration Words is set to ‘1’, the low-voltage ICSP programming entry is enabled. To disable the Low-Voltage ICSP mode, the LVP bit must be programmed to ‘0’.

Entry into the Low-Voltage Programming Entry mode requires the following steps:
1. MCLR is brought to VIL.
2. A 32-bit key sequence is presented on ICSPDAT, while clocking ICSPCLK.

Once the key sequence is complete, MCLR must be held at VIL for as long as Program/Verify mode is to be maintained.

If low-voltage programming is enabled (LVP = 1), the MCLR Reset function is automatically enabled and cannot be disabled. See Section 5.5 “MCLR” for more information.

The LVP bit can only be reprogrammed to ‘0’ by using the High-Voltage Programming mode.

32.3 Common Programming Interfaces

Connection to a target device is typically done through an ICSP header. A commonly found connector on development tools is the RJ-11 in the 6P6C (6-pin, 6-connector) configuration. See Figure 32-1.

FIGURE 32-1: ICD RJ-11 STYLE CONNECTOR INTERFACE

Another connector often found in use with the PICkit™ programmers is a standard 6-pin header with 0.1 inch spacing. Refer to Figure 32-2.

For additional interface recommendations, refer to your specific device programmer manual prior to PCB design.

It is recommended that isolation devices be used to separate the programming pins from other circuitry. The type of isolation is highly dependent on the specific application and may include devices such as resistors, diodes, or even jumpers. See Figure 32-3 for more information.
**FIGURE 32-2: PICkit™ PROGRAMMER STYLE CONNECTOR INTERFACE**

Pin 1 Indicator

<table>
<thead>
<tr>
<th>Pin</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>VPP/MCLR</td>
</tr>
<tr>
<td>2</td>
<td>VDD Target</td>
</tr>
<tr>
<td>3</td>
<td>VSS (ground)</td>
</tr>
<tr>
<td>4</td>
<td>ICSPDAT</td>
</tr>
<tr>
<td>5</td>
<td>ICSPCLK</td>
</tr>
<tr>
<td>6</td>
<td>No Connect</td>
</tr>
</tbody>
</table>

* The 6-pin header (0.100" spacing) accepts 0.025" square pins.

**FIGURE 32-3: TYPICAL CONNECTION FOR ICSP™ PROGRAMMING**

External Programming Signals

- VDD
- VPP
- VSS
- Data
- Clock

Device to be Programmed

- VDD
- MCLR/VPP
- VSS
- ICSPDAT
- ICSPCLK

* Isolation devices (as required).
33.0 INSTRUCTION SET SUMMARY

Each instruction is a 14-bit word containing the operation code (opcode) and all required operands. The opcodes are broken into three broad categories.

- Byte Oriented
- Bit Oriented
- Literal and Control

The literal and control category contains the most varied instruction word format.

Table 33-3 lists the instructions recognized by the MPASM™ assembler.

All instructions are executed within a single instruction cycle, with the following exceptions, which may take two or three cycles:

- Subroutine takes two cycles (CALL, CALLW)
- Returns from interrupts or subroutines take two cycles (RETURN, RETLW, RETFIE)
- Program branching takes two cycles (GOTO, BRA, BRW, BTFSS, BTFSC, DECFSZ, INCFSZ)
- One additional instruction cycle will be used when any instruction references an indirect file register and the file select register is pointing to program memory.

One instruction cycle consists of four oscillator cycles; for an oscillator frequency of 4 MHz, this gives a nominal instruction execution rate of 1 MHz.

All instruction examples use the format ’0xhh’ to represent a hexadecimal number, where ‘h’ signifies a hexadecimal digit.

33.1 Read-Modify-Write Operations

Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (R-M-W) operation. The register is read, the data is modified, and the result is stored according to either the instruction, or the destination designator ‘d’. A read operation is performed on a register even if the instruction writes to that register.

<table>
<thead>
<tr>
<th>Field</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>f</td>
<td>Register file address (0x00 to 0x7F)</td>
</tr>
<tr>
<td>W</td>
<td>Working register (accumulator)</td>
</tr>
<tr>
<td>b</td>
<td>Bit address within an 8-bit file register</td>
</tr>
<tr>
<td>k</td>
<td>Literal field, constant data or label</td>
</tr>
<tr>
<td>x</td>
<td>Don’t care location (= 0 or 1). The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.</td>
</tr>
<tr>
<td>d</td>
<td>Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1.</td>
</tr>
<tr>
<td>n</td>
<td>FSR or INDF number. (0-1)</td>
</tr>
<tr>
<td>mm</td>
<td>Pre-post increment-decrement mode selection</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Field</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>PC</td>
<td>Program Counter</td>
</tr>
<tr>
<td>TO</td>
<td>Time-Out bit</td>
</tr>
<tr>
<td>C</td>
<td>Carry bit</td>
</tr>
<tr>
<td>DC</td>
<td>Digit Carry bit</td>
</tr>
<tr>
<td>Z</td>
<td>Zero bit</td>
</tr>
<tr>
<td>PD</td>
<td>Power-Down bit</td>
</tr>
</tbody>
</table>
### General Format for Instructions

#### Byte-oriented file register operations

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 8 7 6 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>d  f</td>
<td></td>
</tr>
</tbody>
</table>

- d = 0 for destination W
- d = 1 for destination f
- f = 7-bit file register address

#### Bit-oriented file register operations

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 10 9 7 6 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>b  f</td>
<td></td>
</tr>
</tbody>
</table>

- b = 3-bit address
- f = 7-bit file register address

#### Literal and control operations

**General**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 8 7 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>k</td>
<td></td>
</tr>
</tbody>
</table>

- k = 8-bit immediate value

**CALL and GOTO instructions only**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 11 10 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>k</td>
<td></td>
</tr>
</tbody>
</table>

- k = 11-bit immediate value

**MOVLP instruction only**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 7 6 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>k</td>
<td></td>
</tr>
</tbody>
</table>

- k = 7-bit immediate value

**MOVLB instruction only**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 5 4 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>k</td>
<td></td>
</tr>
</tbody>
</table>

- k = 5-bit immediate value

**BRA instruction only**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 9 8 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>k</td>
<td></td>
</tr>
</tbody>
</table>

- k = 9-bit immediate value

**FSR Offset instructions**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 7 6 5 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>n  k</td>
<td></td>
</tr>
</tbody>
</table>

- n = appropriate FSR
- k = 6-bit immediate value

**FSR Increment instructions**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 3 2 1 0</td>
<td>OPCODE</td>
</tr>
<tr>
<td>n  m</td>
<td></td>
</tr>
</tbody>
</table>

- n = appropriate FSR
- m = 2-bit mode value

**OPCODE only**

<table>
<thead>
<tr>
<th>Bit Position</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>13 0</td>
<td>OPCODE</td>
</tr>
</tbody>
</table>

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### TABLE 33-3: PIC16(L)F1717/8/9 INSTRUCTION SET

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDWF f, d</td>
<td>Add W and f</td>
<td>1</td>
<td>00 0111 0fff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>ADDWFC f, d</td>
<td>Add with Carry W and f</td>
<td>1</td>
<td>11 1101 0fff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>ANDWF f, d</td>
<td>AND W with f</td>
<td>1</td>
<td>00 0101 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>ASRF f, d</td>
<td>Arithmetic Right Shift</td>
<td>1</td>
<td>11 0111 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>LSFL f, d</td>
<td>Logical Left Shift</td>
<td>1</td>
<td>11 0101 ffff ffff</td>
<td>C, Z</td>
<td>2</td>
</tr>
<tr>
<td>LSRF f, d</td>
<td>Logical Right Shift</td>
<td>1</td>
<td>11 0110 ffff ffff</td>
<td>C, Z</td>
<td>2</td>
</tr>
<tr>
<td>CLRF f</td>
<td>Clear f</td>
<td>1</td>
<td>00 0001 1fff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>CLRWF –</td>
<td>Clear W</td>
<td>1</td>
<td>00 0000 0fff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>COMF f, d</td>
<td>Complement f</td>
<td>1</td>
<td>00 1001 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>DECF f, d</td>
<td>Decrement f</td>
<td>1</td>
<td>00 0011 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>INCF f, d</td>
<td>Increment f</td>
<td>1</td>
<td>00 1010 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>IORWF f, d</td>
<td>Inclusive OR W with f</td>
<td>1</td>
<td>00 0100 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>MOVF f, d</td>
<td>Move f</td>
<td>1</td>
<td>00 1000 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>MOVWF f</td>
<td>Move W to f</td>
<td>1</td>
<td>00 0000 1fff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>RLF f, d</td>
<td>Rotate Left f through Carry</td>
<td>1</td>
<td>00 1101 ffff ffff</td>
<td>C</td>
<td>2</td>
</tr>
<tr>
<td>RRF f, d</td>
<td>Rotate Right f through Carry</td>
<td>1</td>
<td>00 1100 ffff ffff</td>
<td>C</td>
<td>2</td>
</tr>
<tr>
<td>SUBWF f, d</td>
<td>Subtract W from f</td>
<td>1</td>
<td>00 0010 ffff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>SUBWFB f, d</td>
<td>Subtract with Borrow W from f</td>
<td>1</td>
<td>11 1011 ffff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>SWAPF f, d</td>
<td>Swap nibbles in f</td>
<td>1</td>
<td>00 1110 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>XORWF f, d</td>
<td>Exclusive OR W with f</td>
<td>1</td>
<td>00 0110 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
</tbody>
</table>

#### BYTE-ORIENTED FILE REGISTER OPERATIONS

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>DECFSZ f, d</td>
<td>Decrement f, Skip if 0</td>
<td>1(2)</td>
<td>00 1011 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>INCFSZ f, d</td>
<td>Increment f, Skip if 0</td>
<td>1(2)</td>
<td>00 1111 ffff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
</tbody>
</table>

#### BIT-ORIENTED FILE REGISTER OPERATIONS

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>BCF f, b</td>
<td>Bit Clear f</td>
<td>1</td>
<td>01 00bb bfff ffff</td>
<td></td>
<td>2</td>
</tr>
<tr>
<td>BSF f, b</td>
<td>Bit Set f</td>
<td>1</td>
<td>01 01bb bfff ffff</td>
<td></td>
<td>2</td>
</tr>
</tbody>
</table>

#### BIT-ORIENTED SKIP OPERATIONS

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>BTFSC f, b</td>
<td>Bit Test f, Skip if Clear</td>
<td>1 (2)</td>
<td>01 10bb bfff ffff</td>
<td></td>
<td>2</td>
</tr>
<tr>
<td>BTFSS f, b</td>
<td>Bit Test f, Skip if Set</td>
<td>1 (2)</td>
<td>01 11bb bfff ffff</td>
<td></td>
<td>2</td>
</tr>
</tbody>
</table>

#### LITERAL OPERATIONS

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDLW k</td>
<td>Add literal and W</td>
<td>1</td>
<td>11 1110 kkkk kkkk</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>ANDLW k</td>
<td>AND literal with W</td>
<td>1</td>
<td>11 1001 kkkk kkkk</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>IORLW k</td>
<td>Inclusive OR literal with W</td>
<td>1</td>
<td>11 1000 kkkk kkkk</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>MOVLB k</td>
<td>Move literal to BSR</td>
<td>1</td>
<td>00 0000 001k kkkk</td>
<td></td>
<td>2</td>
</tr>
<tr>
<td>MOVLP k</td>
<td>Move literal to PCLATH</td>
<td>1</td>
<td>11 0001 1kkk kkkk</td>
<td></td>
<td>2</td>
</tr>
<tr>
<td>MOVLL k</td>
<td>Move literal to W</td>
<td>1</td>
<td>11 0000 kkkk kkkk</td>
<td></td>
<td>2</td>
</tr>
<tr>
<td>SUBLW k</td>
<td>Subtract W from literal</td>
<td>1</td>
<td>11 1100 kkkk kkkk</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>XORLW k</td>
<td>Exclusive OR literal with W</td>
<td>1</td>
<td>11 1010 kkkk kkkk</td>
<td>Z</td>
<td>2</td>
</tr>
</tbody>
</table>

#### CONTROL OPERATIONS

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>BRA k</td>
<td>Relative Branch</td>
<td>2</td>
<td>11 001k kkkk kkkk</td>
<td></td>
<td></td>
</tr>
<tr>
<td>BRW –</td>
<td>Relative Branch with W</td>
<td>2</td>
<td>00 0000 0000 1011</td>
<td></td>
<td></td>
</tr>
<tr>
<td>CALL k</td>
<td>Call Subroutine</td>
<td>1</td>
<td>11 kkkk kkkk kkkk</td>
<td></td>
<td></td>
</tr>
<tr>
<td>CALLW –</td>
<td>Call Subroutine with W</td>
<td>2</td>
<td>00 0000 0000 1010</td>
<td></td>
<td></td>
</tr>
<tr>
<td>GOTO k</td>
<td>Go to address</td>
<td>2</td>
<td>11 1kkk kkkk kkkk</td>
<td></td>
<td></td>
</tr>
<tr>
<td>RETEIE k</td>
<td>Return from interrupt</td>
<td>2</td>
<td>00 0000 0000 1001</td>
<td></td>
<td></td>
</tr>
<tr>
<td>RETLW k</td>
<td>Return with literal in W</td>
<td>2</td>
<td>11 0100 kkkk kkkk</td>
<td></td>
<td></td>
</tr>
<tr>
<td>RETURN –</td>
<td>Return from Subroutine</td>
<td>2</td>
<td>00 0000 0000 1000</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Note 1:** If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

**Note 2:** If this instruction addresses an INDF register and the MSb of the corresponding FSR is set, this instruction will require one additional instruction cycle.
### INHERENT OPERATIONS

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLRWDT –</td>
<td>Clear Watchdog Timer</td>
<td>1</td>
<td>00 0000 0110 0100</td>
<td>TO, PD</td>
<td></td>
</tr>
<tr>
<td>NOP –</td>
<td>No Operation</td>
<td>1</td>
<td>00 0000 0000 0000</td>
<td></td>
<td></td>
</tr>
<tr>
<td>OPTION –</td>
<td>Load OPTION_REG register with W</td>
<td>1</td>
<td>00 0000 0110 0010</td>
<td></td>
<td></td>
</tr>
<tr>
<td>RESET –</td>
<td>Software device Reset</td>
<td>1</td>
<td>00 0000 0000 0001</td>
<td></td>
<td></td>
</tr>
<tr>
<td>SLEEP –</td>
<td>Go into Standby mode</td>
<td>1</td>
<td>00 0000 0110 0013</td>
<td>TO, PD</td>
<td></td>
</tr>
<tr>
<td>TRIS f</td>
<td>Load TRIS register with W</td>
<td>1</td>
<td>00 0000 0110 00ff</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

### C-COMPILER OPTIMIZED

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDFSR n, k</td>
<td>Add Literal k to FSRn</td>
<td>1</td>
<td>11 0001 0nkk kkkk</td>
<td>Z</td>
<td>2, 3</td>
</tr>
<tr>
<td>MOVIW n mm</td>
<td>Move Indirect FSRn to W with pre/post inc/dec modifier, mm</td>
<td>1</td>
<td>11 1111 0nkk kkkk</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>MOVWI n mm</td>
<td>Move W to Indirect FSRn with pre/post inc/dec modifier, mm</td>
<td>1</td>
<td>00 0000 0001 1nmm</td>
<td>Z</td>
<td>2, 3</td>
</tr>
<tr>
<td>k[n]</td>
<td>Move INDFn to W, Indexed Indirect.</td>
<td>1</td>
<td>11 1111 1nkk kkkk</td>
<td></td>
<td>2</td>
</tr>
</tbody>
</table>

**Note 1:** If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

**Note 2:** If this instruction addresses an INDF register and the MSb of the corresponding FSR is set, this instruction will require one additional instruction cycle.
### 33.2 Instruction Descriptions

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>ADDFSR</strong> Add Literal to FSRn&lt;br&gt;Syntax: <code>[label] ADDFSR FSRn, k</code>&lt;br&gt;Operands: <code>-32 \leq k \leq 31</code>&lt;br&gt;Operation: <code>FSR(n) + k \rightarrow FSR(n)</code>&lt;br&gt;Status Affected: None&lt;br&gt;Description: The signed 6-bit literal 'k' is added to the contents of the FSRnH:FSRNL register pair.&lt;br&gt;FSRn is limited to the range 0000h-FFFFh. Moving beyond these bounds will cause the FSR to wrap-around.</td>
<td><strong>ADDFSR</strong> Add Literal to FSRn&lt;br&gt;Syntax: <code>[label] ADDFSR FSRn, k</code>&lt;br&gt;Operands: <code>-32 \leq k \leq 31</code>&lt;br&gt;Operation: <code>FSR(n) + k \rightarrow FSR(n)</code>&lt;br&gt;Status Affected: None&lt;br&gt;Description: The signed 6-bit literal 'k' is added to the contents of the FSRnH:FSRNL register pair.&lt;br&gt;FSRn is limited to the range 0000h-FFFFh. Moving beyond these bounds will cause the FSR to wrap-around.</td>
</tr>
<tr>
<td><strong>ADDLW</strong> Add literal and W&lt;br&gt;Syntax: <code>[label] ADDLW k</code>&lt;br&gt;Operands: <code>0 \leq k \leq 255</code>&lt;br&gt;Operation: <code>(W) + k \rightarrow (W)</code>&lt;br&gt;Status Affected: C, DC, Z&lt;br&gt;Description: The contents of the W register are added to the 8-bit literal 'k' and the result is placed in the W register.</td>
<td><strong>ADDWFC</strong> ADD W and CARRY bit to f&lt;br&gt;Syntax: <code>[label] ADDWFC f {,d}</code>&lt;br&gt;Operands: <code>0 \leq f \leq 127</code>&lt;br&gt;d \in [0,1]&lt;br&gt;Operation: <code>(W) + (f) + (C) \rightarrow dest</code>&lt;br&gt;Status Affected: C, DC, Z&lt;br&gt;Description: Add W, the Carry flag and data memory location 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in data memory location 'f'.</td>
</tr>
<tr>
<td><strong>ADDWF</strong> Add W and 'f'&lt;br&gt;Syntax: <code>[label] ADDWF f,d</code>&lt;br&gt;Operands: <code>0 \leq f \leq 127</code>&lt;br&gt;d \in [0,1]&lt;br&gt;Operation: <code>(W) + (f) \rightarrow (destination)</code>&lt;br&gt;Status Affected: C, DC, Z&lt;br&gt;Description: The contents of register 'f' are shifted one bit to the right through the Carry flag. The MSb remains unchanged. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.</td>
<td><strong>ANDLW</strong> AND literal with W&lt;br&gt;Syntax: <code>[label] ANDLW k</code>&lt;br&gt;Operands: <code>0 \leq k \leq 255</code>&lt;br&gt;Operation: <code>(W) .AND. (k) \rightarrow (W)</code>&lt;br&gt;Status Affected: Z&lt;br&gt;Description: The contents of W register are AND'ed with the 8-bit literal 'k'. The result is placed in the W register.</td>
</tr>
<tr>
<td><strong>ANDWF</strong> AND W with 'f'&lt;br&gt;Syntax: <code>[label] ANDWF f,d</code>&lt;br&gt;Operands: <code>0 \leq f \leq 127</code>&lt;br&gt;d \in [0,1]&lt;br&gt;Operation: <code>(W) .AND. (f) \rightarrow (destination)</code>&lt;br&gt;Status Affected: Z&lt;br&gt;Description: AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.</td>
<td><strong>ASRF</strong> Arithmetic Right Shift&lt;br&gt;Syntax: <code>[label] ASRF f {.d}</code>&lt;br&gt;Operands: <code>0 \leq f \leq 127</code>&lt;br&gt;d \in [0,1]&lt;br&gt;Operation: <code>(f&lt;7:0&gt;) \rightarrow dest&lt;7&gt;</code>(f&lt;7:1&gt;) \rightarrow dest&lt;6:0&gt;, (f&lt;0&gt;) \rightarrow C, C&lt;br&gt;Status Affected: C, Z&lt;br&gt;Description: The contents of register 'f' are shifted one bit to the right through the Carry flag. The MSb remains unchanged. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.</td>
</tr>
</tbody>
</table>
### BCF  Bit Clear f

**Syntax:**

\[
\text{[ label ] BCF } \ f,b
\]

**Operands:**

\[
0 \leq f \leq 127 \\
0 \leq b \leq 7
\]

**Operation:**

\[
0 \rightarrow \langle f|b\rangle
\]

**Status Affected:** None

**Description:** Bit 'b' in register 'f' is cleared.

### BRA  Relative Branch

**Syntax:**

\[
\text{[ label ] BRA } \ label \\
\text{[ label ] BRA } \ $+k
\]

**Operands:**

\[
-256 \leq \text{label} - \text{PC} + 1 \leq 255 \\
-256 \leq k \leq 255
\]

**Operation:**

\[
\text{(PC)} + 1 + k \rightarrow \text{PC}
\]

**Status Affected:** None

**Description:** Add the signed 9-bit literal ‘k’ to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + k. This instruction is a 2-cycle instruction. This branch has a limited range.

### BSF  Bit Set f

**Syntax:**

\[
\text{[ label ] BSF } \ f,b
\]

**Operands:**

\[
0 \leq f \leq 127 \\
0 \leq b \leq 7
\]

**Operation:**

\[
1 \rightarrow \langle f|b\rangle
\]

**Status Affected:** None

**Description:** Bit 'b' in register 'f' is set.

### BTFSC  Bit Test f, Skip if Clear

**Syntax:**

\[
\text{[ label ] BTFSC } \ f,b
\]

**Operands:**

\[
0 \leq f \leq 127 \\
0 \leq b \leq 7
\]

**Operation:**

skip if \langle f|b\rangle = 0

**Status Affected:** None

**Description:** If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b', in register 'f', is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.

### BTFSS  Bit Test f, Skip if Set

**Syntax:**

\[
\text{[ label ] BTFSS } \ f,b
\]

**Operands:**

\[
0 \leq f \leq 127 \\
0 \leq b \leq 7
\]

**Operation:**

skip if \langle f|b\rangle = 1

**Status Affected:** None

**Description:** If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.

### BRW  Relative Branch with W

**Syntax:**

\[
\text{[ label ] BRW}
\]

**Operands:** None

**Operation:**

\[
\text{(PC)} + \langle W \rangle \rightarrow \text{PC}
\]

**Status Affected:** None

**Description:** Add the contents of W (unsigned) to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + (W). This instruction is a 2-cycle instruction.
<table>
<thead>
<tr>
<th>CALL</th>
<th>Call Subroutine</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>[ label] CALL ( k )</td>
</tr>
<tr>
<td>Operands:</td>
<td>( 0 \leq k \leq 2047 )</td>
</tr>
<tr>
<td>Operation:</td>
<td>( (PC)+1 \rightarrow \text{TOS}, ) ( k \rightarrow \text{PC}&lt;10:0&gt;, ) ( \text{PCLATH}&lt;6:3&gt; \rightarrow \text{PC}&lt;14:11&gt; )</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>None</td>
</tr>
<tr>
<td>Description:</td>
<td>Call Subroutine. First, return address ((PC + 1)) is pushed onto the stack. The 11-bit immediate address is loaded into PC bits (&lt;10:0&gt;). The upper bits of the PC are loaded from PCLATH. CALL is a 2-cycle instruction.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>CALLW</th>
<th>Subroutine Call With W</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>[ label] CALLW</td>
</tr>
<tr>
<td>Operands:</td>
<td>None</td>
</tr>
<tr>
<td>Operation:</td>
<td>( (PC) +1 \rightarrow \text{TOS}, ) ( W \rightarrow \text{PC}&lt;7:0&gt;, ) ( \text{PCLATH}&lt;6:0&gt; \rightarrow \text{PC}&lt;14:8&gt; )</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>None</td>
</tr>
<tr>
<td>Description:</td>
<td>Subroutine call with W. First, the return address ((PC + 1)) is pushed onto the return stack. Then, the contents of W are loaded into PC bits (&lt;7:0&gt;), and the contents of PCLATH into PC bits (&lt;14:8&gt;). CALLW is a 2-cycle instruction.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>CLRF</th>
<th>Clear f</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>[ label] CLRF ( f )</td>
</tr>
<tr>
<td>Operands:</td>
<td>( 0 \leq f \leq 127 )</td>
</tr>
<tr>
<td>Operation:</td>
<td>( 00h \rightarrow W ) ( f \rightarrow Z )</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>Z</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register ( f ) are cleared and the Z bit is set.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>CLR</th>
<th>Clear W</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>[ label] CLR</td>
</tr>
<tr>
<td>Operands:</td>
<td>None</td>
</tr>
<tr>
<td>Operation:</td>
<td>( 00h \rightarrow W ) ( Z \rightarrow Z )</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>Z</td>
</tr>
<tr>
<td>Description:</td>
<td>W register is cleared. Zero bit ( Z ) is set.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>CLRDT</th>
<th>Clear Watchdog Timer</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>[ label] CLRDT</td>
</tr>
<tr>
<td>Operands:</td>
<td>None</td>
</tr>
<tr>
<td>Operation:</td>
<td>( 00h \rightarrow \text{WDT} ) ( 0 \rightarrow \text{WDT prescaler}, ) ( 1 \rightarrow \text{TO} ) ( 1 \rightarrow \text{PD} )</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>( \text{TO}, \text{PD} )</td>
</tr>
<tr>
<td>Description:</td>
<td>CLRDT instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. Status bits ( \text{TO} ) and ( \text{PD} ) are set.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>COMF</th>
<th>Complement f</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>[ label] COMF ( f,d )</td>
</tr>
<tr>
<td>Operands:</td>
<td>( 0 \leq f \leq 127 ) ( d \in {0,1} )</td>
</tr>
<tr>
<td>Operation:</td>
<td>( \overline{f} \rightarrow \text{destination} )</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>Z</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register ( f ) are complemented. If ( d ) is ( '0' ), the result is stored in ( W ). If ( d ) is ( '1' ), the result is stored back in register ( f ).</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>DECF</th>
<th>Decrement f</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>[ label] DECF ( f,d )</td>
</tr>
<tr>
<td>Operands:</td>
<td>( 0 \leq f \leq 127 ) ( d \in {0,1} )</td>
</tr>
<tr>
<td>Operation:</td>
<td>( \overline{(f)} - 1 \rightarrow \text{destination} )</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>Z</td>
</tr>
<tr>
<td>Description:</td>
<td>Decrement register ( f ). If ( d ) is ( '0' ), the result is stored in the ( W ) register. If ( d ) is ( '1' ), the result is stored back in register ( f ).</td>
</tr>
</tbody>
</table>
### DECFSZ: Decrement f, Skip if 0

**Syntax:**
```assembly
[label]    DECFSZ   f,d
```

**Operands:**
- `0 ≤ f ≤ 127`
- `d ∈ [0,1]`

**Operation:**
- `(f) - 1 → (destination);`
  - skip if result = 0

**Status Affected:** None

**Description:**
The contents of register `f` are decremented. If `d` is '0', the result is placed in the W register. If `d` is '1', the result is placed back in register `f`. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle instruction.

### INCFSZ: Increment f, Skip if 0

**Syntax:**
```assembly
[label]    INCFSZ   f,d
```

**Operands:**
- `0 ≤ f ≤ 127`
- `d ∈ [0,1]`

**Operation:**
- `(f) + 1 → (destination),`
  - skip if result = 0

**Status Affected:** None

**Description:**
The contents of register `f` are incremented. If `d` is '0', the result is placed in the W register. If `d` is '1', the result is placed back in register `f`. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle instruction.

### GOTO: Unconditional Branch

**Syntax:**
```assembly
[label]    GOTO   k
```

**Operands:**
- `0 ≤ k ≤ 2047`

**Operation:**
- `k → PC<10:0>
  - PCLATH<6:3> → PC<14:11>

**Status Affected:** None

**Description:**
`GOTO` is an unconditional branch. The 11-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<6:3>. `GOTO` is a 2-cycle instruction.

### IORLW: Inclusive OR literal with W

**Syntax:**
```assembly
[label]    IORLW   k
```

**Operands:**
- `0 ≤ k ≤ 255`

**Operation:**
- `(W) .OR. k → (W)

**Status Affected:** Z

**Description:**
The contents of the W register are OR'ed with the 8-bit literal 'k'. The result is placed in the W register.

### IORWF: Inclusive OR W with f

**Syntax:**
```assembly
[label]    IORWF    f,d
```

**Operands:**
- `0 ≤ f ≤ 127`
- `d ∈ [0,1]`

**Operation:**
- `(W) .OR. (f) → (destination)

**Status Affected:** Z

**Description:**
Inclusive OR the W register with register 'f'. If `d` is '0', the result is placed in the W register. If `d` is '1', the result is placed back in register 'f'.

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### LSLF Logical Left Shift

**Syntax:** \([ label \] LSLF f\{,d\} \)

**Operands:**
- \(0 \leq f \leq 127\)
- \(d \in [0,1]\)

**Operation:**
- \((f<7>) \rightarrow C\)
- \((f<6:0>) \rightarrow \text{dest}<7:1>\)
- \(0 \rightarrow \text{dest}<0>\)

**Status Affected:** C, Z

**Description:** The contents of register ‘f’ are shifted one bit to the left through the Carry flag. A ‘0’ is shifted into the LSb. If ‘d’ is ‘0’, the result is placed in W. If ‘d’ is ‘1’, the result is stored back in register ‘f’.

### LSRF Logical Right Shift

**Syntax:** \([ label \] LSRF f\{,d\} \)

**Operands:**
- \(0 \leq f \leq 127\)
- \(d \in [0,1]\)

**Operation:**
- \(0 \rightarrow \text{dest}<7>\)
- \((f<7:1>) \rightarrow \text{dest}<6:0>,\)
- \((f<0>) \rightarrow C,\)

**Status Affected:** C, Z

**Description:** The contents of register ‘f’ are shifted one bit to the right through the Carry flag. A ‘0’ is shifted into the MSb. If ‘d’ is ‘0’, the result is placed in W. If ‘d’ is ‘1’, the result is stored back in register ‘f’.

### MOVF Move f

**Syntax:** \([ label \] MOVF f,d \)

**Operands:**
- \(0 \leq f \leq 127\)
- \(d \in [0,1]\)

**Operation:** \((f) \rightarrow (\text{dest})\)

**Status Affected:** Z

**Description:** The contents of register f is moved to a destination dependent upon the status of d. If d = 0, destination is W register. If d = 1, the destination is file register f itself. d = 1 is useful to test a file register since status flag Z is affected.

**Example:**

```
MOV FSR, 0
```

**After Instruction**
- \(W = \) value in FSR register
- \(Z = 1\)
MOVIW  Move INDFn to W
Syntax:  
[ label ] MOVIW ++FSRn
[ label ] MOVIW --FSRn
[ label ] MOVIW FSRn++
[ label ] MOVIW FSRn--
[ label ] MOVIW k[FSRn]
Operands:  
n ∈ [0, 1]
mm ∈ [00, 01, 10, 11]
-32 ≤ k ≤ 31
Operation:  
INDFn → W
Effective address is determined by
• FSR + 1 (preincrement)
• FSR - 1 (predecrement)
• FSR + k (relative offset)
After the Move, the FSR value will be
either:
• FSR + 1 (all increments)
• FSR - 1 (all decrements)
• Unchanged
Status Affected:  Z

MOVLP  Move literal to PCLATH
Syntax:  
[ label ] MOVLP k
Operands:  
0 ≤ k ≤ 127
Operation:  
k → PCLATH
Status Affected:  None
Description:  
The 7-bit literal ‘k’ is loaded into the PCLATH register.

MOVLW  Move literal to W
Syntax:  
[ label ] MOVLW k
Operands:  
0 ≤ k ≤ 255
Operation:  
k → W
Status Affected:  None
Description:  
The 8-bit literal ‘k’ is loaded into W register. The “don’t cares” will assemble as ‘0’s.
Words:  1
Cycles:  1
Example:  
MOVLW 0x5A
After Instruction
W = 0x5A

MOVWF  Move W to f
Syntax:  
[ label ] MOVWF f
Operands:  
0 ≤ f ≤ 127
Operation:  
(W) → (f)
Status Affected:  None
Description:  
Move data from W register to register ‘f’.
Words:  1
Cycles:  1
Example:  
MOVWF OPTION_REG
Before Instruction
OPTION_REG = 0xFF
W = 0x4F
After Instruction
OPTION_REG = 0x4F
W = 0x4F

MOVLB  Move literal to BSR
Syntax:  
[ label ] MOVLB k
Operands:  
0 ≤ k ≤ 31
Operation:  
k → BSR
Status Affected:  None
Description:  
The 5-bit literal ‘k’ is loaded into the Bank Select Register (BSR).
MOVWI Move W to INDFn

Syntax:

[ label ] MOVWI ++FSRn
[ label ] MOVWI --FSRn
[ label ] MOVWI FSRn++
[ label ] MOVWI FSRn--
[ label ] MOVWI k[FSRn]

Operands:

n ∈ [0,1]
mm ∈ [00,01,10,11]
-32 ≤ k ≤ 31

Operation: W → INDFn

Effective address is determined by
- FSR + 1 (preincrement)
- FSR - 1 (predecrement)
- FSR + k (relative offset)

After the Move, the FSR value will be:
- FSR + 1 (all increments)
- FSR - 1 (all decrements)
- Unchanged

Status Affected: None

Description: This instruction is used to move data between W and one of the indirect registers (INDFn). Before/after this move, the pointer (FSRn) is updated by pre/post incrementing/decrementing it.

Note: The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the FSRn.

FSRn is limited to the range 0000h-FFFFh. Incrementing/decrementing it beyond these bounds will cause it to wrap-around.

The increment/decrement operation on FSRn WILL NOT affect any Status bits.

NOP No Operation

Syntax: [ label ] NOP

Operands: None

Operation: No operation

Status Affected: None

Description: No operation

Words: 1

Cycles: 1

Example: NOP

OPTION Load OPTION_REG Register with W

Syntax: [ label ] OPTION

Operands: None

Operation: (W) → OPTION_REG

Status Affected: None

Description: Move data from W register to OPTION_REG register.

Words: 1

Cycles: 1

Example: OPTION

Before Instruction

OPTION_REG = 0xFF
W = 0x4F

After Instruction

OPTION_REG = 0x4F
W = 0x4F

RESET Software Reset

Syntax: [ label ] RESET

Operands: None

Operation: Execute a device Reset. Resets the RI flag of the PCON register.

Status Affected: None

Description: This instruction provides a way to execute a hardware Reset by software.
### RETFIE

**Return from Interrupt**

**Syntax:**

\[
[\text{label}] \quad \text{RETFIE} \ k
\]

**Operands:**

None

**Operation:**

\[
\text{TOS} \rightarrow \text{PC}, \\
1 \rightarrow \text{GIE}
\]

**Status Affected:**

None

**Description:**

Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a 2-cycle instruction.

**Words:** 1

**Cycles:** 2

**Example:**

**After Interrupt**

\[
\begin{align*}
\text{PC} &= \text{TOS} \\
\text{GIE} &= 1
\end{align*}
\]

### RETLW

**Return with literal in W**

**Syntax:**

\[
[\text{label}] \quad \text{RETLW} \ k
\]

**Operands:**

\[
0 \leq k \leq 255
\]

**Operation:**

\[
k \rightarrow (W); \\
\text{TOS} \rightarrow \text{PC}
\]

**Status Affected:**

None

**Description:**

The W register is loaded with the 8-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a 2-cycle instruction.

**Words:** 1

**Cycles:** 2

**Example:**

```
CALL TABLE;W contains table offset value
        ;W now has table value
        
        TABLE
        
ADDWF PC ;W = offset
RETLW k1 ;Begin table
RETLW k2 ;
        
        RETLW kn ; End of table
```

### RLF

**Rotate Left f through Carry**

**Syntax:**

\[
[\text{label}] \quad \text{RLF} \ f,d
\]

**Operands:**

\[
0 \leq f \leq 127 \\
d \in [0,1]
\]

**Operation:**

See description below

**Status Affected:**

C

**Description:**

The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is stored back in register 'f'.

**Words:** 1

**Cycles:** 1

**Example:**

```
RLF REG1,0
```

**Before Instruction**

```
REG1 = 1110 0110 \\
C = 0
```

**After Instruction**

```
REG1 = 1110 0110 \\
W = 1100 1100 \\
C = 1
```
**RRF**

*Rotate Right f through Carry*

**Syntax:** \[
[label]\] RRF \ f,d

**Operands:**
\[
0 \leq f \leq 127
\]
\[
d \in [0,1]
\]

**Operation:** See description below

**Status Affected:** C

**Description:** The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

---

**SLEEP**

*Enter Sleep mode*

**Syntax:** \[
[label]\] SLEEP

**Operands:** None

**Operation:**
\[
00h \rightarrow \text{WDT},
0 \rightarrow \text{WDT prescaler},
1 \rightarrow \text{TO},
0 \rightarrow \text{PD}
\]

**Status Affected:** TO, PD

**Description:** The power-down Status bit, PD, is cleared. Time-out Status bit, TO, is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.

---

**SUBLW**

*Subtract W from literal*

**Syntax:** \[
[label]\] SUBLW \ k

**Operands:**
\[
0 \leq k \leq 255
\]

**Operation:** \[
k - (W) \rightarrow (W)
\]

**Status Affected:** C, DC, Z

**Description:** The W register is subtracted (2’s complement method) from the 8-bit literal 'k'. The result is placed in the W register.

![Status Conditions](image)

---

**SUBWF**

*Subtract W from f*

**Syntax:** \[
[label]\] SUBWF \ f,d

**Operands:**
\[
0 \leq f \leq 127
\]
\[
d \in [0,1]
\]

**Operation:** \[
(f) - (W) \rightarrow \text{destination}
\]

**Status Affected:** C, DC, Z

**Description:** Subtract (2’s complement method) W register from register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

![Status Conditions](image)

---

**SUBWFB**

*Subtract W from f with Borrow*

**Syntax:** SUBWFB \ f \{,d\}

**Operands:**
\[
0 \leq f \leq 127
\]
\[
d \in [0,1]
\]

**Operation:** \[
(f) - (W) - (B) \rightarrow \text{dest}
\]

**Status Affected:** C, DC, Z

**Description:** Subtract W and the BORROW flag (CARRY) from register 'f' (2’s complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

![Status Conditions](image)
### SWAPF
**Swap Nibbles in f**

**Syntax:**

\[
\text{[ label ] SWAPF f,d}
\]

**Operands:**

- \(0 \leq f \leq 127\)
- \(d \in [0,1]\)

**Operation:**

- \((f<3:0>) \rightarrow (\text{destination}<7:4>)\)
- \((f<7:4>) \rightarrow (\text{destination}<3:0>)\)

**Status Affected:** None

**Description:** The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.

### TRIS
**Load TRIS Register with W**

**Syntax:**

\[
\text{[ label ] TRIS f}
\]

**Operands:**

- \(5 \leq f \leq 7\)

**Operation:**

\((W) \rightarrow \text{TRIS register 'f'}\)

**Status Affected:** None

**Description:** Move data from W register to TRIS register.
- When 'f' = 5, TRISA is loaded.
- When 'f' = 6, TRISB is loaded.
- When 'f' = 7, TRISC is loaded.

### XORLW
**Exclusive OR literal with W**

**Syntax:**

\[
\text{[ label ] XORLW k}
\]

**Operands:**

- \(0 \leq k \leq 255\)

**Operation:**

\((W) \text{ XOR k} \rightarrow (W)\)

**Status Affected:** Z

**Description:** The contents of the W register are XOR'ed with the 8-bit literal 'k'. The result is placed in the W register.

### XORWF
**Exclusive OR W with f**

**Syntax:**

\[
\text{[ label ] XORWF f,d}
\]

**Operands:**

- \(0 \leq f \leq 127\)
- \(d \in [0,1]\)

**Operation:**

\((W) \text{ XOR (f)} \rightarrow (\text{destination})\)

**Status Affected:** Z

**Description:** Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

---

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34.0 ELECTRICAL SPECIFICATIONS

34.1 Absolute Maximum Ratings(†)

Ambient temperature under bias .......................................................... -40°C to +125°C
Storage temperature ........................................................................... -65°C to +150°C

Voltage on pins with respect to Vss

<table>
<thead>
<tr>
<th>Pin</th>
<th>Specification</th>
</tr>
</thead>
<tbody>
<tr>
<td>on VDD pin</td>
<td>PIC16F1717/8/9 -0.3V to +6.5V</td>
</tr>
<tr>
<td></td>
<td>PIC16LF1717/8/9 -0.3V to +4.0V</td>
</tr>
<tr>
<td>on MCLR pin</td>
<td>-0.3V to +9.0V</td>
</tr>
<tr>
<td>on all other pins</td>
<td>-0.3V to (VDD + 0.3V)</td>
</tr>
</tbody>
</table>

Maximum current

<table>
<thead>
<tr>
<th>Pin</th>
<th>Specification</th>
</tr>
</thead>
<tbody>
<tr>
<td>on VSS pin(†)</td>
<td>PIC16F1717/8/9 -40°C ≤ TA ≤ +85°C 340 mA</td>
</tr>
<tr>
<td></td>
<td>-40°C ≤ TA ≤ +125°C 140 mA</td>
</tr>
<tr>
<td>on VDD pin(†)</td>
<td>PIC16(L)F1718 only -40°C ≤ TA ≤ +85°C 250 mA</td>
</tr>
<tr>
<td></td>
<td>-40°C ≤ TA ≤ +125°C 85 mA</td>
</tr>
<tr>
<td>on VDD pin(†)</td>
<td>PIC16(L)F1717/9 only -40°C ≤ TA ≤ +85°C 350 mA</td>
</tr>
<tr>
<td></td>
<td>-40°C ≤ TA ≤ +125°C 120 mA</td>
</tr>
<tr>
<td>Sunk by any standard I/O pin</td>
<td>50 mA</td>
</tr>
<tr>
<td>Sourced by any standard I/O pin</td>
<td>50 mA</td>
</tr>
<tr>
<td>Sourced by any Op Amp output pin</td>
<td>100 mA</td>
</tr>
</tbody>
</table>

Clamp current, IK (VPIN < 0 or VPIN > VDD) ±20 mA

Total power dissipation(2) 800 mW

Note 1: Maximum current rating requires even load distribution across I/O pins. Maximum current rating may be limited by the device package power dissipation characterizations, see Table 34-6 to calculate device specifications.

2: Power dissipation is calculated as follows:

\[ P_{dis} = VDD \{I_{dd} - \sum I_{oh}\} + \sum (VDD - V_{oh})I_{oh} + \sum (V_{ol}I_{ol}) \]

† NOTICE: Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for extended periods may affect device reliability.
34.2 Standard Operating Conditions

The standard operating conditions for any device are defined as:

- **Operating Voltage:** \( V_{DD\text{MIN}} \leq V_{DD} \leq V_{DD\text{MAX}} \)
- **Operating Temperature:** \( T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}} \)

**Vdd — Operating Supply Voltage\(^{(1)}\)**

<table>
<thead>
<tr>
<th>PIC16LF1717/8/9</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>VDDMIN (Fosc (\leq) 16 MHz)</td>
<td>( \geq +1.8V )</td>
</tr>
<tr>
<td>VDDMIN (Fosc &gt; 16 MHz)</td>
<td>( \geq +2.5V )</td>
</tr>
<tr>
<td>VDDMAX</td>
<td>( \leq +3.6V )</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>PIC16F1717/8/9</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>VDDMIN (Fosc (\leq) 16 MHz)</td>
<td>( \geq +2.3V )</td>
</tr>
<tr>
<td>VDDMIN (&gt; 16 MHz)</td>
<td>( \geq +2.5V )</td>
</tr>
<tr>
<td>VDDMAX</td>
<td>( \leq +5.5V )</td>
</tr>
</tbody>
</table>

**Ta — Operating Ambient Temperature Range**

- **Industrial Temperature**
  - \( T_{A\text{MIN}} \): \(-40^\circ C\)  
  - \( T_{A\text{MAX}} \): \(+85^\circ C\)

- **Extended Temperature**
  - \( T_{A\text{MIN}} \): \(-40^\circ C\)  
  - \( T_{A\text{MAX}} \): \(+125^\circ C\)

**Note 1:** See Parameter **D001**, DS Characteristics: Supply Voltage.
FIGURE 34-1: VOLTAGE FREQUENCY GRAPH, -40°C ≤ TA ≤ +125°C, PIC16F1717/8/9 ONLY

Note 1: The shaded region indicates the permissible combinations of voltage and frequency.
2: Refer to Table 34-7 for each Oscillator mode’s supported frequencies.

FIGURE 34-2: VOLTAGE FREQUENCY GRAPH, -40°C ≤ TA ≤ +125°C, PIC16LF1717/8/9 ONLY

Note 1: The shaded region indicates the permissible combinations of voltage and frequency.
2: Refer to Table 34-7 for each Oscillator mode’s supported frequencies.
### 34.3 DC Characteristics

#### TABLE 34-1: SUPPLY VOLTAGE

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>D001</td>
<td>VDD</td>
<td>Supply Voltage</td>
<td>VDDMIN</td>
<td>1.8</td>
<td>—</td>
<td>V</td>
<td>FOSC ≤ 16 MHz</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>VDDMAX</td>
<td>3.6</td>
<td>—</td>
<td>V</td>
<td>FOSC &gt; 32 MHz (Note 2)</td>
</tr>
<tr>
<td>D001</td>
<td></td>
<td></td>
<td>2.3</td>
<td>—</td>
<td>5.5</td>
<td>V</td>
<td>FOSC ≤ 16 MHz</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>2.5</td>
<td>—</td>
<td>5.5</td>
<td>V</td>
<td>FOSC &gt; 32 MHz (Note 2)</td>
</tr>
<tr>
<td>D002*</td>
<td>VDR</td>
<td>RAM Data Retention Voltage(1)</td>
<td>1.5</td>
<td>—</td>
<td>—</td>
<td>V</td>
<td>Device in Sleep mode</td>
</tr>
<tr>
<td>D002*</td>
<td></td>
<td></td>
<td>1.7</td>
<td>—</td>
<td>—</td>
<td>V</td>
<td>Device in Sleep mode</td>
</tr>
<tr>
<td>D002A*</td>
<td>VPOR</td>
<td>Power-on Reset Release Voltage(3)</td>
<td>—</td>
<td>1.6</td>
<td>—</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>D002A*</td>
<td></td>
<td></td>
<td>—</td>
<td>1.6</td>
<td>—</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>D002B*</td>
<td>VPORR*</td>
<td>Power-on Reset Rearm Voltage(3)</td>
<td>—</td>
<td>0.8</td>
<td>—</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>D002B*</td>
<td></td>
<td></td>
<td>—</td>
<td>1.5</td>
<td>—</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>D003</td>
<td>VFVR</td>
<td>Fixed Voltage Reference Voltage</td>
<td>-4</td>
<td>—</td>
<td>+4</td>
<td>%</td>
<td>1x Gain, 1.024, VDD ≥ 2.5V, -40°C to 85°C</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-4</td>
<td>—</td>
<td>+4</td>
<td>%</td>
<td>2x Gain, 2.048, VDD ≥ 2.5V, -40°C to 85°C</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-5</td>
<td>—</td>
<td>+5</td>
<td>%</td>
<td>4x Gain, 4.096, VDD ≥ 4.75V, -40°C to 85°C</td>
</tr>
<tr>
<td>D004*</td>
<td>SVDD</td>
<td>VDD Rise Rate(2)</td>
<td>0.05</td>
<td>—</td>
<td>—</td>
<td>V/ms</td>
<td>Ensures that the Power-on Reset signal is released properly.</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note**
1: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.
2: PLL required for 32 MHz operation.
3: See Figure 34-3: POR and POR Rearm with Slow Rising VDD.
FIGURE 34-3: POR AND POR REARM WITH SLOW RISING VDD

Note 1: When NPOR is low, the device is held in Reset.
2: TPOR 1 μs typical.
3: TVLOW 2.7 μs typical.
## TABLE 34-2: SUPPLY CURRENT (I_{DD})^{(1,2)}

<table>
<thead>
<tr>
<th>PIC16LF1717/8/9</th>
<th>Standard Operating Conditions (unless otherwise stated)</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Param. No.</strong></td>
<td><strong>Device Characteristics</strong></td>
</tr>
<tr>
<td>D009</td>
<td>LDO Regulator</td>
</tr>
<tr>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
</tr>
<tr>
<td>D010</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
</tr>
<tr>
<td>D010</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
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<tr>
<td></td>
<td></td>
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<tr>
<td>D012</td>
<td>—</td>
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<tr>
<td>D012</td>
<td>—</td>
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<td></td>
<td></td>
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<tr>
<td>D014</td>
<td>—</td>
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<tr>
<td></td>
<td></td>
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<tr>
<td>D014</td>
<td>—</td>
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<tr>
<td></td>
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<tr>
<td></td>
<td></td>
</tr>
<tr>
<td>D015</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
</tr>
<tr>
<td>D015</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
</tr>
</tbody>
</table>

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** The test conditions for all I_{DD} measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

**Note 2:** The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

**Note 3:** For RC oscillator configurations, current through R_{EXT} is not included. The current through the resistor can be extended by the formula I_{R} = VDD/2R_{EXT} (mA) with R_{EXT} in kΩ.

**Note 4:** FVR and BOR are disabled.

**Note 5:** 0.1 μF capacitor on V_{CAP}.

**Note 6:** 8 MHz clock with 4x PLL enabled.
#### TABLE 34-2: SUPPLY CURRENT (I\(\text{DD}\))(1,2) (CONTINUED)

<table>
<thead>
<tr>
<th>PIC16LF1717/8/9</th>
<th>Standard Operating Conditions (unless otherwise stated)</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>PIC16F1717/8/9</strong></td>
<td></td>
</tr>
<tr>
<td><strong>Param. No.</strong></td>
<td><strong>Device Characteristics</strong></td>
</tr>
<tr>
<td>D017</td>
<td>—</td>
</tr>
<tr>
<td>D017</td>
<td>—</td>
</tr>
<tr>
<td>D019</td>
<td>—</td>
</tr>
<tr>
<td>D019</td>
<td>—</td>
</tr>
<tr>
<td>D020</td>
<td>—</td>
</tr>
<tr>
<td>D020</td>
<td>—</td>
</tr>
<tr>
<td>D022</td>
<td>—</td>
</tr>
<tr>
<td>D022</td>
<td>—</td>
</tr>
<tr>
<td>† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.</td>
<td></td>
</tr>
</tbody>
</table>

**Note 1:** The test conditions for all I\(\text{DD}\) measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to V\(\text{DD}\); MCLR = V\(\text{DD}\); WDT disabled.

**Note 2:** The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

**Note 3:** For RC oscillator configurations, current through R\(\text{EXT}\) is not included. The current through the resistor can be extended by the formula I\(\text{R}\) = V\(\text{DD}/2\times\text{R}\(\text{EXT}\) (mA) with R\(\text{EXT}\) in kΩ.

**Note 4:** FVR and BOR are disabled.

**Note 5:** 0.1 µF capacitor on V\(\text{CAP}\).

**Note 6:** 8 MHz clock with 4x PLL enabled.
## TABLE 34-3: POWER-DOWN CURRENTS (I\text{PD})^{(1,2)}

<table>
<thead>
<tr>
<th>PIC16LF1717/8/9</th>
<th>Standard Operating Conditions (unless otherwise stated)</th>
<th>Low-Power Sleep Mode, VREGPM = 1</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Param. No.</strong></td>
<td><strong>Device Characteristics</strong></td>
<td><strong>Min.</strong></td>
</tr>
<tr>
<td>D023</td>
<td>Base I\text{PD}</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td>D023</td>
<td>Base I\text{PD}</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td>D023A</td>
<td>Base I\text{PD}</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td>D024</td>
<td></td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td>D024</td>
<td></td>
<td>—</td>
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<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td>D025</td>
<td></td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td>D025</td>
<td></td>
<td>—</td>
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<td></td>
<td>—</td>
</tr>
<tr>
<td></td>
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<td>—</td>
</tr>
<tr>
<td>D026</td>
<td></td>
<td>—</td>
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<tr>
<td>D026</td>
<td></td>
<td>—</td>
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<tr>
<td></td>
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<td>—</td>
</tr>
<tr>
<td>D027</td>
<td></td>
<td>—</td>
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<td>D027</td>
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</tr>
<tr>
<td>D028</td>
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<td>—</td>
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<td></td>
<td></td>
<td>—</td>
</tr>
<tr>
<td>D028</td>
<td></td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** The peripheral current is the sum of the base I\text{PD} and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base I\text{DD} or I\text{PD} current from this limit. Max values should be used when calculating total current consumption.

**2:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to V\text{SS}.

**3:** ADC clock source is FRC.
TABLE 34-3: POWER-DOWN CURRENTS (IPD)(1,2) (CONTINUED)

<table>
<thead>
<tr>
<th>PIC16LF1717/8/9</th>
<th>Standard Operating Conditions (unless otherwise stated)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC16F1717/8/9</td>
<td>Low-Power Sleep Mode, VREGPM = 1</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Device Characteristics</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max. +85°C</th>
<th>Max. +125°C</th>
<th>Units</th>
<th>Conditions</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>D029</td>
<td></td>
<td></td>
<td>0.05</td>
<td>2</td>
<td>9</td>
<td>µA</td>
<td>1.8</td>
<td>ADC Current (Note 3), no conversion in progress</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0.08</td>
<td>3</td>
<td>10</td>
<td>µA</td>
<td>3.0</td>
<td>ADC Current (Note 3), no conversion in progress</td>
</tr>
<tr>
<td>D029</td>
<td></td>
<td></td>
<td>0.3</td>
<td>4</td>
<td>12</td>
<td>µA</td>
<td>2.3</td>
<td>ADC Current (Note 3), no conversion in progress</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0.4</td>
<td>5</td>
<td>13</td>
<td>µA</td>
<td>3.0</td>
<td>ADC Current (Note 3), no conversion in progress</td>
</tr>
<tr>
<td>D030</td>
<td></td>
<td></td>
<td>0.5</td>
<td>7</td>
<td>16</td>
<td>µA</td>
<td>5.0</td>
<td>ADC Current (Note 3), conversion in progress</td>
</tr>
<tr>
<td>D030</td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>—</td>
<td>µA</td>
<td>1.8</td>
<td>ADC Current (Note 3), conversion in progress</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>—</td>
<td>µA</td>
<td>3.0</td>
<td>ADC Current (Note 3), conversion in progress</td>
</tr>
<tr>
<td>D030</td>
<td></td>
<td></td>
<td>280</td>
<td>—</td>
<td>—</td>
<td>µA</td>
<td>2.3</td>
<td>ADC Current (Note 3), conversion in progress</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>280</td>
<td>—</td>
<td>—</td>
<td>µA</td>
<td>3.0</td>
<td>ADC Current (Note 3), conversion in progress</td>
</tr>
<tr>
<td>D031</td>
<td></td>
<td></td>
<td>280</td>
<td>—</td>
<td>—</td>
<td>µA</td>
<td>5.0</td>
<td>ADC Current (Note 3), conversion in progress</td>
</tr>
<tr>
<td>D031</td>
<td></td>
<td></td>
<td>250</td>
<td>650</td>
<td>—</td>
<td>µA</td>
<td>3.0</td>
<td>Op Amp (High power)</td>
</tr>
<tr>
<td>D031</td>
<td></td>
<td></td>
<td>250</td>
<td>650</td>
<td>—</td>
<td>µA</td>
<td>3.0</td>
<td>Op Amp (High power)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>350</td>
<td>650</td>
<td>—</td>
<td>µA</td>
<td>5.0</td>
<td>Op Amp (High power)</td>
</tr>
<tr>
<td>D032</td>
<td></td>
<td></td>
<td>250</td>
<td>600</td>
<td>—</td>
<td>µA</td>
<td>1.8</td>
<td>Comparator, CxSP = 0</td>
</tr>
<tr>
<td>D032</td>
<td></td>
<td></td>
<td>300</td>
<td>650</td>
<td>—</td>
<td>µA</td>
<td>3.0</td>
<td>Comparator, CxSP = 0 VREGPM = 0</td>
</tr>
<tr>
<td>D032</td>
<td></td>
<td></td>
<td>280</td>
<td>600</td>
<td>—</td>
<td>µA</td>
<td>2.3</td>
<td>Comparator, CxSP = 0 VREGPM = 0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>300</td>
<td>650</td>
<td>—</td>
<td>µA</td>
<td>3.0</td>
<td>Comparator, CxSP = 0 VREGPM = 0</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** The peripheral current is the sum of the base IPD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

**Note 2:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to Vss.

**Note 3:** ADC clock source is FRC.
### TABLE 34-4: I/O PORTS

**Standard Operating Conditions** (unless otherwise stated)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>VIL</td>
<td>Input Low Voltage</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D034</td>
<td></td>
<td>I/O PORT:</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>with TTL buffer</td>
<td>—</td>
<td>—</td>
<td>0.8</td>
<td>V</td>
<td>4.5V ≤ VDD ≤ 5.5V</td>
</tr>
<tr>
<td>D034A</td>
<td></td>
<td>—</td>
<td>—</td>
<td>0.15 VDD</td>
<td>V</td>
<td>1.8V ≤ VDD &lt; 4.5V</td>
<td></td>
</tr>
<tr>
<td>D035</td>
<td></td>
<td>—</td>
<td>—</td>
<td>0.2 VDD</td>
<td>V</td>
<td>2.0V ≤ VDD ≤ 5.5V</td>
<td></td>
</tr>
<tr>
<td>D035A</td>
<td></td>
<td>—</td>
<td>—</td>
<td>0.3 VDD</td>
<td>V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D036</td>
<td></td>
<td>—</td>
<td>—</td>
<td>0.8 VDD</td>
<td>V</td>
<td>2.7V ≤ VDD ≤ 5.5V</td>
<td></td>
</tr>
<tr>
<td>D036A</td>
<td></td>
<td>—</td>
<td>—</td>
<td>0.2 VDD</td>
<td>V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D036B</td>
<td></td>
<td>—</td>
<td>—</td>
<td>0.3 VDD</td>
<td>V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D040</td>
<td></td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>V</td>
<td>4.5V ≤ VDD ≤ 5.5V</td>
</tr>
<tr>
<td>D040A</td>
<td></td>
<td>—</td>
<td>0.25 VDD + 0.8</td>
<td>V</td>
<td>1.8V ≤ VDD &lt; 4.5V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D041</td>
<td></td>
<td>—</td>
<td>0.8 VDD</td>
<td>V</td>
<td>2.0V ≤ VDD ≤ 5.5V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D041A</td>
<td></td>
<td>—</td>
<td>0.7 VDD</td>
<td>V</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D042</td>
<td></td>
<td>—</td>
<td>2.1</td>
<td>—</td>
<td>—</td>
<td>V</td>
<td>2.7V ≤ VDD ≤ 5.5V</td>
</tr>
<tr>
<td>D042A</td>
<td></td>
<td>—</td>
<td>0.8 VDD</td>
<td>V</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D043A</td>
<td></td>
<td>—</td>
<td>0.7 VDD</td>
<td>V</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D043B</td>
<td></td>
<td>—</td>
<td>0.9 VDD</td>
<td>V</td>
<td>VDD &gt; 2.0V (Note 1)</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D060</td>
<td></td>
<td>—</td>
<td>—</td>
<td>± 5</td>
<td>± 125 nA</td>
<td>VSS ≤ VPIN ≤ VDD, Pin at high-impedance, 85°C</td>
<td></td>
</tr>
<tr>
<td>D061</td>
<td></td>
<td>—</td>
<td>—</td>
<td>± 5</td>
<td>± 200 nA</td>
<td>VSS ≤ VPIN ≤ VDD, Pin at high-impedance, 85°C</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
<td>—</td>
<td>± 1000 nA</td>
<td>VSS ≤ VPIN ≤ VDD, Pin at high-impedance, 125°C</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** In EXTRC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in EXTRC mode.

2: Negative current is defined as current sourced by the pin.

3: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: Including OSC2 in CLKOUT mode.
### Table 34-4: I/O Ports (Continued)

#### Standard Operating Conditions (unless otherwise stated)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>D070*</td>
<td>IPUR</td>
<td>Weak Pull-up Current</td>
<td>25</td>
<td>100</td>
<td>200</td>
<td>μA</td>
<td>VDD = 3.3V, VPIN = VSS</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>25</td>
<td>140</td>
<td>300</td>
<td>μA</td>
<td>VDD = 5.0V, VPIN = VSS</td>
</tr>
<tr>
<td>D080 Vol</td>
<td></td>
<td>Output Low Voltage(4)</td>
<td>—</td>
<td>—</td>
<td>0.6</td>
<td>V</td>
<td>IOL = 8mA, VDD = 5V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>I/O ports</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>IOL = 6mA, VDD = 3.3V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>IOL = 1.8mA, VDD = 1.8V</td>
</tr>
<tr>
<td>D090 Vol</td>
<td></td>
<td>Output High Voltage(4)</td>
<td></td>
<td></td>
<td></td>
<td>V</td>
<td>IOH = 3.5mA, VDD = 5V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>I/O ports</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>IOH = 3mA, VDD = 3.3V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>IOH = 1mA, VDD = 1.8V</td>
</tr>
<tr>
<td>D101*</td>
<td>COSC2</td>
<td>Capacitive Loading Specs on Output Pins</td>
<td>—</td>
<td>—</td>
<td>15</td>
<td>pF</td>
<td>In XT, HS and LP modes when external clock is used to drive OSC1</td>
</tr>
<tr>
<td>D101A*</td>
<td>CIO</td>
<td>All I/O pins</td>
<td>—</td>
<td>—</td>
<td>50</td>
<td>pF</td>
<td>—</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** In EXTRC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in EXTRC mode.

**2:** Negative current is defined as current sourced by the pin.

**3:** The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

**4:** Including OSC2 in CLKOUT mode.
### TABLE 34-5: MEMORY PROGRAMMING SPECIFICATIONS

**(Standard Operating Conditions) (unless otherwise stated)**

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>D110</td>
<td>VHH</td>
<td>Voltage on MCLR/VPP pin</td>
<td>8.0</td>
<td></td>
<td>9.0</td>
<td>V</td>
<td>(Note 2)</td>
</tr>
<tr>
<td>D111</td>
<td>IDDP</td>
<td>Supply Current during Programming</td>
<td>—</td>
<td>—</td>
<td>10</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>D112</td>
<td>VBE</td>
<td>VDD for Bulk Erase</td>
<td>2.7</td>
<td>—</td>
<td>VDDMAX</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>D113</td>
<td>VPEW</td>
<td>VDD for Write or Read Erase</td>
<td>VDDMIN</td>
<td>—</td>
<td>VDDMAX</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>D114</td>
<td>IPPGM</td>
<td>Current on MCLR/VPP during Erase/Write</td>
<td>—</td>
<td>1.0</td>
<td>—</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>D115</td>
<td>IDDPM</td>
<td>Current on VDD during Erase/Write</td>
<td>—</td>
<td>5.0</td>
<td>—</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>D121</td>
<td>EP</td>
<td>Cell Endurance</td>
<td>10K</td>
<td>—</td>
<td>—</td>
<td>E/W</td>
<td>-40°C ≤ TA ≤ +85°C (Note 1)</td>
</tr>
<tr>
<td>D122</td>
<td>VPRW</td>
<td>VDD for Read/Write</td>
<td>VDDMIN</td>
<td>—</td>
<td>VDDMAX</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>D123</td>
<td>TiW</td>
<td>Self-timed Write Cycle Time</td>
<td>—</td>
<td>2</td>
<td>2.5</td>
<td>ms</td>
<td>Provided no other specifications are violated</td>
</tr>
<tr>
<td>D124</td>
<td>TRETD</td>
<td>Characteristic Retention</td>
<td>—</td>
<td>40</td>
<td>—</td>
<td>Year</td>
<td></td>
</tr>
<tr>
<td>D125</td>
<td>EHEFC</td>
<td>High-Endurance Flash Cell</td>
<td>100K</td>
<td>—</td>
<td>—</td>
<td>E/W</td>
<td>-0°C ≤ TA ≤ +60°C, Lower byte last 128 addresses</td>
</tr>
</tbody>
</table>

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Self-write and Block Erase.

2: Required only if single-supply programming is disabled.
## TABLE 34-6: THERMAL CHARACTERISTICS

**Standard Operating Conditions** (unless otherwise stated)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Typ.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>TH01</td>
<td>θJA</td>
<td>Thermal Resistance Junction to Ambient</td>
<td>60.0</td>
<td>°C/W</td>
<td>28-pin SPDIP package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>80.3</td>
<td>°C/W</td>
<td>28-pin SOIC package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>90.0</td>
<td>°C/W</td>
<td>28-pin SSOP package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>36.0</td>
<td>°C/W</td>
<td>28-pin QFN 6x6 mm package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>48.0</td>
<td>°C/W</td>
<td>28-pin UQFN 4x4 mm package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>47.2</td>
<td>°C/W</td>
<td>40-pin PDIP package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>46.0</td>
<td>°C/W</td>
<td>44-pin TQFP</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>41.0</td>
<td>°C/W</td>
<td>40-pin UQFN 5x5 mm package</td>
</tr>
<tr>
<td>TH02</td>
<td>θJC</td>
<td>Thermal Resistance Junction to Case</td>
<td>31.4</td>
<td>°C/W</td>
<td>28-pin SPDIP package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>24.0</td>
<td>°C/W</td>
<td>28-pin SOIC package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>24.0</td>
<td>°C/W</td>
<td>28-pin SSOP package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>6.0</td>
<td>°C/W</td>
<td>28-pin QFN 6x6 mm package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>12.0</td>
<td>°C/W</td>
<td>28-pin UQFN 4x4 mm package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>24.7</td>
<td>°C/W</td>
<td>40-pin PDIP package</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>14.5</td>
<td>°C/W</td>
<td>44-pin TQFP</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>50.5</td>
<td>°C/W</td>
<td>40-pin UQFN 5x5 mm package</td>
</tr>
<tr>
<td>TH03</td>
<td>TJMAX</td>
<td>Maximum Junction Temperature</td>
<td>150</td>
<td>°C</td>
<td></td>
</tr>
<tr>
<td>TH04</td>
<td>PD</td>
<td>Power Dissipation</td>
<td>—</td>
<td>W</td>
<td>PD = PINTERNAL + PI/O</td>
</tr>
<tr>
<td>TH05</td>
<td>PINTERNAL</td>
<td>Internal Power Dissipation</td>
<td>—</td>
<td>W</td>
<td>PINTERNAL = IDD x VDD(1)</td>
</tr>
<tr>
<td>TH06</td>
<td>PI/O</td>
<td>I/O Power Dissipation</td>
<td>—</td>
<td>W</td>
<td>PI/O = Σ (IOL * VOL) + Σ (IOH * (VDD - VOH))</td>
</tr>
<tr>
<td>TH07</td>
<td>PDER</td>
<td>Derated Power</td>
<td>—</td>
<td>W</td>
<td>PDER = PDMAX (TJ - TA)/θJA(2)</td>
</tr>
</tbody>
</table>

**Note 1:** IDD is current to run the chip alone without driving any load on the output pins.

**Note 2:** Ta = Ambient Temperature, TJ = Junction Temperature
34.4 AC Characteristics

Timing Parameter Symbology has been created with one of the following formats:

1. $T_{ppS2ppS}$
2. $T_{ppS}$

<table>
<thead>
<tr>
<th>T</th>
<th>Frequency</th>
<th>T</th>
<th>Time</th>
</tr>
</thead>
</table>

Lowercase letters (pp) and their meanings:

<table>
<thead>
<tr>
<th>pp</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>cc</td>
<td>CCP1</td>
</tr>
<tr>
<td>ck</td>
<td>CLKOUT</td>
</tr>
<tr>
<td>cs</td>
<td>CS</td>
</tr>
<tr>
<td>di</td>
<td>SDI</td>
</tr>
<tr>
<td>do</td>
<td>SDO</td>
</tr>
<tr>
<td>dt</td>
<td>Data in</td>
</tr>
<tr>
<td>io</td>
<td>I/O PORT</td>
</tr>
<tr>
<td>mc</td>
<td>MCLR</td>
</tr>
</tbody>
</table>

Uppercase letters and their meanings:

<table>
<thead>
<tr>
<th>S</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>F</td>
<td>Fall</td>
</tr>
<tr>
<td>H</td>
<td>High</td>
</tr>
<tr>
<td>I</td>
<td>Invalid (High-impedance)</td>
</tr>
<tr>
<td>L</td>
<td>Low</td>
</tr>
</tbody>
</table>

Legend:

- $\text{CL} = 50 \text{ pF for all pins}$
FIGURE 34-5: CLOCK TIMING

Note 1: See Table 34-10.

TABLE 34-7: CLOCK OSCILLATOR TIMING REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>OS01</td>
<td>Fosc</td>
<td>External CLKN Frequency(1)</td>
<td>DC</td>
<td>—</td>
<td>0.5</td>
<td>MHz</td>
<td>External Clock (ECL)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>DC</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>External Clock (ECM)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>DC</td>
<td>—</td>
<td>32</td>
<td>MHz</td>
<td>External Clock (ECH)</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Oscillator Frequency(1)</td>
<td>—</td>
<td>32.768</td>
<td>—</td>
<td>kHz</td>
<td>LP Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0.1</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>XT Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>HS Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1</td>
<td>—</td>
<td>20</td>
<td>MHz</td>
<td>HS Oscillator, VDD &gt; 2.7V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>DC</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>EXTRC, VDD &gt; 2.0V</td>
</tr>
<tr>
<td>OS02</td>
<td>Tosc</td>
<td>External CLKN Period(1)</td>
<td>27</td>
<td>—</td>
<td>—</td>
<td>μs</td>
<td>LP Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>XT Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>50</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>HS Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Oscillator Period(1)</td>
<td>—</td>
<td>30.5</td>
<td>—</td>
<td>μs</td>
<td>LP Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>10,000</td>
<td>ns</td>
<td>XT Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>50</td>
<td>—</td>
<td>1,000</td>
<td>ns</td>
<td>HS Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>EXTRC</td>
</tr>
<tr>
<td>OS03</td>
<td>Tcy</td>
<td>Instruction Cycle Time(1)</td>
<td>125</td>
<td>Tcy</td>
<td>DC</td>
<td>ns</td>
<td>Tcy = 4/Fosc</td>
</tr>
<tr>
<td>OS04*</td>
<td>ToSH,</td>
<td>External CLKN High,</td>
<td>2</td>
<td>—</td>
<td>—</td>
<td>μs</td>
<td>LP Oscillator</td>
</tr>
<tr>
<td></td>
<td>ToSL</td>
<td>External CLKN Low</td>
<td>100</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>XT Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>HS Oscillator</td>
</tr>
<tr>
<td>OS05*</td>
<td>ToSR,</td>
<td>External CLKN Rise,</td>
<td>0</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>LP Oscillator</td>
</tr>
<tr>
<td></td>
<td>ToSF</td>
<td>External CLKN Fall</td>
<td>0</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>XT Oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>HS Oscillator</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (Tcy) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.
### TABLE 34-8: OSCILLATOR PARAMETERS

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>OS08</td>
<td>HFOSC</td>
<td>Internal Calibrated HFINTOSC Frequency(1)</td>
<td>±2%</td>
<td>—</td>
<td>16.0</td>
<td>—</td>
<td>MHz</td>
<td>VDD = 3.0V, TA = 25°C, (Note 2)</td>
</tr>
<tr>
<td>OS08A</td>
<td>MFOSC</td>
<td>Internal Calibrated MFINTOSC Frequency(1)</td>
<td>±2%</td>
<td>—</td>
<td>500</td>
<td>—</td>
<td>kHz</td>
<td>VDD = 3.0V, TA = 25°C, (Note 2)</td>
</tr>
<tr>
<td>OS09</td>
<td>LFOSC</td>
<td>Internal LFINTOSC Frequency</td>
<td>—</td>
<td>—</td>
<td>31</td>
<td>—</td>
<td>kHz</td>
<td>-40°C ≤ TA ≤ +125°C, (Note 3)</td>
</tr>
<tr>
<td>OS10*</td>
<td>T WARM</td>
<td>HFINTOSC Wake-up from Sleep Start-up Time</td>
<td>—</td>
<td>—</td>
<td>3.2</td>
<td>8</td>
<td>μs</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>MFINTOSC Wake-up from Sleep Start-up Time</td>
<td>—</td>
<td>—</td>
<td>24</td>
<td>35</td>
<td>μs</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>LFINTOSC Wake-up from Sleep Start-up Time</td>
<td>—</td>
<td>—</td>
<td>0.5</td>
<td>—</td>
<td>ms</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.

2: See Figure 34-6.
3: See Figure 35-57: LFINTOSC Frequency, PIC16LF1717/8/9 Only., and Figure 35-58: LFINTOSC Frequency, PIC16F1717/8/9 Only..

### FIGURE 34-6: HFINTOSC FREQUENCY ACCURACY OVER DEVICE VDD AND TEMPERATURE

![HFINTOSC Frequency Accuracy Over Device VDD and Temperature](image-url)
TABLE 34-9: PLL CLOCK TIMING SPECIFICATIONS

**Standard Operating Conditions** (unless otherwise stated)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>F10</td>
<td>FOSC</td>
<td>Oscillator Frequency Range</td>
<td>4</td>
<td>—</td>
<td>8</td>
<td>MHz</td>
<td></td>
</tr>
<tr>
<td>F11</td>
<td>FSYS</td>
<td>On-Chip VCO System Frequency</td>
<td>16</td>
<td>—</td>
<td>32</td>
<td>MHz</td>
<td></td>
</tr>
<tr>
<td>F12</td>
<td>TRC</td>
<td>PLL Start-up Time (Lock Time)</td>
<td>—</td>
<td>—</td>
<td>2</td>
<td>ms</td>
<td></td>
</tr>
<tr>
<td>F13*</td>
<td>ΔCLK</td>
<td>CLKOUT Stability (Jitter)</td>
<td>-0.25%</td>
<td>—</td>
<td>+0.25%</td>
<td>%</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
### FIGURE 34-7: CLKOUT AND I/O TIMING

![CLKOUT and I/O Timing Diagram](image)

### TABLE 34-10: CLKOUT AND I/O TIMING PARAMETERS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>OS11</td>
<td>ToscH2ckL</td>
<td>Fosc↑ to CLKOUT↓&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>—</td>
<td>—</td>
<td>70</td>
<td>ns</td>
<td>3.3V ≤ VDD ≤ 5.0V</td>
</tr>
<tr>
<td>OS12</td>
<td>ToscH2ckH</td>
<td>Fosc↑ to CLKOUT↑&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>—</td>
<td>—</td>
<td>72</td>
<td>ns</td>
<td>3.3V ≤ VDD ≤ 5.0V</td>
</tr>
<tr>
<td>OS13</td>
<td>TclkL2ioV</td>
<td>CLKOUT↓ to Port out valid&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>—</td>
<td>—</td>
<td>20</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>OS14</td>
<td>Tiov2ckH</td>
<td>Port input valid before CLKOUT↑&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>—</td>
<td>Tosc + 200 ns</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td>OS15</td>
<td>ToscH2ioV</td>
<td>Fosc↑ (Q1 cycle) to Port out valid</td>
<td>—</td>
<td>50</td>
<td>70*</td>
<td>ns</td>
<td>3.3V ≤ VDD ≤ 5.0V</td>
</tr>
<tr>
<td>OS16</td>
<td>ToscH2ioI</td>
<td>Fosc↑ (Q2 cycle) to Port input invalid</td>
<td>50</td>
<td>—</td>
<td></td>
<td></td>
<td>3.3V ≤ VDD ≤ 5.0V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>(I/O in hold time)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>OS17</td>
<td>Tiov2osh</td>
<td>Port input valid to Fosc↑ (Q2 cycle)</td>
<td>20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>(I/O in setup time)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>OS18*</td>
<td>TioR</td>
<td>Port output rise time&lt;sup&gt;(2)&lt;/sup&gt;</td>
<td>—</td>
<td>40</td>
<td>72</td>
<td>32</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>15</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>OS19*</td>
<td>TioF</td>
<td>Port output fall time&lt;sup&gt;(2)&lt;/sup&gt;</td>
<td>—</td>
<td>28</td>
<td>55</td>
<td>30</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>15</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>OS20*</td>
<td>Tinp</td>
<td>INT pin input high or low time</td>
<td>25</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>OS21*</td>
<td>TioC</td>
<td>Interrupt-on-Change new input level</td>
<td>25</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated.

**Note 1:** Measurements are taken in EXTRC mode where CLKOUT output is 4 x Tosc.

**Note 2:** Slew rate limited.
FIGURE 34-8: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

Note 1: Asserted low.
### TABLE 34-11: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET PARAMETERS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>30</td>
<td>TMCL</td>
<td>MCLR Pulse Width (low)</td>
<td>2</td>
<td>—</td>
<td>—</td>
<td>μs</td>
<td></td>
</tr>
<tr>
<td>31</td>
<td>TWDTLP</td>
<td>Low-Power Watchdog Timer Time-out Period</td>
<td>10</td>
<td>16</td>
<td>27</td>
<td>ms</td>
<td>VDD = 3.3V-5V 1:16 Prescaler used</td>
</tr>
<tr>
<td>32</td>
<td>TOST</td>
<td>Oscillator Start-up Timer Period(1)</td>
<td>—</td>
<td>1024</td>
<td>—</td>
<td>ToSC</td>
<td></td>
</tr>
<tr>
<td>33*</td>
<td>TPWRT</td>
<td>Power-up Timer Period, PWRTE = 0</td>
<td>40</td>
<td>65</td>
<td>140</td>
<td>ms</td>
<td></td>
</tr>
<tr>
<td>34*</td>
<td>TIOZ</td>
<td>I/O high-impedance from MCLR Low or Watchdog Timer Reset</td>
<td>—</td>
<td>—</td>
<td>2.0</td>
<td>μs</td>
<td></td>
</tr>
<tr>
<td>35</td>
<td>VBOR</td>
<td>Brown-out Reset Voltage</td>
<td>2.55</td>
<td>2.70</td>
<td>2.85</td>
<td>V</td>
<td>BORV = 0</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>2.30</td>
<td>2.45</td>
<td>2.60</td>
<td>V</td>
<td>BORV = 1 (PIC16F1717/8/9)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1.80</td>
<td>1.90</td>
<td>2.10</td>
<td>V</td>
<td>BORV = 1 (PIC16LF1717/8/9)</td>
</tr>
<tr>
<td>35A</td>
<td>VLPBOR</td>
<td>Low-Power Brown-out</td>
<td>1.8</td>
<td>2.1</td>
<td>2.5</td>
<td>V</td>
<td>LPBOR = 1</td>
</tr>
<tr>
<td>36*</td>
<td>VHYST</td>
<td>Brown-out Reset Hysteresis</td>
<td>0</td>
<td>25</td>
<td>75</td>
<td>mV</td>
<td>-40°C ≤ TA ≤ +85°C</td>
</tr>
<tr>
<td>37*</td>
<td>TBORDC</td>
<td>Brown-out Reset DC Response Time</td>
<td>1</td>
<td>3</td>
<td>35</td>
<td>μs</td>
<td>VDD ≤ VBOR</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** By design, the Oscillator Start-up Timer (OST) counts the first 1024 cycles, independent of frequency.

**Note 2:** To ensure these voltage tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.
FIGURE 34-9: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

FIGURE 34-10: BROWN-OUT RESET TIMING AND CHARACTERISTICS

Note 1: 64 ms delay only if PWRTE bit in the Configuration Word register is programmed to ‘0’.
2 ms delay if PWRTE = 0.
### TABLE 34-12: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Conditions</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>40*</td>
<td>T0H</td>
<td>T0CKI High Pulse Width</td>
<td>No Prescaler</td>
<td>0.5 Tcy + 20</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>With Prescaler</td>
<td>10</td>
<td>—</td>
</tr>
<tr>
<td>41*</td>
<td>T0L</td>
<td>T0CKI Low Pulse Width</td>
<td>No Prescaler</td>
<td>0.5 Tcy + 20</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>With Prescaler</td>
<td>10</td>
<td>—</td>
</tr>
<tr>
<td>42*</td>
<td>T0P</td>
<td>T0CKI Period</td>
<td>Greater of: 20 or Tcy + 40 N</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>45*</td>
<td>T1H</td>
<td>T1CKI High Time</td>
<td>Synchronous, No Prescaler</td>
<td>0.5 Tcy + 20</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Synchronous, with Prescaler</td>
<td>15</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Asynchronous</td>
<td>30</td>
<td>—</td>
</tr>
<tr>
<td>46*</td>
<td>T1L</td>
<td>T1CKI Low Time</td>
<td>Synchronous, No Prescaler</td>
<td>0.5 Tcy + 20</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Synchronous, with Prescaler</td>
<td>15</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Asynchronous</td>
<td>30</td>
<td>—</td>
</tr>
<tr>
<td>47*</td>
<td>T1P</td>
<td>T1CKI Input Period</td>
<td>Synchronous</td>
<td>Greater of: 30 or Tcy + 40 N</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Asynchronous</td>
<td>60</td>
<td>—</td>
</tr>
<tr>
<td>48</td>
<td>F1</td>
<td>Secondary Oscillator Input Frequency Range</td>
<td>(oscillator enabled by setting bit T1OSCEN)</td>
<td>32.4</td>
<td>32.76</td>
</tr>
<tr>
<td>49*</td>
<td>TCKEZTMR1</td>
<td>Delay from External Clock Edge to Timer Increment</td>
<td>2 Tosc</td>
<td>—</td>
<td>7 Tosc</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
FIGURE 34-11: CAPTURE/COMPARE/PWM TIMINGS (CCP)

Note: Refer to Figure 34-4 for load conditions.

TABLE 34-13: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP)

<table>
<thead>
<tr>
<th>Standard Operating Conditions (unless otherwise stated)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operating Temperature: -40°C ≤ TA ≤ +125°C</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>CC01*</td>
<td>TcCL</td>
<td>CCPx Input Low Time</td>
<td>No Prescaler 0.5TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>With Prescaler 20</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>CC02*</td>
<td>TcCH</td>
<td>CCPx Input High Time</td>
<td>No Prescaler 0.5TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>With Prescaler 20</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>CC03*</td>
<td>TcCP</td>
<td>CCPx Input Period</td>
<td>3TCY + 40 N</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>N = prescale value</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
FIGURE 34-12: CLC PROPAGATION TIMING

Note 1: See Figure 19-1: CLCx Simplified Block Diagram to identify specific CLC signals.

TABLE 34-14: CONFIGURATION LOGIC CELL (CLC) CHARACTERISTICS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLC01*</td>
<td>TCLCIN</td>
<td>CLC input time</td>
<td>—</td>
<td>7</td>
<td>OS17</td>
<td>ns</td>
<td>(Note 1)</td>
</tr>
<tr>
<td>CLC02*</td>
<td>TCLC</td>
<td>CLC module input to output propagation time</td>
<td>—</td>
<td>24</td>
<td>—</td>
<td>ns</td>
<td>VDD = 1.8V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>—</td>
<td>12</td>
<td>—</td>
<td>ns</td>
<td>VDD &gt; 3.6V</td>
</tr>
<tr>
<td>CLC03*</td>
<td>TCLCOUT</td>
<td>CLC output time</td>
<td>—</td>
<td>OS18</td>
<td>—</td>
<td>—</td>
<td>(Note 1)</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Rise Time</td>
<td>—</td>
<td>OS19</td>
<td>—</td>
<td>—</td>
<td>(Note 1)</td>
</tr>
<tr>
<td>CLC04*</td>
<td>FCLCMAX</td>
<td>CLC maximum switching frequency</td>
<td>—</td>
<td>45</td>
<td>—</td>
<td>MHz</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: See Table 34-10 for OS17, OS18 and OS19 rise and fall times.
### TABLE 34-15: ANALOG-TO-DIGITAL CONVERTER (ADC) CHARACTERISTICS\(^{(1,2,3,4)}\)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>AD01</td>
<td>NR</td>
<td>Resolution</td>
<td></td>
<td></td>
<td>10</td>
<td>bit</td>
<td></td>
</tr>
<tr>
<td>AD02</td>
<td>EI</td>
<td>Integral Error</td>
<td></td>
<td></td>
<td>±1.7</td>
<td>LSb</td>
<td>VREF = 3.0V</td>
</tr>
<tr>
<td>AD03</td>
<td>EDI</td>
<td>Differential Error</td>
<td></td>
<td></td>
<td>±1</td>
<td>LSb</td>
<td>No missing codes, VREF = 3.0V</td>
</tr>
<tr>
<td>AD04</td>
<td>EOFF</td>
<td>Offset Error</td>
<td></td>
<td></td>
<td>±2.5</td>
<td>LSb</td>
<td>VREF = 3.0V</td>
</tr>
<tr>
<td>AD05</td>
<td>EGN</td>
<td>Gain Error</td>
<td></td>
<td></td>
<td>±2.0</td>
<td>LSb</td>
<td>VREF = 3.0V</td>
</tr>
<tr>
<td>AD06</td>
<td>VREF</td>
<td>Reference Voltage</td>
<td>1.8</td>
<td>VDD</td>
<td>VREF</td>
<td></td>
<td>(VREF+ minus VREF-)</td>
</tr>
<tr>
<td>AD07</td>
<td>VAIN</td>
<td>Full-Scale Range</td>
<td>Vss</td>
<td>VREF</td>
<td>V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>AD08</td>
<td>ZAIN</td>
<td>Recommended Impedance of Analog Voltage Source</td>
<td></td>
<td>10</td>
<td>kΩ</td>
<td>Can go higher if external 0.01μF capacitor is present on input pin.</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Total Absolute Error includes integral, differential, offset and gain errors.

2: The ADC conversion result never decreases with an increase in the input voltage and has no missing codes.

3: ADC VREF is from external VREF+ pin, VDD pin or FVR, whichever is selected as reference input.

4: See Section 35.0 “DC and AC Characteristics Graphs and Charts”\(^{5}\) for operating characterization.

### TABLE 34-16: ADC CONVERSION REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>AD130</td>
<td>TAD</td>
<td>ADC Clock Period (TADC)</td>
<td>1.0</td>
<td>—</td>
<td>9.0</td>
<td>μs</td>
<td>Fosc-based</td>
</tr>
<tr>
<td></td>
<td></td>
<td>ADC Internal FRC Oscillator Period (TFRC)</td>
<td>1.0</td>
<td>2</td>
<td>6.0</td>
<td>μs</td>
<td>ADCS&lt;1:0&gt; = 11 (ADC FRC mode)</td>
</tr>
<tr>
<td>AD131</td>
<td>TCNV</td>
<td>Conversion Time (not including Acquisition Time)(^{(1)})</td>
<td>—</td>
<td>11</td>
<td>—</td>
<td>TAD</td>
<td>Set GO/DONE bit to conversion complete</td>
</tr>
<tr>
<td>AD132</td>
<td>TACQ</td>
<td>Acquisition Time</td>
<td></td>
<td>5.0</td>
<td>—</td>
<td>μs</td>
<td></td>
</tr>
<tr>
<td>AD133</td>
<td>THCD</td>
<td>Holding Capacitor Disconnect Time</td>
<td></td>
<td>1/2 TAD</td>
<td>—</td>
<td>ADCS&lt;2:0&gt; ≠ x11 (Fosc-based)</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>1/2 TAD + 1TCY</td>
<td>—</td>
<td>ADCS&lt;2:0&gt; = x11 (FRC-based)</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** The ADRES register may be read on the following TCY cycle.
FIGURE 34-13: ADC CONVERSION TIMING (ADC CLOCK Fosc-BASED)

FIGURE 34-14: ADC CONVERSION TIMING (ADC CLOCK FROM FRC)

Note 1: If the ADC clock source is selected as FRC, a time of TCY is added before the ADC clock starts. This allows the SLEEP instruction to be executed.
### TABLE 34-17: OPERATIONAL AMPLIFIER (OPA)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristics</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>OPA01*</td>
<td>GBWP</td>
<td>Gain Bandwidth Product</td>
<td>—</td>
<td>2</td>
<td>—</td>
<td>MHz</td>
<td></td>
</tr>
<tr>
<td>OPA02*</td>
<td>TON</td>
<td>Turn on Time</td>
<td>—</td>
<td>10</td>
<td>—</td>
<td>µs</td>
<td></td>
</tr>
<tr>
<td>OPA03*</td>
<td>PM</td>
<td>Phase Margin</td>
<td>—</td>
<td>40</td>
<td>—</td>
<td>degrees</td>
<td></td>
</tr>
<tr>
<td>OPA04*</td>
<td>SR</td>
<td>Slew Rate</td>
<td>—</td>
<td>3</td>
<td>—</td>
<td>V/µs</td>
<td></td>
</tr>
<tr>
<td>OPA05</td>
<td>OFF</td>
<td>Offset</td>
<td>—</td>
<td>±3</td>
<td>±9</td>
<td>mV</td>
<td></td>
</tr>
<tr>
<td>OPA06</td>
<td>CMRR</td>
<td>Common Mode Rejection Ratio</td>
<td>55</td>
<td>70</td>
<td>—</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>OPA07*</td>
<td>AOL</td>
<td>Open Loop Gain</td>
<td>—</td>
<td>90</td>
<td>—</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>OPA08</td>
<td>VICM</td>
<td>Input Common Mode Voltage</td>
<td>0</td>
<td>—</td>
<td>VDD</td>
<td>V</td>
<td>VDD &gt; 2.5V</td>
</tr>
<tr>
<td>OPA09*</td>
<td>PSRR</td>
<td>Power Supply Rejection Ratio</td>
<td>—</td>
<td>80</td>
<td>—</td>
<td>dB</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

### TABLE 34-18: COMPARATOR SPECIFICATIONS

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>CM01</td>
<td>VIOFF</td>
<td>Input Offset Voltage</td>
<td>—</td>
<td>±2.5</td>
<td>±5</td>
<td>mV</td>
<td>CxSP = 1, VICM = VDD/2</td>
</tr>
<tr>
<td>CM02</td>
<td>VICM</td>
<td>Input Common Mode Voltage</td>
<td>0</td>
<td>—</td>
<td>VDD</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>CM03</td>
<td>CMRR</td>
<td>Common Mode Rejection Ratio</td>
<td>40</td>
<td>50</td>
<td>—</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>CM04A</td>
<td>TRESP</td>
<td>Response Time Rising Edge</td>
<td>—</td>
<td>60</td>
<td>85</td>
<td>ns</td>
<td>CxSP = 1</td>
</tr>
<tr>
<td>CM04B</td>
<td>TRESP</td>
<td>Response Time Falling Edge</td>
<td>—</td>
<td>60</td>
<td>90</td>
<td>ns</td>
<td>CxSP = 1</td>
</tr>
<tr>
<td>CM04C</td>
<td>TRESP</td>
<td>Response Time Rising Edge</td>
<td>—</td>
<td>85</td>
<td>—</td>
<td>ns</td>
<td>CxSP = 0</td>
</tr>
<tr>
<td>CM04D</td>
<td>TRESP</td>
<td>Response Time Falling Edge</td>
<td>—</td>
<td>85</td>
<td>—</td>
<td>ns</td>
<td>CxSP = 0</td>
</tr>
<tr>
<td>CM05*</td>
<td>TMC2OV</td>
<td>Comparator Mode Change to Output Valid*</td>
<td>—</td>
<td>—</td>
<td>10</td>
<td>µs</td>
<td></td>
</tr>
<tr>
<td>CM06</td>
<td>CHYSTER</td>
<td>Comparator Hysteresis</td>
<td>20</td>
<td>45</td>
<td>75</td>
<td>mV</td>
<td>CxHYS = 1, CxSP = 1</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

**Note 1:** Response time measured with one comparator input at VDD/2, while the other input transitions from VSS to VDD.
### TABLE 34-19: 8-BIT DIGITAL-TO-ANALOG CONVERTER (DAC) SPECIFICATIONS

**Standard Operating Conditions (unless otherwise stated)**

\[ V_{DD} = 3.0V, \; TA = 25°C \]

See Section 35.0 “DC and AC Characteristics Graphs and Charts” for operating characterization.

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC01*</td>
<td>CLSB</td>
<td>Step Size</td>
<td>VDD/256</td>
<td></td>
<td></td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>DAC02*</td>
<td>CACC</td>
<td>Absolute Accuracy</td>
<td>± 1.5 LSb</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>DAC03*</td>
<td>CR</td>
<td>Unit Resistor Value (R)</td>
<td>600</td>
<td></td>
<td></td>
<td>Ω</td>
<td></td>
</tr>
<tr>
<td>DAC04*</td>
<td>CST</td>
<td>Settling TimeVal(1)</td>
<td>10</td>
<td></td>
<td></td>
<td>μs</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

**Note 1:** Settling time measured while DACR<7:0> transitions from '0x00' to '0xFF'.

### TABLE 34-20: 5-BIT DIGITAL-TO-ANALOG CONVERTER (DAC) SPECIFICATIONS

**Standard Operating Conditions (unless otherwise stated)**

\[ V_{DD} = 3.0V, \; TA = 25°C \]

See Section 35.0 “DC and AC Characteristics Graphs and Charts” for operating characterization.

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC05*</td>
<td>CLSB</td>
<td>Step Size</td>
<td>VDD/32</td>
<td></td>
<td></td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>DAC06*</td>
<td>CACC</td>
<td>Absolute Accuracy</td>
<td>± 0.5 LSb</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>DAC07*</td>
<td>CR</td>
<td>Unit Resistor Value (R)</td>
<td>6000</td>
<td></td>
<td></td>
<td>Ω</td>
<td></td>
</tr>
<tr>
<td>DAC08*</td>
<td>CST</td>
<td>Settling TimeVal(1)</td>
<td>10</td>
<td></td>
<td></td>
<td>μs</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

**Note 1:** Settling time measured while DACR<7:0> transitions from '0x00' to '0xFF'.

### TABLE 34-21: ZERO-CROSS PIN SPECIFICATIONS

**Standard Operating Conditions (unless otherwise stated)**

\[ V_{DD} = 3.0V, \; TA = 25°C \]

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>ZC01</td>
<td>ZCPINV</td>
<td>Voltage on Zero-Cross Pin</td>
<td>0.75</td>
<td></td>
<td></td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>ZC02</td>
<td>ZCSRC</td>
<td>Source current</td>
<td>300</td>
<td></td>
<td></td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td>ZC03</td>
<td>ZCSNK</td>
<td>Sink current</td>
<td>300</td>
<td></td>
<td></td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td>ZC04</td>
<td>ZCISW</td>
<td>Response Time Rising Edge</td>
<td>1</td>
<td></td>
<td></td>
<td>μs</td>
<td></td>
</tr>
<tr>
<td>ZC05</td>
<td>ZCOUT</td>
<td>Response Time Rising Edge</td>
<td>1</td>
<td></td>
<td></td>
<td>μs</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
FIGURE 34-15:  EUSART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

Note: Refer to Figure 34-4 for load conditions.

TABLE 34-22:  EUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

<table>
<thead>
<tr>
<th>Standard Operating Conditions (unless otherwise stated)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Param. No.</td>
</tr>
<tr>
<td>US120</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>US121</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>US122</td>
</tr>
<tr>
<td></td>
</tr>
</tbody>
</table>

FIGURE 34-16:  EUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

Note: Refer to Figure 34-4 for load conditions.

TABLE 34-23:  EUSART SYNCHRONOUS RECEIVE REQUIREMENTS

<table>
<thead>
<tr>
<th>Standard Operating Conditions (unless otherwise stated)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Param. No.</td>
</tr>
<tr>
<td>US125</td>
</tr>
<tr>
<td></td>
</tr>
<tr>
<td>US126</td>
</tr>
</tbody>
</table>
FIGURE 34-17: SPI MASTER MODE TIMING (CKE = 0, SMP = 0)

Note: Refer to Figure 34-4 for load conditions.

FIGURE 34-18: SPI MASTER MODE TIMING (CKE = 1, SMP = 1)

Note: Refer to Figure 34-4 for load conditions.
FIGURE 34-19: SPI SLAVE MODE TIMING (CKE = 0)

FIGURE 34-20: SPI SLAVE MODE TIMING (CKE = 1)

Note: Refer to Figure 34-4 for load conditions.
### TABLE 34-24: SPI MODE REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP70*</td>
<td>TSSL2SCH, TSSL2SCL</td>
<td>SS↓ to SCK↓ or SCK↑ input</td>
<td>2.25 TCY</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP71*</td>
<td>TsCh</td>
<td>SCK input high time (Slave mode)</td>
<td>TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP72*</td>
<td>TsCL</td>
<td>SCK input low time (Slave mode)</td>
<td>TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP73*</td>
<td>TdiV2SCH, TdiV2SCL</td>
<td>Setup time of SDI data input to SCK edge</td>
<td>100</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP74*</td>
<td>TsCH2DIL, TsCL2DIL</td>
<td>Hold time of SDI data input to SCK edge</td>
<td>100</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP75*</td>
<td>TdoR</td>
<td>SDO data output rise time</td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
<td>3.0V ≤ VDD ≤ 5.5V</td>
</tr>
<tr>
<td>SP76*</td>
<td>TdoF</td>
<td>SDO data output fall time</td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP77*</td>
<td>TssH2doZ</td>
<td>SS↑ to SDO output high-impedance</td>
<td>10</td>
<td>—</td>
<td>50</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP78*</td>
<td>TscR</td>
<td>SCK output rise time (Master mode)</td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
<td>3.0V ≤ VDD ≤ 5.5V</td>
</tr>
<tr>
<td>SP79*</td>
<td>TscF</td>
<td>SCK output fall time (Master mode)</td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP80*</td>
<td>Tsch2doV, TscL2doV</td>
<td>SDO data output valid after SCK edge</td>
<td>—</td>
<td>—</td>
<td>50</td>
<td>ns</td>
<td>3.0V ≤ VDD ≤ 5.5V</td>
</tr>
<tr>
<td>SP81*</td>
<td>TdoV2SCH, TdoV2SCL</td>
<td>SDO data output setup to SCK edge</td>
<td>1 TCY</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP82*</td>
<td>TSSL2doV</td>
<td>SDO data output valid after SS↓ edge</td>
<td>—</td>
<td>—</td>
<td>50</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP83*</td>
<td>TsCH2ssh, TsCL2ssh</td>
<td>SS ↑ after SCK edge</td>
<td>1.5 TCY + 40</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
FIGURE 34-21: \(\text{i}^2\text{C} \) BUS START/STOP BITS TIMING

![I²C Bus Timing Diagram](image)

**Note:** Refer to Figure 34-4 for load conditions.

TABLE 34-25: \(\text{i}^2\text{C} \) BUS START/STOP BITS REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP90*</td>
<td>TSU:STA</td>
<td>Start condition</td>
<td>4700</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>Only relevant for Repeated Start condition</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Setup time</td>
<td>400</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP91*</td>
<td>THD:STA</td>
<td>Start condition</td>
<td>4000</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>After this period, the first clock pulse is generated</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Hold time</td>
<td>400</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP92*</td>
<td>TSU:STO</td>
<td>Stop condition</td>
<td>4700</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Setup time</td>
<td>400</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP93</td>
<td>THD:STO</td>
<td>Stop condition</td>
<td>4000</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Hold time</td>
<td>600</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

FIGURE 34-22: \(\text{i}^2\text{C} \) BUS DATA TIMING

![I²C Bus Timing Diagram](image)

**Note:** Refer to Figure 34-4 for load conditions.
### TABLE 34-26: I^2C BUS DATA REQUIREMENTS

**Standard Operating Conditions** (unless otherwise stated)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP100*</td>
<td>T_HIGH</td>
<td>Clock high time</td>
<td>100 kHz mode</td>
<td>4.0</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>0.6</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>SSP module</td>
<td>1.5T_CY</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>SP101*</td>
<td>T_LOW</td>
<td>Clock low time</td>
<td>100 kHz mode</td>
<td>4.7</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>1.3</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>SSP module</td>
<td>1.5T_CY</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>SP102*</td>
<td>TR</td>
<td>SDA and SCL rise time</td>
<td>100 kHz mode</td>
<td>—</td>
<td>1000</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>20 + 0.1C_B</td>
<td>300</td>
<td>ns</td>
</tr>
<tr>
<td>SP103*</td>
<td>TF</td>
<td>SDA and SCL fall time</td>
<td>100 kHz mode</td>
<td>—</td>
<td>250</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>20 + 0.1C_B</td>
<td>250</td>
<td>ns</td>
</tr>
<tr>
<td>SP106*</td>
<td>TH_D:DAT</td>
<td>Data input hold time</td>
<td>100 kHz mode</td>
<td>0</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>0</td>
<td>0.9</td>
<td>μs</td>
</tr>
<tr>
<td>SP107*</td>
<td>TSU:D_AT</td>
<td>Data input setup time</td>
<td>100 kHz mode</td>
<td>250</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>100</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td>SP109*</td>
<td>T_AA</td>
<td>Output valid from clock</td>
<td>100 kHz mode</td>
<td>—</td>
<td>3500</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td>SP110*</td>
<td>T_BUF</td>
<td>Bus free time</td>
<td>100 kHz mode</td>
<td>4.7</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>1.3</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td>SP111</td>
<td>C_B</td>
<td>Bus capacitive loading</td>
<td>—</td>
<td>—</td>
<td>400</td>
<td>pF</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

**Note 1:** As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

**Note 2:** A Fast mode (400 kHz) I^2C bus device can be used in a Standard mode (100 kHz) I^2C bus system, but the requirement TSU:D\_AT ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the low period of the SCL signal. If such a device does stretch the low period of the SCL signal, it must output the next data bit to the SDA line TR max. + TSU:D\_AT = 1000 + 250 = 1250 ns (according to the Standard mode I^2C bus specification), before the SCL line is released.
35.0 DC AND AC CHARACTERISTICS GRAPHS AND CHARTS

The graphs and tables provided in this section are for design guidance and are not tested.

In some graphs or tables, the data presented are outside specified operating range (i.e., outside specified VDD range). This is for information only and devices are ensured to operate properly only within the specified range.

Unless otherwise noted, all graphs apply to both the L and LF devices.

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

"Typical" represents the mean of the distribution at 25°C. “Maximum”, “Max.”, “Minimum” or “Min.” represents (mean + 3σ) or (mean - 3σ) respectively, where σ is a standard deviation, over each temperature range.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, CIN = 0.1 µF, TA = 25°C.

**FIGURE 35-1:** IDD, LP Oscillator Mode, Fosc = 32 kHz, PIC16LF1717/8/9 Only.

**FIGURE 35-2:** IDD, LP Oscillator Mode, Fosc = 32 kHz, PIC16F1717/8/9 Only.

**FIGURE 35-3:** IDD Typical, XT and EXTRC Oscillator, PIC16LF1717/8/9 Only.

**FIGURE 35-4:** IDD Maximum, XT and EXTRC Oscillator, PIC16LF1717/8/9 Only.

**FIGURE 35-5:** IDD Typical, XT and EXTRC Oscillator, PIC16F1717/8/9 Only.

**FIGURE 35-6:** IDD Maximum, XT and EXTRC Oscillator, PIC16F1717/8/9 Only.
Note: Unless otherwise noted, \(V_{IN} = 5\text{ V}, F_{OSC} = 500\text{ kHz}, C_{IN} = 0.1\text{ \mu F}, T_A = 25^\circ\text{C}\).

**FIGURE 35-7:** \(I_{DD}\), EC Oscillator LP Mode, \(F_{OSC} = 32\text{ kHz}, \text{PIC16LF1717/8/9 Only.}\)

**FIGURE 35-8:** \(I_{DD}\), EC Oscillator LP Mode, \(F_{OSC} = 32\text{ kHz}, \text{PIC16F1717/8/9 Only.}\)

**FIGURE 35-9:** \(I_{DD}\), EC Oscillator LP Mode, \(F_{OSC} = 500\text{ kHz}, \text{PIC16LF1717/8/9 Only.}\)

**FIGURE 35-10:** \(I_{DD}\), EC Oscillator LP Mode, \(F_{OSC} = 500\text{ kHz}, \text{PIC16F1717/8/9 Only.}\)

**FIGURE 35-11:** \(I_{DD}\) Typical, EC Oscillator MP Mode, \(\text{PIC16LF1717/8/9 Only.}\)

**FIGURE 35-12:** \(I_{DD}\) Maximum, EC Oscillator MP Mode, \(\text{PIC16LF1717/8/9 Only.}\)
**Note:** Unless otherwise noted, \( V_{IN} = 5 \, \text{V} \), \( F_{OSC} = 500 \, \text{kHz} \), \( C_{IN} = 0.1 \, \mu \text{F} \), \( T_A = 25^\circ \text{C} \).

**FIGURE 35-13:** \( I_{DD} \) Typical, EC Oscillator MP Mode, PIC16F1717/8/9 Only.

**FIGURE 35-14:** \( I_{DD} \) Maximum, EC Oscillator MP Mode, PIC16F1717/8/9 Only.

**FIGURE 35-15:** \( I_{DD} \) Typical, EC Oscillator HP Mode, PIC16F1717/8/9 Only.

**FIGURE 35-16:** \( I_{DD} \) Maximum, EC Oscillator HP Mode, PIC16F1717/8/9 Only.

**FIGURE 35-17:** \( I_{DD} \) Typical, EC Oscillator HP Mode, PIC16F1717/8/9 Only.

**FIGURE 35-18:** \( I_{DD} \) Maximum, EC Oscillator HP Mode, PIC16F1717/8/9 Only.
Note: Unless otherwise noted, \( V_{\text{IN}} = 5 \text{V} \), \( F_{\text{OSC}} = 500 \text{kHz} \), \( C_{\text{IN}} = 0.1 \mu\text{F} \), \( T_A = 25^\circ\text{C} \).

**FIGURE 35-19:** \( I_{\text{DD}} \), LFINTOSC Mode, \( F_{\text{osc}} = 31 \text{kHz} \), PIC16LF1717/8/9 Only.

**FIGURE 35-20:** \( I_{\text{DD}} \), LFINTOSC Mode, \( F_{\text{osc}} = 31 \text{kHz} \), PIC16F1717/8/9 Only.

**FIGURE 35-21:** \( I_{\text{DD}} \), MFINTOSC Mode, \( F_{\text{osc}} = 500 \text{kHz} \), PIC16LF1717/8/9 Only.

**FIGURE 35-22:** \( I_{\text{DD}} \), MFINTOSC Mode, \( F_{\text{osc}} = 500 \text{kHz} \), PIC16F1717/8/9 Only.

**FIGURE 35-23:** \( I_{\text{DD}} \) Typical, HFINTOSC Mode, PIC16LF1717/8/9 Only.

**FIGURE 35-24:** \( I_{\text{DD}} \) Maximum, HFINTOSC Mode, PIC16LF1717/8/9 Only.
Note: Unless otherwise noted, $V_{IN} = 5V$, $F_{OSC} = 500$ kHz, $C_{IN} = 0.1 \mu F$, $T_A = 25°C$.

**FIGURE 35-25:** $I_{DD}$ Typical, HFINTOSC Mode, PIC16F1717/8/9 Only.

**FIGURE 35-26:** $I_{DD}$ Maximum, HFINTOSC Mode, PIC16F1717/8/9 Only.

**FIGURE 35-27:** $I_{DD}$ Typical, HS Oscillator, 25°C, PIC16F1717/8/9 Only.

**FIGURE 35-28:** $I_{DD}$ Maximum, HS Oscillator, 25°C, PIC16F1717/8/9 Only.

**FIGURE 35-29:** $I_{DD}$ Typical, HS Oscillator, 25°C, PIC16LF1717/8/9 Only.

**FIGURE 35-30:** $I_{DD}$ Maximum, HS Oscillator, PIC16F1717/8/9 Only.
Note: Unless otherwise noted, \( V_{IN} = 5\, V \), \( F_{OSC} = 500\, \text{kHz} \), \( C_{IN} = 0.1\, \mu F \), \( T_A = 25^\circ C \).

**FIGURE 35-31:** \( I_{DD} \), HS Oscillator, 32 MHz (8 MHz + 4x PLL), PIC16LF1717/8/9 Only.

**FIGURE 35-32:** \( I_{DD} \), HS Oscillator, 32 MHz (8 MHz + 4x PLL), PIC16F1717/8/9 Only.

**FIGURE 35-33:** \( I_{PD} \), Base, LP Sleep Mode, PIC16LF1717/8/9 Only.

**FIGURE 35-34:** \( I_{PD} \), Base, LP Sleep Mode (VREGPM = 1), PIC16F1717/8/9 Only.

**FIGURE 35-35:** \( I_{PD} \), Watchdog Timer (WDT), PIC16LF1717/8/9 Only.

**FIGURE 35-36:** \( I_{PD} \), Watchdog Timer (WDT), PIC16F1717/8/9 Only.
Note: Unless otherwise noted, \( V_{IN} = 5 \text{ V}, \ FOSC = 500 \text{ kHz}, \ C_{IN} = 0.1 \mu \text{F}, \ TA = 25^\circ \text{C}. \)

**FIGURE 35-37:** \( I_{PD}, \) Fixed Voltage Reference (FVR), PIC16LF1717/8/9 Only.

**FIGURE 35-38:** \( I_{PD}, \) Fixed Voltage Reference (FVR), PIC16F1717/8/9 Only.

**FIGURE 35-39:** \( I_{PD}, \) Brown-out Reset (BOR), \( BORV = 1, \) PIC16LF1717/8/9 Only.

**FIGURE 35-40:** \( I_{PD}, \) Brown-out Reset (BOR), \( BORV = 1, \) PIC16F1717/8/9 Only.

**FIGURE 35-41:** \( I_{PD}, \) LP Brown-out Reset (LPBOR = 0), PIC16LF1717/8/9 Only.

**FIGURE 35-42:** \( I_{PD}, \) LP Brown-out Reset (LPBOR = 0), PIC16F1717/8/9 Only.
Note: Unless otherwise noted, \(V_{\text{IN}} = 5\, \text{V}, \, F_{\text{OSC}} = 500\, \text{kHz}, \, C_{\text{IN}} = 0.1\, \mu\text{F}, \, T_{\text{A}} = 25^\circ\text{C}\).
Note: Unless otherwise noted, $V_{IN} = 5V$, $F_{OSC} = 500 \text{ kHz}$, $C_{IN} = 0.1 \mu F$, $TA = 25^\circ C$.

**FIGURE 35-49:** $I_{PD}$, Comparator, NP Mode ($C_{XSP} = 1$), PIC16LF1717/8/9 Only.

**FIGURE 35-50:** $I_{PD}$, Comparator, NP Mode ($C_{XSP} = 1$), PIC16F1717/8/9 Only.

**FIGURE 35-51:** $V_{OH}$ vs. $I_{OH}$ Over Temperature, $V_{DD} = 5.0V$, PIC16F1717/8/9 Only.

**FIGURE 35-52:** $V_{OL}$ vs. $I_{OL}$ Over Temperature, $V_{DD} = 5.0V$, PIC16F1717/8/9 Only.

**FIGURE 35-53:** $V_{OH}$ vs. $I_{OH}$ Over Temperature, $V_{DD} = 3.0V$.

**FIGURE 35-54:** $V_{OL}$ vs. $I_{OL}$ Over Temperature, $V_{DD} = 3.0V$. 
Note: Unless otherwise noted, $V_{IN} = 5\, V$, $F_{OSC} = 500\, kHz$, $C_{IN} = 0.1\, \mu F$, $T_A = 25^\circ C$.

**FIGURE 35-55:** VOH vs. IOH Over Temperature, $V_{DD} = 1.8\, V$, PIC16LF1717/8/9 Only.

**FIGURE 35-56:** VOL vs. IOL Over Temperature, $V_{DD} = 1.8\, V$, PIC16LF1717/8/9 Only.

**FIGURE 35-57:** LFINTOSC Frequency, PIC16LF1717/8/9 Only.

**FIGURE 35-58:** LFINTOSC Frequency, PIC16F1717/8/9 Only.

**FIGURE 35-59:** WDT Time-Out Period, PIC16F1717/8/9 Only.

**FIGURE 35-60:** WDT Time-Out Period, PIC16LF1717/8/9 Only.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, CIN = 0.1 μF, TA = 25°C.

**FIGURE 35-61:** Brown-out Reset Voltage, Low Trip Point (BORV = 1), PIC16LF1717/8/9 Only.

**FIGURE 35-62:** Brown-out Reset Hysteresis, Low Trip Point (BORV = 1), PIC16LF1717/8/9 Only.

**FIGURE 35-63:** Brown-out Reset Voltage, Low Trip Point (BORV = 1), PIC16F1717/8/9 Only.

**FIGURE 35-64:** Brown-out Reset Hysteresis, Low Trip Point (BORV = 1), PIC16F1717/8/9 Only.

**FIGURE 35-65:** Brown-out Reset Voltage, High Trip Point (BORV = 0).

**FIGURE 35-66:** Brown-out Reset Hysteresis, High Trip Point (BORV = 0).
Note: Unless otherwise noted, $V_{IN} = 5\, V$, $F_{OSC} = 500\, kHz$, $C_{IN} = 0.1\, \mu F$, $T_A = 25^\circ C$.

**FIGURE 35-67:** LPBOR Reset Voltage.

**FIGURE 35-68:** LPBOR Reset Hysteresis.

**FIGURE 35-69:** PWRT Period, PIC16F1717/8/9 Only.

**FIGURE 35-70:** PWRT Period, PIC16LF1717/8/9 Only.

**FIGURE 35-71:** POR Release Voltage.

**FIGURE 35-72:** POR Rearm Voltage, NP Mode ($V_{REGPM1} = 0$), PIC16F1717/8/9 Only.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, CIN = 0.1 µF, TA = 25°C.


FIGURE 35-74: Wake From Sleep, VREGPM = 0.

FIGURE 35-75: Wake From Sleep, VREGPM = 1.


FIGURE 35-77: ADC 10-bit Mode, Single-Ended DNL, VDD = 3.0V, TAD = 1 µS, 25°C.

FIGURE 35-78: ADC 10-bit Mode, Single-Ended DNL, VDD = 3.0V, TAD = 4 µS, 25°C.
Note: Unless otherwise noted, \( V_{\text{IN}} = 5 \text{V}, \ F_{\text{OSC}} = 500 \text{kHz}, \ C_{\text{IN}} = 0.1 \mu \text{F}, \ T_{\text{A}} = 25^\circ \text{C} \).

**FIGURE 35-79:**  ADC 10-bit Mode, Single-Ended INL, \( V_{\text{DD}} = 3.0 \text{V}, \ T_{\text{AD}} = 1 \mu \text{s}, 25^\circ \text{C} \).

**FIGURE 35-80:**  ADC 10-bit Mode, Single-Ended INL, \( V_{\text{DD}} = 3.0 \text{V}, \ T_{\text{AD}} = 4 \mu \text{s}, 25^\circ \text{C} \).

**FIGURE 35-81:**  ADC 10-bit Mode, Single-Ended DNL, \( V_{\text{DD}} = 3.0 \text{V}, \ T_{\text{AD}} = 1 \mu \text{s}, 25^\circ \text{C} \).

**FIGURE 35-82:**  ADC 10-bit Mode, Single-Ended INL, \( V_{\text{DD}} = 3.0 \text{V}, \ V_{\text{REF}} = 3.0 \text{V} \).

**FIGURE 35-83:**  ADC 10-bit Mode, Single-Ended DNL, \( V_{\text{DD}} = 3.0 \text{V}, \ T_{\text{AD}} = 1 \mu \text{s} \).

**FIGURE 35-84:**  ADC 10-bit Mode, Single-Ended INL, \( V_{\text{DD}} = 3.0 \text{V}, \ T_{\text{AD}} = 1 \mu \text{s} \).
Note: Unless otherwise noted, 
Vin = 5V, Fosc = 500 kHz, Cin = 0.1µF, TA = 25°C.


**FIGURE 35-87:** Temp. Indicator Initial Offset, Low Range, Temp. = 20°C, PIC16LF1717/8/9 Only.

**FIGURE 35-88:** Temp. Indicator Slope Normalized to 20°C, High Range, VDD = 5.5V, PIC16F1717/8/9 Only.


Note: Unless otherwise noted, \( V_{IN} = 5\, V \), \( F_{OSC} = 500\, kHz \), \( C_{IN} = 0.1\, \mu F \), \( T_A = 25^\circ C \).

**FIGURE 35-91:** Temp. Indicator Slope
Normalized to 20°C, Low Range, \( V_{DD} = 1.8V \), PIC16LF1717/8/9 Only.

**FIGURE 35-92:** Temp. Indicator Slope
Normalized to 20°C, Low Range, \( V_{DD} = 3.0V \), PIC16LF1717/8/9 Only.

**FIGURE 35-93:** Temp. Indicator Slope
Normalized to 20°C, High Range, \( V_{DD} = 3.6V \), PIC16LF1717/8/9 Only.

**FIGURE 35-94:** Op Amp, Common Mode Rejection Ratio (CMRR), \( V_{DD} = 3.0V \).

**FIGURE 35-95:** Op Amp, Output Voltage Histogram, \( V_{DD} = 3.0V \), \( V_{CM} = V_{DD}/2 \).

**FIGURE 35-96:** Op Amp, Offset Over Common Mode Voltage, \( V_{DD} = 3.0V \), Temp. = 25°C.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, CIN = 0.1 µF, TA = 25°C.


**FIGURE 35-98:** Op Amp, Output Slew Rate, Rising Edge, PIC16F1717/8/9 Only.

**FIGURE 35-99:** Op Amp, Output Slew Rate, Falling Edge, PIC16F1717/8/9 Only.

**FIGURE 35-100:** Comparator Hysteresis, NP Mode (CxSP = 1), VDD = 3.0V, Typical Measured Values.

**FIGURE 35-101:** Comparator Offset, NP Mode (CxSP = 1), VDD = 3.0V, Typical Measured Values at 25°C.

**FIGURE 35-102:** Comparator Offset, NP Mode (CxSP = 1), VDD = 3.0V, Typical Measured Values From -40°C to 125°C.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, CIN = 0.1 µF, TA = 25°C.

FIGURE 35-103: Comparator Hysteresis, NP Mode (CxSP = 1), VDD = 5.5V, Typical Measured Values, PIC16F1717/8/9 Only.

FIGURE 35-104: Comparator Offset, NP Mode (CxSP = 1), VDD = 5.0V, Typical Measured Values at 25°C, PIC16F1717/8/9 Only.

FIGURE 35-105: Comparator Offset, NP Mode (CxSP = 1), VDD = 5.5V, Typical Measured Values from -40°C to 125°C, PIC16F1717/8/9 Only.

FIGURE 35-106: Comparator Response Time Over Voltage, NP Mode (CxSP = 1), Typical Measured Values, PIC16LF1717/8/9 Only.

FIGURE 35-107: Comparator Response Time Over Voltage, NP Mode (CxSP = 1), Typical Measured Values, PIC16F1717/8/9 Only.

FIGURE 35-108: Comparator Output Filter Delay Time Over Temp., NP Mode (CxSP = 1), Typical Measured Values, PIC16LF1717/8/9 Only.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, CIN = 0.1 µF, TA = 25°C.

**FIGURE 35-109:** Comparator Output Filter Delay Time Over Temp., NP Mode (CxSP = 1), Typical Measured Values, PIC16F1717/8/9 Only.

**FIGURE 35-110:** Typical DAC DNL Error, VDD = 3.0V, VREF = External 3V.

**FIGURE 35-111:** Typical DAC INL Error, VDD = 3.0V, VREF = External 3V.

**FIGURE 35-112:** Typical DAC INL Error, VDD = 5.0V, VREF = External 5V, PIC16F1717/8/9 Only.

**FIGURE 35-113:** Typical DAC INL Error, VDD = 5.0V, VREF = External 5V, PIC16F1717/8/9 Only.

**FIGURE 35-114:** DAC INL Error, VDD = 3.0V, PIC16LF1717/8/9 Only.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, CIN = 0.1 µF, TA = 25°C.

**FIGURE 35-115:** ZCD Pin Voltage, Typical Measured Values

**FIGURE 35-116:** ZCD Response Time Over Voltage, Typical Measured Values From -40°C to 125°C.

**FIGURE 35-117:** ZCD Pin Current Over ZCD Pin Voltage, Typical Measured Values From -40°C to 125°C.

**FIGURE 35-118:** ZCD Pin Response Time Over Current, Typical Measured Values From -40°C to 125°C.

**FIGURE 35-119:** COG Dead-Band Delay, DBR/DBF = 32, Typical Measured Values

**FIGURE 35-120:** COG Dead-Band DBR/DBF Delay Per Step, Typical Measured Values.
Note: Unless otherwise noted, VIN = 5V, FOSC = 500 kHz, Cin = 0.1 µF, TA = 25°C.

**FIGURE 35-121:** COG Dead Band Delay Per Step, Typical Measured Values

**FIGURE 35-122:** COG Dead Band Delay Per Step, Zoomed to First 10 Codes, Typical Measured Values.
36.0 DEVELOPMENT SUPPORT

The PIC microcontrollers (MCU) and dsPIC® digital signal controllers (DSC) are supported with a full range of software and hardware development tools:

- Integrated Development Environment
  - MPLAB® X IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB XC Compiler
  - MPASM™ Assembler
  - MPLINK™ Object Linker/
    MPLIB™ Object Librarian
  - MPLAB Assembler/Linker/Librarian for
    Various Device Families
- Simulators
  - MPLAB X SIM Software Simulator
- Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers/Programmers
  - MPLAB ICD 3
  - PICkit 3
- Device Programmers
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards,
  Evaluation Kits and Starter Kits
- Third-party development tools

36.1 MPLAB X Integrated Development Environment Software

The MPLAB X IDE is a single, unified graphical user interface for Microchip and third-party software, and hardware development tool that runs on Windows®, Linux and Mac OS® X. Based on the NetBeans IDE, MPLAB X IDE is an entirely new IDE with a host of free software components and plug-ins for high-performance application development and debugging. Moving between tools and upgrading from software simulators to hardware debugging and programming tools is simple with the seamless user interface.

With complete project management, visual call graphs, a configurable watch window and a feature-rich editor that includes code completion and context menus, MPLAB X IDE is flexible and friendly enough for new users. With the ability to support multiple tools on multiple projects with simultaneous debugging, MPLAB X IDE is also suitable for the needs of experienced users.

Feature-Rich Editor:

- Color syntax highlighting
- Smart code completion makes suggestions and
  provides hints as you type
- Automatic code formatting based on user-defined
  rules
- Live parsing

User-Friendly, Customizable Interface:

- Fully customizable interface: toolbars, toolbar
  buttons, windows, window placement, etc.
- Call graph window

Project-Based Workspaces:

- Multiple projects
- Multiple tools
- Multiple configurations
- Simultaneous debugging sessions

File History and Bug Tracking:

- Local file history feature
- Built-in support for Bugzilla issue tracker
36.2 MPLAB XC Compilers

The MPLAB XC Compilers are complete ANSI C compilers for all of Microchip’s 8, 16, and 32-bit MCU and DSC devices. These compilers provide powerful integration capabilities, superior code optimization and ease of use. MPLAB XC Compilers run on Windows, Linux or MAC OS X.

For easy source level debugging, the compilers provide debug information that is optimized to the MPLAB X IDE.

The free MPLAB XC Compiler editions support all devices and commands, with no time or memory restrictions, and offer sufficient code optimization for most applications.

MPLAB XC Compilers include an assembler, linker and utilities. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. MPLAB XC Compiler uses the assembler to produce its object file. Notable features of the assembler include:

- Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility

36.3 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel® standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code, and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB X IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multipurpose source files
- Directives that allow complete control over the assembly process

36.4 MPLINK Object Linker/
MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

36.5 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC DSC devices. MPLAB XC Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility
36.6 MPLAB X SIM Software Simulator

The MPLAB X SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB X SIM Software Simulator fully supports symbolic debugging using the MPLAB XC Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

36.7 MPLAB REAL ICE In-Circuit Emulator System

The MPLAB REAL ICE In-Circuit Emulator System is Microchip’s next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs all 8, 16 and 32-bit MCU, and DSC devices with the easy-to-use, powerful graphical user interface of the MPLAB X IDE.

The emulator is connected to the design engineer’s PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ-11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradeable through future firmware downloads in MPLAB X IDE. MPLAB REAL ICE offers significant advantages over competitive emulators including full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, logic probes, a ruggedized probe interface and long (up to three meters) interconnection cables.

36.8 MPLAB ICD 3 In-Circuit Debugger System

The MPLAB ICD 3 In-Circuit Debugger System is Microchip’s most cost-effective, high-speed hardware debugger/programmer for Microchip Flash DSC and MCU devices. It debugs and programs PIC Flash microcontrollers and dsPIC DSCs with the powerful, yet easy-to-use graphical user interface of the MPLAB IDE.

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer’s PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

36.9 PICkit 3 In-Circuit Debugger/Programmer

The MPLAB PICkit 3 allows debugging and programming of PIC and dsPIC Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB IDE. The MPLAB PICkit 3 is connected to the design engineer’s PC using a full-speed USB interface and can be connected to the target via a Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the Reset line to implement in-circuit debugging and In-Circuit Serial Programming™ (ICSP™).

36.10 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages, and a modular, detachable socket assembly to support various package types. The ICSP cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices, and incorporates an MMC card for file storage and data applications.
36.11 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM™ and dsPICDEM™ demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOG® security ICs, CAN, IrDA®, PowerSmart battery management, SEEVAL® evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

36.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent® and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika®
37.0 PACKAGING INFORMATION

37.1 Package Marking Information

Legend:

XX...X Customer-specific information
Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week ‘01’)
NNN Alphanumeric traceability code
Pb-free JEDEC® designator for Matte Tin (Sn)
* This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.
Package Marking Information (Continued)

28-Lead UQFN (4x4x0.5 mm) Example

```
PIN 1

• XXXXX
XXXXXX
XXXXXX
YWWNNN
```

28-Lead QFN (6x6x0.9 mm) Example

```
PIN 1

XXXXXXXX
XXXXXXXX
YYWWNNN
```

40-Lead PDIP (600 mil) Example

```
XXXXXXXXXXXXXXXXXX
XXXXXXXXXXXXXXXXXX
XXXXXXXXXXXXXXXXXX
XX...X
```

Legend:
- XX...X Customer-specific information
- Y Year code (last digit of calendar year)
- YY Year code (last 2 digits of calendar year)
- WW Week code (week of January 1 is week ‘01’)
- NNN Alphanumeric traceability code
- e3 Pb-free JEDEC® designator for Matte Tin (Sn)
- * This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.
Package Marking Information (Continued)

40-Lead UQFN (5x5x0.5 mm) Example

Example

Legend:  
XX...X Customer-specific information
Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')
NNN Alphanumeric traceability code
  e3 Pb-free JEDEC® designator for Matte Tin (Sn)
* This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.
37.2 Package Details

The following sections give the technical details of the packages.

28-Lead Skinny Plastic Dual In-Line (SP) – 300 mil Body [SPDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

---

**NOTE 1**

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### Table: Dimension Limits

<table>
<thead>
<tr>
<th>Units</th>
<th>INCHES</th>
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<tbody>
<tr>
<td>Dimension</td>
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<table>
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<td>.290</td>
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<tr>
<td>Overall Row Spacing</td>
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**Notes:**
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010” per side.
4. Dimensioning and tolerancing per ASME Y14.5M.
   
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.
28-Lead Plastic Small Outline (SO) - Wide, 7.50 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

<table>
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<td>0*</td>
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<td>-</td>
</tr>
<tr>
<td>Foot Angle</td>
<td>φ</td>
<td>0*</td>
<td>-</td>
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<tr>
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<td>β</td>
<td>5*</td>
<td>-</td>
<td>15*</td>
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Notes:
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic
3. Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
4. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.
   REF: Reference Dimension, usually without tolerance, for information purposes only.
5. Datums A & B to be determined at Datum H.
Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

RECOMMENDED LAND PATTERN

<table>
<thead>
<tr>
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Notes:
1. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2052A
28-Lead Plastic Shrink Small Outline (SS) – 5.30 mm Body [SSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

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**Notes:**
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.20 mm per side.
3. Dimensioning and tolerancing per ASME Y14.5M.
   - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
   - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-073B
28-Lead Plastic Shrink Small Outline (SS) - 5.30 mm Body [SSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

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<table>
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<tr>
<td>Contact Pad Spacing</td>
<td>C</td>
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<td>Contact Pad Width (X28)</td>
<td>X1</td>
</tr>
<tr>
<td>Contact Pad Length (X28)</td>
<td>Y1</td>
</tr>
<tr>
<td>Distance Between Pads</td>
<td>G</td>
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Notes:
1. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2073A
28-Lead Plastic Ultra Thin Quad Flat, No Lead Package (MV) – 4x4x0.5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging
Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

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<tr>
<td>Pitch</td>
<td>e</td>
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<td>A</td>
</tr>
<tr>
<td>Standoff</td>
<td>A1</td>
</tr>
<tr>
<td>Contact Thickness</td>
<td>A3</td>
</tr>
<tr>
<td>Overall Width</td>
<td>E</td>
</tr>
<tr>
<td>Exposed Pad Width</td>
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<td>Overall Length</td>
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</tr>
<tr>
<td>Contact-to-Exposed Pad</td>
<td>K</td>
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</table>

Notes:
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Package is saw singulated.
3. Dimensioning and tolerancing per ASME Y14.5M.
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.
   REF: Reference Dimension, usually without tolerance, for information purposes only.
28-Lead Ultra Thin Plastic Quad Flat, No Lead Package (MV) - 4x4 mm Body [UQFN] With 0.40 mm Contact Length

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

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</tr>
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</tr>
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<tr>
<td>Contact Pad Spacing</td>
<td>C1</td>
</tr>
<tr>
<td>Contact Pad Spacing</td>
<td>C2</td>
</tr>
<tr>
<td>Contact Pad Width (X28)</td>
<td>X1</td>
</tr>
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<td>Contact Pad Length (X28)</td>
<td>Y1</td>
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<td>G</td>
</tr>
</tbody>
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**Notes:**
1. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2152A
28-Lead Plastic Quad Flat, No Lead Package (MM) - 6x6x0.9mm Body [QFN-S]
With 0.40 mm Terminal Length

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging
28-Lead Plastic Quad Flat, No Lead Package (MM) - 6x6x0.9mm Body [QFN-S]  
With 0.40 mm Terminal Length

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

![28-Lead Plastic Quad Flat, No Lead Package](image)

<table>
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<td>A</td>
</tr>
<tr>
<td>Standoff</td>
<td>A1</td>
</tr>
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<td>Terminal Thickness</td>
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<td>E</td>
</tr>
<tr>
<td>Exposed Pad Width</td>
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<td>Terminal-to-Exposed Pad</td>
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**Notes:**
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Package is saw singulated
3. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-124C Sheet 2 of 2
28-Lead Plastic Quad Flat, No Lead Package (MM) – 6x6x0.9 mm Body [QFN-S] with 0.40 mm Contact Length

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com-packaging

RECOMMENDED LAND PATTERN

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<td>Contact Pad Width (X28)</td>
<td>X1</td>
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<tr>
<td>Contact Pad Length (X28)</td>
<td>Y1</td>
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<tr>
<td>Distance Between Pads</td>
<td>G</td>
</tr>
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Notes:
1. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2124A
40-Lead Plastic Dual In-Line (P) – 600 mil Body [PDIP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

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<td>Pitch</td>
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<td>Top to Seating Plane</td>
<td>A</td>
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<tr>
<td>Molded Package Thickness</td>
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</tr>
<tr>
<td>Base to Seating Plane</td>
<td>A1</td>
</tr>
<tr>
<td>Shoulder to Shoulder Width</td>
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<td>Overall Length</td>
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</tr>
<tr>
<td>Tip to Seating Plane</td>
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<td>Lead Thickness</td>
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<td>Overall Row Spacing</td>
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**Notes:**
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
4. Dimensioning and tolerancing per ASME Y14.5M.
   - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

---

Microchip Technology Drawing C04-016B
40-Lead Ultra Thin Plastic Quad Flat, No Lead Package (MV) – 5x5x0.5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging
40-Lead Ultra Thin Plastic Quad Flat, No Lead Package (MV) – 5x5x0.5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging
40-Lead Plastic Ultra Thin Quad Flat, No Lead Package (MV) - 5x5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com包aaging

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</tr>
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<td>W2</td>
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<tr>
<td>Contact Pad Spacing</td>
<td>C2</td>
</tr>
<tr>
<td>Contact Pad Width (X40)</td>
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<tr>
<td>Contact Pad Length (X40)</td>
<td>Y1</td>
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<tr>
<td>Distance Between Pads</td>
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</tr>
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</table>

Notes:
1. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2156B
44-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com-packaging

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<td>α</td>
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<tr>
<td>Mold Draft Angle Bottom</td>
<td>β</td>
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</table>

**Notes:**
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Chamfers at corners are optional; size may vary.
3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.
4. Dimensioning and tolerancing per ASME Y14.5M.
   - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
   - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-076B
44-Lead Plastic Thin Quad Flatpack (PT) 10X10X1 mm Body, 2.00 mm Footprint [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

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<td>Limits</td>
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<td>C1</td>
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<tr>
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Notes:
1. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.
APPENDIX A: DATA SHEET

REVISION HISTORY

Revision A (02/2014)
Initial release of the document.

Revision B (12/2015)
Chapter 18.0 COG rewritten.
Added High-endurance Flash Data Memory information to cover pages and Memory chapter. Added Section 11.3.1 heading. Added Section 6.3.5: Clock Switching Before Sleep.
Deleted Section 24.4: Low-Power Voltage State.
Updated the Family Types Table and the Product Identification System page.
Minor typos corrected.

Revision C (9/2018)
Updated Figures 1-1, 21-1, and 23-1; Sections 11.7.3 and 25.0; and Tables 1, 28-1, 34-4 and 34-7.
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PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<table>
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<th>PART NO.</th>
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<td>Device</td>
<td>Tape and Reel Option</td>
<td>Temperature Range</td>
<td>Package</td>
<td>Pattern</td>
<td></td>
</tr>
</tbody>
</table>

**Device:**
- PIC16F1717, PIC16LF1717,
- PIC16F1718, PIC16LF1718
- PIC16F1719, PIC16LF1719

**Tape and Reel Option:**
- Blank = Standard packaging (tube or tray)
- T = Tape and Reel \(^{(1)}\)

**Temperature Range:**
- I = -40°C to +85°C (Industrial)
- E = -40°C to +125°C (Extended)

**Package:** \(^{(2)}\)
- MV = UQFN, 28-lead 4x4x0.5mm
- MV = UQFN, 40-lead 5x5x0.5mm
- MM = QFN-S, 28-lead 6x6x0.9mm
- P = PDIP, 40-lead
- PT = TQFP, 44-lead 10x10x1mm
- SO = SOIC, 28-lead
- SP = SPDIP, 28-lead
- SS = SSOP, 28-lead

**Pattern:**
- QTP, SQTP, Code or Special Requirements
  (blank otherwise)

**Examples:**

a) PIC16LF1717- I/P  
Industrial temperature
PDIP package

b) PIC16F1718- E/SS  
Extended temperature,  
SSOP package

**Note 1:** Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.

2: Small form-factor packaging options may be available. Please check www.microchip.com/packaging for small-form factor package availability, or contact your local Sales Office.
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- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip’s Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as “unbreakable.”

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