High-Performance RISC CPU:
- Only 49 Instructions to Learn:
  - All single-cycle instructions except branches
- Operating Speed:
  - DC – 32 MHz oscillator/clock input
  - DC – 125 ns instruction cycle
- Interrupt Capability with Automatic Context Saving
- 16-Level Deep Hardware Stack with Optional Overflow/Underflow Reset
- Direct, Indirect and Relative Addressing modes:
  - Two full 16-bit File Select Registers (FSRs)
  - FSRs can read program and data memory

Flexible Oscillator Structure:
- Precision 32 MHz Internal Oscillator Block:
  - Factory calibrated to ± 1%, typical
  - Software selectable frequencies range of 31 kHz to 32 MHz
- 31 kHz Low-Power Internal Oscillator
- Four Crystal modes up to 32 MHz
- Three External Clock modes up to 32 MHz
- 4X Phase Lock Loop (PLL)
- Fail-Safe Clock Monitor:
  - Allows for safe shutdown if peripheral clock stops
- Two-Speed Oscillator Start-up
- Reference Clock module:
  - Programmable clock output frequency and duty-cycle

Special Microcontroller Features:
- 1.8V-3.6V Operation
- Self-Reprogrammable under Software Control
- Power-on Reset (POR), Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Programmable Brown-out Reset (BOR)
- Extended Watchdog Timer (WDT)
- In-Circuit Serial Programming™ (ICSP™) via Two Pins
- In-Circuit Debug (ICD) via Two Pins
- Enhanced Low-Voltage Programming (LVP)
- Programmable Code Protection
- Power-Saving Sleep mode

Low-Power Features:
- Standby Current:
  - 170 nA @ 1.8V, typical, RF Sleep
- Operating Current:
  - 35 µA @ 1 MHz, 1.8V, RF Sleep, typical
  - 9 mA @ 1 MHz, 1.8V, RF on at 0 dBm, typical
  - 16.5 mA @ 1 MHz, 1.8V, RF on at +10 dBm, typical
- Low-Power Watchdog Timer Current:
  - 500 nA @ 3.0V, typical

RF Transmitter:
- Fully Integrated Transmitter
- FSK Operation up to 100 kbps
- OOK Operation up to 10 kbps
- Frequency-Agile Operation in 310, 433, 868 and 915 MHz Bands
- +10 dBm or 0 dBm Configurable Output Power

Analog Features:
- Analog-to-Digital Converter (ADC) module:
  - 10-bit resolution, four channels
  - Conversion available during Sleep
- Analog Comparator module:
  - One rail-to-rail analog comparator
  - Power mode control
  - Software controllable hysteresis
- Voltage Reference module:
  - Fixed Voltage Reference (FVR) with 1.024V and 2.048V output levels
- 5-bit rail-to-rail resistive DAC with positive and negative reference selection

Peripheral Features:
- Five I/O Pins and one Input Only Pin:
  - High-current sink/source 25 mA/25 mA
  - Programmable weak pull-ups
  - Programmable interrupt-on-change pins
- Timer0: 8-Bit Timer/Counter with 8-Bit Prescaler
- Enhanced Timer1:
  - 16-bit timer/counter with prescaler
  - External Gate Input mode
  - Dedicated, low-power 32 kHz oscillator driver
- Timer2: 8-Bit Timer/Counter with 8-Bit Period Register, Prescaler and Postscaler
Peripheral Features: (Continued)

- Enhanced CCP (ECCP) module:
  - Software selectable time bases
  - Auto-shutdown and auto-restart
  - PWM steering
- Master Synchronous Serial Port (MSSP) with SPI and I²C™ with:
  - 7-bit address masking
  - SMBus/PMBus™ compatibility
- Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module:
  - RS-232, RS-485 and LIN compatible
  - Auto-Baud Detect
- Capacitive Sensing (CPS) module (mTouch™ sensing solution):
  - Four input channels
- Data Signal Modulator module:
  - Selectable modulator and carrier sources
- SR Latch:
  - Multiple Set/Reset input options
  - Emulates 555 Timer applications

PIC12LF1840T39A Family Types

<table>
<thead>
<tr>
<th>Device</th>
<th>Data Sheet Index</th>
<th>Program Memory</th>
<th>Data Memory</th>
<th>Data EEPROM</th>
<th>I/Os</th>
<th>10-bit ADC</th>
<th>CapSense</th>
<th>Comparators</th>
<th>Timers (8/16-bit)</th>
<th>EUSART</th>
<th>MSSP</th>
<th>ECCP (Full-Bridge)</th>
<th>ECCP (Half-Bridge)</th>
<th>SR Latch</th>
<th>RF</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC12LF1840T39A</td>
<td>(1)</td>
<td>4K</td>
<td>256</td>
<td>256</td>
<td>6</td>
<td>4</td>
<td>4</td>
<td>1</td>
<td>1/1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>Yes</td>
<td>Yes</td>
</tr>
</tbody>
</table>

Note 1: One pin is input only.

Data Sheet Index: (Unshaded devices are described in this document.)
1: DS40001636 PIC12LF1840T39A 8-Bit Flash Microcontroller with XLP Technology

Note: For other small form-factor package availability and marking information, please visit http://www.microchip.com/packaging or contact your local sales office.
### Pin Diagram

**FIGURE 1: 14-PIN DIAGRAM TSSOP PIC12LF1840T39A**

<table>
<thead>
<tr>
<th>Pin Function</th>
<th>Pin Number</th>
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<tbody>
<tr>
<td>VDD</td>
<td>1</td>
</tr>
<tr>
<td>RX((f))</td>
<td>2</td>
</tr>
<tr>
<td>DT((f))</td>
<td>3</td>
</tr>
<tr>
<td>CCP1((f))</td>
<td>4</td>
</tr>
<tr>
<td>P1A((f))</td>
<td>5</td>
</tr>
<tr>
<td>SRNQ</td>
<td>6</td>
</tr>
<tr>
<td>T1CKI</td>
<td>7</td>
</tr>
<tr>
<td>T1OSI</td>
<td>8</td>
</tr>
<tr>
<td>OSC1</td>
<td>9</td>
</tr>
<tr>
<td>CLkin</td>
<td>10</td>
</tr>
<tr>
<td>RA5</td>
<td>11</td>
</tr>
<tr>
<td>MDCIN2((f))</td>
<td>12</td>
</tr>
<tr>
<td>T1G((f))</td>
<td>13</td>
</tr>
<tr>
<td>P1B(1)</td>
<td>14</td>
</tr>
<tr>
<td>RX((f))</td>
<td>2</td>
</tr>
<tr>
<td>DT((f))</td>
<td>3</td>
</tr>
<tr>
<td>CCP1((f))</td>
<td>4</td>
</tr>
<tr>
<td>P1A((f))</td>
<td>5</td>
</tr>
<tr>
<td>SRNQ</td>
<td>6</td>
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<tr>
<td>T1CKI</td>
<td>7</td>
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<td>T1OSI</td>
<td>8</td>
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<td>OSC1</td>
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</tr>
<tr>
<td>CLkin</td>
<td>10</td>
</tr>
<tr>
<td>RA5</td>
<td>11</td>
</tr>
<tr>
<td>MDCIN2((f))</td>
<td>12</td>
</tr>
<tr>
<td>T1G((f))</td>
<td>13</td>
</tr>
<tr>
<td>P1B(1)</td>
<td>14</td>
</tr>
</tbody>
</table>

**Note 1:** Pin function is selectable via the APFCON register.
<table>
<thead>
<tr>
<th>I/O</th>
<th>14-Pin TSSOP</th>
<th>A/D</th>
<th>Reference</th>
<th>CapSense</th>
<th>Comparator</th>
<th>SR Latch</th>
<th>Timers</th>
<th>ECCP</th>
<th>EUSART</th>
<th>MSSP</th>
<th>Interrupt</th>
<th>Modulator</th>
<th>Pull-up</th>
<th>Basic</th>
<th>RF</th>
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<tbody>
<tr>
<td>RA0</td>
<td>13</td>
<td>AN0</td>
<td>DACOUT</td>
<td>CPS0</td>
<td>C1IN+</td>
<td>—</td>
<td>—</td>
<td>P1B(1)</td>
<td>TX(1)</td>
<td>SDQ(1)</td>
<td>IOC</td>
<td>MDOUT</td>
<td>Y</td>
<td>ICSPDAT</td>
<td>ICDDAT</td>
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<tr>
<td>RA1</td>
<td>12</td>
<td>AN1</td>
<td>VREF</td>
<td>CPS1</td>
<td>C1IN0-</td>
<td>SRI</td>
<td>—</td>
<td>—</td>
<td>RX(1)</td>
<td>SCL</td>
<td>IOC</td>
<td>MDMIN</td>
<td>Y</td>
<td>ICSPCLK</td>
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<tr>
<td>RA2</td>
<td>11</td>
<td>AN2</td>
<td>—</td>
<td>CPS2</td>
<td>C1OUT</td>
<td>SRQ</td>
<td>T0CKI</td>
<td>CCP1(1)</td>
<td>P1A(1)</td>
<td>FLT0</td>
<td>—</td>
<td>SDA</td>
<td>INT</td>
<td>IOC</td>
<td>MDCIN1</td>
</tr>
<tr>
<td>RA3</td>
<td>4</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>T1G(1)</td>
<td>—</td>
<td>SS(1)</td>
<td>IOC</td>
<td>—</td>
<td>Y</td>
<td>MCLR</td>
<td>VPP</td>
<td>—</td>
</tr>
<tr>
<td>RA4</td>
<td>3</td>
<td>AN3</td>
<td>—</td>
<td>CPS3</td>
<td>C1IN1-</td>
<td>—</td>
<td>T1G(1)</td>
<td>T1OSO</td>
<td>P1B(1)</td>
<td>TX(1)</td>
<td>SDO(1)</td>
<td>IOC</td>
<td>MDCIN2</td>
<td>Y</td>
<td>OSC2 CLKOUT</td>
</tr>
<tr>
<td>RA5</td>
<td>2</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>SRNQ(1)</td>
<td>T1CKI</td>
<td>T1OSI</td>
<td>CCP1(1)</td>
<td>P1A(1)</td>
<td>RX(1)</td>
<td>DT(1)</td>
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<td>IOC</td>
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<tr>
<td>VDD</td>
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<td>CTRL</td>
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<td>—</td>
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<td>RFOUT</td>
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<tr>
<td>DATA</td>
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<td>—</td>
<td>—</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>DATA</td>
</tr>
<tr>
<td>XTAL</td>
<td>10</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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</tr>
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<td>VDDRF</td>
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<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
<td>—</td>
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<td>VSSRF</td>
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<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>Vss</td>
</tr>
</tbody>
</table>

**Note 1:** Pin function is selectable via the APFCON register.
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An errata sheet, describing minor operational differences from the data sheet and recommended workarounds, may exist for current devices. As device/documentation issues become known to us, we will publish an errata sheet. The errata will specify the revision of silicon and revision of document to which it applies.

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• Your local Microchip sales office (see last page)

When contacting a sales office, please specify which device, revision of silicon and data sheet (include literature number) you are using.

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1.0 DEVICE OVERVIEW

The PIC12LF1840T39A device is described within this data sheet. It is available in 14-pin packages. Figure 1-1 shows a block diagram of the PIC12LF1840T39A device. Table 1-2 shows the pinout descriptions.

Reference Table 1-1 for peripherals available per device.

TABLE 1-1: DEVICE PERIPHERAL SUMMARY

<table>
<thead>
<tr>
<th>Peripheral</th>
<th>PIC12LF1840T39A</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADC</td>
<td>●</td>
</tr>
<tr>
<td>Capacitive Sensing (CPS) Module</td>
<td>●</td>
</tr>
<tr>
<td>Data EEPROM</td>
<td>●</td>
</tr>
<tr>
<td>Digital-to-Analog Converter (DAC)</td>
<td>●</td>
</tr>
<tr>
<td>Digital Signal Modulator (DSM)</td>
<td>●</td>
</tr>
<tr>
<td>EUSART</td>
<td>●</td>
</tr>
<tr>
<td>Fixed Voltage Reference (FVR)</td>
<td>●</td>
</tr>
<tr>
<td>SR Latch</td>
<td>●</td>
</tr>
<tr>
<td>Capture/Compare/PWM Modules</td>
<td>•</td>
</tr>
<tr>
<td>ECCP1</td>
<td>●</td>
</tr>
<tr>
<td>Comparators</td>
<td>C1</td>
</tr>
<tr>
<td>Master Synchronous Serial Ports</td>
<td>MSSP</td>
</tr>
<tr>
<td>Timers</td>
<td>Timer0</td>
</tr>
<tr>
<td></td>
<td>Timer1</td>
</tr>
<tr>
<td></td>
<td>Timer2</td>
</tr>
</tbody>
</table>
FIGURE 1-1: PIC12LF1840T39A BLOCK DIAGRAM

Note 1: See applicable chapters for more information on peripherals.
Note 2: See Table 1-1 for peripherals available on specific devices.
<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA0/AN0/CPS0/C1IN+/DACOUT/TX(1)/CK(1)/SS(1)/P1B(1)/MDOUT/ICSPDAT/</td>
<td>RA0</td>
<td>TTL</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN0</td>
<td>AN</td>
<td>—</td>
<td>A/D Channel 0 input.</td>
<td></td>
</tr>
<tr>
<td>CPS0</td>
<td>AN</td>
<td>—</td>
<td>Capacitive sensing input 0.</td>
<td></td>
</tr>
<tr>
<td>C1IN+</td>
<td>AN</td>
<td>—</td>
<td>Comparator C1 positive input.</td>
<td></td>
</tr>
<tr>
<td>DACOUT</td>
<td>—</td>
<td>AN</td>
<td>Digital-to-Analog Converter output.</td>
<td></td>
</tr>
<tr>
<td>TX</td>
<td>—</td>
<td>CMOS</td>
<td>USART asynchronous transmit.</td>
<td></td>
</tr>
<tr>
<td>CK</td>
<td>ST</td>
<td>CMOS</td>
<td>USART synchronous clock.</td>
<td></td>
</tr>
<tr>
<td>SDO</td>
<td>—</td>
<td>CMOS</td>
<td>SPI data output.</td>
<td></td>
</tr>
<tr>
<td>SS</td>
<td>ST</td>
<td>—</td>
<td>Slave Select input.</td>
<td></td>
</tr>
<tr>
<td>P1B</td>
<td>—</td>
<td>CMOS</td>
<td>PWM output.</td>
<td></td>
</tr>
<tr>
<td>MDOUT</td>
<td>—</td>
<td>CMOS</td>
<td>Modulator output.</td>
<td></td>
</tr>
<tr>
<td>ICSPDAT</td>
<td>ST</td>
<td>CMOS</td>
<td>ICSP™ Data I/O.</td>
<td></td>
</tr>
<tr>
<td>RA1/AN1/CPS1/VREF/C1IN0-/SRI/RX(1)/DT(1)/SCL/SCK/MDMIN/ICSPCLK</td>
<td>RA1</td>
<td>TTL</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN1</td>
<td>AN</td>
<td>—</td>
<td>A/D Channel 1 input.</td>
<td></td>
</tr>
<tr>
<td>CPS1</td>
<td>AN</td>
<td>—</td>
<td>Capacitive sensing input 1.</td>
<td></td>
</tr>
<tr>
<td>VREF</td>
<td>AN</td>
<td>—</td>
<td>A/D and DAC Positive Voltage Reference input.</td>
<td></td>
</tr>
<tr>
<td>C1IN0-</td>
<td>AN</td>
<td>—</td>
<td>Comparator C1 negative input.</td>
<td></td>
</tr>
<tr>
<td>SRI</td>
<td>ST</td>
<td>—</td>
<td>SR Latch input.</td>
<td></td>
</tr>
<tr>
<td>RX</td>
<td>ST</td>
<td>—</td>
<td>USART asynchronous input.</td>
<td></td>
</tr>
<tr>
<td>DT</td>
<td>ST</td>
<td>CMOS</td>
<td>USART synchronous data.</td>
<td></td>
</tr>
<tr>
<td>SCL</td>
<td>I²C™</td>
<td>OD</td>
<td>I²C™ clock.</td>
<td></td>
</tr>
<tr>
<td>SCK</td>
<td>ST</td>
<td>CMOS</td>
<td>SPI clock.</td>
<td></td>
</tr>
<tr>
<td>MDMIN</td>
<td>ST</td>
<td>—</td>
<td>Modulator source input.</td>
<td></td>
</tr>
<tr>
<td>ICSPCLK</td>
<td>ST</td>
<td>—</td>
<td>Serial Programming Clock.</td>
<td></td>
</tr>
<tr>
<td>RA2/AN2/CPS2/C1OUT/SRQ/T0CKI/CCP1(1)/P1A(1)/FLT0/SDA/SDL/INT/MDCIN1</td>
<td>RA2</td>
<td>ST</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN2</td>
<td>AN</td>
<td>—</td>
<td>A/D Channel 2 input.</td>
<td></td>
</tr>
<tr>
<td>CPS2</td>
<td>AN</td>
<td>—</td>
<td>Capacitive sensing input 2.</td>
<td></td>
</tr>
<tr>
<td>C1OUT</td>
<td>—</td>
<td>CMOS</td>
<td>Comparator C1 output.</td>
<td></td>
</tr>
<tr>
<td>SRQ</td>
<td>—</td>
<td>CMOS</td>
<td>SR Latch non-inverting output.</td>
<td></td>
</tr>
<tr>
<td>T0CKI</td>
<td>ST</td>
<td>—</td>
<td>Timer0 clock input.</td>
<td></td>
</tr>
<tr>
<td>CCP1</td>
<td>ST</td>
<td>CMOS</td>
<td>Capture/Compare/PWM 1.</td>
<td></td>
</tr>
<tr>
<td>P1A</td>
<td>—</td>
<td>CMOS</td>
<td>PWM output.</td>
<td></td>
</tr>
<tr>
<td>FLT0</td>
<td>ST</td>
<td>—</td>
<td>ECCP Auto-Shutdown Fault input.</td>
<td></td>
</tr>
<tr>
<td>SDA</td>
<td>I²C™</td>
<td>OD</td>
<td>I²C™ data input/output.</td>
<td></td>
</tr>
<tr>
<td>SDI</td>
<td>CMOS</td>
<td>—</td>
<td>SPI data input.</td>
<td></td>
</tr>
<tr>
<td>INT</td>
<td>ST</td>
<td>—</td>
<td>External interrupt.</td>
<td></td>
</tr>
<tr>
<td>MDCIN1</td>
<td>ST</td>
<td>—</td>
<td>Modulator Carrier Input 1.</td>
<td></td>
</tr>
<tr>
<td>RA3/SS(1)/T1G(1)/VPP/MCLR</td>
<td>RA3</td>
<td>TTL</td>
<td>—</td>
<td>General purpose input.</td>
</tr>
<tr>
<td>SS</td>
<td>ST</td>
<td>—</td>
<td>Slave Select input.</td>
<td></td>
</tr>
<tr>
<td>T1G</td>
<td>ST</td>
<td>—</td>
<td>Timer1 Gate input.</td>
<td></td>
</tr>
<tr>
<td>VPP</td>
<td>HV</td>
<td>—</td>
<td>Programming voltage.</td>
<td></td>
</tr>
<tr>
<td>MCLR</td>
<td>ST</td>
<td>—</td>
<td>Master Clear with internal pull-up.</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- **AN** = Analog input or output
- **TTL** = TTL compatible input
- **CMOS** = CMOS compatible input or output
- **OD** = Open Drain
- **ST** = Schmitt Trigger input with CMOS levels
- **I²C™** = Schmitt Trigger input with I²C levels
- **HV** = High Voltage
- **XTAL** = Crystal levels

**Note 1:** Pin functions can be assigned to one of two pin locations via software. See the APFCON register (Register 12-1).
### TABLE 1-2: PIC12LF1840T39A PINOUT DESCRIPTION (CONTINUED)

<table>
<thead>
<tr>
<th>Name</th>
<th>Function</th>
<th>Input Type</th>
<th>Output Type</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA4/AN3/CPS3/OSC2/CLKOUT/T1OSO/C1IN1/-CLKR/SDO(1)/CK(1)/TX(1)/P1B(1)/T1G(1)/MDCIN2</td>
<td>RA4</td>
<td>TTL</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>AN3</td>
<td>AN</td>
<td>—</td>
<td>—</td>
<td>A/D Channel 3 input.</td>
</tr>
<tr>
<td>CPS3</td>
<td>AN</td>
<td>—</td>
<td>—</td>
<td>Capacitive sensing input 3.</td>
</tr>
<tr>
<td>OSC2</td>
<td>—</td>
<td>XTAL</td>
<td>Crystal/Resonator (LP, XT, HS modes).</td>
<td></td>
</tr>
<tr>
<td>CLKOUT</td>
<td>—</td>
<td>CMOS</td>
<td>Fosc/4 output.</td>
<td></td>
</tr>
<tr>
<td>T1OSO</td>
<td>XTAL</td>
<td>XTAL</td>
<td>Timer1 oscillator connection.</td>
<td></td>
</tr>
<tr>
<td>C1IN1-</td>
<td>AN</td>
<td>—</td>
<td>—</td>
<td>Comparator C1 negative input.</td>
</tr>
<tr>
<td>CLKR</td>
<td>—</td>
<td>CMOS</td>
<td>Clock Reference output.</td>
<td></td>
</tr>
<tr>
<td>SDO</td>
<td>—</td>
<td>CMOS</td>
<td>SPI data output.</td>
<td></td>
</tr>
<tr>
<td>CK</td>
<td>ST</td>
<td>CMOS</td>
<td>USART synchronous clock.</td>
<td></td>
</tr>
<tr>
<td>TX</td>
<td>—</td>
<td>CMOS</td>
<td>USART asynchronous transmit.</td>
<td></td>
</tr>
<tr>
<td>P1B</td>
<td>—</td>
<td>CMOS</td>
<td>PWM output.</td>
<td></td>
</tr>
<tr>
<td>T1G</td>
<td>ST</td>
<td>—</td>
<td>—</td>
<td>Timer1 Gate input.</td>
</tr>
<tr>
<td>MDCIN2</td>
<td>ST</td>
<td>—</td>
<td>—</td>
<td>Modulator Carrier Input 2.</td>
</tr>
<tr>
<td>RA5/CLKIN/OSC1/T1OSI/T1CKI/SRNQ/P1A(1)/CCP1(1)/DT(1)/RX(1)</td>
<td>RA5</td>
<td>TTL</td>
<td>CMOS</td>
<td>General purpose I/O.</td>
</tr>
<tr>
<td>CLKIN</td>
<td>CMOS</td>
<td>—</td>
<td>—</td>
<td>External clock input (EC mode).</td>
</tr>
<tr>
<td>OSC1</td>
<td>XTAL</td>
<td>—</td>
<td>—</td>
<td>Crystal/Resonator (LP, XT, HS modes).</td>
</tr>
<tr>
<td>T1OSI</td>
<td>XTAL</td>
<td>XTAL</td>
<td>Timer1 oscillator connection.</td>
<td></td>
</tr>
<tr>
<td>T1CKI</td>
<td>ST</td>
<td>—</td>
<td>—</td>
<td>Timer1 clock input.</td>
</tr>
<tr>
<td>SRNQ</td>
<td>—</td>
<td>CMOS</td>
<td>SR Latch inverting output.</td>
<td></td>
</tr>
<tr>
<td>P1A</td>
<td>—</td>
<td>CMOS</td>
<td>PWM output.</td>
<td></td>
</tr>
<tr>
<td>CCP1</td>
<td>ST</td>
<td>CMOS</td>
<td>Capture/Compare/PWM 1.</td>
<td></td>
</tr>
<tr>
<td>DT</td>
<td>ST</td>
<td>CMOS</td>
<td>USART synchronous data.</td>
<td></td>
</tr>
<tr>
<td>RX</td>
<td>ST</td>
<td>—</td>
<td>—</td>
<td>USART asynchronous input.</td>
</tr>
<tr>
<td>VDD</td>
<td>VDD</td>
<td>Power</td>
<td>—</td>
<td>Positive supply.</td>
</tr>
<tr>
<td>VSS</td>
<td>VSS</td>
<td>Power</td>
<td>—</td>
<td>Ground reference.</td>
</tr>
<tr>
<td>VDDRF</td>
<td>VDDRF</td>
<td>Power</td>
<td>—</td>
<td>Positive supply.</td>
</tr>
<tr>
<td>VSSRF</td>
<td>VSSRF</td>
<td>Power</td>
<td>—</td>
<td>Ground reference.</td>
</tr>
</tbody>
</table>

**Legend:**
- **AN** = Analog input or output
- **CMOS** = CMOS compatible input or output
- **OCTM** = Schmitt Trigger input with CMOS levels
- **I2C™** = Schmitt Trigger input with I2C levels
- **TTL** = TTL compatible input
- **ST** = Schmitt Trigger input with CMOS levels
- **HV** = High Voltage
- **XTAL** = Crystal

**Note 1:** Pin functions can be assigned to one of two pin locations via software. See the APFCON register (Register 12-1).
2.0 ENHANCED MID-RANGE CPU

This family of devices contain an enhanced mid-range 8-bit CPU core. The CPU has 49 instructions. Interrupt capability includes automatic context saving. The hardware stack is 16 levels deep and has Overflow and Underflow Reset capability. Direct, Indirect, and Relative Addressing modes are available. Two File Select Registers (FSRs) provide the ability to read program and data memory.

- Automatic Interrupt Context Saving
- 16-level Stack with Overflow and Underflow
- File Select Registers
- Instruction Set

2.1 Automatic Interrupt Context Saving

During interrupts, certain registers are automatically saved in shadow registers and restored when returning from the interrupt. This saves stack space and user code. See Section 8.5 “Automatic Context Saving”, for more information.

2.2 16-Level Stack with Overflow and Underflow

These devices have an external stack memory 15 bits wide and 16 words deep. A Stack Overflow or Underflow will set the appropriate bit (STKOVF or STKUNF) in the PCON register, and if enabled will cause a software Reset. See Section 3.4 “Stack” for more details.

2.3 File Select Registers

There are two 16-bit File Select Registers (FSR). FSRs can access all file registers and program memory, which allows one Data Pointer for all memory. When an FSR points to program memory, there is one additional instruction cycle in instructions using INDF to allow the data to be fetched. General purpose memory can now also be addressed linearly, providing the ability to access contiguous data larger than 80 bytes. There are also new instructions to support the FSRs. See Section 3.5 “Indirect Addressing” for more details.

2.4 Instruction Set

There are 49 instructions for the enhanced mid-range CPU to support the features of the CPU. See Section 30.0 “Instruction Set Summary” for more details.
FIGURE 2-1: CORE BLOCK DIAGRAM

- **Configuration**
- **Flash Program Memory**
- **Program Counter**
- **16-Level Stack (15-bit)**
- **Program Memory Read (PMR)**
- **Address MUX**
- **Status Register**
- **Address MUX**
- **IR Register**
- **Direct Address 7**
- **BSR Register**
- **FSR Register**
- **Multiplexer**
- **RAM**
- **RAM Address**
- **Instruction Decode and Control**
- **Timing Generation**
- **Internal Oscillator Block**
- **Power-up Timer**
- **Oscillator Start-up Timer**
- **Power-on Reset**
- **Watchdog Timer**
- **Brown-out Reset**
- **W Register**
- **Power-up Timer**
- **OSC1/CLKIN**
- **OSC2/CLKOUT**
- **VDD**
- **VSS**
3.0 MEMORY ORGANIZATION

There are three types of memory in PIC12LF1840T39A devices: Data Memory, Program Memory and Data EEPROM Memory(1).

- Program Memory
- Data Memory
  - Core Registers
  - Special Function Registers
  - General Purpose RAM
  - Common RAM
  - Device Memory Maps
  - Special Function Registers Summary
- Data EEPROM memory(1)

The following features are associated with access and control of program memory and data memory:
- PCL and PCLATH
- Stack
- Indirect Addressing

3.1 Program Memory Organization

The enhanced mid-range core has a 15-bit program counter capable of addressing a 32K x 14 program memory space. Table 3-1 shows the memory sizes implemented for the PIC12LF1840T39A family. Accessing a location above these boundaries will cause a wrap-around within the implemented memory space. The Reset vector is at 0000h and the interrupt vector is at 0004h (see Figure 3-1).

Note 1: The Data EEPROM Memory and the method to access Flash memory through the EECON registers is described in Section 11.0 “Data EEPROM and Flash Program Memory Control”.

<table>
<thead>
<tr>
<th>Device</th>
<th>Program Memory Space (Words)</th>
<th>Last Program Memory Address</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC12LF1840T39A</td>
<td>4,096</td>
<td>0FFh</td>
</tr>
</tbody>
</table>
3.1.1 READING PROGRAM MEMORY AS DATA

There are two methods of accessing constants in program memory. The first method is to use tables of RETLW instructions. The second method is to set an FSR to point to the program memory.

3.1.1.1 RETLW Instruction

The RETLW instruction can be used to provide access to tables of constants. The recommended way to create such a table is shown in Example 3-1.

```
EXAMPLE 3-1: RETLW INSTRUCTION

constants
  BRW ;Add Index in W to ;program counter to ;select data
  RETLW DATA0 ;Index0 data
  RETLW DATA1 ;Index1 data
  RETLW DATA2
  RETLW DATA3

my_function
  ;-- LOTS OF CODE--
  MOVlw DATA_INDEX
  CALL constants
  ;-- THE CONSTANT IS IN W
```

The BRW instruction makes this type of table very simple to implement. If your code must remain portable with previous generations of microcontrollers, then the BRW instruction is not available so the older table read method must be used.
3.1.1.2 Indirect Read with FSR

The program memory can be accessed as data by setting bit 7 of the FSR\(x\)H register and reading the matching INDF\(x\) register. The MOV\(I\) instruction will place the lower eight bits of the addressed word in the W register. Writes to the program memory cannot be performed via the INDF registers. Instructions that access the program memory via the FSR require one extra instruction cycle to complete. Example 3-2 demonstrates accessing the program memory via an FSR.

The HIGH directive will set bit 7 if a label points to a location in program memory.

**EXAMPLE 3-2: ACCESSING PROGRAM MEMORY VIA FSR**

```assembly
constants
  RETLW DATA0  ;Index0 data
  RETLW DATA1  ;Index1 data
  RETLW DATA2
  RETLW DATA3
my_function
  ;... LOTS OF CODE...
  MOVLW LOW constants
  MOVWF FSR1L
  MOVLW HIGH constants
  MOVWF FSR1H
  MOVWI 0[FSR1]
;THE PROGRAM MEMORY IS IN W
```

3.2 Data Memory Organization

The data memory is partitioned in 32 memory banks with 128 bytes in a bank. Each bank consists of (see Figure 3-2):

- 12 core registers
- 20 Special Function Registers (SFR)
- Up to 80 bytes of General Purpose RAM (GPR)
- 16 bytes of common RAM

The active bank is selected by writing the bank number into the Bank Select Register (BSR). Unimplemented memory will read as '0'. All data memory can be accessed either directly (via instructions that use the file registers) or indirectly via the two File Select Registers (FSR). See Section 3.5 “Indirect Addressing” for more information.
3.2.1.1 STATUS Register

The STATUS register, shown in Register 3-1, contains:

- the arithmetic status of the ALU
- the Reset status

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS will clear the upper three bits and set the Z bit. This leaves the STATUS register as '000u uluu' (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits (Refer to Section 30.0 “Instruction Set Summary”).

**Note 1:** The C and DC bits operate as Borrow and Digit Borrow out bits, respectively, in subtraction.

### REGISTER 3-1: STATUS: STATUS REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R-1/q</th>
<th>R-1/q</th>
<th>R/W-0/u</th>
<th>R/W-0/u</th>
<th>R/W-0/u</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC(1)</td>
<td>C(1)</td>
</tr>
</tbody>
</table>

**Legend:**
R = Readable bit
W = Writable bit
U = Unimplemented bit, read as ‘0’
u = Bit is unchanged
x = Bit is unknown
‘1’ = Bit is set
‘0’ = Bit is cleared
q = Value depends on condition

**bit 7 to 0**

**Unimplemented:** Read as ‘0’

**bit 4**

TO: Time-out bit
1 = After power-up, CLRWDt instruction or SLEEP instruction
0 = A WDT time-out occurred

**bit 3**

PD: Power-down bit
1 = After power-up or by the CLRWDt instruction
0 = By execution of the SLEEP instruction

**bit 2**

Z: Zero bit
1 = The result of an arithmetic or logic operation is zero
0 = The result of an arithmetic or logic operation is not zero

**bit 1**

DC: Digit Carry/Digit Borrow bit(1)
1 = A carry-out from the 4th low-order bit of the result occurred
0 = No carry-out from the 4th low-order bit of the result

**bit 0**

C: Carry/Borrow bit(1)
1 = A carry-out from the Most Significant bit of the result occurred
0 = No carry-out from the Most Significant bit of the result occurred

**Note 1:** For Borrow, the polarity is reversed. A subtraction is executed by adding the two’s complement of the second operand.
3.2.2 SPECIAL FUNCTION REGISTER
The Special Function Registers are registers used by the application to control the desired operation of peripheral functions in the device. The registers associated with the operation of the peripherals are described in the appropriate peripheral chapter of this data sheet.

3.2.3 GENERAL PURPOSE RAM
There are up to 80 bytes of GPR in each data memory bank.

3.2.3.1 Linear Access to GPR
The general purpose RAM can be accessed in a non-banked method via the FSRs. This can simplify access to large memory structures. See Section 3.5.2 “Linear Data Memory” for more information.

3.2.4 COMMON RAM
There are 16 bytes of common RAM accessible from all banks.

3.2.5 DEVICE MEMORY MAPS
The memory maps for the device family are as shown in Table 3-2.

<table>
<thead>
<tr>
<th>Device</th>
<th>Banks</th>
<th>Table No.</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC12LF1840T39A</td>
<td>0-7</td>
<td>Table 3-3</td>
</tr>
<tr>
<td></td>
<td>8-23</td>
<td>Table 3-4</td>
</tr>
<tr>
<td></td>
<td>24-31</td>
<td>Table 3-5</td>
</tr>
<tr>
<td></td>
<td>31</td>
<td>Table 3-4</td>
</tr>
</tbody>
</table>

FIGURE 3-2: BANKED MEMORY PARTITIONING

<table>
<thead>
<tr>
<th>7-bit Bank Offset</th>
<th>Memory Region</th>
</tr>
</thead>
<tbody>
<tr>
<td>00h</td>
<td>Core Registers (12 bytes)</td>
</tr>
<tr>
<td>0Bh</td>
<td>Special Function Registers (20 bytes maximum)</td>
</tr>
<tr>
<td>0Ch</td>
<td>General Purpose RAM (80 bytes maximum)</td>
</tr>
<tr>
<td>1Fh 20h</td>
<td></td>
</tr>
<tr>
<td>6Fh 70h</td>
<td>Common RAM (16 bytes)</td>
</tr>
<tr>
<td>7Fh</td>
<td></td>
</tr>
<tr>
<td>BANK 0</td>
<td>BANK 1</td>
</tr>
<tr>
<td>-------</td>
<td>-------</td>
</tr>
<tr>
<td>000h</td>
<td>080h</td>
</tr>
<tr>
<td>00Bh</td>
<td>08Bh</td>
</tr>
<tr>
<td>00Ch</td>
<td>PORTA</td>
</tr>
<tr>
<td>00Dh</td>
<td>—</td>
</tr>
<tr>
<td>00Eh</td>
<td>—</td>
</tr>
<tr>
<td>00Fh</td>
<td>—</td>
</tr>
<tr>
<td>010h</td>
<td>—</td>
</tr>
<tr>
<td>011h</td>
<td>PIR1</td>
</tr>
<tr>
<td>012h</td>
<td>PIR2</td>
</tr>
<tr>
<td>013h</td>
<td>—</td>
</tr>
<tr>
<td>014h</td>
<td>—</td>
</tr>
<tr>
<td>015h</td>
<td>TMRO</td>
</tr>
<tr>
<td>016h</td>
<td>TMR1L</td>
</tr>
<tr>
<td>017h</td>
<td>TMR1H</td>
</tr>
<tr>
<td>018h</td>
<td>T1CON</td>
</tr>
<tr>
<td>019h</td>
<td>T2CON</td>
</tr>
<tr>
<td>01Ah</td>
<td>TMRO</td>
</tr>
<tr>
<td>01Bh</td>
<td>PR2</td>
</tr>
<tr>
<td>01Ch</td>
<td>ADRESL</td>
</tr>
<tr>
<td>01Dh</td>
<td>ADRESH</td>
</tr>
<tr>
<td>01Eh</td>
<td>CPSCON0</td>
</tr>
<tr>
<td>020h</td>
<td>CPSCON1</td>
</tr>
</tbody>
</table>

**Legend:**

- **= Unimplemented data memory locations, read as '0'.**
- **Available only on PIC12F1840.**
### TABLE 3-3: PIC12(L)F1840 MEMORY MAP (CONTINUED)

<table>
<thead>
<tr>
<th>BANK 8</th>
<th>BANK 9</th>
<th>BANK 10</th>
<th>BANK 11</th>
<th>BANK 12</th>
<th>BANK 13</th>
<th>BANK 14</th>
<th>BANK 15</th>
</tr>
</thead>
<tbody>
<tr>
<td>400h</td>
<td>40Bh</td>
<td>40Ch</td>
<td>Unimplemented Read as '0'</td>
<td>40Ch</td>
<td>Unimplemented Read as '0'</td>
<td>500h</td>
<td>50Bh</td>
</tr>
<tr>
<td>40Ch</td>
<td>48Fh</td>
<td>4EFh</td>
<td>Unimplemented Read as '0'</td>
<td>48Fh</td>
<td>Unimplemented Read as '0'</td>
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<td>8EFh</td>
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<td>90Bh</td>
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<td>8EFh</td>
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<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
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<td>98Fh</td>
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<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
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<td>98Fh</td>
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<td>8EFh</td>
<td>8EFh</td>
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<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
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<td>98Fh</td>
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<td>8EFh</td>
<td>8EFh</td>
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<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
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<td>8EFh</td>
<td>8EFh</td>
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<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
<td>90Fh</td>
<td>98Fh</td>
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<td>8EFh</td>
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<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
<td>90Fh</td>
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<td>Unimplemented Read as '0'</td>
<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
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<td>98Fh</td>
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<td>8EFh</td>
<td>8EFh</td>
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<td>8EFh</td>
<td>Unimplemented Read as '0'</td>
<td>90Fh</td>
<td>98Fh</td>
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**Legend:**
- **= Unimplemented data memory locations, read as '0'**
TABLE 3-4: PIC12LF1840T39A MEMORY MAP, BANK 31

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<tr>
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<tr>
<td>FA0h</td>
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<td>FE3h</td>
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| FE4h    | Unimplemented Read as '0'
| FE5h    |
| FE6h    |
| FE7h    |
| FE8h    |
| FE9h    |
| FEAh    |
| FEBh    |
| FEDh    |
| FEEh    | —
| FEFh    |

Legend: = Unimplemented data memory locations, read as '0'.

3.2.6 SPECIAL FUNCTION REGISTERS SUMMARY

The Special Function Register Summary for the device family are as follows:

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## TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY

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<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
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</table>

Legend:  x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, r = reserved.

Note 1: These registers can be addressed from any bank.
Note 2: Unimplemented, read as '1'.

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### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

<table>
<thead>
<tr>
<th>Address</th>
<th>Name</th>
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<th>Bit 6</th>
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<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
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<th>Value on all other Resets</th>
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**Legend:**  
- **x** = unknown, **u** = unchanged, **q** = value depends on condition, **-** = unimplemented, **r** = reserved.  
Shaded locations are unimplemented, read as '0'.

**Note 1:** These registers can be addressed from any bank.

**Note 2:** Unimplemented, read as '1'.
### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

<table>
<thead>
<tr>
<th>Address (H)</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
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<th>Value on POR, BOR</th>
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<td>TSRNG</td>
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<td>Unimplemented</td>
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</tr>
</tbody>
</table>

**Legend:**  
\( \times \) = unknown, \( u \) = unchanged, \( q \) = value depends on condition, \( - \) = unimplemented, \( r \) = reserved.  
Shaded locations are unimplemented, read as ‘0’.  

**Note:**  
1. These registers can be addressed from any bank.  
2. Unimplemented, read as ‘1’.  

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DS40001636B-page 23
### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

<table>
<thead>
<tr>
<th>Address</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
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</thead>
<tbody>
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<td>INDF0</td>
<td>Addressing this location uses contents of FSR0H/FSR0L to address data memory (not a physical register)</td>
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<td>xxxx</td>
<td>xxxx</td>
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<td>xxxx</td>
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<td>xxxx</td>
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<td>xxxx</td>
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<td>PCL</td>
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**Legend:**
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, r = reserved.
- Shaded locations are unimplemented, read as ‘0’.

**Note:**
1. These registers can be addressed from any bank.
2. Unimplemented, read as ‘1’.
### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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<thead>
<tr>
<th>Address</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
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<th>Bit 4</th>
<th>Bit 3</th>
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**Legend:**  
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, c = reserved.  
- Shaded locations are unimplemented, read as '0'.

**Note:**  
1: These registers can be addressed from any bank.  
2: Unimplemented, read as '1.'
### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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<th>Address</th>
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<th>Bit 3</th>
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**Legend:**  
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, c = reserved.  
- Shaded locations are unimplemented, read as ‘0’.  

**Note 1:** These registers can be addressed from any bank.  
**Note 2:** Unimplemented, read as ‘1’.
### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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**Legend:**
- x = unknown,
- u = unchanged,
- q = value depends on condition,
- - = unimplemented,
- r = reserved.

**Note**
1. These registers can be addressed from any bank.
2. Unimplemented, read as ‘1’.
### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

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</tr>
<tr>
<td>399h</td>
<td>—</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>—</td>
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</tr>
<tr>
<td>39Ah</td>
<td>CLKRCON</td>
<td>CLKREN</td>
<td>CLKROE</td>
<td>CLKRSLR</td>
<td>CLKRDIC&lt;1:0&gt;</td>
<td>CLKRDIR&lt;2:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>0011 0000 0011 0000</td>
<td></td>
</tr>
<tr>
<td>39Bh</td>
<td>—</td>
<td></td>
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<td></td>
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<td></td>
<td></td>
<td></td>
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</tr>
<tr>
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<td>MDCON</td>
<td>MDEN</td>
<td>MDOE</td>
<td>MDSLR</td>
<td>MDPOL</td>
<td>MOUT</td>
<td>—</td>
<td>—</td>
<td>MDBIT</td>
<td>0010 —---0 0010 —---0</td>
<td></td>
</tr>
<tr>
<td>39Dh</td>
<td>MDSRC</td>
<td>MDMSODIS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MDMS&lt;3:0&gt;</td>
<td>x--- xxxx u--- uuuu</td>
<td></td>
<td></td>
</tr>
<tr>
<td>39Eh</td>
<td>MDCARL</td>
<td>MDCDIS</td>
<td>MDCPOL</td>
<td>MDCSYSNC</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MDC&lt;3:0&gt;</td>
<td>xxxx--- xxxx yyyy uuuu</td>
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<td>MDCODIS</td>
<td>MDCHPOL</td>
<td>MDCHSYSNC</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MDC&lt;3:0&gt;</td>
<td>xxxx--- xxxx yyyy uuuu</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- x = unknown,
- u = unchanged,
- q = value depends on condition,
- - = unimplemented,
- r = reserved.

**Note 1:** Shaded locations are unimplemented, read as ‘0’.

**Note 2:** Some registers can be addressed from any bank.

**Note 3:** Unimplemented, read as ‘1’.

---

**Legend:**
- x = unknown,
- u = unchanged,
- q = value depends on condition,
- - = unimplemented,
- r = reserved.

**Note 1:** These registers can be addressed from any bank.

**Note 2:** Unimplemented, read as ‘1’.

---

**Note 3:** Shaded locations are unimplemented, read as ‘0’.

---

**Note 4:** Some registers can be addressed from any bank.

**Note 5:** Unimplemented, read as ‘1’.
### TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

<table>
<thead>
<tr>
<th>Address</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
</tr>
</thead>
<tbody>
<tr>
<td>x00h/</td>
<td>INDF0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>xxxxx xxxxx</td>
<td>xxxxx xxxxx xxxxx xxxxx</td>
</tr>
<tr>
<td>x00h/</td>
<td>INDF1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>xxxxx xxxxx</td>
<td>xxxxx xxxxx xxxxx xxxxx</td>
</tr>
<tr>
<td>x02h/</td>
<td>PCL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0000 0000</td>
<td>0000 0000</td>
</tr>
<tr>
<td>x03h/</td>
<td>STATUS</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>------</td>
<td>---1 1000</td>
</tr>
<tr>
<td>x04h/</td>
<td>FSR0L</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0000 0000</td>
<td>uuuu uuuu</td>
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<td>x05h/</td>
<td>FSR0H</td>
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<td></td>
<td></td>
<td></td>
<td>0000 0000</td>
<td>0000 0000</td>
</tr>
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<td>x06h/</td>
<td>FSR1L</td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td>0000 0000</td>
<td>uuuu uuuu</td>
</tr>
<tr>
<td>x07h/</td>
<td>FSR1H</td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td>0000 0000</td>
<td>0000 0000</td>
</tr>
<tr>
<td>x08h/</td>
<td>BSR</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>-----</td>
<td>---0 0000</td>
<td>----0 0000</td>
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<tr>
<td>x09h/</td>
<td>WREG</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td>0000 0000</td>
<td>uuuu uuuu</td>
</tr>
<tr>
<td>x0Ah/</td>
<td>PCLATH</td>
<td>-----</td>
<td>Write Buffer for the upper 7 bits of the Program Counter</td>
<td>0000 0000</td>
<td>-000 0000</td>
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<td></td>
<td></td>
<td></td>
<td></td>
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<tr>
<td>x0Bh/</td>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMROIE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>0000 000x</td>
<td>0000 000u</td>
</tr>
<tr>
<td>x0Ch/</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
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<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- x = unknown,
- u = unchanged,
- q = value depends on condition,
- - = unimplemented,
- v = reserved.

**Note**
1. These registers can be addressed from any bank.
2. Unimplemented, read as ‘1’.
<table>
<thead>
<tr>
<th>Bank 31</th>
<th>Address</th>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Value on POR, BOR</th>
<th>Value on all other Resets</th>
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</thead>
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<td></td>
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<td>xxxx</td>
</tr>
<tr>
<td></td>
<td>F81h</td>
<td>INDF1</td>
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<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td></td>
<td></td>
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<td>xxxx</td>
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<td>F82h</td>
<td>PCL</td>
<td>0000</td>
<td>0000</td>
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<td>0000</td>
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<td></td>
<td>0000</td>
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<td></td>
<td>F83h</td>
<td>STATUS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>Z</td>
<td>DC</td>
<td>C</td>
<td>—1 1000</td>
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<td>F84h</td>
<td>FSR0L</td>
<td>Indirect Data Memory Address 0 Low Pointer</td>
<td>0000</td>
<td>0000</td>
<td>0000</td>
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<td></td>
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<td>FSR0H</td>
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<td>0000</td>
<td>0000</td>
<td>0000</td>
<td>0000</td>
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<td></td>
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<tr>
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<td>—</td>
<td>—</td>
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<td>—</td>
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<td>BSR&lt;4:0&gt;</td>
<td></td>
<td>—-0 0000</td>
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<td>WREG</td>
<td>Working Register</td>
<td>0000</td>
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<td>0000</td>
<td>0000</td>
<td></td>
<td></td>
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<td>0000</td>
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<tr>
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<td>PCLATH</td>
<td>Write Buffer for the upper 7 bits of the Program Counter</td>
<td>-000</td>
<td>0000</td>
<td>-000</td>
<td>0000</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>-000</td>
</tr>
<tr>
<td></td>
<td>F8Bh</td>
<td>INTCON</td>
<td>GIE PEIE TMR0IF INTE IOCIE TMR0IE INTF IOCIF</td>
<td>0000</td>
<td>000x</td>
<td>0000</td>
<td>000u</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0000</td>
</tr>
<tr>
<td></td>
<td>F8Ch</td>
<td>—</td>
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<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
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</tr>
<tr>
<td></td>
<td>FE3h</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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</tr>
<tr>
<td></td>
<td>FE4h</td>
<td>STATUS_SHAD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>Z_SHAD</td>
<td>DC_SHAD</td>
<td>C_SHAD</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td>FE5h</td>
<td>WREG_SHAD</td>
<td>Working Register Shadow</td>
<td>0000</td>
<td>0000</td>
<td>0000</td>
<td>0000</td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td>—</td>
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<tr>
<td></td>
<td>FE7h</td>
<td>PCLATH_SHAD</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
<td>Program Counter Latch High Register Shadow</td>
<td>—xxx</td>
<td>xxxx</td>
<td>uuuu</td>
</tr>
<tr>
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<td>FE8h</td>
<td>FSR0L_SHAD</td>
<td>Indirect Data Memory Address 0 Low Pointer Shadow</td>
<td>xxxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
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<td>xxxxx</td>
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<td>FE9h</td>
<td>FSR0H_SHAD</td>
<td>Indirect Data Memory Address 0 High Pointer Shadow</td>
<td>xxxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
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<td>FEAh</td>
<td>FSR1L_SHAD</td>
<td>Indirect Data Memory Address 1 Low Pointer Shadow</td>
<td>xxxxx</td>
<td>xxxx</td>
<td>xxxx</td>
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<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
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<td>FE Bh</td>
<td>FSR1H_SHAD</td>
<td>Indirect Data Memory Address 1 High Pointer Shadow</td>
<td>xxxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
</tr>
<tr>
<td></td>
<td>FECh</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td></td>
<td>FE Dh</td>
<td>STKPTR</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<tr>
<td></td>
<td>FEEh</td>
<td>TOSL</td>
<td>Top-of-Stack Low byte</td>
<td>xxxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
<td>xxxxx</td>
</tr>
<tr>
<td></td>
<td>FE Fh</td>
<td>TOSH</td>
<td>Top-of-Stack High byte</td>
<td>—xxx</td>
<td>xxxx</td>
<td>—xxx</td>
<td>—xxx</td>
<td>—xxx</td>
<td>—xxx</td>
<td>—xxx</td>
<td>—xxx</td>
<td>—xxx</td>
</tr>
</tbody>
</table>

**Legend:**  
- x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, r = reserved.  
Shaded locations are unimplemented, read as '0'.

**Note:**  
1: These registers can be addressed from any bank.  
2: Unimplemented, read as '1'.

---

**TABLE 3-5: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)**
3.3 PCL and PCLATH

The Program Counter (PC) is 15 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<14:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 3-3 shows the five situations for the loading of the PC.

3.3.1 MODIFYING PCL

Executing any instruction with the PCL register as the destination simultaneously causes the Program Counter PC<14:8> bits (PCH) to be replaced by the contents of the PCLATH register. This allows the entire contents of the program counter to be changed by writing the desired upper seven bits to the PCLATH register. When the lower eight bits are written to the PCL register, all 15 bits of the program counter will change to the values contained in the PCLATH register and those being written to the PCL register.

3.3.2 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When performing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to Application Note AN556, “Implementing a Table Read” (DS00556).

3.3.3 COMPUTED FUNCTION CALLS

A computed function CALL allows programs to maintain tables of functions and provide another way to execute state machines or look-up tables. When performing a table read using a computed function CALL, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block).

If using the CALL instruction, the PCH<2:0> and PCL registers are loaded with the operand of the CALL instruction. PCH<6:3> is loaded with PCLATH<6:3>.

The CALLW instruction enables computed calls by combining PCLATH and W to form the destination address. A computed CALLW is accomplished by loading the W register with the desired address and executing CALLW. The PCL register is loaded with the value of W and PCH is loaded with PCLATH.

3.3.4 BRANCHING

The branching instructions add an offset to the PC. This allows relocatable code and code that crosses page boundaries. There are two forms of branching, BRW and BRA. The PC will have incremented to fetch the next instruction in both cases. When using either branching instruction, a PCL memory boundary may be crossed.

If using BRW, load the W register with the desired unsigned address and execute BRW. The entire PC will be loaded with the address PC + 1 + W.

If using BRA, the entire PC will be loaded with PC + 1 +, the signed value of the operand of the BRA instruction.
3.4 Stack

All devices have a 16-level x 15-bit wide hardware stack (refer to Figures 3-4 through 3-7). The stack space is not part of either program or data space. The PC is PUSHed onto the stack when CALL or CALLW instructions are executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer if the STVREN bit is programmed to '0' (Configuration Word 2). This means that after the stack has been PUSHed sixteen times, the seventeenth PUSH overwrites the value that was stored from the first PUSH. The eighteenth PUSH overwrites the second PUSH (and so on). The STKOVF and STKUNF flag bits will be set on an Overflow/Underflow, regardless of whether the Reset is enabled.

Note 1: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, CALLW, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

3.4.1 ACCESSING THE STACK

The stack is available through the TOSH, TOSL and STKPTR registers. STKPTR is the current value of the Stack Pointer. TOSH:TOSL register pair points to the top of the stack. Both registers are read/writable. TOS is split into TOSH and TOSL due to the 15-bit size of the PC. To access the stack, adjust the value of STKPTR, which will position TOSH:TOSL, then read/write to TOSH:TOSL. STKPTR is five bits to allow detection of overflow and underflow.

Note: Care should be taken when modifying the STKPTR while interrupts are enabled.

During normal program operation, CALL, CALLW and Interrupts will increment STKPTR while RETLW, RETURN, and RETFIE will decrement STKPTR. At any time STKPTR can be inspected to see how much stack is left. The STKPTR always points at the currently used place on the stack. Therefore, a CALL or CALLW will increment the STKPTR and then write the PC, and a return will unload the PC and then decrement STKPTR.

Reference Figure 3-4 through Figure 3-7 for examples of accessing the stack.

FIGURE 3-4: ACCESSING THE STACK EXAMPLE 1

Note: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, CALLW, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.
This figure shows the stack configuration after the first CALL or a single interrupt. If a RETURN instruction is executed, the return address will be placed in the Program Counter and the Stack Pointer decremented to the empty state (0x1F).

After seven CALLs or six CALLs and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and POP the stack.
3.4.2 OVERFLOW/UNDERFLOW RESET

If the STVREN bit in Configuration Word 2 is programmed to ‘1’, the device will be reset if the stack is PUSHed beyond the sixteenth level or POPed beyond the first level, setting the appropriate bits (STKOVF or STKUNF, respectively) in the PCON register.

3.5 Indirect Addressing

The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the File Select Registers (FSR). If the FSRn address specifies one of the two INDFn registers, the read will return ‘0’ and the write will not occur (though Status bits may be affected). The FSRn register value is created by the pair FSRnH and FSRnL.

The FSR registers form a 16-bit address that allows an addressing space with 65536 locations. These locations are divided into three memory regions:

- Traditional Data Memory
- Linear Data Memory
- Program Flash Memory

When the stack is full, the next CALL or an interrupt will set the Stack Pointer to 0x10. This is identical to address 0x00 so the stack will wrap and overwrite the return address at 0x00. If the Stack Overflow/Underflow Reset is enabled, a Reset will occur and location 0x00 will not be overwritten.
FIGURE 3-8: INDIRECT ADDRESSING

Note: Not all memory regions are completely implemented. Consult device memory tables for memory limits.
3.5.1 TRADITIONAL DATA MEMORY

The traditional data memory is a region from FSR address 0x000 to FSR address 0xFFF. The addresses correspond to the absolute addresses of all SFR, GPR and common registers.

**FIGURE 3-9: TRADITIONAL DATA MEMORY MAP**

<table>
<thead>
<tr>
<th>Direct Addressing</th>
<th>Indirect Addressing</th>
</tr>
</thead>
<tbody>
<tr>
<td>4 BSR 0 6 From Opcode 0</td>
<td>7 FSRxH 0 7 FSRxL 0</td>
</tr>
<tr>
<td>Bank Select</td>
<td>Bank Select</td>
</tr>
<tr>
<td>Location Select</td>
<td>Location Select</td>
</tr>
<tr>
<td>0x00</td>
<td>00000 00001 00010</td>
</tr>
<tr>
<td>0x7F</td>
<td>11111</td>
</tr>
</tbody>
</table>

Bank 0  Bank 1  Bank 2  Bank 31
3.5.2 LINEAR DATA MEMORY

The linear data memory is the region from FSR address 0x2000 to FSR address 0x29AF. This region is a virtual region that points back to the 80-byte blocks of GPR memory in all the banks.

Unimplemented memory reads as 0x00. Use of the linear data memory region allows buffers to be larger than 80 bytes because incrementing the FSR beyond one bank will go directly to the GPR memory of the next bank.

The 16 bytes of common memory are not included in the linear data memory region.

3.5.3 PROGRAM FLASH MEMORY

To make constant data access easier, the entire program Flash memory is mapped to the upper half of the FSR address space. When the MSB of FSRnH is set, the lower 15 bits are the address in program memory which will be accessed through INDF. Only the lower eight bits of each memory location is accessible via INDF. Writing to the program Flash memory cannot be accomplished via the FSR/INDF interface. All instructions that access program Flash memory via the FSR/INDF interface will require one additional instruction cycle to complete.
4.0 DEVICE CONFIGURATION

Device Configuration consists of Configuration Word 1 and Configuration Word 2, Code Protection and Device ID.

4.1 Configuration Words

There are several Configuration Word bits that allow different oscillator and memory protection options. These are implemented as Configuration Word 1 at 8007h and Configuration Word 2 at 8008h.

Note: The DEBUG bit in Configuration Word 2 is managed automatically by device development tools including debuggers and programmers. For normal device operation, this bit should be maintained as a '1'.

## REGISTER 4-1: CONFIGURATION WORD 1

<table>
<thead>
<tr>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>FCMEN</td>
<td>IESO</td>
<td>CLKOUTEN</td>
<td>BOREN1</td>
<td>BOREN0</td>
<td>CPD</td>
<td>CP</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>MCLRE</td>
<td>PWRTE</td>
<td>WDETE1</td>
<td>WDETE0</td>
<td>FOSC2</td>
<td>FOSC1</td>
<td>FOSC0</td>
</tr>
</tbody>
</table>

### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘1’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **P** = Programmable bit

### Bit Descriptions

- **bit 13** 
  - **FCMEN**: Fail-Safe Clock Monitor Enable bit
  - 1 = Fail-Safe Clock Monitor is enabled
  - 0 = Fail-Safe Clock Monitor is disabled

- **bit 12** 
  - **IESO**: Internal External Switchover bit
  - 1 = Internal/External Switchover mode is enabled
  - 0 = Internal/External Switchover mode is disabled

- **bit 11** 
  - **CLKOUTEN**: Clock Out Enable bit
  - If FOSC Configuration bits are set to LP, XT, HS modes:
  - This bit is ignored, CLKOUT function is disabled. Oscillator function on the CLKOUT pin.
  - All other FOSC modes:
    - 1 = CLKOUT function is disabled. I/O function on the CLKOUT pin.
    - 0 = CLKOUT function is enabled on the CLKOUT pin

- **bit 10-9** 
  - **BOREN<1:0>**: Brown-out Reset Enable bits
    - 11 = BOR enabled
    - 10 = BOR enabled during operation and disabled in Sleep
    - 01 = BOR controlled by SBOREN bit of the BORCON register
    - 00 = BOR disabled

- **bit 8** 
  - **CPD**: Data Code Protection bit
    - 1 = Data memory code protection is disabled
    - 0 = Data memory code protection is enabled

- **bit 7** 
  - **CP**: Code Protection bit
    - 1 = Program memory code protection is disabled
    - 0 = Program memory code protection is enabled

- **bit 6** 
  - **MCLRE**: RA3/MCLR/VPP Pin Function Select bit
    - If LVP bit = 1:
      - This bit is ignored.
    - If LVP bit = 0:
      - 1 = MCLR/VPP pin function is MCLR; Weak pull-up enabled.
      - 0 = MCLR/VPP pin function is digital input; MCLR internally disabled; Weak pull-up under control of WPUA register.

- **bit 5** 
  - **PWRTE**: Power-up Timer Enable bit
    - 1 = PWRT disabled
    - 0 = PWRT enabled

- **bit 4-3** 
  - **WDETE<1:0>**: Watchdog Timer Enable bit
    - 11 = WDT enabled
    - 10 = WDT enabled while running and disabled in Sleep
    - 01 = WDT controlled by the SWDTEN bit in the WDTCON register
    - 00 = WDT disabled

- **bit 2-0** 
  - **FOSC<2:0>**: Oscillator Selection bits
    - 111 = ECH: External Clock, High-Power mode (4-32 MHz): device clock supplied to CLKIN pin
    - 110 = ECM: External Clock, Medium-Power mode (0.5-4 MHz): device clock supplied to CLKIN pin
    - 101 = ECL: External Clock, Low-Power mode (0-0.5 MHz): device clock supplied to CLKIN pin
    - 100 = INTOSC oscillator: I/O function on CLKIN pin
    - 011 = EXTRC oscillator: External RC circuit connected to CLKIN pin
    - 010 = HS oscillator: High-speed crystal/resonator connected between OSC1 and OSC2 pins
    - 001 = XT oscillator: Crystal/resonator connected between OSC1 and OSC2 pins
    - 000 = LP oscillator: Low-power crystal connected between OSC1 and OSC2 pins

### Notes:
1. Enabling Brown-out Reset does not automatically enable Power-up Timer.
2. The entire EEPROM will be erased when the code protection is turned off during an erase.
3. The entire program memory will be erased when the code protection is turned off.
## REGISTER 4-2: CONFIGURATION WORD 2

<table>
<thead>
<tr>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>U-1</th>
<th>R/P-1/1</th>
<th>R/P-1/1</th>
<th>U-1</th>
<th>bit 7</th>
</tr>
</thead>
<tbody>
<tr>
<td>LVP&lt;1&gt;</td>
<td>DEBUG&lt;2&gt;</td>
<td>—</td>
<td>BORV&lt;3&gt;</td>
<td>STVREN</td>
<td>PLLEN</td>
<td>—</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>U-1</td>
<td>U-1</td>
<td>R-1</td>
<td>U-1</td>
<td>U-1</td>
<td>R/P-1/1</td>
<td>R/P-1/1</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘1’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **P** = Programmable bit

### Bit Descriptions

**Bit 13**
- **LVP**: Low-Voltage Programming Enable bit<sup>(1)</sup>
  - 1 = Low-voltage programming enabled
  - 0 = High-voltage on MCLR must be used for programming

**Bit 12**
- **DEBUG**: In-Circuit Debugger mode bit<sup>(2)</sup>
  - 1 = In-Circuit Debugger disabled, ICSPCLK and ICSPDAT are general purpose I/O pins
  - 0 = In-Circuit Debugger enabled, ICSPCLK and ICSPDAT are dedicated to the debugger

**Bit 11**
- **BORV**: Brown-out Reset Voltage Selection bit<sup>(3)</sup>
  - 1 = Brown-out Reset voltage (V<sub>b</sub>or), low trip point selected
  - 0 = Brown-out Reset voltage (V<sub>b</sub>or), high trip point selected

**Bit 10**
- **STVREN**: Stack Overflow/Underflow Reset Enable bit
  - 1 = Stack Overflow or Underflow will cause a Reset
  - 0 = Stack Overflow or Underflow will not cause a Reset

**Bit 9**
- **PLLEN**: PLL Enable bit
  - 1 = 4xPLL enabled
  - 0 = 4xPLL disabled

**Bit 8-5**
- **Reserved**: Read as ‘1’

**Bit 4**
- **Reserved**: This location should be programmed to a ‘1’.

**Bit 3-2**
- **Reserved**: Read as ‘1’

**Bit 1-0**
- **WRT<1:0>**: Flash Memory Self-Write Protection bits
  - 11 = Write protection off
  - 10 = 000h to 1FFh write-protected, 200h to FFFh may be modified by EECON control
  - 01 = 000h to 7FFh write-protected, 800h to FFFh may be modified by EECON control
  - 00 = 000h to FFFh write-protected, no addresses may be modified by EECON control

### Notes
1. The LVP bit cannot be programmed to ‘0’ when Programming mode is entered via LVP.
2. The **DEBUG** bit in Configuration Word is managed automatically by device development tools including debuggers and programmers. For normal device operation, this bit should be maintained as a ‘1’.
3. See V<sub>b</sub>or parameter for specific trip point voltages.
4.2 Code Protection

Code protection allows the device to be protected from unauthorized access. Program memory protection and data EEPROM protection are controlled independently. Internal access to the program memory and data EEPROM are unaffected by any code protection setting.

4.2.1 PROGRAM MEMORY PROTECTION

The entire program memory space is protected from external reads and writes by the CP bit in Configuration Word 1. When CP = 0, external reads and writes of program memory are inhibited and a read will return all '0's. The CPU can continue to read program memory, regardless of the protection bit settings. Writing the program memory is dependent upon the write protection setting. See Section 4.3 “Write Protection” for more information.

4.2.2 DATA EEPROM PROTECTION

The entire data EEPROM is protected from external reads and writes by the CPD bit. When CPD = 0, external reads and writes of data EEPROM are inhibited. The CPU can continue to read and write data EEPROM regardless of the protection bit settings.

4.3 Write Protection

Write protection allows the device to be protected from unintended self-writes. Applications, such as bootloader software, can be protected while allowing other regions of the program memory to be modified.

The WRT<1:0> bits in Configuration Word 2 define the size of the program memory block that is protected.

4.4 User ID

Four memory locations (8000h-8003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are readable and writable during normal execution. See Section 11.5 “User ID, Device ID and Configuration Word Access” for more information on accessing these memory locations. For more information on checksum calculation, see the “PIC12LF1840T48A/T39A Programming Specification” (DS41595).
4.5 Device ID and Revision ID

The memory location 8006h is where the Device ID and Revision ID are stored. The upper nine bits hold the Device ID. The lower five bits hold the Revision ID. See Section 11.5 “User ID, Device ID and Configuration Word Access” for more information on accessing these memory locations.

Development tools, such as device programmers and debuggers, may be used to read the Device ID and Revision ID.

REGISTER 4-3: DEVICEID: DEVICE ID REGISTER(1)

<table>
<thead>
<tr>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
</tr>
</thead>
<tbody>
<tr>
<td>DEV8</td>
<td>DEV7</td>
<td>DEV6</td>
<td>DEV5</td>
<td>DEV4</td>
<td>DEV3</td>
<td>DEV2</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 13</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>bit 7</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
<th>R</th>
</tr>
</thead>
<tbody>
<tr>
<td>DEV1</td>
<td>DEV0</td>
<td>REV4</td>
<td>REV3</td>
<td>REV2</td>
<td>REV1</td>
<td>REV0</td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 6</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>bit 0</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- U = Unimplemented bit, read as ‘0’
- R = Readable bit
- W = Writable bit
- ‘0’ = Bit is cleared
- ‘1’ = Bit is set
- x = Bit is unknown

bit 13-5 DEVID<8:0>: Device ID bits
011011110 = PIC12LF1840T39A

bit 4-0 REV<4:0>: Revision ID bits
These bits are used to identify the revision.

Note 1: This location cannot be written.
5.0 OSCILLATOR MODULE (WITH FAIL-SAFE CLOCK MONITOR)

5.1 Overview

The oscillator module has a wide variety of clock sources and selection features that allow it to be used in a wide range of applications while maximizing performance and minimizing power consumption. Figure 5-1 illustrates a block diagram of the oscillator module.

Clock sources can be supplied from external oscillators, quartz crystal resonators, ceramic resonators and Resistor-Capacitor (RC) circuits. In addition, the system clock source can be supplied from one of two internal oscillators and PLL circuits, with a choice of speeds selectable via software. Additional clock features include:

- Selectable system clock source between external or internal sources via software.
- Two-Speed Start-up mode, which minimizes latency between external oscillator start-up and code execution.
- Fail-Safe Clock Monitor (FSCM) designed to detect a failure of the external clock source (LP, XT, HS, EC or RC modes) and switch automatically to the internal oscillator.
- Oscillator Start-up Timer (OST) ensures stability of crystal oscillator sources.

The oscillator module can be configured in one of eight clock modes.

1. ECL – External Clock Low-Power mode (0 MHz to 0.5 MHz)
2. ECM – External Clock Medium-Power mode (0.5 MHz to 4 MHz)
3. ECH – External Clock High-Power mode (4 MHz to 32 MHz)
4. LP – 32 kHz Low-Power Crystal mode.
5. XT – Medium Gain Crystal or Ceramic Resonator Oscillator mode (up to 4 MHz)
6. HS – High Gain Crystal or Ceramic Resonator mode (4 MHz to 20 MHz)
8. INTOSC – Internal oscillator (31 kHz to 32 MHz).

Clock source modes are selected by the FOSC<2:0> bits in the Configuration Word 1. The FOSC bits determine the type of oscillator that will be used when the device is first powered.

The EC clock mode relies on an external logic level signal as the device clock source. The LP, XT and HS clock modes require an external crystal or resonator to be connected to the device. Each mode is optimized for a different frequency range. The RC clock mode requires an external resistor and capacitor to set the oscillator frequency.

The INTOSC internal oscillator block produces low, medium, and high-frequency clock sources, designated LFINTOSC, MFINTOSC, and HFINTOSC. (see Internal Oscillator Block, Figure 5-1). A wide selection of device clock frequencies may be derived from these three clock sources.
FIGURE 5-1: SIMPLIFIED PIC® MCU CLOCK SOURCE BLOCK DIAGRAM

- Oscillator T1OSC
- Oscillator T1OSO
- Oscillator T1OSI
- Oscillator OSC1
- Oscillator OSC2
- Oscillator OSC

- Internal Oscillator Block
  - HFPLL
  - 500 kHz Source
  - 31 kHz Source
  - 500 kHz Source (HFINTOSC)
  - 31 kHz Source (LFINTOSC)

- Postscaler
- MUX
- IRCF<3:0>
- FOSC<2:0> = 100
- T1OSC
- 4xPLL

- LP, XT, HS, RC, EC
- Sleep
- CPU and Peripherals

- Clock Control
- Clock Source Option
- WDT, PWRT, Fail-Safe Clock Monitor
- Two-Speed Start-up and other modules
5.2 Clock Source Types

Clock sources can be classified as external or internal. External clock sources rely on external circuitry for the clock source to function. Examples are: oscillator modules (EC mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (RC) mode circuits.

Internal clock sources are contained internally within the oscillator module. The internal oscillator block has two internal oscillators and a dedicated Phase-Locked Loop (HFPLL) that are used to generate three internal system clock sources: the 16 MHz High-Frequency Internal Oscillator (HFINTOSC), 500 kHz (MFINTOSC) and the 31 kHz Low-Frequency Internal Oscillator (LFINTOSC).

The system clock can be selected between external or internal clock sources via the System Clock Select (SCS) bits in the OSCCON register. See Section 5.3 “Clock Switching” for additional information.

5.2.1 EXTERNAL CLOCK SOURCES

An external clock source can be used as the device system clock by performing one of the following actions:

- Program the FOSC<2:0> bits in the Configuration Word 1 to select an external clock source that will be used as the default system clock upon a device Reset.
- Write the SCS<1:0> bits in the OSCCON register to switch the system clock source to:
  - Timer1 Oscillator during run-time, or
  - An external clock source determined by the value of the FOSC bits.

See Section 5.3 “Clock Switching” for more information.

5.2.1.1 EC Mode

The External Clock (EC) mode allows an externally generated logic level signal to be the system clock source. When operating in this mode, an external clock source is connected to the OSC1 input. OSC2/CLKOUT is available for general purpose I/O or CLKOUT. Figure 5-2 shows the pin connections for EC mode.

EC mode has three power modes to select from through Configuration Word 1:

- High power, 4-32 MHz (FOSC = 111)
- Medium power, 0.5-4 MHz (FOSC = 110)
- Low power, 0-0.5 MHz (FOSC = 101)

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC® MCU design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.

5.2.1.2 LP, XT, HS Modes

The LP, XT and HS modes support the use of quartz crystal resonators or ceramic resonators connected to OSC1 and OSC2 (Figure 5-3). The three modes select a low, medium or high gain setting of the internal inverter-amplifier to support various resonator types and speed.

1. **LP** Oscillator mode selects the lowest gain setting of the internal inverter-amplifier. LP mode current consumption is the least of the three modes. This mode is designed to drive only 32.768 kHz tuning-fork type crystals (watch crystals).

2. **XT** Oscillator mode selects the intermediate gain setting of the internal inverter-amplifier. XT mode current consumption is the medium of the three modes. This mode is best suited for resonators with a medium drive level specification.

3. **HS** Oscillator mode selects the highest gain setting of the internal inverter-amplifier. HS mode current consumption is the highest of the three modes. This mode is best suited for resonators that require a high drive setting.

Figure 5-3 and Figure 5-4 show typical circuits for quartz crystal and ceramic resonators, respectively.
5.2.1.3 Oscillator Start-up Timer (OST)

If the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the oscillator module.

In order to minimize latency between external oscillator start-up and code execution, the Two-Speed Clock Start-up mode can be selected (see Section 5.4 “Two-Speed Clock Start-up Mode”).

5.2.1.4 4xPLL

The oscillator module contains a 4xPLL that can be used with both external and internal clock sources to provide a system clock source. The input frequency for the 4xPLL must fall within specifications. See the PLL Clock Timing Specifications in Section 31.0 “Electrical Specifications”.

The 4xPLL may be enabled for use by one of two methods:

1. Program the PLLEN bit in Configuration Word 2 to a ‘1’.
2. Write the SPLLEN bit in the OSCCON register to a ‘1’. If the PLLEN bit in Configuration Word 2 is programmed to a ‘1’, then the value of SPLLEN is ignored.

Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.

2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

3: For oscillator design assistance, reference the following Microchip Applications Notes:
   • AN826, “Crystal Oscillator Basics and Crystal Selection for rPIC® and PIC® Devices” (DS00826)
   • AN849, “Basic PIC® Oscillator Design” (DS00849)
   • AN943, “Practical PIC® Oscillator Analysis and Design” (DS00943)
   • AN949, “Making Your Oscillator Work” (DS00949)
5.2.1.5 TIMER1 Oscillator

The Timer1 oscillator is a separate crystal oscillator that is associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the T1OSO and T1OSI device pins.

The Timer1 oscillator can be used as an alternate system clock source and can be selected during run-time using clock switching. Refer to Section 5.3 “Clock Switching” for more information.

FIGURE 5-5: QUARTZ CRYSTAL OPERATION (TIMER1 OSCILLATOR)

**Note 1:** Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.

2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

3: For oscillator design assistance, reference the following Microchip Applications Notes:

- AN826, “Crystal Oscillator Basics and Crystal Selection for rPIC® and PIC® Devices” (DS00826)
- AN849, “Basic PIC® Oscillator Design” (DS00849)
- AN943, “Practical PIC® Oscillator Analysis and Design” (DS00943)
- AN949, “Making Your Oscillator Work” (DS00949)
- TB097, “Interfacing a Micro Crystal MS1V-T1K 32.768 kHz Tuning Fork Crystal to a PIC16F690/SS” (DS91097)
- AN1288, “Design Practices for Low-Power External Oscillators” (DS01288)

5.2.1.6 External RC Mode

The external Resistor-Capacitor (RC) modes support the use of an external RC circuit. This allows the designer maximum flexibility in frequency choice while keeping costs to a minimum when clock accuracy is not required.

The RC circuit connects to OSC1. OSC2/CLKOUT is available for general purpose I/O or CLKOUT. The function of the OSC2/CLKOUT pin is determined by the state of the CLKOUTEN bit in Configuration Word 1.

Figure 5-6 shows the external RC mode connections.

FIGURE 5-6: EXTERNAL RC MODES

The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. Other factors affecting the oscillator frequency are:

- Threshold voltage variation
- Component tolerances
- Packaging variations in capacitance

The user also needs to take into account variation due to tolerance of external RC components used.

Note 1: Output depends upon CLKOUTEN bit of the Configuration Word 1.
5.2.2 INTERNAL CLOCK SOURCES

The device may be configured to use the internal oscillator block as the system clock by performing one of the following actions:

- Program the FOSC<2:0> bits in Configuration Word 1 to select the INTOSC clock source, which will be used as the default system clock upon a device Reset.
- Write the SCS<1:0> bits in the OSCCON register to switch the system clock source to the internal oscillator during run-time. See Section 5.3 “Clock Switching” for more information.

In INTOSC mode, OSC1/CLKIN is available for general purpose I/O. OSC2/CLKOUT is available for general purpose I/O or CLKOUT.

The function of the OSC2/CLKOUT pin is determined by the state of the CLKOUTEN bit in Configuration Word 1.

The internal oscillator block has two independent oscillators and a dedicated Phase-Locked Loop, HFPLL that can produce one of three internal system clock sources.

1. The HFINTOSC (High-Frequency Internal Oscillator) is factory calibrated and operates at 16 MHz. The HFINTOSC source is generated from the 500 kHz MFINTOSC source and the dedicated Phase-Locked Loop, HFPLL. The frequency of the HFINTOSC can be user-adjusted via software using the OSCTUNE register (Register 5-3).

2. The MFINTOSC (Medium-Frequency Internal Oscillator) is factory-calibrated and operates at 500 kHz. The frequency of the MFINTOSC can be user-adjusted via software using the OSCTUNE register (Register 5-3).

3. The LFINTOSC (Low-Frequency Internal Oscillator) is uncalibrated and operates at 31 kHz.

5.2.2.1 HFINTOSC

The High-Frequency Internal Oscillator (HFINTOSC) is a factory-calibrated 16 MHz internal clock source. The frequency of the HFINTOSC can be altered via software using the OSCTUNE register (Register 5-3).

The output of the HFINTOSC connects to a postscaler and multiplexer (see Figure 5-1). One of nine frequencies derived from the HFINTOSC can be selected via software using the IRCF<3:0> bits of the OSCCON register. See Section 5.2.2.7 “Internal Oscillator Clock Switch Timing” for more information.

The HFINTOSC is enabled by:

- Configure the IRCF<3:0> bits of the OSCCON register for the desired HF frequency, and
- FOSC<2:0> = 100, or
- Set the System Clock Source (SCS) bits of the OSCCON register to '1x'.

The High-Frequency Internal Oscillator Ready bit (HFIOFR) of the OSCSTAT register indicates when the HFINTOSC is running and can be utilized.

The High-Frequency Internal Oscillator Status Locked bit (HFIOFL) of the OSCSTAT register indicates when the HFINTOSC is running within 2% of its final value.

The High-Frequency Internal Oscillator Status Stable bit (HFIOFS) of the OSCSTAT register indicates when the HFINTOSC is running within 0.5% of its final value.

5.2.2.2 MFINTOSC

The Medium-Frequency Internal Oscillator (MFINTOSC) is a factory-calibrated 500 kHz internal clock source. The frequency of the MFINTOSC can be altered via software using the OSCTUNE register (Register 5-3).

The output of the MFINTOSC connects to a postscaler and multiplexer (see Figure 5-1). One of nine frequencies derived from the MFINTOSC can be selected via software using the IRCF<3:0> bits of the OSCCON register. See Section 5.2.2.7 “Internal Oscillator Clock Switch Timing” for more information.

The MFINTOSC is enabled by:

- Configure the IRCF<3:0> bits of the OSCCON register for the desired HF frequency, and
- FOSC<2:0> = 100, or
- Set the System Clock Source (SCS) bits of the OSCCON register to '1x'.

The Medium-Frequency Internal Oscillator Ready bit (MFIOFR) of the OSCSTAT register indicates when the MFINTOSC is running and can be utilized.
5.2.2.3 Internal Oscillator Frequency Adjustment

The 500 kHz internal oscillator is factory-calibrated. This internal oscillator can be adjusted in software by writing to the OSCTUNE register (Register 5-3). Since the HFINTOSC and MFINTOSC clock sources are derived from the 500 kHz internal oscillator, a change in the OSCTUNE register value will apply to both.

The default value of the OSCTUNE register is '0'. The value is a 6-bit two’s complement number. A value of 1Fh will provide an adjustment to the maximum frequency. A value of 20h will provide an adjustment to the minimum frequency.

When the OSCTUNE register is modified, the oscillator frequency will begin shifting to the new frequency. Code execution continues during this shift. There is no indication that the shift has occurred.

OSCTUNE does not affect the LFINTOSC frequency. Operation of features that depend on the LFINTOSC clock source frequency, such as the Power-up Timer (PWRT), Watchdog Timer (WDT), Fail-Safe Clock Monitor (FSCM) and peripherals, are not affected by the change in frequency.

5.2.2.4 LFINTOSC

The Low-Frequency Internal Oscillator (LFINTOSC) is an uncalibrated 31 kHz internal clock source.

The output of the LFINTOSC connects to a postscaler and multiplexer (see Figure 5-1). Select 31 kHz, via software, using the IRCF<3:0> bits of the OSCCON register. See Section 5.2.2.7 “Internal Oscillator Clock Switch Timing” for more information. The LFINTOSC is also the frequency for the Power-up Timer (PWRT), Watchdog Timer (WDT) and Fail-Safe Clock Monitor (FSCM).

The LFINTOSC is enabled by selecting 31 kHz (IRCF<3:0> bits of the OSCCON register = 000) as the system clock source (SCS bits of the OSCCON register = 1x), or when any of the following are enabled:

- Configure the IRCF<3:0> bits of the OSCCON register for the desired LF frequency, and
- FOSC<2:0> = 100, or
- Set the System Clock Source (SCS) bits of the OSCCON register to '1x'

Peripherals that use the LFINTOSC are:

- Power-up Timer (PWRT)
- Watchdog Timer (WDT)
- Fail-Safe Clock Monitor (FSCM)

The Low-Frequency Internal Oscillator Ready bit (LFIOFR) of the OSCSTAT register indicates when the LFINTOSC is running and can be utilized.

5.2.2.5 Internal Oscillator Frequency Selection

The system clock speed can be selected via software using the Internal Oscillator Frequency Select bits IRCF<3:0> of the OSCCON register.

The output of the 16 MHz HFINTOSC and 31 kHz LFINTOSC connects to a postscaler and multiplexer (see Figure 5-1). The Internal Oscillator Frequency Select bits IRCF<3:0> of the OSCCON register select the frequency output of the internal oscillators. One of the following frequencies can be selected via software:

- 32 MHz (requires 4xPLL)
- 16 MHz
- 8 MHz
- 4 MHz
- 2 MHz
- 1 MHz
- 500 kHz (default after Reset)
- 250 kHz
- 125 kHz
- 62.5 kHz
- 31.25 kHz
- 31 kHz (LFINTOSC)

Note: Following any Reset, the IRCF<3:0> bits of the OSCCON register are set to '0111' and the frequency selection is set to 500 kHz. The user can modify the IRCF bits to select a different frequency.

The IRCF<3:0> bits of the OSCCON register allow duplicate selections for some frequencies. These duplicate choices can offer system design trade-offs. Lower-power consumption can be obtained when changing oscillator sources for a given frequency. Faster transition times can be obtained between frequency changes that use the same oscillator source.
5.2.2.6 32 MHz Internal Oscillator Frequency Selection

The Internal Oscillator Block can be used with the 4xPLL associated with the External Oscillator Block to produce a 32 MHz internal system clock source. The following settings are required to use the 32 MHz internal clock source:

• The FOSC bits in Configuration Word 1 must be set to use the INTOSC source as the device system clock (FOSC<2:0> = 100).
• The SCS bits in the OSCCON register must be cleared to use the clock determined by FOSC<2:0> in Configuration Word 1 (SCS<1:0> = 00).
• The IRCF bits in the OSCCON register must be set to the 8 MHz HFINTOSC set to use (IRCF<3:0> = 1110).
• The SPLLEN bit in the OSCCON register must be set to enable the 4xPLL, or the PLLEN bit of the Configuration Word 2 must be programmed to a ‘1’.

Note: When using the PLLEN bit of the Configuration Word 2, the 4xPLL cannot be disabled by software and the 8 MHz HFINTOSC option will no longer be available.

The 4xPLL is not available for use with the internal oscillator when the SCS bits of the OSCCON register are set to ‘1x’. The SCS bits must be set to ‘00’ to use the 4xPLL with the internal oscillator.

5.2.2.7 Internal Oscillator Clock Switch Timing

When switching between the HFINTOSC, MFINTOSC and the LFINTOSC, the new oscillator may already be shut down to save power (see Figure 5-7). If this is the case, there is a delay after the IRCF<3:0> bits of the OSCCON register are modified before the frequency selection takes place. The OSCSTAT register will reflect the current active status of the HFINTOSC, MFINTOSC and LFINTOSC oscillators. The sequence of a frequency selection is as follows:

1. IRCF<3:0> bits of the OSCCON register are modified.
2. If the new clock is shut down, a clock start-up delay is started.
3. Clock switch circuitry waits for a falling edge of the current clock.
4. The current clock is held low and the clock switch circuitry waits for a rising edge in the new clock.
5. The new clock is now active.
6. The OSCSTAT register is updated as required.
7. Clock switch is complete.

See Figure 5-7 for more details.

If the internal oscillator speed is switched between two clocks of the same source, there is no start-up delay before the new frequency is selected. Clock switching time delays are shown in Table 5-1.

Start-up delay specifications are located in the oscillator tables of Section 31.0 “Electrical Specifications”
FIGURE 5-7: INTERNAL OSCILLATOR SWITCH TIMING

HFINTOSC/→ LFINTOSC (FSCM and WDT disabled)

HFINTOSC/ MFINTOSC

HFINTOSC/ MFINTOSC

LFINTOSC

IRCF <3:0> ≠ 0 = 0

System Clock

HFINTOSC/→ LFINTOSC (Either FSCM or WDT enabled)

HFINTOSC/ MFINTOSC

HFINTOSC/ MFINTOSC

LFINTOSC

IRCF <3:0> ≠ 0 = 0

System Clock

LFINTOSC → HFINTOSC/MFINTOSC

LFINTOSC

HFINTOSC/ MFINTOSC

IRCF <3:0> = 0 ≠ 0

System Clock

LFINTOSC turns off unless WDT or FSCM is enabled
5.3 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the System Clock Select (SCS) bits of the OSCCON register. The following clock sources can be selected using the SCS bits:

- Default system oscillator determined by FOSC bits in Configuration Word 1
- Timer1 32 kHz crystal oscillator
- Internal Oscillator Block (INTOSC)

5.3.1 SYSTEM CLOCK SELECT (SCS) BITS

The System Clock Select (SCS) bits of the OSCCON register selects the system clock source that is used for the CPU and peripherals.

- When the SCS bits of the OSCCON register = 00, the system clock source is determined by value of the FOSC<2:0> bits in the Configuration Word 1.
- When the SCS bits of the OSCCON register = 01, the system clock source is the Timer1 oscillator.
- When the SCS bits of the OSCCON register = 1x, the system clock source is chosen by the internal oscillator frequency selected by the IRCF<3:0> bits of the OSCCON register. After a Reset, the SCS bits of the OSCCON register are always cleared.

Note: Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bits of the OSCCON register. The user can monitor the OSTS bit of the OSCSTAT register to determine the current system clock source.

When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 5-1.

5.3.2 OSCILLATOR START-UP TIME-OUT STATUS (OSTS) BIT

The Oscillator Start-up Time-out Status (OSTS) bit of the OSCSTAT register indicates whether the system clock is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Word 1, or from the internal clock source. In particular, OSTS indicates that the Oscillator Start-up Timer (OST) has timed out for LP, XT or HS modes. The OST does not reflect the status of the Timer1 oscillator.

5.3.3 TIMER1 OSCILLATOR

The Timer1 oscillator is a separate crystal oscillator associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the T1OSO and T1OSI device pins.

The Timer1 oscillator is enabled using the T1OSCEN control bit in the T1CON register. See Section 21.0 “Timer1 Module with Gate Control” for more information about the Timer1 peripheral.

5.3.4 TIMER1 OSCILLATOR READY (T1OSCR) BIT

The user must ensure that the Timer1 oscillator is ready to be used before it is selected as a system clock source. The Timer1 Oscillator Ready (T1OSCR) bit of the OSCSTAT register indicates whether the Timer1 oscillator is ready to be used. After the T1OSCR bit is set, the SCS bits can be configured to select the Timer1 oscillator.

Note:

Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bits of the OSCCON register. The user can monitor the OSTS bit of the OSCSTAT register to determine the current system clock source.
5.4 Two-Speed Clock Start-up Mode

Two-Speed Start-up mode provides additional power savings by minimizing the latency between external oscillator start-up and code execution. In applications that make heavy use of the Sleep mode, Two-Speed Start-up will remove the external oscillator start-up time from the time spent awake and can reduce the overall power consumption of the device. This mode allows the application to wake-up from Sleep, perform a few instructions using the INTOSC Internal Oscillator block as the clock source and go back to Sleep without waiting for the external oscillator to become stable.

Two-Speed Start-up provides benefits when the oscillator module is configured for LP, XT or HS modes. The Oscillator Start-up Timer (OST) is enabled for these modes and must count 1024 oscillations before the oscillator can be used as the system clock source.

If the oscillator module is configured for any mode other than LP, XT or HS mode, then Two-Speed Start-up is disabled. This is because the external clock oscillator does not require any stabilization time after POR or an exit from Sleep.

If the OST count reaches 1024 before the device enters Sleep mode, the OSTS bit of the OSCSTAT register is set and program execution switches to the external oscillator. However, the system may never operate from the external oscillator if the time spent awake is very short.

### 5.4.1 TWO-SPEED START-UP MODE CONFIGURATION

Two-Speed Start-up mode is configured by the following settings:

- IESO (of the Configuration Word 1) = 1; Internal/External Switchover bit (Two-Speed Start-up mode enabled).
- SCS (of the OSCCON register) = 00.
- FOSC<2:0> bits in the Configuration Word 1 configured for LP, XT or HS mode.

Two-Speed Start-up mode is entered after:

- Power-on Reset (POR) and, if enabled, after Power-up Timer (PWRT) has expired, or
- Wake-up from Sleep.

<table>
<thead>
<tr>
<th>Switch From</th>
<th>Switch To</th>
<th>Frequency</th>
<th>Oscillator Delay</th>
</tr>
</thead>
<tbody>
<tr>
<td>Sleep/POR</td>
<td>LFINTOSC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>31 kHz</td>
<td>Oscillator Warm-up Delay (TWARM)</td>
</tr>
<tr>
<td></td>
<td>MFINTOSC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>31.25 kHz - 500 kHz</td>
<td></td>
</tr>
<tr>
<td></td>
<td>HFINTOSC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>31.25 kHz - 16 MHz</td>
<td></td>
</tr>
<tr>
<td>Sleep/POR</td>
<td>EC, RC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>DC – 32 MHz</td>
<td>2 cycles</td>
</tr>
<tr>
<td>LFINTOSC</td>
<td>EC, RC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>DC – 32 MHz</td>
<td>1 cycle of each</td>
</tr>
<tr>
<td>Sleep/POR</td>
<td>Timer1 Oscillator</td>
<td>32 kHz - 20 MHz</td>
<td>1024 Clock Cycles (OST)</td>
</tr>
<tr>
<td></td>
<td>LP, XT, HS&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Any clock source</td>
<td>MFINTOSC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>31.25 kHz - 500 kHz</td>
<td>2 μs (approx.)</td>
</tr>
<tr>
<td>Any clock source</td>
<td>HFINTOSC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>31.25 kHz - 16 MHz</td>
<td></td>
</tr>
<tr>
<td>Any clock source</td>
<td>LFINTOSC&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>31 kHz</td>
<td>1 cycle of each</td>
</tr>
<tr>
<td>Any clock source</td>
<td>Timer1 Oscillator</td>
<td>32 kHz</td>
<td>1024 Clock Cycles (OST)</td>
</tr>
<tr>
<td>PLL inactive</td>
<td>PLL active</td>
<td>16 - 32 MHz</td>
<td>2 ms (approx.)</td>
</tr>
</tbody>
</table>

**Note:** PLL inactive.

---

Note: 
Executing a SLEEP instruction will abort the oscillator start-up time and will cause the OSTS bit of the OSCSTAT register to remain clear.
5.4.2 TWO-SPEED START-UP SEQUENCE

1. Wake-up from Power-on Reset or Sleep.
2. Instructions begin execution by the internal oscillator at the frequency set in the IRCF<3:0> bits of the OSCCON register.
3. OST enabled to count 1024 clock cycles.
4. OST timed out, wait for falling edge of the internal oscillator.
5. OSTS is set.
6. System clock held low until the next falling edge of new clock (LP, XT or HS mode).
7. System clock is switched to external clock source.

5.4.3 CHECKING TWO-SPEED CLOCK STATUS

Checking the state of the OSTS bit of the OSCSTAT register will confirm if the microcontroller is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Word 1, or the internal oscillator.

FIGURE 5-8: TWO-SPEED START-UP
5.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the device to continue operating should the external oscillator fail. The FSCM can detect oscillator failure any time after the Oscillator Start-up Timer (OST) has expired. The FSCM is enabled by setting the FCMEN bit in the Configuration Word 1. The FSCM is applicable to all external Oscillator modes (LP, XT, HS, EC, Timer1 Oscillator and RC).

**FIGURE 5-9: FSCM BLOCK DIAGRAM**

<table>
<thead>
<tr>
<th>External Clock</th>
<th>LFINTOSC Oscillator</th>
</tr>
</thead>
<tbody>
<tr>
<td>+ 64</td>
<td>Sample Clock</td>
</tr>
<tr>
<td></td>
<td>Clock Monitor Latch</td>
</tr>
<tr>
<td></td>
<td>Q</td>
</tr>
<tr>
<td></td>
<td>R</td>
</tr>
<tr>
<td></td>
<td>S</td>
</tr>
</tbody>
</table>

5.5.1 FAIL-SAFE DETECTION

The FSCM module detects a failed oscillator by comparing the external oscillator to the FSCM sample clock. The sample clock is generated by dividing the LFINTOSC by 64. See Figure 5-9. Inside the fail detector block is a latch. The external clock sets the latch on each falling edge of the external clock. The sample clock clears the latch on each rising edge of the sample clock. A failure is detected when an entire half-cycle of the sample clock elapses before the external clock goes low.

5.5.2 FAIL-SAFE OPERATION

When the external clock fails, the FSCM switches the device clock to an internal clock source and sets the bit flag OSFIF of the PIR2 register. Setting this flag will generate an interrupt if the OSFIE bit of the PIE2 register is also set. The device firmware can then take steps to mitigate the problems that may arise from a failed clock. The system clock will continue to be sourced from the internal clock source until the device firmware successfully restarts the external oscillator and switches back to external operation.

The internal clock source chosen by the FSCM is determined by the IRCF<3:0> bits of the OSCCON register. This allows the internal oscillator to be configured before a failure occurs.

5.5.3 FAIL-SAFE CONDITION CLEARING

The Fail-Safe condition is cleared after a Reset, executing a SLEEP instruction or changing the SCS bits of the OSCCON register. When the SCS bits are changed, the OST is restarted. While the OST is running, the device continues to operate from the INTOSC selected in OSCCON. When the OST times out, the Fail-Safe condition is cleared and the device will be operating from the external clock source. The Fail-Safe condition must be cleared before the OSFIF flag can be cleared.

5.5.4 RESET OR WAKE-UP FROM SLEEP

The FSCM is designed to detect an oscillator failure after the Oscillator Start-up Timer (OST) has expired. The OST is used after waking up from Sleep and after any type of Reset. The OST is not used with the EC or RC Clock modes so that the FSCM will be active as soon as the Reset or wake-up has completed. When the FSCM is enabled, the Two-Speed Start-up is also enabled. Therefore, the device will always be executing code while the OST is operating.

**Note:** Due to the wide range of oscillator start-up times, the Fail-Safe circuit is not active during oscillator start-up (i.e., after exiting Reset or Sleep). After an appropriate amount of time, the user should check the Status bits in the OSCSTAT register to verify the oscillator start-up and that the system clock switchover has successfully completed.
FIGURE 5-10: FSCM TIMING DIAGRAM

Note: The system clock is normally at a much higher frequency than the sample clock. The relative frequencies in this example have been chosen for clarity.
5.6 Oscillator Control Registers

REGISTER 5-1: OSCCON: OSCILLATOR CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>SPLLEN</td>
<td>IRCF&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
<td>SCS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- '1' = Bit is set
- '0' = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

**bit 7** SPLLEN: Software PLL Enable bit
- If PLLEN in Configuration Word 1 = 1:
  - SPLLEN bit is ignored. 4xPLL is always enabled (subject to oscillator requirements)
- If PLLEN in Configuration Word 1 = 0:
  - 1 = 4xPLL is enabled
  - 0 = 4xPLL is disabled

**bit 6-3** IRCF<3:0>: Internal Oscillator Frequency Select bits
- 000x = 31 kHz LF
- 0010 = 31.25 kHz MF
- 0011 = 31.25 kHz HF(1)
- 0100 = 62.5 kHz MF
- 0101 = 125 kHz MF
- 0110 = 250 kHz MF
- 0111 = 500 kHz MF (default upon Reset)
- 1000 = 125 kHz HF(1)
- 1001 = 250 kHz HF(1)
- 1010 = 500 kHz HF(1)
- 1011 = 1 MHz HF
- 1100 = 2 MHz HF
- 1101 = 4 MHz HF
- 1110 = 8 MHz or 32 MHz HF (see Section 5.2.2.1 “HFINTOSC”)
- 1111 = 16 MHz HF

**bit 2** Unimplemented: Read as ‘0’

**bit 1-0** SCS<1:0>: System Clock Select bits
- 1x = Internal oscillator block
- 01 = Timer1 oscillator
- 00 = Clock determined by FOSC<2:0> in Configuration Word 1.

**Note 1:** Duplicate frequency derived from HFINTOSC.
REGISTER 5-2:  OSCSTAT: OSCILLATOR STATUS REGISTER

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>T1OSCR</td>
<td>PLLR</td>
<td>OSTS</td>
<td>HFIOSR</td>
<td>HFIOSL</td>
<td>MFIOSR</td>
<td>LFIOSR</td>
<td>HFIOSF</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared
- q = Conditional

bit 7  
T1OSCR: Timer1 Oscillator Ready bit
If T1OSCEN = 1:
1 = Timer1 oscillator is ready
0 = Timer1 oscillator is not ready
If T1OSCEN = 0:
1 = Timer1 clock source is always ready

bit 6  
PLLR 4xPLL Ready bit
1 = 4xPLL is ready
0 = 4xPLL is not ready

bit 5  
OSTS: Oscillator Start-up Time-out Status bit
1 = Running from the clock defined by the FOSC<2:0> bits of the Configuration Word 1
0 = Running from an internal oscillator (FOSC<2:0> = 100)

bit 4  
HFIOSR: High-Frequency Internal Oscillator Ready bit
1 = HFINTOSC is ready
0 = HFINTOSC is not ready

bit 3  
HFIOSL: High-Frequency Internal Oscillator Locked bit
1 = HFINTOSC is at least 2% accurate
0 = HFINTOSC is not 2% accurate

bit 2  
MFIOSR: Medium-Frequency Internal Oscillator Ready bit
1 = MFINTOSC is ready
0 = MFINTOSC is not ready

bit 1  
LFIOSR: Low-Frequency Internal Oscillator Ready bit
1 = LFINTOSC is ready
0 = LFINTOSC is not ready

bit 0  
HFIOSF: High-Frequency Internal Oscillator Stable bit
1 = HFINTOSC is at least 0.5% accurate
0 = HFINTOSC is not 0.5% accurate
### REGISTER 5-3: OSCTUNE: OSCILLATOR TUNING REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>TUN&lt;5:0&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **’1’** = Bit is set
- **’0’** = Bit is cleared

- **bit 7-6** Unimplemented: Read as ‘0’
- **bit 5-0** **TUN<5:0>: Frequency Tuning bits**
  - 011111 = Maximum frequency
  - 011110 = ...
  - ...
  - 000001 = 000000 = Oscillator module is running at the factory-calibrated frequency.
  - 111111 = ...
  - ...
  - 100000 = Minimum frequency

### TABLE 5-2: SUMMARY OF REGISTERS ASSOCIATED WITH CLOCK SOURCES

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>OSCCON</td>
<td>SPLLEN</td>
<td>—</td>
<td>IRCF&lt;3:0&gt;</td>
<td>—</td>
<td>SCS&lt;1:0&gt;</td>
<td>57</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>OSCSTAT</td>
<td>T1OSCR</td>
<td>PLLR</td>
<td>OSTS</td>
<td>HFIOR</td>
<td>HFIORL</td>
<td>MFIOFR</td>
<td>LFIOFR</td>
<td>HFIOFS</td>
<td>58</td>
</tr>
<tr>
<td>OSCTUNE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>59</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>—</td>
<td>C1IE</td>
<td>EEIE</td>
<td>BCL1IE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>79</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>—</td>
<td>C1IF</td>
<td>EEIF</td>
<td>BCL1IF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>81</td>
</tr>
<tr>
<td>T1CON</td>
<td>TMR1CS&lt;1:0&gt;</td>
<td>—</td>
<td>T1CKPS&lt;1:0&gt;</td>
<td>T1OSCEN</td>
<td>T1SYNC</td>
<td>—</td>
<td>TMR1ON</td>
<td>157</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used by clock sources.

### TABLE 5-3: SUMMARY OF CONFIGURATION WORD WITH CLOCK SOURCES

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit -7</th>
<th>Bit -6</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG1</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>FCEN</td>
<td>IESO</td>
<td>CKOUTEN</td>
<td>BOREN&lt;1:0&gt;</td>
<td>CPD</td>
<td>39</td>
<td></td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>CP</td>
<td>MCLRE</td>
<td>PWRT</td>
<td>WDTE&lt;1:0&gt;</td>
<td>FOSC&lt;2:0&gt;</td>
<td>—</td>
<td>—</td>
<td>39</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ‘0’. Shaded cells are not used by clock sources.
6.0 REFERENCE CLOCK MODULE

The reference clock module provides the ability to send a divided clock to the clock output pin of the device (CLKR) and provide a secondary internal clock source to the modulator module. This module is available in all oscillator configurations and allows the user to select a greater range of clock submultiples to drive external devices in the application. The reference clock module includes the following features:

- System clock is the source
- Available in all oscillator configurations
- Programmable clock divider
- Output enable to a port pin
- Selectable duty cycle
- Slew rate control

The reference clock module is controlled by the CLKRCON register (Register 6-1) and is enabled when setting the CLKREN bit. To output the divided clock signal to the CLKR port pin, the CLKROE bit must be set. The CLKRDIV<2:0> bits enable the selection of eight different clock divider options. The CLKRDC<1:0> bits can be used to modify the duty cycle of the output clock. The CLKRSLR bit controls slew rate limiting.

Note 1: If the base clock rate is selected without a divider, the output clock will always have a duty cycle equal to that of the source clock, unless a 0% duty cycle is selected. If the clock divider is set to base clock/2, then 25% and 75% duty cycle accuracy will be dependent upon the source clock.

For information on using the reference clock output with the modulator module, see Section 23.0 “Data Signal Modulator”.

6.1 Slew rate

The slew rate limitation on the output port pin can be disabled. The Slew Rate limitation can be removed by clearing the CLKRSLR bit in the CLKRCON register.

6.2 Effects of a Reset

Upon any device Reset, the reference clock module is disabled. The user’s firmware is responsible for initializing the module before enabling the output. The registers are reset to their default values.

6.3 Conflicts with the CLKR pin

There are two cases when the reference clock output signal cannot be output to the CLKR pin, if:

- LP, XT or HS Oscillator mode is selected.
- CLKOUT function is enabled.

Even if either of these cases are true, the module can still be enabled and the reference clock signal may be used in conjunction with the modulator module.

6.3.1 OSCILLATOR MODES

If LP, XT or HS Oscillator modes are selected, the OSC2/CLKR pin must be used as an oscillator input pin and the CLKR output cannot be enabled. See Section 5.2 “Clock Source Types” for more information on different oscillator modes.

6.3.2 CLKOUT FUNCTION

The CLKOUT function has a higher priority than the reference clock module. Therefore, if the CLKOUT function is enabled by the CLKOUTEN bit in Configuration Word 1, Fosc/4 will always be output on the port pin. Reference Section 4.0 “Device Configuration” for more information.

6.4 Operation During Sleep

As the reference clock module relies on the system clock as its source, and the system clock is disabled in Sleep, the module does not function in Sleep, even if an external clock source or the Timer1 clock source is configured as the system clock. The module outputs will remain in their current state until the device exits Sleep.

Note 1: If the base clock rate is selected without a divider, the output clock will always have a duty cycle equal to that of the source clock, unless a 0% duty cycle is selected. If the clock divider is set to base clock/2, then 25% and 75% duty cycle accuracy will be dependent upon the source clock.
## REGISTER 6-1: CLKRCON: REFERENCE CLOCK CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLKREN</td>
<td>CLKROE</td>
<td>CLKRSLR</td>
<td>CLKRD1</td>
<td>CLKRDC0</td>
<td>CLKRDIV2</td>
<td>CLKRDIV1</td>
<td>CLKRDIV0</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4-3</th>
<th>bit 2-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLKREN: Reference Clock Module Enable bit&lt;br&gt;1 = Reference clock module is enabled&lt;br&gt;0 = Reference clock module is disabled</td>
<td>CLKROE: Reference Clock Output Enable Enable bit(3)&lt;br&gt;1 = Reference clock output is enabled on CLKR pin&lt;br&gt;0 = Reference clock output disabled on CLKR pin</td>
<td>CLKRSLR: Reference Clock Slew Rate Control Limiting Enable bit&lt;br&gt;1 = Slew rate limiting is enabled&lt;br&gt;0 = Slew rate limiting is disabled</td>
<td>CLKRDC&lt;1:0&gt;: Reference Clock Duty Cycle bits&lt;br&gt;11 = Clock outputs duty cycle of 75%&lt;br&gt;10 = Clock outputs duty cycle of 50%&lt;br&gt;01 = Clock outputs duty cycle of 25%&lt;br&gt;00 = Clock outputs duty cycle of 0%</td>
<td>CLKRDIV&lt;2:0&gt;: Reference Clock Divider bits&lt;br&gt;111 = Base clock value divided by 128&lt;br&gt;110 = Base clock value divided by 64&lt;br&gt;101 = Base clock value divided by 32&lt;br&gt;100 = Base clock value divided by 16&lt;br&gt;011 = Base clock value divided by 8&lt;br&gt;010 = Base clock value divided by 4&lt;br&gt;001 = Base clock value divided by 2(1)&lt;br&gt;000 = Base clock value(2)</td>
</tr>
</tbody>
</table>

### Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

### Notes:
1. In this mode, the 25% and 75% duty cycle accuracy will be dependent on the source clock duty cycle.
2. In this mode, the duty cycle will always be equal to the source clock duty cycle, unless a duty cycle of 0% is selected.
3. To route CLKR to pin, CLKOUTEN of Configuration Word 1 = 1 is required. CLKOUTEN of Configuration Word 1 = 0 will result in FOSC/4. See Section 6.3 “Conflicts with the CLKR pin” for details.
### TABLE 6-1: SUMMARY OF REGISTERS ASSOCIATED WITH REFERENCE CLOCK SOURCES

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLKCON</td>
<td>CLKREN</td>
<td>CLKROE</td>
<td>CLKRSLR</td>
<td>CLKRDC1</td>
<td>CLKRDC0</td>
<td>CLKRDC2</td>
<td>CLKRDC1</td>
<td>CLKRDC0</td>
<td>61</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented locations read as ‘0’. Shaded cells are not used by reference clock sources.

### TABLE 6-2: SUMMARY OF CONFIGURATION WORD WITH REFERENCE CLOCK SOURCES

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit -7</th>
<th>Bit -6</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG1</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>FCMEN</td>
<td>IESO</td>
<td>CLKOUTEN</td>
<td>BOREN1</td>
<td>BOREN0</td>
<td>CPD</td>
<td>39</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented locations read as ‘0’. Shaded cells are not used by reference clock sources.
7.0 RESETS

There are multiple ways to reset this device:

• Power-on Reset (POR)
• Brown-out Reset (BOR)
• MCLR Reset
• WDT Reset
• RESET instruction
• Stack Overflow
• Stack Underflow
• Programming mode exit

To allow Vdd to stabilize, an optional Power-up Timer can be enabled to extend the Reset time after a BOR or POR event.

A simplified block diagram of the on-chip Reset circuit is shown in Figure 7-1.

**FIGURE 7-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT**
7.1 Power-on Reset (POR)

The POR circuit holds the device in Reset until VDD has reached an acceptable level for minimum operation. Slow rising VDD, fast operating speeds or analog performance may require greater than minimum VDD. The PWRT, BOR or MCLR features can be used to extend the start-up period until all device operation conditions have been met.

7.1.1 POWER-UP TIMER (PWRT)

The Power-up Timer provides a nominal 64 ms time-out on POR or Brown-out Reset.

The device is held in Reset as long as PWRT is active. The PWRT delay allows additional time for the VDD to rise to an acceptable level. The Power-up Timer is enabled by clearing the PWRTEN bit in Configuration Word 1.

The Power-up Timer starts after the release of the POR and BOR.

For additional information, refer to Application Note AN607, “Power-up Trouble Shooting” (DS00607).

7.2 Brown-Out Reset (BOR)

The BOR circuit holds the device in Reset when VDD reaches a selectable minimum level. Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

The Brown-out Reset module has four operating modes controlled by the BOREN<1:0> bits in Configuration Word 1. The four operating modes are:

- BOR is always On
- BOR is off when in Sleep
- BOR is controlled by software
- BOR is always Off

Refer to Table 7-1 for more information.

The Brown-out Reset voltage level is selectable by configuring the BORV bit in Configuration Word 2.

A VDD noise rejection filter prevents the BOR from triggering on small events. If VDD falls below VBOR for a duration greater than parameter TBORDC, the device will reset. See Figure 7-3 for more information.

<table>
<thead>
<tr>
<th>BOREN Config bits</th>
<th>SBOREN</th>
<th>Device Mode</th>
<th>BOR Mode</th>
<th>Device Operation upon release of POR</th>
<th>Device Operation upon wake-up from Sleep</th>
</tr>
</thead>
<tbody>
<tr>
<td>BOR_ON (11)</td>
<td>X</td>
<td>X</td>
<td>Active</td>
<td>Waits for BOR ready(^{(1)})</td>
<td></td>
</tr>
<tr>
<td>BOR_NSLEEP (10)</td>
<td>X</td>
<td>Awake</td>
<td>Active</td>
<td>Waits for BOR ready</td>
<td></td>
</tr>
<tr>
<td>BOR_NSLEEP (10)</td>
<td>X</td>
<td>Sleep</td>
<td>Disabled</td>
<td></td>
<td></td>
</tr>
<tr>
<td>BOR_SBOREN (01)</td>
<td>1</td>
<td>X</td>
<td>Active</td>
<td>Begins immediately</td>
<td></td>
</tr>
<tr>
<td>BOR_SBOREN (01)</td>
<td>0</td>
<td>X</td>
<td>Disabled</td>
<td>Begins immediately</td>
<td></td>
</tr>
<tr>
<td>BOR_OFF (00)</td>
<td>X</td>
<td>X</td>
<td>Disabled</td>
<td>Begins immediately</td>
<td></td>
</tr>
</tbody>
</table>

\(^{(1)}\) Even though this case specifically waits for the BOR, the BOR is already operating, so there is no delay in start-up.

7.2.1 BOR IS ALWAYS ON

When the BOREN bits of Configuration Word 1 are set to ‘11’, the BOR is always On. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is active during Sleep. The BOR does not delay wake-up from Sleep.

7.2.2 BOR IS OFF IN SLEEP

When the BOREN bits of Configuration Word 1 are set to ‘10’, the BOR is On, except in Sleep. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is not active during Sleep. The device wake-up will be delayed until the BOR is ready.

7.2.3 BOR CONTROLLED BY SOFTWARE

When the BOREN bits of Configuration Word 1 are set to ‘01’, the BOR is controlled by the SBOREN bit of the BORCON register. The device start-up is not delayed by the BOR ready condition or the VDD level.

BOR protection begins as soon as the BOR circuit is ready. The status of the BOR circuit is reflected in the BORRDY bit of the BORCON register.

BOR protection is unchanged by Sleep.
FIGURE 7-2: BROWN-OUT READY

SBOREN

BORRDY

TBORRDY

BOR Protection Active

FIGURE 7-3: BROWN-OUT SITUATIONS

VDD

Internal
Reset

VBOR

TPWRT(1)

VDD

Internal
Reset

VBOR

< TPWRT(1)

VDD

Internal
Reset

VBOR

< TPWRT(1)

Note 1: TPWRT delay only if PWRT bit is programmed to '0'.

REGISTER 7-1: BORCON: BROWN-OUT RESET CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-1/u</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R-q/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>SBOREN</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>BORRDY</td>
</tr>
<tr>
<td>bit 7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>bit 0</td>
</tr>
</tbody>
</table>

Legend:

R = Readable bit  
W = Writable bit  
U = Unimplemented bit, read as ‘0’  
u = Bit is unchanged  
x = Bit is unknown  
-n/n = Value at POR and BOR/Value at all other Resets

bit 7  
SBOREN: Software Brown-out Reset Enable bit
If BOREN <1:0> in Configuration Word 1 = 01:
SBOREN is read/write, but has no effect on the BOR.
If BOREN <1:0> in Configuration Word 1 = 01:
1 = BOR Enabled  
0 = BOR Disabled

bit 6-1  
Unimplemented: Read as '0'

bit 0  
BORRDY: Brown-out Reset Circuit Ready Status bit  
1 = The Brown-out Reset circuit is active  
0 = The Brown-out Reset circuit is inactive
7.3 MCLR

The MCLR is an optional external input that can reset the device. The MCLR function is controlled by the MCLRE bit of Configuration Word 1 and the LVP bit of Configuration Word 2 (Table 7-2).

<table>
<thead>
<tr>
<th>MCLRE</th>
<th>LVP</th>
<th>MCLR</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>Disabled</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Enabled</td>
</tr>
<tr>
<td>x</td>
<td>x</td>
<td>Enabled</td>
</tr>
</tbody>
</table>

7.3.1 MCLR ENABLED

When MCLR is enabled and the pin is held low, the device is held in Reset. The MCLR pin is connected to VDD through an internal weak pull-up.

The device has a noise filter in the MCLR Reset path. The filter will detect and ignore small pulses.

Note: A Reset does not drive the MCLR pin low.

7.3.2 MCLR DISABLED

When MCLR is disabled, the pin functions as a general purpose input and the internal weak pull-up is under software control. See Section 12.2 “PORTA Registers” for more information.

7.4 Watchdog Timer (WDT) Reset

The Watchdog Timer generates a Reset if the firmware does not issue a CLRWD instruction within the time-out period. The TO and PD bits in the STATUS register are changed to indicate the WDT Reset. See Section 10.0 “Watchdog Timer (WDT)” for more information.

7.5 RESET Instruction

A RESET instruction will cause a device Reset. The RI bit in the PCON register will be set to ‘0’. See Table 7-4 for default conditions after a RESET instruction has occurred.

7.6 Stack Overflow/Underflow Reset

The device can reset when the Stack Overflows or Underflows. The STKOVF or STKUNF bits of the PCON register indicate the Reset condition. These Resets are enabled by setting the STVREN bit in Configuration Word 2. See Section 3.4.2 “Overflow/Underflow Reset” for more information.

7.7 Programming Mode Exit

Upon exit of Programming mode, the device will behave as if a POR had just occurred.

7.8 Power-Up Timer

The Power-up Timer optionally delays device execution after a BOR or POR event. This timer is typically used to allow VDD to stabilize before allowing the device to start running.

The Power-up Timer is controlled by the PWRTE bit of Configuration Word 1.

7.9 Start-up Sequence

Upon the release of a POR or BOR, the following must occur before the device will begin executing:

1. Power-up Timer runs to completion (if enabled).
2. Oscillator start-up timer runs to completion (if required for oscillator source).
3. MCLR must be released (if enabled).

The total time-out will vary based on oscillator Configuration and Power-up Timer Configuration. See Section 5.0 “Oscillator Module (With Fail-Safe Clock Monitor)” for more information.

The Power-up Timer and oscillator start-up timer run independently of MCLR Reset. If MCLR is kept low long enough, the Power-up Timer and oscillator start-up timer will expire. Upon bringing MCLR high, the device will begin execution immediately (see Figure 7-4). This is useful for testing purposes or to synchronize more than one device operating in parallel.
FIGURE 7-4:  RESET START-UP SEQUENCE

VDD
Internal POR
Power-Up Timer
MCLR
Internal RESET

Oscillator Modes
External Crystal
Oscillator Start-Up Timer
Oscillator
Fosc

Internal Oscillator
Oscillator
Fosc

External Clock (EC)
CLKIN
Fosc
7.10 Determining the Cause of a Reset

Upon any Reset, multiple bits in the STATUS and PCON register are updated to indicate the cause of the Reset. Table 7-3 and Table 7-4 show the Reset conditions of these registers.

TABLE 7-3: RESET STATUS BITS AND THEIR SIGNIFICANCE

<table>
<thead>
<tr>
<th>STKOVF</th>
<th>STKUNF</th>
<th>RMCLR</th>
<th>RI</th>
<th>POR</th>
<th>BOR</th>
<th>TO</th>
<th>PD</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>x</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>x</td>
<td>0</td>
<td>x</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>x</td>
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<td>u</td>
<td>u</td>
<td>u</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Condition</th>
<th>Program Counter</th>
<th>STATUS Register</th>
<th>PCON Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>Power-on Reset</td>
<td>0000h</td>
<td>---1 1000</td>
<td>00-- 110x</td>
</tr>
<tr>
<td>MCLR Reset during normal operation</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>uu-- 0uuu</td>
</tr>
<tr>
<td>MCLR Reset during Sleep</td>
<td>0000h</td>
<td>---1 0uuu</td>
<td>uu-- 0uuu</td>
</tr>
<tr>
<td>WDT Reset</td>
<td>0000h</td>
<td>---0 uuuu</td>
<td>uu-- uuuu</td>
</tr>
<tr>
<td>WDT Wake-up from Sleep</td>
<td>PC + 1</td>
<td>---0 0uuu</td>
<td>uu-- uuuu</td>
</tr>
<tr>
<td>Brown-out Reset</td>
<td>0000h</td>
<td>---1 1uuu</td>
<td>00-- 11u0</td>
</tr>
<tr>
<td>Interrupt Wake-up from Sleep</td>
<td>PC + 1(1)</td>
<td>---1 0uuu</td>
<td>uu-- uuuu</td>
</tr>
<tr>
<td>RESET Instruction Executed</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>uu-- u0uuu</td>
</tr>
<tr>
<td>Stack Overflow Reset (STVREN = 1)</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>1u-- uuuu</td>
</tr>
<tr>
<td>Stack Underflow Reset (STVREN = 1)</td>
<td>0000h</td>
<td>---u uuuu</td>
<td>u1-- uuuu</td>
</tr>
</tbody>
</table>

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

Note 1: When the wake-up is due to an interrupt and Global Enable bit (GIE) is set, the return address is pushed on the stack and PC is loaded with the interrupt vector (0004h) after execution of PC + 1.

2: If a Status bit is not implemented, that bit will be read as '0'.

TABLE 7-4: RESET CONDITION FOR SPECIAL REGISTERS(2)

(2)
7.11 Power Control (PCON) Register

The Power Control (PCON) register contains flag bits to differentiate between a:

- Power-on Reset (POR)
- Brown-out Reset (BOR)
- Reset Instruction Reset (RI)
- Stack Overflow Reset (STKOVF)
- Stack Underflow Reset (STKUNF)
- MCLR Reset (RMCLR)

The PCON register bits are shown in Register 7-2.

**REGISTER 7-2: PCON: POWER CONTROL REGISTER**

<table>
<thead>
<tr>
<th>R/W/HS-0/q</th>
<th>R/W/HS-0/q</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W/HC-1/q</th>
<th>R/W/HC-1/q</th>
<th>R/W/HC-q/u</th>
<th>R/W/HC-q/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>STKOVF</td>
<td>STKUNF</td>
<td>—</td>
<td>—</td>
<td>RMCLR</td>
<td>RI</td>
<td>POR</td>
<td>BOR</td>
</tr>
</tbody>
</table>

**Legend:**

- HC = Bit is cleared by hardware
- HS = Bit is set by hardware
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- °m/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- q = Value depends on condition

**bit 7**

- **STKOVF**: Stack Overflow Flag bit
  - 1 = A Stack Overflow occurred
  - 0 = A Stack Overflow has not occurred or set to ‘0’ by firmware

**bit 6**

- **STKUNF**: Stack Underflow Flag bit
  - 1 = A Stack Underflow occurred
  - 0 = A Stack Underflow has not occurred or set to ‘0’ by firmware

**bit 5-4**

- **Unimplemented**: Read as ‘0’

**bit 3**

- **RMCLR**: MCLR Reset Flag bit
  - 1 = A MCLR Reset has not occurred or set to ‘1’ by firmware
  - 0 = A MCLR Reset has occurred (set to ‘0’ in hardware when a MCLR Reset occurs)

**bit 2**

- **RI**: Reset Instruction Flag bit
  - 1 = A Reset instruction has not been executed or set to ‘1’ by firmware
  - 0 = A Reset instruction has been executed (set to ‘0’ in hardware upon executing a Reset instruction)

**bit 1**

- **POR**: Power-on Reset Status bit
  - 1 = No Power-on Reset occurred
  - 0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)

**bit 0**

- **BOR**: Brown-out Reset Status bit
  - 1 = No Brown-out Reset occurred
  - 0 = A Brown-out Reset occurred (must be set in software after a Power-on Reset or Brown-out Reset occurs)
### TABLE 7-5: SUMMARY OF REGISTERS ASSOCIATED WITH RESETS

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BORCON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>BORRDY 65</td>
</tr>
<tr>
<td>PCON</td>
<td>STKOVF</td>
<td>STKUNF</td>
<td>—</td>
<td>—</td>
<td>RMCLR</td>
<td>RI</td>
<td>POR</td>
<td>BOR</td>
<td>69</td>
</tr>
<tr>
<td>STATUS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC</td>
<td>C 16</td>
</tr>
<tr>
<td>WDTCON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>WDTPS&lt;4:0&gt;</td>
<td>SWDTEN</td>
<td>87</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented bit, reads as ‘0’. Shaded cells are not used by Resets.

**Note 1:** Other (non Power-up) Resets include MCLR Reset and Watchdog Timer Reset during normal operation.
8.0 INTERRUPTS

The interrupt feature allows certain events to preempt normal program flow. Firmware is used to determine the source of the interrupt and act accordingly. Some interrupts can be configured to wake the MCU from Sleep mode.

This chapter contains the following information for Interrupts:

- Operation
- Interrupt Latency
- Interrupts During Sleep
- INT Pin
- Automatic Context Saving

Many peripherals produce interrupts. Refer to the corresponding chapters for details.

A block diagram of the interrupt logic is shown in Figure 8-1 and Figure 8-2.

FIGURE 8-1: INTERRUPT LOGIC
FIGURE 8-2: PERIPHERAL INTERRUPT LOGIC

To Interrupt Logic (Figure 8-1)
8.1 Operation

Interrupts are disabled upon any device Reset. They are enabled by setting the following bits:

- GIE bit of the INTCON register
- Interrupt Enable bit(s) for the specific interrupt event(s)
- PEIE bit of the INTCON register (if the Interrupt Enable bit of the interrupt event is contained in the PIEx register)

The INTCON, PIR1 and PIR2 registers record individual interrupts via interrupt flag bits. Interrupt flag bits will be set, regardless of the status of the GIE, PEIE and individual interrupt enable bits.

The following events happen when an interrupt event occurs while the GIE bit is set:

- Current pre-fetched instruction is flushed
- GIE bit is cleared
- Current Program Counter (PC) is pushed onto the stack
- Critical registers are automatically saved to the shadow registers (See Section 8.5 “Automatic Context Saving”)
- PC is loaded with the interrupt vector 0004h

The firmware within the Interrupt Service Routine (ISR) should determine the source of the interrupt by polling the interrupt flag bits. The interrupt flag bits must be cleared before exiting the ISR to avoid repeated interrupts. Because the GIE bit is cleared, any interrupt that occurs while executing the ISR will be recorded through its interrupt flag, but will not cause the processor to redirect to the interrupt vector.

The RETFIE instruction exits the ISR by popping the previous address from the stack, restoring the saved context from the shadow registers and setting the GIE bit.

For additional information on a specific interrupt's operation, refer to its peripheral chapter.

| Note 1: Individual interrupt flag bits are set, regardless of the state of any other enable bits. |
| Note 2: All interrupts will be ignored while the GIE bit is cleared. Any interrupt occurring while the GIE bit is clear will be serviced when the GIE bit is set again. |

8.2 Interrupt Latency

Interrupt latency is defined as the time from when the interrupt event occurs to the time code execution at the interrupt vector begins. The latency for synchronous interrupts is three or four instruction cycles. For asynchronous interrupts, the latency is three to five instruction cycles, depending on when the interrupt occurs. See Figure 8-3 and Figure 8-4 for more details.
FIGURE 8-3: INTERRUPT LATENCY

OSC1

CLKOUT

Interrupt

GIE

PC

Execute

1 Cycle Instruction at PC

2 Cycle Instruction at PC

3 Cycle Instruction at PC

3 Cycle Instruction at PC

Interrupt Sampled during Q1

1 Cycle Instruction at PC

2 Cycle Instruction at PC

3 Cycle Instruction at PC

NOP

NOP

Inst(0004h)

NOP

NOP

Inst(0004h)

NOP

NOP

Inst(0004h)

NOP

NOP

NOP

Inst(0004h)

NOP

NOP

NOP

Inst(0004h)

NOP

NOP

NOP

Inst(0004h)

Inst(0004h)

Inst(0004h)

Inst(0004h)

Inst(0004h)

Inst(0005h)
FIGURE 8-4: INT PIN INTERRUPT TIMING

<table>
<thead>
<tr>
<th>OSC1</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLKOUT (3)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>INT pin</td>
<td>(1)</td>
<td>(1)</td>
<td></td>
<td></td>
<td>(4)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>INTF</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>(1)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>GIE</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>INSTRUCTION FLOW</td>
<td>PC</td>
<td>PC + 1</td>
<td>PC + 1</td>
<td>0004h</td>
<td>0005h</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Instruction Fetched</td>
<td>Inst (PC)</td>
<td>Inst (PC + 1)</td>
<td>—</td>
<td>Inst (0004h)</td>
<td>Inst (0005h)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Instruction Executed</td>
<td>Inst (PC – 1)</td>
<td>Inst (PC)</td>
<td>Dummy Cycle</td>
<td>Dummy Cycle</td>
<td>Inst (0004h)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Note 1: INTF flag is sampled here (every Q1).
Note 2: Asynchronous interrupt latency = 3-5 TCY. Synchronous latency = 3-4 TCY, where TCY = instruction cycle time. Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.
Note 3: CLKOUT not available in all oscillator modes.
Note 4: For minimum width of INT pulse, refer to AC specifications in Section 31.0 “Electrical Specifications”.
Note 5: INTF is enabled to be set any time during the Q4-Q1 cycles.
8.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR. Refer to the Section 9.0 “Power-Down Mode (Sleep)” for more details.

8.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. This interrupt is enabled by setting the INTE bit of the INTCON register. The INTEDG bit of the OPTION register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector.

8.5 Automatic Context Saving

Upon entering an interrupt, the return PC address is saved on the stack. Additionally, the following registers are automatically saved in the Shadow registers:

- W register
- STATUS register (except for TO and PD)
- BSR register
- FSR registers
- PCLATH register

Upon exiting the Interrupt Service Routine, these registers are automatically restored. Any modifications to these registers during the ISR will be lost. If modifications to any of these registers are desired, the corresponding Shadow register should be modified and the value will be restored when exiting the ISR. The Shadow registers are available in Bank 31 and are readable and writable. Depending on the user’s application, other registers may also need to be saved.
8.5.1 INTCON REGISTER

The INTCON register is a readable and writable register, which contains the various enable and flag bits for TMR0 register overflow, interrupt-on-change and external INT pin interrupts.

### Note:

Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

### REGISTER 8-1: INTCON: INTERRUPT CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
</tr>
<tr>
<td>bit 7</td>
<td>bit 6</td>
<td>bit 5</td>
<td>bit 4</td>
<td>bit 3</td>
<td>bit 2</td>
<td>bit 1</td>
<td>bit 0</td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

- **1** = Enables all active interrupts
- **0** = Disables all interrupts

- **1** = Enables all active peripheral interrupts
- **0** = Disables all peripheral interrupts

- **1** = Enables the Timer0 interrupt
- **0** = Disables the Timer0 interrupt

- **1** = Enables the INT external interrupt
- **0** = Disables the INT external interrupt

- **1** = Enables the interrupt-on-change
- **0** = Disables the interrupt-on-change

- **1** = The INT external interrupt occurred
- **0** = The INT external interrupt did not occur

- **1** = When at least one of the interrupt-on-change pins changed state
- **0** = None of the interrupt-on-change pins have changed state

**Note 1:** The IOCIF Flag bit is read-only and cleared when all the Interrupt-on-Change flags in the IOCAF register have been cleared by software.
8.5.2 PIE1 REGISTER

The PIE1 register contains the interrupt enable bits, as shown in Register 8-2.

**REGISTER 8-2: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
</table>
| 7   | TMR1GIE: Timer1 Gate Interrupt Enable bit | Enables the Timer1 Gate Acquisition interrupt
| 6   | ADIE: A/D Converter (ADC) Interrupt Enable bit | Enables the ADC interrupt
| 5   | RCIE: USART Receive Interrupt Enable bit | Enables the USART receive interrupt
| 4   | TXIE: USART Transmit Interrupt Enable bit | Enables the USART transmit interrupt
| 3   | SSP1IE: Synchronous Serial Port (MSSP) Interrupt Enable bit | Enables the MSSP interrupt
| 2   | CCP1IE: CCP1 Interrupt Enable bit | Enables the CCP1 interrupt
| 1   | TMR2IE: TMR2 to PR2 Match Interrupt Enable bit | Enables the Timer2 to PR2 match interrupt
| 0   | TMR1IE: Timer1 Overflow Interrupt Enable bit | Enables the Timer1 overflow interrupt

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **’1’ =** Bit is set
- **’0’ =** Bit is cleared

**Note:** Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.
8.5.3 PIE2 REGISTER

The PIE2 register contains the interrupt enable bits, as shown in Register 8-3.

**Note:** Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

**REGISTER 8-3: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2**

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>OSFIE</th>
<th>C1IE</th>
<th>EEIE</th>
<th>BCL1IE</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>bit 6</td>
<td>bit 5</td>
<td>bit 4</td>
<td>bit 3</td>
<td>bit 2-0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **1** = Bit is set
- **0** = Bit is cleared

**bit 7**  **OSFIE**: Oscillator Fail Interrupt Enable bit
- 1 = Enables the Oscillator Fail interrupt
- 0 = Disables the Oscillator Fail interrupt

**bit 6**  **Unimplemented**: Read as ‘0’

**bit 5**  **C1IE**: Comparator C1 Interrupt Enable bit
- 1 = Enables the Comparator C1 interrupt
- 0 = Disables the Comparator C1 interrupt

**bit 4**  **EEIE**: EEPROM Write Completion Interrupt Enable bit
- 1 = Enables the EEPROM Write Completion interrupt
- 0 = Disables the EEPROM Write Completion interrupt

**bit 3**  **BCL1IE**: MSSP Bus Collision Interrupt Enable bit
- 1 = Enables the MSSP Bus Collision interrupt
- 0 = Disables the MSSP Bus Collision interrupt

**bit 2-0**  **Unimplemented**: Read as ‘0’
8.5.4 PIR1 REGISTER

The PIR1 register contains the interrupt flag bits, as shown in Register 8-4.

**Note:** Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

### REGISTER 8-4: PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>TMR1GIF</td>
<td>Timer1 Gate Interrupt Flag bit</td>
</tr>
<tr>
<td>6</td>
<td>ADIF</td>
<td>A/D Converter Interrupt Flag bit</td>
</tr>
<tr>
<td>5</td>
<td>RCIF</td>
<td>USART Receive Interrupt Flag bit</td>
</tr>
<tr>
<td>4</td>
<td>TXIF</td>
<td>USART Transmit Interrupt Flag bit</td>
</tr>
<tr>
<td>3</td>
<td>SSP1IF</td>
<td>Synchronous Serial Port (MSSP) Interrupt Flag bit</td>
</tr>
<tr>
<td>2</td>
<td>CCP1IF</td>
<td>CCP1 Interrupt Flag bit</td>
</tr>
<tr>
<td>1</td>
<td>TMR2IF</td>
<td>Timer2 to PR2 Interrupt Flag bit</td>
</tr>
<tr>
<td>0</td>
<td>TMR1IF</td>
<td>Timer1 Overflow Interrupt Flag bit</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets

- **‘1’ =** Bit is set
- **‘0’ =** Bit is cleared

- **1 =** Interrupt is pending
- **0 =** Interrupt is not pending
8.5.5 PIR2 REGISTER

The PIR2 register contains the interrupt flag bits, as shown in Register 8-5.

**Note:** Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

**REGISTER 8-5: PIR2: PERIPHERAL INTERRUPT REQUEST REGISTER 2**

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
</tr>
</thead>
<tbody>
<tr>
<td>OSFIF</td>
<td>—</td>
<td>C1IF</td>
<td>EEIF</td>
<td>BCL1IF</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

- **bit 7** **OSFIF:** Oscillator Fail Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 6** **Unimplemented:** Read as ‘0’

- **bit 5** **C1IF:** Comparator C1 Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 4** **EEIF:** EEPROM Write Completion Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 3** **BCL1IF:** MSSP Bus Collision Interrupt Flag bit
  - 1 = Interrupt is pending
  - 0 = Interrupt is not pending

- **bit 2-0** **Unimplemented:** Read as ‘0’
### TABLE 8-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>OPTION_REG</td>
<td>WPUE</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS2</td>
<td>PS1</td>
<td>PS0</td>
<td>147</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1E</td>
<td>CCP1E</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>—</td>
<td>C1IE</td>
<td>EEIE</td>
<td>BCL1IE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>79</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1F</td>
<td>CCP1F</td>
<td>TMR2F</td>
<td>TMR1F</td>
<td>80</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>—</td>
<td>C1IF</td>
<td>EEIF</td>
<td>BCL1F</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>81</td>
</tr>
</tbody>
</table>

**Legend:**
- — = unimplemented locations read as '0'.
- Shaded cells are not used by Interrupts.
9.0 POWER-DOWN MODE (SLEEP)

The Power-Down mode is entered by executing a SLEEP instruction.

Upon entering Sleep mode, the following conditions exist:
1. WDT will be cleared but keeps running, if enabled for operation during Sleep.
2. PD bit of the STATUS register is cleared.
3. TO bit of the STATUS register is set.
4. CPU clock is disabled.
5. 31 kHz LFINTOSC is unaffected and peripherals that operate from it may continue operation in Sleep.
6. Timer1 oscillator is unaffected and peripherals that operate from it may continue operation in Sleep.
7. ADC is unaffected, if the dedicated FRC clock is selected.
8. Capacitive Sensing oscillator is unaffected.
9. I/O ports maintain the status they had before SLEEP was executed (driving high, low or high-impedance).
10. Resets other than WDT are not affected by Sleep mode.

Refer to individual chapters for more details on peripheral operation during Sleep.

To minimize current consumption, the following conditions should be considered:
• I/O pins should not be floating
• External circuitry sinking current from I/O pins
• Internal circuitry sourcing current from I/O pins
• Current draw from pins with internal weak pull-ups
• Modules using 31 kHz LFINTOSC
• Modules using Timer1 oscillator

I/O pins that are high-impedance inputs should be pulled to VDD or VSS externally to avoid switching currents caused by floating inputs.

Examples of internal circuitry that might be sourcing current include modules such as the DAC and FVR modules. See Section 17.0 “Digital-to-Analog Converter (DAC) Module” and Section 14.0 “Fixed Voltage Reference (FVR)” for more information on these modules.

9.1 Wake-up from Sleep

The device can wake-up from Sleep through one of the following events:
1. External Reset input on MCLR pin, if enabled
2. BOR Reset, if enabled
3. POR Reset
4. Watchdog Timer, if enabled
5. Any external interrupt
6. Interrupts by peripherals capable of running during Sleep (see individual peripheral for more information)

The first three events will cause a device Reset. The last three events are considered a continuation of program execution. To determine whether a device Reset or wake-up event occurred, refer to Section 7.10 “Determining the Cause of a Reset”.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is pre-fetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be enabled. Wake-up will occur regardless of the state of the GIE bit. If the GIE bit is disabled, the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is enabled, the device executes the instruction after the SLEEP instruction, the device will call the Interrupt Service Routine. In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

The WDT is cleared when the device wakes up from Sleep, regardless of the source of wake-up.
9.1.1 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a **SLEEP** instruction
  - **SLEEP** instruction will execute as a **NOP**.
  - WDT and WDT prescaler will not be cleared
  - **TO** bit of the STATUS register will not be set
  - **PD** bit of the STATUS register will not be cleared.

- If the interrupt occurs **during or after** the execution of a **SLEEP** instruction
  - **SLEEP** instruction will be completely executed
  - Device will immediately wake-up from Sleep
  - WDT and WDT prescaler will be cleared
  - **TO** bit of the STATUS register will be set
  - **PD** bit of the STATUS register will be cleared.

Even if the flag bits were checked before executing a **SLEEP** instruction, it may be possible for flag bits to become set before the **SLEEP** instruction completes. To determine whether a **SLEEP** instruction executed, test the **PD** bit. If the **PD** bit is set, the **SLEEP** instruction was executed as a **NOP**.

---

**FIGURE 9-1: WAKE-UP FROM SLEEP THROUGH INTERRUPT**

**TABLE 9-1: SUMMARY OF REGISTERS ASSOCIATED WITH POWER-DOWN MODE**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTCN</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCF</td>
<td>77</td>
</tr>
<tr>
<td>IOCAN</td>
<td></td>
<td></td>
<td>IOCAN4</td>
<td>IOCAN3</td>
<td>IOCAN2</td>
<td>IOCAN1</td>
<td></td>
<td>IACF0</td>
<td>110</td>
</tr>
<tr>
<td>IOCAP</td>
<td></td>
<td></td>
<td>IOCAP4</td>
<td>IOCAP3</td>
<td>IOCAP2</td>
<td>IOCAP1</td>
<td></td>
<td>IACF0</td>
<td>110</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1E</td>
<td>CCP1E</td>
<td>TMR2E</td>
<td>TMR1E</td>
<td>78</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td></td>
<td>C11E</td>
<td>EEIE</td>
<td>BCL1E</td>
<td></td>
<td></td>
<td></td>
<td>79</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>R CIF</td>
<td>TXIF</td>
<td>SSP1F</td>
<td>CCP1F</td>
<td>TMR2F</td>
<td>TMR1F</td>
<td>80</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td></td>
<td>C11F</td>
<td>EEIF</td>
<td>BCL1F</td>
<td></td>
<td></td>
<td></td>
<td>81</td>
</tr>
<tr>
<td>STATUS</td>
<td></td>
<td></td>
<td></td>
<td>TO</td>
<td>PD</td>
<td>Z</td>
<td>DC</td>
<td>C</td>
<td>16</td>
</tr>
<tr>
<td>WDTCON</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>WDTPS4</td>
<td>WDTPS3</td>
<td>WDTPS2</td>
<td>WDTPS1</td>
<td>WDTPS0</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented, read as ‘0’. Shaded cells are not used in Power-Down mode.
**10.0 WATCHDOG TIMER (WDT)**

The Watchdog Timer is a system timer that generates a Reset if the firmware does not issue a `CLRWDT` instruction within the time-out period. The Watchdog Timer is typically used to recover the system from unexpected events.

The WDT has the following features:

- Independent clock source
- Multiple operating modes
  - WDT is always On
  - WDT is Off when in Sleep
  - WDT is controlled by software
  - WDT is always Off
- Configurable time-out period is from 1 ms to 256 seconds (typical)
- Multiple Reset conditions
- Operation during Sleep

**FIGURE 10-1: WATCHDOG TIMER BLOCK DIAGRAM**
10.1 Independent Clock Source

The WDT derives its time base from the 31 kHz LFINTOSC internal oscillator.

10.2 WDT Operating Modes

The Watchdog Timer module has four operating modes controlled by the WDTE<1:0> bits in Configuration Word 1. See Table 10-1.

10.2.1 WDT IS ALWAYS ON

When the WDTE bits of Configuration Word 1 are set to '11', the WDT is always On. WDT protection is active during Sleep.

10.2.2 WDT IS OFF IN SLEEP

When the WDTE bits of Configuration Word 1 are set to '10', the WDT is Off, except in Sleep. WDT protection is not active during Sleep.

10.2.3 WDT CONTROLLED BY SOFTWARE

When the WDTE bits of Configuration Word 1 are set to '01', the WDT is controlled by the SWDTEN bit of the WDTCN register. WDT protection is unchanged by Sleep. See Table 10-1 for more details.

10.3 Time-Out Period

The WDTPS bits of the WDTCN register set the time-out period from 1 ms to 256 seconds. After a Reset, the default time-out period is two seconds.

10.4 Clearing the WDT

The WDT is cleared when any of the following conditions occur:

- Any Reset
- CLRWD T instruction is executed
- Device enters Sleep
- Device wakes up from Sleep
- Oscillator fail event
- WDT is disabled
- OST is running

See Table 10-2 for more information.

10.5 Operation During Sleep

When the device enters Sleep, the WDT is cleared. If the WDT is enabled during Sleep, the WDT resumes counting.

When the device exits Sleep, the WDT is cleared again. The WDT remains clear until the OST, if enabled, completes. See Section 5.0 “Oscillator Module (With Fail-Safe Clock Monitor)” for more information on the OST.

When a WDT time-out occurs while the device is in Sleep, no Reset is generated. Instead, the device wakes up and resumes operation. The TG and PD bits in the STATUS register are changed to indicate the event. See Section 3.0 “Memory Organization” and The STATUS register (Register 3-1) for more information.

### Table 10-1: WDT Operating Modes

<table>
<thead>
<tr>
<th>WDTE Config bits</th>
<th>SWDTEN</th>
<th>Device Mode</th>
<th>WDT Mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>WD_ON (11)</td>
<td>X</td>
<td>X</td>
<td>Active</td>
</tr>
<tr>
<td>WDT_NSLEEP (10)</td>
<td>X</td>
<td>Awake</td>
<td>Active</td>
</tr>
<tr>
<td>WDT_NSLEEP (10)</td>
<td>X</td>
<td>Sleep</td>
<td>Disabled</td>
</tr>
<tr>
<td>WDT_SWDTEN (01)</td>
<td>1</td>
<td>X</td>
<td>Active</td>
</tr>
<tr>
<td>WDT_SWDTEN (01)</td>
<td>0</td>
<td>X</td>
<td>Disabled</td>
</tr>
<tr>
<td>WDT_OFF (00)</td>
<td>X</td>
<td>X</td>
<td>Disabled</td>
</tr>
</tbody>
</table>

### Table 10-2: WDT Clearing Conditions

<table>
<thead>
<tr>
<th>Conditions</th>
<th>WDT</th>
</tr>
</thead>
<tbody>
<tr>
<td>WDTE&lt;1:0&gt; = 00</td>
<td>Cleared</td>
</tr>
<tr>
<td>WDTE&lt;1:0&gt; = 01 and SWDTEN = 0</td>
<td>Cleared</td>
</tr>
<tr>
<td>WDTE&lt;1:0&gt; = 10 and enter Sleep</td>
<td>Cleared until the end of OST</td>
</tr>
<tr>
<td>CLRWD T Command</td>
<td>Cleared</td>
</tr>
<tr>
<td>Oscillator Fail Detected</td>
<td>Cleared</td>
</tr>
<tr>
<td>Exit Sleep + System Clock = T1OSC, EXTRC, INTOSC, EXTCLK</td>
<td>Cleared</td>
</tr>
<tr>
<td>Exit Sleep + System Clock = XT, HS, LP</td>
<td>Cleared</td>
</tr>
<tr>
<td>Change INTOSC divider (IRCF bits)</td>
<td>Unaffected</td>
</tr>
</tbody>
</table>
REGISTER 10-1: WDTCON: WATCHDOG TIMER CONTROL REGISTER

| bit 7-6 | Unimplemented: Read as '0' |
| bit 5-1 | WDTPS<4:0>: Watchdog Timer Period Select bits |

Bit Value = Prescale Rate
00000 = 1:32 (Interval 1 ms typ)  
00001 = 1:64 (Interval 2 ms typ)  
00010 = 1:128 (Interval 4 ms typ)  
00011 = 1:256 (Interval 8 ms typ)  
00100 = 1:512 (Interval 16 ms typ)  
00101 = 1:1024 (Interval 32 ms typ)  
00110 = 1:2048 (Interval 64 ms typ)  
00111 = 1:4096 (Interval 128 ms typ)  
01000 = 1:8192 (Interval 256 ms typ)  
01001 = 1:16384 (Interval 512 ms typ)  
01010 = 1:32768 (Interval 1s typ)  
01011 = 1:65536 (Interval 2s typ) (Reset value)  
01100 = 1:131072 (2^17) (Interval 4s typ)  
01101 = 1:262144 (2^18) (Interval 8s typ)  
01110 = 1:524288 (2^19) (Interval 16s typ)  
01111 = 1:1048576 (2^20) (Interval 32s typ)  
10000 = 1:2097152 (2^21) (Interval 64s typ)  
10001 = 1:4194304 (2^22) (Interval 128s typ)  
10010 = 1:8388608 (2^23) (Interval 256s typ)  
10011 = Reserved. Results in minimum interval (1:32)  
10100 = 1:131072 (2^17) (Interval 4s typ)  
10101 = 1:262144 (2^18) (Interval 8s typ)  
10110 = 1:524288 (2^19) (Interval 16s typ)  
10111 = 1:1048576 (2^20) (Interval 32s typ)  
11000 = 1:2097152 (2^21) (Interval 64s typ)  
11001 = 1:4194304 (2^22) (Interval 128s typ)  
11010 = 1:8388608 (2^23) (Interval 256s typ)  
11011 = Reserved. Results in minimum interval (1:32)  
11100 = 1:131072 (2^17) (Interval 4s typ)  
11101 = 1:262144 (2^18) (Interval 8s typ)  
11110 = 1:524288 (2^19) (Interval 16s typ)  
11111 = 1:1048576 (2^20) (Interval 32s typ)  

bit 0 SWDTEN: Software Enable/Disable for Watchdog Timer bit
- If WDTE<1:0> = 0:0:  
  This bit is ignored.  
- If WDTE<1:0> = 0:1:  
  1 = WDT is turned on  
  0 = WDT is turned off  
- If WDTE<1:0> = 1:x:  
  This bit is ignored.
11.0 DATA EEPROM AND FLASH PROGRAM MEMORY CONTROL

The Data EEPROM and Flash program memory are readable and writable during normal operation (full VDD range). These memories are not directly mapped in the register file space. Instead, they are indirectly addressed through the Special Function Registers (SFRs). There are six SFRs used to access these memories:

- EECON1
- EECON2
- EEDATL
- EEDATH
- EEADRL
- EEADRH

When interfacing the data memory block, EEDATL holds the 8-bit data for read/write, and EEADRL holds the address of the EEDATL location being accessed. These devices have 256 bytes of data EEPROM with an address range from 0h to 0FFh.

When accessing the program memory block, the EEDATH:EEDATL register pair forms a 2-byte word that holds the 14-bit data for read/write, and the EEADRL and EEADRH registers form a 2-byte word that holds the 15-bit address of the program memory location being read.

The EEPROM data memory allows byte read and write. An EEPROM byte write automatically erases the location and writes the new data (erase before write).

The write time is controlled by an on-chip timer. The write/erase voltages are generated by an on-chip charge pump rated to operate over the voltage range of the device for byte or word operations.

Depending on the setting of the Flash Program Memory Self-Write Enable bits WRT<1:0> of the Configuration Word 2, the device may or may not be able to write certain blocks of the program memory. However, reads from the program memory are always allowed.

When the device is code-protected, the device programmer can no longer access data or program memory. When code-protected, the CPU may continue to read and write the data EEPROM memory and Flash program memory.

11.1 EEADRL and EEADRH Registers

The EEADRH:EEADRL register pair can address up to a maximum of 256 bytes of data EEPROM or up to a maximum of 32K words of program memory.

When selecting a program address value, the MSB of the address is written to the EEADRH register and the LSB is written to the EEADRL register. When selecting an EEPROM address value, only the LSB of the address is written to the EEADRL register.

11.1.1 EECON1 AND EECON2 REGISTERS

EECON1 is the control register for EE memory accesses.

Control bit EEPGD determines if the access will be a program or data memory access. When clear, any subsequent operations will operate on the EEPROM memory. When set, any subsequent operations will operate on the program memory. On Reset, EEPROM is selected by default.

Control bits RD and WR initiate read and write, respectively. These bits cannot be cleared, only set, in software. They are cleared in hardware at completion of the read or write operation. The inability to clear the WR bit in software prevents the accidental, premature termination of a write operation.

The WREN bit, when set, will allow a write operation to occur. On power-up, the WREN bit is clear. The WRERR bit is set when a write operation is interrupted by a Reset during normal operation. In these situations, following Reset, the user can check the WRERR bit and execute the appropriate error handling routine.

Interrupt flag bit EEIF of the PIR2 register is set when the write is complete. It must be cleared in the software.

Reading EECON2 will read all '0's. The EECON2 register is used exclusively in the data EEPROM write sequence. To enable writes, a specific pattern must be written to EECON2.
11.2 Using the Data EEPROM

The data EEPROM is a high-endurance, byte-addressable array that has been optimized for the storage of frequently changing information (e.g., program variables or other data that are updated often). When variables in one section change frequently, while variables in another section do not change, it is possible to exceed the total number of write cycles to the EEPROM without exceeding the total number of write cycles to a single byte. Refer to Section 31.0 “Electrical Specifications”. If this is the case, then a refresh of the array must be performed. For this reason, variables that change infrequently (such as constants, IDs, calibration, etc.) should be stored in Flash program memory.

11.2.1 READING THE DATA EEPROM MEMORY

To read a data memory location, the user must write the address to the EEA DR register, clear the EEPGD and CFGS control bits of the EECON1 register, and then set control bit RD. The data is available at the very next cycle, in the EEDATL register; therefore, it can be read in the next instruction. EEDATL will hold this value until another read or until it is written to by the user (during a write operation).

EXAMPLE 11-1: DATA EEPROM READ

```
BANKSEL EEA DR
MOVLW DATA_EE_ADDR
MOVWF EEA DR ;Address to read
BCF EECON1, CFGS ;Deselect Config space
BCF EECON1, EEPGD ;Point to DATA memory
BSF EECON1, RD ;EE Read
MOVF EEDATL, W ;W = EEDATL
```

Note: Data EEPROM can be read regardless of the setting of the CPD bit.

11.2.2 WRITING TO THE DATA EEPROM MEMORY

To write an EEPROM data location, the user must first write the address to the EEA DR register and the data to the EEDATL register. Then the user must follow a specific sequence to initiate the write for each byte. The write will not initiate if the above sequence is not followed exactly (write 55h to EECON2, write AAh to EECON2, then set WR bit) for each byte. Interrupts should be disabled during this code segment.

Additionally, the WREN bit in EECON1 must be set to enable write. This mechanism prevents accidental writes to data EEPROM due to errant (unexpected) code execution (i.e., lost programs). The user should keep the WREN bit clear at all times, except when updating EEPROM. The WREN bit is not cleared by hardware.

After a write sequence has been initiated, clearing the WREN bit will not affect this write cycle. The WR bit will be inhibited from being set unless the WREN bit is set.

At the completion of the write cycle, the WR bit is cleared in hardware and the EE Write Complete Interrupt Flag bit (EEIF) is set. The user can either enable this interrupt or poll this bit. EEIF must be cleared by software.

11.2.3 PROTECTION AGAINST SPURIOUS WRITE

There are conditions when the user may not want to write to the data EEPROM memory. To protect against spurious EEPROM writes, various mechanisms have been built-in. On power-up, WREN is cleared. Also, the Power-up Timer (64 ms duration) prevents EEPROM write.

The write initiate sequence and the WREN bit together help prevent an accidental write during:
- Brown-out
- Power Glitch
- Software Malfunction

11.2.4 DATA EEPROM OPERATION DURING CODE-PROTECT

Data memory can be code-protected by programming the CPD bit in the Configuration Word 1 (Register 4-1) to '0'.

When the data memory is code-protected, only the CPU is able to read and write data to the data EEPROM. It is recommended to code-protect the program memory when code-protecting data memory. This prevents anyone from replacing your program with a program that will access the contents of the data EEPROM.
EXAMPLE 11-2: DATA EEPROM WRITE

BANKSEL EEADRL ;
MOVLW DATA_EE_ADDR ;
MOVWF EEADRL ;Data Memory Address to write
MOVLW DATA_EE_DATA ;
MOVWF EEDATL ;Data Memory Value to write
BCF EECON1, CFGS ;Deselect Configuration space
BCF EECON1, EEPGD ;Point to DATA memory
BSF EECON1, WREN ;Enable writes

BCF INTCON, GIE ;Disable INTs.
MOVLW 55h ;
MOVWF EECON2 ;Write 55h
MOVLW 0AAh ;
MOVWF EECON2 ;Write AAh
BSF EECON1, WR ;Set WR bit to begin write
BSF INTCON, GIE ;Enable Interrupts
BCF EECON1, WREN ;Disable writes
BTFSC EECON1, WR ;Wait for write to complete
GOTO $-2 ;Done

FIGURE 11-1: FLASH PROGRAM MEMORY READ CYCLE EXECUTION

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>PC</td>
<td>PC + 1</td>
<td>EEADRH,EEADRL</td>
<td>PC + 3</td>
<td>PC + 4</td>
<td>PC + 5</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Flash ADDR</td>
<td>Flash Data</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>INSTR (PC)</td>
<td>INSTR (PC + 1)</td>
<td>EEDATH,EEDATL</td>
<td>INSTR (PC + 3)</td>
<td>INSTR (PC + 4)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>RD bit</th>
<th>EEDATL</th>
<th>EEDATH</th>
<th>EERHLT</th>
</tr>
</thead>
<tbody>
<tr>
<td>INSTR(PC - 1) executed here</td>
<td>BSF EECON1, RD executed here</td>
<td>INSTR(PC + 1) executed here</td>
<td>Forced NOP executed here</td>
</tr>
</tbody>
</table>
11.3 Flash Program Memory Overview

It is important to understand the Flash program memory structure for erase and programming operations. Flash program memory is arranged in rows. A row consists of a fixed number of 14-bit program memory words. A row is the minimum block size that can be erased by user software.

Flash program memory may only be written or erased if the destination address is in a segment of memory that is not write-protected, as defined in bits WRT<1:0> of Configuration Word 2.

After a row has been erased, the user can reprogram all or a portion of this row. Data to be written into the program memory row is written to 14-bit wide data write latches. These write latches are not directly accessible to the user, but may be loaded via sequential writes to the EEDATH:EEDATL register pair.

The number of data write latches may not be equivalent to the number of row locations. During programming, user software may need to fill the set of write latches and initiate a programming operation multiple times in order to fully reprogram an erased row. For example, a device with a row size of 32 words and eight write latches will need to load the write latches with data and initiate a programming operation four times.

The size of a program memory row and the number of program memory write latches may vary by device. See Table 11-1 for details.

<table>
<thead>
<tr>
<th>Device</th>
<th>Erase Block (Row) Size/ Boundary</th>
<th>Number of Write Latches/ Boundary</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC12LF1840T39A</td>
<td>32 words, EEADRL&lt;4:0&gt; = 00000</td>
<td>32 words, EEADRL&lt;4:0&gt; = 00000</td>
</tr>
</tbody>
</table>

11.3.1 READING THE FLASH PROGRAM MEMORY

To read a program memory location, the user must:

1. Write the Least and Most Significant address bits to the EEADRH:EEADRL register pair.
2. Clear the CFGS bit of the EECON1 register.
3. Set the EEPOD control bit of the EECON1 register.
4. Then, set control bit RD of the EECON1 register.

Once the read control bit is set, the program memory Flash controller will use the second instruction cycle to read the data. This causes the second instruction immediately following the “BSF EECON1, RD” instruction to be ignored. The data is available in the very next cycle, in the EEDATH:EEDATL register pair; therefore, it can be read as two bytes in the following instructions.

EEDATH:EEDATL register pair will hold this value until another read or until it is written to by the user.

Note: If the user wants to modify only a portion of a previously programmed row, then the contents of the entire row must be read and saved in RAM prior to the erase.

The two instructions following a program memory read are required to be NOPs. This prevents the user from executing a two-cycle instruction on the next instruction after the RD bit is set.

2: Flash program memory can be read regardless of the setting of the CP bit.
EXAMPLE 11-3: FLASH PROGRAM MEMORY READ

* This code block will read 1 word of program memory at the memory address:
* PROG_ADDR_HI: PROG_ADDR_LO
* data will be returned in the variables;
* PROG_DATA_HI, PROG_DATA_LO

BANKSEL EEADRL ; Select Bank for EEPROM registers
MOVLW PROG_ADDR_LO ;
MOVWF EEADRL ; Store LSB of address
MOVLW PROG_ADDR_HI ;
MOVWF EEADRH ; Store MSB of address
BCF EECON1,CFGS ; Do not select Configuration Space
BSF EECON1,EEPGD ; Select Program Memory
BCF INTCON,GIE ; Disable interrupts
BSF EECON1,RD ; Initiate read
NOP ; Executed (Figure 11-1)
NOP ; Ignored (Figure 11-1)
BSF INTCON,GIE ; Restore interrupts
MOVF EEDATL,W ; Get LSB of word
MOVWF PROG_DATA_LO ; Store in user location
MOVF EEDATH,W ; Get MSB of word
MOVWF PROG_DATA_HI ; Store in user location
11.3.2 ERASING FLASH PROGRAM MEMORY

While executing code, program memory can only be erased by rows. To erase a row:
1. Load the EEADRH:EEADRL register pair with the address of new row to be erased.
2. Clear the CFGS bit of the EECON1 register.
3. Set the EEPGD, FREE and WREN bits of the EECON1 register.
4. Write 55h, then AAh, to EECON2 (Flash programming unlock sequence).
5. Set control bit WR of the EECON1 register to begin the erase operation.
6. Poll the FREE bit in the EECON1 register to determine when the row erase has completed.

See Example 11-4.

After the "BSF EECON1,WR" instruction, the processor requires two cycles to set up the erase operation. The user must place two NOP instructions after the WR bit is set. The processor will halt internal operations for the typical 2 ms erase time. This is not Sleep mode as the clocks and peripherals will continue to run. After the erase cycle, the processor will resume operation with the third instruction after the EECON1 write instruction.

11.3.3 WRITING TO FLASH PROGRAM MEMORY

Program memory is programmed using the following steps:
1. Load the starting address of the word(s) to be programmed.
2. Load the write latches with data.
3. Initiate a programming operation.
4. Repeat steps 1 through 3 until all data is written.

Before writing to program memory, the word(s) to be written must be erased or previously unwritten. Program memory can only be erased one row at a time. No automatic erase occurs upon the initiation of the write.

Program memory can be written one or more words at a time. The maximum number of words written at one time is equal to the number of write latches. See Figure 11-2 (block writes to program memory with 32 write latches) for more details. The write latches are aligned to the address boundary defined by EEADRL as shown in Table 11-1. Write operations do not cross these boundaries. At the completion of a program memory write operation, the write latches are reset to contain 0x3FFF.

The following steps should be completed to load the write latches and program a block of program memory. These steps are divided into two parts. First, all write latches are loaded with data except for the last program memory location. Then, the last write latch is loaded and the programming sequence is initiated. A special unlock sequence is required to load a write latch with data or initiate a Flash programming operation. This unlock sequence should not be interrupted.
1. Set the EEPGD and WREN bits of the EECON1 register.
2. Clear the CFGS bit of the EECON1 register.
3. Set the LWLO bit of the EECON1 register. When the LWLO bit of the EECON1 register is '1', the write sequence will only load the write latches and will not initiate the write to Flash program memory.
4. Load the EEADRH:EEADRL register pair with the address of the location to be written.
5. Load the EEDATH:EEDATL register pair with the program memory data to be written.
6. Write 55h, then AAh, to EECON2, then set the WR bit of the EECON1 register (Flash programming unlock sequence). The write latch is now loaded.
7. Increment the EEADRH:EEADRL register pair to point to the next location.
8. Repeat steps 5 through 7 until all but the last write latch has been loaded.
9. Clear the LWLO bit of the EECON1 register. When the LWLO bit of the EECON1 register is '0', the write sequence will initiate the write to Flash program memory.
10. Load the EEDATH:EEDATL register pair with the program memory data to be written.
11. Write 55h, then AAh, to EECON2, then set the WR bit of the EECON1 register (Flash programming unlock sequence). The entire latch block is now written to Flash program memory.

It is not necessary to load the entire write latch block with user program data. However, the entire write latch block will be written to program memory.

An example of the complete write sequence for 32 words is shown in Example 11-5. The initial address is loaded into the EEADRH:EEADRL register pair; the 32 words of data are loaded using indirect addressing.
After the “BSF EECON1, WR” instruction, the processor requires two cycles to set up the write operation. The user must place two NOP instructions after the WR bit is set. The processor will halt internal operations for the typical 2 ms, only during the cycle in which the write takes place (i.e., the last word of the block write). This is not Sleep mode as the clocks and peripherals will continue to run. The processor does not stall when LWLO = 1, loading the write latches. After the write cycle, the processor will resume operation with the third instruction after the EECON1 write instruction.

**FIGURE 11-2: BLOCK WRITES TO FLASH PROGRAM MEMORY WITH 32 WRITE LATCHES**

```
<table>
<thead>
<tr>
<th>Buffer Register</th>
<th>Buffer Register</th>
<th>Buffer Register</th>
<th>Buffer Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>EEADRL&lt;4:0&gt; = 00000</td>
<td>EEADRL&lt;4:0&gt; = 00001</td>
<td>EEADRL&lt;4:0&gt; = 00010</td>
<td>EEADRL&lt;4:0&gt; = 11111</td>
</tr>
<tr>
<td>First word of block to be written</td>
<td>Last word of block to be written</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>
```

Program Memory
EXAMPLE 11-4: ERASING ONE ROW OF PROGRAM MEMORY

; This row erase routine assumes the following:
; 1. A valid address within the erase block is loaded in ADDRH:ADDRL
; 2. ADDRH and ADDRL are located in shared data memory 0x70 - 0x7F

BCF INTCON,GIE ; Disable ints so required sequences will execute properly
BANKSEL EEADRL

MOVF ADDRL,W ; Load lower 8 bits of erase address boundary
MOVWF EEADRL

MOVF ADDRH,W ; Load upper 6 bits of erase address boundary
MOVWF EEADRH

BSF EECON1,EEPGD ; Point to program memory
BCF EECON1,CFGS ; Not configuration space
BSF EECON1,FREE ; Specify an erase operation
BSF EECON1,WREN ; Enable writes

MOVLW 55h ; Start of required sequence to initiate erase
MOVWF EECON2 ; Write 55h
MOVLW 0AAh ;
MOVWF EECON2 ; Write AAh
BSF EECON1,NR ; Set WR bit to begin erase
NOP ; Any instructions here are ignored as processor
; halts to begin erase sequence
NOP ; Processor will stop here and wait for erase complete.
; after erase processor continues with 3rd instruction

BCF EECON1,WREN ; Disable writes
BSF INTCON,GIE ; Enable interrupts
EXAMPLE 11-5: WRITING TO FLASH PROGRAM MEMORY

; This write routine assumes the following:
; 1. The 64 bytes of data are loaded, starting at the address in DATA_ADDR
; 2. Each word of data to be written is made up of two adjacent bytes in DATA_ADDR, 
;    stored in little endian format
; 3. A valid starting address (the least significant bits = 00000) is loaded in ADDRH:ADDRL
; 4. ADDRH and ADDRL are located in shared data memory 0x70 - 0x7F
;
; BCF INTCON,GIE ; Disable ints so required sequences will execute properly
BANKSEL EEADRH ; Bank 3
MOVF ADDRH,W ; Load initial address
MOVFEEADRH ;
MOVF ADDRL,W ;
MOVFEEADRL ;
MOVWL LOW DATA_ADDR ; Load initial data address
MOVF FSR0L ;
MOVWL HIGH DATA_ADDR ; Load initial data address
MOVF FSR0H ;
BSF EECON1,EEPGD ; Point to program memory
BCF EECON1,CFGS ; Not configuration space
BSF EECON1,WREN ; Enable writes
BSF EECON1,LWLO ; Only Load Write Latches

LOOP
MOVIW FSR0++ ; Load first data byte into lower
MOVFEEDATL ;
MOVIW FSR0++ ; Load second data byte into upper
MOVFEEDATH ;

MOVF EEA DRL, W ; Check if lower bits of address are '00000'
XORLW 0x1F ; Check if we're on the last of 32 addresses
ANDLW 0x1F ;
BTFSC STATUS,Z ; Exit if last of 32 words,
GOTO START_WRITE ;

MOVWL 55h ; Start of required write sequence:
MOVF EECON2 ; Write 55h
MOVLW 0AAh ;
MOVF EECON2 ; Write AAh
BSF EECON1,WR ; Set WR bit to begin write
NOP ; Any instructions here are ignored as processor
; halts to begin write sequence
NOP ; Processor will stop here and wait for write to complete.
; After write processor continues with 3rd instruction.

INCF EEADRL,F ; Still loading latches Increment address
GOTO LOOP ; Write next latches

START_WRITE
BCF EECON1,LWLO ; No more loading latches - Actually start Flash program
; memory write

MOVWL 55h ; Start of required write sequence:
MOVF EECON2 ; Write 55h
MOVLW 0AAh ;
MOVF EECON2 ; Write AAh
BSF EECON1,WR ; Set WR bit to begin write
NOP ; Any instructions here are ignored as processor
; halts to begin write sequence
NOP ; Processor will stop here and wait for write to complete.
; after write processor continues with 3rd instruction
BCF EECON1,WREN ; Disable writes
BSF INTCON,GIE ; Enable interrupts
11.4 Modifying Flash Program Memory

When modifying existing data in a program memory row, and data within that row must be preserved, it must first be read and saved in a RAM image. Program memory is modified using the following steps:

1. Load the starting address of the row to be modified.
2. Read the existing data from the row into a RAM image.
3. Modify the RAM image to contain the new data to be written into program memory.
4. Load the starting address of the row to be rewritten.
5. Erase the program memory row.
6. Load the write latches with data from the RAM image.
7. Initiate a programming operation.
8. Repeat steps 6 and 7 as many times as required to reprogram the erased row.

11.5 User ID, Device ID and Configuration Word Access

Instead of accessing program memory or EEPROM data memory, the User ID’s, Device ID/Revision ID and Configuration Words can be accessed when CFGS = 1 in the EECON1 register. This is the region that would be pointed to by PC<15> = 1, but not all addresses are accessible. Different access may exist for reads and writes. Refer to Table 11-2.

When read access is initiated on an address outside the parameters listed in Table 11-2, the EEDATH:EEDATL register pair is cleared.

<table>
<thead>
<tr>
<th>Address</th>
<th>Function</th>
<th>Read Access</th>
<th>Write Access</th>
</tr>
</thead>
<tbody>
<tr>
<td>8000h-8003h</td>
<td>User IDs</td>
<td>Yes</td>
<td>Yes</td>
</tr>
<tr>
<td>8006h</td>
<td>Device ID/Revision ID</td>
<td>Yes</td>
<td>No</td>
</tr>
<tr>
<td>8007h-8008h</td>
<td>Configuration Words 1 and 2</td>
<td>Yes</td>
<td>No</td>
</tr>
</tbody>
</table>

EXAMPLE 11-3: Configuration Word and Device ID Access

* This code block will read 1 word of program memory at the memory address:
* PROG_ADDR_LO (must be 00h-08h) data will be returned in the variables;
* PROG_DATA_HI, PROG_DATA_LO

BANKSEL EEADRL ; Select correct Bank
MOVLW PROG_ADDR_LO  ;
MOVF EEADRL,W  ; Store LSB of address
CLRF EEADRH  ; Clear MSB of address
BSF EECON1,CFGs  ; Select Configuration Space
BCF INTCON,GIE  ; Disable interrupts
BSF EECON1,RD  ; Initiate read
NOP  ; Executed (See Figure 11-1)
NOP  ; Ignored (See Figure 11-1)
BSF INTCON,GIE  ; Restore interrupts
MOVF EEDATL,W  ; Get LSB of word
MOVF PROG_DATA_LO  ; Store in user location
MOVF EEDATL,W  ; Get MSB of word
MOVF PROG_DATA_HI  ; Store in user location
11.6 Write Verify

Depending on the application, good programming practice may dictate that the value written to the data EEPROM or program memory should be verified (see Example 11-6) to the desired value to be written. Example 11-6 shows how to verify a write to EEPROM.

**EXAMPLE 11-6: EEPROM WRITE VERIFY**

```
BANKSEL EEDATL ;
MOVF EEDATL, W ;EEDATL not changed
;from previous write
BSF EECON1, RD ;YES, Read the
;value written
XORWF EEDATL, W ;
BTFSS STATUS, Z ;Is data the same
GOTO WRITE_ERR ;No, handle error
: ;Yes, continue
```
**REGISTER 11-1: EEDATL: EEPROM DATA REGISTER**

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

**bit 7-0**

**EEDAT<7:0>:** Read/write value for EEPROM data byte or Least Significant bits of program memory

**REGISTER 11-2: EEDATH: EEPROM DATA HIGH-BYTE REGISTER**

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

**bit 7-6**

**Unimplemented:** Read as ‘0’

**bit 5-0**

**EEDAT<13:8>:** Read/write value for Most Significant bits of program memory

**REGISTER 11-3: EEADRL: EEPROM ADDRESS REGISTER**

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

**bit 7-0**

**EEADR<7:0>:** Specifies the Least Significant bits for program memory address or EEPROM address

**REGISTER 11-4: EEADRH: EEPROM ADDRESS HIGH-BYTE REGISTER**

<table>
<thead>
<tr>
<th>U-1</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

**bit 7**

**Unimplemented:** Read as ‘1’

**bit 6-0**

**EEADR<14:8>:** Specifies the Most Significant bits for program memory address or EEPROM address
REGISTER 11-5: EECON1: EEPROM CONTROL 1 REGISTER

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>EEPGD</td>
<td>CFGS</td>
<td>LWLO</td>
<td>FREE</td>
<td>WRERR</td>
<td>WREN</td>
<td>WR</td>
<td>RD</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- S = Bit can only be set
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets

`'1' = Bit is set
'0' = Bit is cleared
HC = Bit is cleared by hardware

**bit 7**

**EEPGD**: Flash Program/Data EEPROM Memory Select bit
1 = Accesses program space Flash memory
0 = Accesses data EEPROM memory

**bit 6**

**CFGS**: Flash Program/Data EEPROM or Configuration Select bit
1 = Accesses Configuration, User ID and Device ID Registers
0 = Accesses Flash program or data EEPROM memory

**bit 5**

**LWLO**: Load Write Latches Only bit

If CFGS = 1 (Configuration space) OR CFGS = 0 and EEPGD = 1 (program Flash):
1 = The next WR command does not initiate a write; only the program memory latches are updated.
0 = The next WR command writes a value from EEDATH:EEDATL into program memory latches and initiates a write of all the data stored in the program memory latches.

If CFGS = 0 and EEPGD = 0: (Accessing data EEPROM)
LWLO is ignored. The next WR command initiates a write to the data EEPROM.

**bit 4**

**FREE**: Program Flash Erase Enable bit

If CFGS = 1 (Configuration space) OR CFGS = 0 and EEPGD = 1 (program Flash):
1 = Performs an erase operation on the next WR command (cleared by hardware after completion of erase).
0 = Performs a write operation on the next WR command.

If EEPGD = 0 and CFGS = 0: (Accessing data EEPROM)
FREE is ignored. The next WR command will initiate both an erase cycle and a write cycle.

**bit 3**

**WRERR**: EEPROM Error Flag bit
1 = Condition indicates an improper program or erase sequence attempt or termination (bit is set automatically on any set attempt (write '1') of the WR bit).
0 = The program or erase operation completed normally.

**bit 2**

**WREN**: Program/Erase Enable bit
1 = Allows program/erase cycles
0 = Inhibits programming/erasing of program Flash and data EEPROM

**bit 1**

**WR**: Write Control bit
1 = Initiates a program Flash or data EEPROM program/erase operation.
The operation is self-timed and the bit is cleared by hardware once operation is complete.
The WR bit can only be set (not cleared) in software.
0 = Program/erase operation to the Flash or data EEPROM is complete and inactive.

**bit 0**

**RD**: Read Control bit
1 = Initiates a program Flash or data EEPROM read. Read takes one cycle. RD is cleared in hardware. The RD bit can only be set (not cleared) in software.
0 = Does not initiate a program Flash or data EEPROM data read.
### REGISTER 11-6: EECON2: EEPROM CONTROL 2 REGISTER

<table>
<thead>
<tr>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
<th>W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

EEPROM Control Register 2

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 0</th>
</tr>
</thead>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **S** = Bit can only be set
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7-0**

**Data EEPROM Unlock Pattern bits**

To unlock writes, a 55h must be written first, followed by an AAh, before setting the WR bit of the EECON1 register. The value written to this register is used to unlock the writes. There are specific timing requirements on these writes. Refer to Section 11.2.2 “Writing to the Data EEPROM Memory” for more information.

### TABLE 11-3: SUMMARY OF REGISTERS ASSOCIATED WITH DATA EEPROM

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>EECON1</td>
<td>EEPGD</td>
<td>CFGS</td>
<td>LWLO</td>
<td>FREE</td>
<td>WRERR</td>
<td>WREN</td>
<td>WR</td>
<td>RD</td>
<td>100</td>
</tr>
<tr>
<td>EECON2</td>
<td>EEPROM Control Register 2 (not a physical register)</td>
<td>101*</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>EEADRL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>99</td>
</tr>
<tr>
<td>EEADRH</td>
<td>_{(1)}</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>99</td>
</tr>
<tr>
<td>EEDATL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>99</td>
</tr>
<tr>
<td>EEDATH</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>99</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>—</td>
<td>C1IE</td>
<td>EEIE</td>
<td>BCL1IE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>79</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>—</td>
<td>C1IF</td>
<td>EEIF</td>
<td>BCL1IF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>81</td>
</tr>
</tbody>
</table>

**Legend:**
- — = unimplemented location, read as ‘0’.
- Shaded cells are not used by data EEPROM module.
- * Page provides register information.

**Note 1:** Unimplemented, read as ‘1’.
12.0 I/O PORTS

In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

The port has three registers for its operation. These registers are:

- TRISA register (data direction register)
- PORTA register (reads the levels on the pins of the device)
- LATA register (output latch)

PORTA has the following additional registers. They are:

- ANSELA (analog select)
- WPUA (weak pull-up)

The Data Latch (LATA register) is useful for read-modify-write operations on the value that the I/O pins are driving.

A write operation to the LATA register has the same affect as a write to the corresponding PORTA register. A read of the LATA register reads of the values held in the I/O port latches, while a read of the PORTA register reads the actual I/O pin value.

The port has analog functions and has an ANSELA register which can disable the digital input and save power. A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 12-1.

12.1 Alternate Pin Function

The Alternate Pin Function Control (APFCON) register is used to steer specific peripheral input and output functions between different pins. The APFCON register is shown in Register 12-1. For this device family, the following functions can be moved between different pins.

- RX/DT
- TX/CK
- SDO
- S (Slave Select)
- T1G
- P1B
- CCP1/P1A

These bits have no effect on the values of any TRIS register. PORT and TRIS overrides will be routed to the correct pin. The unselected pin will be unaffected.
REGISTER 12-1: APFCON: ALTERNATE PIN FUNCTION CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>RXDTSEL</td>
<td>SDOSEL</td>
<td>SSSEL</td>
<td>—</td>
<td>T1GSEL</td>
<td>TXCKSEL</td>
<td>P1BSEL</td>
<td>CCP1SEL</td>
</tr>
</tbody>
</table>

Legend:

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>RXDTSEL: Pin Selection bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>RX/DT function is on RA1</td>
</tr>
<tr>
<td>1</td>
<td>RX/DT function is on RA5</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 6</th>
<th>SDOSEL: Pin Selection bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>SDO function is on RA0</td>
</tr>
<tr>
<td>1</td>
<td>SDO function is on RA4</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 5</th>
<th>SSSEL: Pin Selection bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>SS function is on RA3</td>
</tr>
<tr>
<td>1</td>
<td>SS function is on RA0</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 4</th>
<th>Unimplemented: Read as ‘0’</th>
</tr>
</thead>
</table>

<table>
<thead>
<tr>
<th>Bit 3</th>
<th>T1GSEL: Pin Selection bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>T1G function is on RA4</td>
</tr>
<tr>
<td>1</td>
<td>T1G function is on RA3</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 2</th>
<th>TXCKSEL: Pin Selection bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>TX/CK function is on RA0</td>
</tr>
<tr>
<td>1</td>
<td>TX/CK function is on RA4</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 1</th>
<th>P1BSEL: Pin Selection bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>P1B function is on RA0</td>
</tr>
<tr>
<td>1</td>
<td>P1B function is on RA4</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 0</th>
<th>CCP1SEL: Pin Selection bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>CCP1/P1A function is on RA2</td>
</tr>
<tr>
<td>1</td>
<td>CCP1/P1A function is on RA5</td>
</tr>
</tbody>
</table>
12.2 PORTA Registers

PORTA is a 6-bit wide, bidirectional port. The corresponding data direction register is TRISA (Register 12-3). Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., disable the output driver). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). The exception is RA3, which is input only and its TRIS bit will always read as ‘1’. Example 12-1 shows how to initialize PORTA.

Reading the PORTA register (Register 12-2) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATA).

The TRISA register (Register 12-3) controls the PORTA pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISA register are maintained set when using them as analog inputs. I/O pins configured as analog input always read ‘0’.

12.2.1 ANSELA REGISTER

The ANSELA register (Register 12-5) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELA bit high will cause all digital reads on the pin to be read as ‘0’ and allow analog functions on the pin to operate correctly.

The state of the ANSELA bits has no affect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note: The ANSELA register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read ‘0’.

EXAMPLE 12-1: INITIALIZING PORTA

```
BANKSEL PORTA ;
CLRF PORTA ;Init PORTA
BANKSEL LATA ;Data Latch
CLRF LATA ;
BANKSEL ANSELA ;
CLRF ANSELA ;digital I/O
BANKSEL TRISA ;
MOVLW B'00111000' ;Set RA<5:3> as inputs
MOVWF TRISA ;and set RA<2:0> as outputs
```
12.2.2 PORTA FUNCTIONS AND OUTPUT PRIORITIES

Each PORTA pin is multiplexed with other functions. The pins, their combined functions and their output priorities are briefly described here. For additional information, refer to the appropriate section in this data sheet.

When multiple outputs are enabled, the actual pin control goes to the peripheral with the lowest number in the following lists.

Analog input functions, such as ADC, comparator and CapSense inputs, are not shown in the priority lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELx registers. Digital output functions may control the pin when it is in Analog mode with the priority shown below.

RA0
1. ICSPDAT
2. ICDDAT
3. DACOUT (DAC)
4. MDOUT
5. TX/CK (EUSART)
6. SDO
7. P1B

RA1
1. ICSPCLK
2. ICDCLK
3. SCL
4. RX/DT (EUSART)
5. SCK

RA2
1. SRQ
2. C1OUT (Comparator)
3. SDA
4. CCP1/P1A

RA3
No output priorities. Input only pin.

RA4
1. OSC2
2. CLKOUT
3. T1OSO
4. CLKR
5. TX/CK
6. SDO
7. P1B

RA5
1. OSC1
2. T1OSI (Timer1 Oscillator)
3. SRNQ
4. RX/DT
5. CCP1/P1A
### REGISTER 12-2: PORTA: PORTA REGISTER

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/x</th>
<th>R/W-x/x</th>
<th>R-x/x</th>
<th>R/W-x/x</th>
<th>R/W-x/x</th>
<th>R/W-x/x</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>—</td>
<td>—</td>
<td>RA5</td>
<td>RA4</td>
<td>RA3</td>
<td>RA2</td>
<td>RA1</td>
<td>RA0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

**bit 7-6** Unimplemented: Read as '0'

**bit 5-0**
- **RA<5:0>:** PORTA I/O Value bits
  - 1 = Port pin is $\geq$ $V_{IH}$
  - 0 = Port pin is $\leq$ $V_{IL}$

**Note 1:** Writes to PORTA are actually written to corresponding LATA register. Reads from PORTA register is return of actual I/O pin values.

### REGISTER 12-3: TRISA: PORTA TRI-STATE REGISTER

<table>
<thead>
<tr>
<th></th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

**bit 7-6** Unimplemented: Read as '0'

**bit 5-4**
- **TRISA<5:4>:** PORTA Tri-State Control bits
  - 1 = PORTA pin configured as an input (tri-stated)
  - 0 = PORTA pin configured as an output

**bit 3**
- **TRISA3:** RA3 Port Tri-State Control bit
  - This bit is always '1' as RA3 is an input only

**bit 2-0**
- **TRISA<2:0>:** PORTA Tri-State Control bits
  - 1 = PORTA pin configured as an input (tri-stated)
  - 0 = PORTA pin configured as an output
## REGISTER 12-4: LATA: PORTA DATA LATCH REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>LATA5</td>
<td>LATA4</td>
<td>—</td>
<td>LATA2</td>
<td>LATA1</td>
<td>LATA0</td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **’0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

bit 7-6: **Unimplemented**: Read as '0'

bit 5-4: **LATA<5:4>: RA<5:4> Output Latch Value bits**(1)

bit 3: **Unimplemented**: Read as '0'

bit 2-0: **LATA<2:0>: RA<2:0> Output Latch Value bits**(1)

**Note 1:** Writes to PORTA are actually written to corresponding LATA register. Reads from PORTA register is return of actual I/O pin values.

## REGISTER 12-5: ANSELA: PORTA ANALOG SELECT REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-1/1</th>
<th>U-0</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA4</td>
<td>—</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **’1’** = Bit is set
- **’0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

bit 7-5: **Unimplemented**: Read as '0'

bit 4: **ANSA4: Analog Select between Analog or Digital Function on pins RA4, respectively**

- **0** = Digital I/O. Pin is assigned to port or digital special function.
- **1** = Analog input. Pin is assigned as analog input**(1)**. Digital input buffer disabled.

bit 3: **Unimplemented**: Read as '0'

bit 2-0: **ANSA<2:0>: Analog Select between Analog or Digital Function on pins RA<2:0>, respectively**

- **0** = Digital I/O. Pin is assigned to port or digital special function.
- **1** = Analog input. Pin is assigned as analog input**(1)**. Digital input buffer disabled.

**Note 1:** When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.
**TABLE 12-1: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA4</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
</tr>
<tr>
<td>APFCON</td>
<td>RXDSEL</td>
<td>SDSEL</td>
<td>SSSEL</td>
<td>—</td>
<td>T1GSEL</td>
<td>TXKSEL</td>
<td>PSSEL</td>
<td>P1BSEL</td>
<td>CCP1SEL</td>
</tr>
<tr>
<td>LAT</td>
<td>—</td>
<td>—</td>
<td>LATA5</td>
<td>LATA4</td>
<td>—</td>
<td>LATA2</td>
<td>LATA1</td>
<td>LATA0</td>
<td>107</td>
</tr>
<tr>
<td>OPTION_REG</td>
<td>WPUEN</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS&lt;2:0&gt;</td>
<td>107</td>
<td></td>
<td></td>
</tr>
<tr>
<td>PORTA</td>
<td>—</td>
<td>—</td>
<td>RA5</td>
<td>RA4</td>
<td>RA3</td>
<td>RA2</td>
<td>RA1</td>
<td>RA0</td>
<td>106</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRIS5</td>
<td>TRIS4</td>
<td>TRIS3</td>
<td>TRIS2</td>
<td>TRIS1</td>
<td>TRIS0</td>
<td>106</td>
</tr>
<tr>
<td>WPUA</td>
<td>—</td>
<td>—</td>
<td>WPUA5</td>
<td>WPUA4</td>
<td>WPUA3</td>
<td>WPUA2</td>
<td>WPUA1</td>
<td>WPUA0</td>
<td>106</td>
</tr>
</tbody>
</table>

**Legend:**
- \( x \) = unknown, \( u \) = unchanged, = unimplemented locations read as ‘0’. Shaded cells are not used by PORTA.

**TABLE 12-2: SUMMARY OF CONFIGURATION WORD WITH PORTA**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bits</th>
<th>Bit -7</th>
<th>Bit -6</th>
<th>Bit 13/5</th>
<th>Bit 12/4</th>
<th>Bit 11/3</th>
<th>Bit 10/2</th>
<th>Bit 9/1</th>
<th>Bit 8/0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CONFIG1</td>
<td>13:8</td>
<td>—</td>
<td>—</td>
<td>FCMEN</td>
<td>IESO</td>
<td>CLKOUTEN</td>
<td>BOREN&lt;1:0&gt;</td>
<td>CPD</td>
<td>39</td>
<td></td>
</tr>
<tr>
<td></td>
<td>7:0</td>
<td>CP</td>
<td>MCLRE</td>
<td>PWLTE</td>
<td>WDTE&lt;1:0&gt;</td>
<td>FOSC&lt;2:0&gt;</td>
<td>CPD</td>
<td>39</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- = unimplemented location, read as ‘0’. Shaded cells are not used by PORTA.
13.0 INTERRUPT-ON-CHANGE

The PORTA pins can be configured to operate as Interrupt-On-Change (IOC) pins. An interrupt can be generated by detecting a signal that has either a rising edge or a falling edge. Any individual PORTA pin, or combination of PORTA pins, can be configured to generate an interrupt. The interrupt-on-change module has the following features:

- Interrupt-on-Change enable (Master Switch)
- Individual pin configuration
- Rising and falling edge detection
- Individual pin interrupt flags

Figure 13-1 is a block diagram of the IOC module.

13.1 Enabling the Module

To allow individual PORTA pins to generate an interrupt, the IOCIE bit of the INTCON register must be set. If the IOCIE bit is disabled, the edge detection on the pin will still occur, but an interrupt will not be generated.

13.2 Individual Pin Configuration

For each PORTA pin, a rising edge detector and a falling edge detector are present. To enable a pin to detect a rising edge, the associated IOCAPx bit of the IOCAP register is set. To enable a pin to detect a falling edge, the associated IOCANx bit of the IOCAN register is set.

A pin can be configured to detect rising and falling edges simultaneously by setting both the IOCAPx bit and the IOCANx bit of the IOCAP and IOCAN registers, respectively.

13.3 Interrupt Flags

The IOCAFx bits located in the IOCAF register are status flags that correspond to the interrupt-on-change pins of PORTA. If an expected edge is detected on an appropriately enabled pin, then the status flag for that pin will be set, and an interrupt will be generated if the IOCIE bit is set. The IOCF bit of the INTCON register reflects the status of all IOCAFx bits.

13.4 Clearing Interrupt Flags

The individual status flags, (IOCAFx bits), can be cleared by resetting them to zero. If another edge is detected during this clearing operation, the associated status flag will be set at the end of the sequence, regardless of the value actually being written.

In order to ensure that no detected edge is lost while clearing flags, only AND operations masking out known changed bits should be performed. The following sequence is an example of what should be performed.

EXAMPLE 13-1: CLEARING INTERRUPT-ON-CHANGE FLAGS (PORTA EXAMPLE)

```
MOVlw 0xff
XORwf IOCAF, W
ANDwf IOCAF, F
```

13.5 Operation in Sleep

The interrupt-on-change interrupt sequence will wake the device from Sleep mode, if the IOCIE bit is set.

If an edge is detected while in Sleep mode, the IOCAF register will be updated prior to the first instruction executed out of Sleep.

---

**FIGURE 13-1: INTERRUPT-ON-CHANGE BLOCK DIAGRAM**
REGISTER 13-1: IOCAP: INTERRUPT-ON-CHANGE PORTA POSITIVE EDGE REGISTER

<table>
<thead>
<tr>
<th></th>
<th></th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>IOCAP5</td>
<td>IOCAP4</td>
<td>IOCAP3</td>
<td>IOCAP2</td>
<td>IOCAP1</td>
<td>IOCAP0</td>
</tr>
</tbody>
</table>

definitions:
- **R** = Readable bit
- **W** =Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

Legend:
- **Unimplemented**: Read as ‘0’
- **bit 7-6**: Unimplemented: Read as ‘0’
- **bit 5-0**: IOCAP<5:0>: Interrupt-on-Change PORTA Positive Edge Enable bits
  - 1 = Interrupt-on-Change enabled on the pin for a positive going edge. Associated Status bit and interrupt flag will be set upon detecting an edge.
  - 0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-2: IOCAN: INTERRUPT-ON-CHANGE PORTA NEGATIVE EDGE REGISTER

<table>
<thead>
<tr>
<th></th>
<th></th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>IOCAN5</td>
<td>IOCAN4</td>
<td>IOCAN3</td>
<td>IOCAN2</td>
<td>IOCAN1</td>
<td>IOCAN0</td>
</tr>
</tbody>
</table>

definitions:
- **R** = Readable bit
- **W** =Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

Legend:
- **Unimplemented**: Read as ‘0’
- **bit 7-6**: Unimplemented: Read as ‘0’
- **bit 5-0**: IOCAN<5:0>: Interrupt-on-Change PORTA Negative Edge Enable bits
  - 1 = Interrupt-on-Change enabled on the pin for a negative going edge. Associated Status bit and interrupt flag will be set upon detecting an edge.
  - 0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-3: IOCAF: INTERRUPT-ON-CHANGE PORTA FLAG REGISTER

<table>
<thead>
<tr>
<th></th>
<th></th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
<th>R/W/HS-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>IOCAF5</td>
<td>IOCAF4</td>
<td>IOCAF3</td>
<td>IOCAF2</td>
<td>IOCAF1</td>
<td>IOCAF0</td>
</tr>
</tbody>
</table>

definitions:
- **R** = Readable bit
- **W** =Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **HS** - Bit is set in hardware

Legend:
- **Unimplemented**: Read as ‘0’
- **bit 7-6**: Unimplemented: Read as ‘0’
- **bit 5-0**: IOCAF<5:0>: Interrupt-on-Change PORTA Flag bits
  - 1 = An enabled change was detected on the associated pin.
    Set when IOCAPx = 1 and a rising edge was detected on RAx, or when IOCANx = 1 and a falling edge was detected on RAx.
  - 0 = No change was detected, or the user cleared the detected change.
### TABLE 13-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPT-ON-CHANGE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA4</td>
<td>—</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>107</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>IOCAF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>IOCAF5</td>
<td>IOCAF4</td>
<td>IOCAF3</td>
<td>IOCAF2</td>
<td>IOCAF1</td>
<td>IOCAF0</td>
</tr>
<tr>
<td>IOCAN</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>IOCAN5</td>
<td>IOCAN4</td>
<td>IOCAN3</td>
<td>IOCAN2</td>
<td>IOCAN1</td>
<td>IOCAN0</td>
</tr>
<tr>
<td>IOCAP</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>IOCAP5</td>
<td>IOCAP4</td>
<td>IOCAP3</td>
<td>IOCAP2</td>
<td>IOCAP1</td>
<td>IOCAP0</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used by interrupt-on-change.
14.0 FIXED VOLTAGE REFERENCE (FVR)

The Fixed Voltage Reference, or FVR, is a stable voltage reference, independent of VDD, with 1.024V, 2.048V or 4.096V selectable output levels. The output of the FVR can be configured to supply a reference voltage to the following:

- ADC input channel
- ADC positive reference
- Comparator positive input
- Digital-to-Analog Converter (DAC)
- Capacitive Sensing (CPS) module

The FVR can be enabled by setting the FVREN bit of the FVRCON register.

14.1 Independent Gain Amplifiers

The output of the FVR supplied to the ADC, Comparators, DAC and CPS module is routed through two independent programmable gain amplifiers. Each amplifier can be configured to amplify the reference voltage by 1x, 2x or 4x, to produce the three possible voltage levels.

The ADFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the ADC module. Reference Section 16.0 “Analog-to-Digital Converter (ADC) Module” for additional information.

The CDAFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the Comparators, DAC, and CPS module. Reference Section 17.0 “Digital-to-Analog Converter (DAC) Module”, Section 19.0 “Comparator Module” and Section 17.0 “Digital-to-Analog Converter (DAC) Module” for additional information.

14.2 FVR Stabilization Period

When the Fixed Voltage Reference module is enabled, it requires time for the reference and amplifier circuits to stabilize. Once the circuits stabilize and are ready for use, the FVRRDY bit of the FVRCON register will be set. See Section 31.0 “Electrical Specifications” for the minimum delay requirement.

---

**FIGURE 14-1: VOLTAGE REFERENCE BLOCK DIAGRAM**

![Block Diagram of Fixed Voltage Reference (FVR)](image-url)
## REGISTER 14-1: FVRCON: FIXED VOLTAGE REFERENCE CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>FVREN</td>
<td>Fixed Voltage Reference Enable bit</td>
</tr>
<tr>
<td></td>
<td>FVREN</td>
<td>0 = Fixed Voltage Reference is disabled</td>
</tr>
<tr>
<td></td>
<td>FVREN</td>
<td>1 = Fixed Voltage Reference is enabled</td>
</tr>
<tr>
<td>6</td>
<td>FVRDDY</td>
<td>Fixed Voltage Reference Ready Flag bit</td>
</tr>
<tr>
<td></td>
<td>FVRDDY</td>
<td>0 = Fixed Voltage Reference output is not ready or not enabled</td>
</tr>
<tr>
<td></td>
<td>FVRDDY</td>
<td>1 = Fixed Voltage Reference output is ready for use</td>
</tr>
<tr>
<td>5</td>
<td>TSEN</td>
<td>Temperature Indicator Enable bit</td>
</tr>
<tr>
<td></td>
<td>TSEN</td>
<td>0 = Temperature Indicator is disabled</td>
</tr>
<tr>
<td></td>
<td>TSEN</td>
<td>1 = Temperature Indicator is enabled</td>
</tr>
<tr>
<td>4</td>
<td>TSRNG</td>
<td>Temperature Indicator Range Selection bit</td>
</tr>
<tr>
<td></td>
<td>TSRNG</td>
<td>0 = VOUT = VDD - 2VT (Low Range)</td>
</tr>
<tr>
<td></td>
<td>TSRNG</td>
<td>1 = VOUT = VDD - 4VT (High Range)</td>
</tr>
<tr>
<td>3-2</td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>Comparator and DAC Fixed Voltage Reference Selection bits</td>
</tr>
<tr>
<td></td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>00 = Comparator, DAC and CPS module Fixed Voltage Reference Peripheral output is off</td>
</tr>
<tr>
<td></td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>01 = Comparator, DAC and CPS module Fixed Voltage Reference Peripheral output is 1x (1.024V)</td>
</tr>
<tr>
<td></td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>10 = Comparator, DAC and CPS module Fixed Voltage Reference Peripheral output is 2x (2.048V)</td>
</tr>
<tr>
<td></td>
<td>CDAFVR&lt;1:0&gt;</td>
<td>11 = Reserved</td>
</tr>
<tr>
<td>1-0</td>
<td>ADFVR&lt;1:0&gt;</td>
<td>ADC Fixed Voltage Reference Selection bits</td>
</tr>
<tr>
<td></td>
<td>ADFVR&lt;1:0&gt;</td>
<td>00 = ADC Fixed Voltage Reference Peripheral output is off</td>
</tr>
<tr>
<td></td>
<td>ADFVR&lt;1:0&gt;</td>
<td>01 = ADC Fixed Voltage Reference Peripheral output is 1x (1.024V)</td>
</tr>
<tr>
<td></td>
<td>ADFVR&lt;1:0&gt;</td>
<td>10 = ADC Fixed Voltage Reference Peripheral output is 2x (2.048V)</td>
</tr>
<tr>
<td></td>
<td>ADFVR&lt;1:0&gt;</td>
<td>11 = Reserved</td>
</tr>
</tbody>
</table>

**Legend:**
- `R` = Readable bit
- `W` = Writable bit
- `U` = Unimplemented bit, read as ‘0’
- `u` = Bit is unchanged
- `x` = Bit is unknown
- `-n/n` = Value at POR and BOR/Value at all other Resets
- `'1'` = Bit is set
- `'0'` = Bit is cleared
- `q` = Value depends on condition

**Note 1:** Fixed Voltage Reference output cannot exceed VDD.
15.0 TEMPERATURE INDICATOR MODULE

This family of devices is equipped with a temperature circuit designed to measure the operating temperature of the silicon die. The circuit's range of operating temperature falls between -40°C and +85°C. The output is a voltage that is proportional to the device temperature. The output of the temperature indicator is internally connected to the device ADC.

The circuit may be used as a temperature threshold detector or a more accurate temperature indicator, depending on the level of calibration performed. A one-point calibration allows the circuit to indicate a temperature closely surrounding that point. A two-point calibration allows the circuit to sense the entire range of temperature more accurately. Reference Application Note AN1333, “Use and Calibration of the Internal Temperature Indicator” (DS01333) for more details regarding the calibration process.

15.1 Circuit Operation

Figure 15-1 shows a simplified block diagram of the temperature circuit. The proportional voltage output is achieved by measuring the forward voltage drop across multiple silicon junctions.

Equation 15-1 describes the output characteristics of the temperature indicator.

**EQUATION 15-1: Vout Ranges**

<table>
<thead>
<tr>
<th>Range</th>
<th>Equation</th>
</tr>
</thead>
<tbody>
<tr>
<td>High Range</td>
<td>Vout = VDD - 4VT</td>
</tr>
<tr>
<td>Low Range</td>
<td>Vout = VDD - 2VT</td>
</tr>
</tbody>
</table>

The temperature sense circuit is integrated with the Fixed Voltage Reference (FVR) module. See Section 14.0 “Fixed Voltage Reference (FVR)” for more information.

The circuit is enabled by setting the TSEN bit of the FVRCON register. When disabled, the circuit draws no current.

The circuit operates in either high or low range. The high range, selected by setting the TSRNG bit of the FVRCON register, provides a wider output voltage. This provides more resolution over the temperature range, but may be less consistent from part to part. This range requires a higher bias voltage to operate and thus, a higher VDD is needed.

The low range is selected by clearing the TSRNG bit of the FVRCON register. The low range generates a lower voltage drop and thus, a lower bias voltage is needed to operate the circuit. The low range is provided for low voltage operation.

15.2 Minimum Operating VDD vs. Minimum Sensing Temperature

When the temperature circuit is operated in low range, the device may be operated at any operating voltage that is within specifications.

When the temperature circuit is operated in high range, the device operating voltage, VDD, must be high enough to ensure that the temperature circuit is correctly biased.

Table 15-1 shows the recommended minimum VDD vs. range setting.

**TABLE 15-1: RECOMMENDED VDD VS. RANGE**

<table>
<thead>
<tr>
<th>Minimum VDD, TSRNG = 1</th>
<th>Minimum VDD, TSRNG = 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>3.6V</td>
<td>1.8V</td>
</tr>
</tbody>
</table>

15.3 Temperature Output

The output of the circuit is measured using the internal Analog-to-Digital Converter. Channel 29 is reserved for the temperature circuit output. Refer to Section 16.0 “Analog-to-Digital Converter (ADC) Module” for detailed information.

15.3.1 ACQUISITION TIME

The user must wait at least 200 μsec when the ADC input multiplexer is switched to the temperature indicator output. This wait time is also required between sequential conversions of the temperature indicator output voltage.
16.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE

The Analog-to-Digital Converter (ADC) allows conversion of an analog input signal to a 10-bit binary representation of that signal. This device uses analog inputs, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates a 10-bit binary result via successive approximation and stores the conversion result into the ADC result registers (ADRESH:ADRESL register pair). Figure 16-1 shows the block diagram of the ADC.

The ADC voltage reference is software selectable to be either internally generated or externally supplied.

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.

FIGURE 16-1: ADC BLOCK DIAGRAM

Note 1: When ADON = 0, all multiplexer inputs are disconnected.
16.1 ADC Configuration

When configuring and using the ADC the following functions must be considered:

- Port configuration
- Channel selection
- ADC voltage reference selection
- ADC conversion clock source
- Interrupt control
- Result formatting

16.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin should be configured for analog by setting the associated TRIS and ANSEL bits. Refer to Section 12.0 “I/O Ports” for more information.

16.1.2 CHANNEL SELECTION

There are seven channel selections available:

- AN<3:0> pins
- Temperature Indicator
- DAC Output
- FVR (Fixed Voltage Reference) Output

Refer to Section 17.0 “Digital-to-Analog Converter (DAC) Module”, Section 14.0 “Fixed Voltage Reference (FVR)” and Section 15.0 “Temperature Indicator Module” for more information on these channel selections.

The CHS bits of the ADCON0 register determine which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion. Refer to Section 16.2 “ADC Operation” for more information.

16.1.3 ADC VOLTAGE REFERENCE

The ADPREF bits of the ADCON1 register provides control of the positive voltage reference. The positive voltage reference can be:

- VREF+ pin
- VDD
- FVR 2.048V

See Section 14.0 “Fixed Voltage Reference (FVR)” for more details on the Fixed Voltage Reference.

16.1.4 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCS bits of the ADCON1 register. There are seven possible clock options:

- Fosc/2
- Fosc/4
- Fosc/8
- Fosc/16
- Fosc/32
- Fosc/64
- FRC (dedicated internal oscillator)

The time to complete one bit conversion is defined as TAD. One full 10-bit conversion requires 11.5 TAD periods as shown in Figure 16-2.

For correct conversion, the appropriate TAD specification must be met. Refer to the A/D conversion requirements in Section 31.0 “Electrical Specifications” for more information. Table 16-1 gives examples of appropriate ADC clock selections.

Note: Analog voltages on any pin that is defined as a digital input may cause the input buffer to conduct excess current.

Note: Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.
**TABLE 16-1: ADC CLOCK PERIOD (TAD) VS. DEVICE OPERATING FREQUENCIES**

<table>
<thead>
<tr>
<th>ADC Clock Source</th>
<th>ADCS&lt;2:0&gt;</th>
<th>Device Frequency (Fosc)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>32 MHz</td>
</tr>
<tr>
<td>Fosc/2</td>
<td>000</td>
<td>62.5 ns(2)</td>
</tr>
<tr>
<td>Fosc/4</td>
<td>100</td>
<td>125 ns(2)</td>
</tr>
<tr>
<td>Fosc/8</td>
<td>001</td>
<td>0.5 µs(2)</td>
</tr>
<tr>
<td>Fosc/16</td>
<td>101</td>
<td>800 ns</td>
</tr>
<tr>
<td>Fosc/32</td>
<td>010</td>
<td>1.0 µs</td>
</tr>
<tr>
<td>Fosc/64</td>
<td>110</td>
<td>2.0 µs</td>
</tr>
<tr>
<td>FRC</td>
<td>x11</td>
<td>1.0-6.0 µs(1,4)</td>
</tr>
</tbody>
</table>

**Legend:** Shaded cells are outside of recommended range.

**Note 1:** The FRC source has a typical TAD time of 1.6 µs for VDD.

**Note 2:** These values violate the minimum required TAD time.

**Note 3:** For faster conversion times, the selection of another clock source is recommended.

**Note 4:** The ADC clock period (TAD) and total ADC conversion time can be minimized when the ADC clock is derived from the system clock FOSC. However, the FRC clock source must be used when conversions are to be performed with the device in Sleep mode.

**FIGURE 16-2: ANALOG-TO-DIGITAL CONVERSION TAD CYCLES**

Conversion starts

Holding capacitor is disconnected from analog input (typically 100 ns)

Set GO bit

On the following cycle:

ADRESH:ADRESL is loaded, GO bit is cleared,
ADIF bit is set, holding capacitor is connected to analog input.
16.1.5 INTERRUPTS
The ADC module allows for the ability to generate an interrupt upon completion of an Analog-to-Digital conversion. The ADC Interrupt Flag is the ADIF bit in the PIR1 register. The ADC Interrupt Enable is the ADIE bit in the PIE1 register. The ADIF bit must be cleared in software.

Note 1: The ADIF bit is set at the completion of every conversion, regardless of whether or not the ADC interrupt is enabled.

2: The ADC operates during Sleep only when the FrC oscillator is selected.

This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the GIE and PEIE bits of the INTCON register must be disabled. If the GIE and PEIE bits of the INTCON register are enabled, execution will switch to the Interrupt Service Routine.

16.1.6 RESULT FORMATTING
The 10-bit A/D conversion result can be supplied in two formats, left justified or right justified. The ADFM bit of the ADCON1 register controls the output format.

Figure 16-3 shows the two output formats.
16.2 ADC Operation

16.2.1 STARTING A CONVERSION
To enable the ADC module, the ADON bit of the ADCON0 register must be set to a ‘1’. Setting the GO/DONE bit of the ADCON0 register to a ‘1’ will start the Analog-to-Digital conversion.

**Note:** The GO/DONE bit should not be set in the same instruction that turns on the ADC. Refer to Section 16.2.6 “A/D Conversion Procedure”.

16.2.2 COMPLETION OF A CONVERSION
When the conversion is complete, the ADC module will:
- Clear the GO/DONE bit
- Set the ADIF Interrupt Flag bit
- Update the ADRESH and ADRESL registers with new conversion result

16.2.3 TERMINATING A CONVERSION
If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRESH and ADRESL registers will be updated with the partially complete Analog-to-Digital conversion sample. Incomplete bits will match the last bit converted.

**Note:** A device Reset forces all registers to their Reset state. Thus, the ADC module is turned off and any pending conversion is terminated.

16.2.4 ADC OPERATION DURING SLEEP
The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC clock source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

16.2.5 SPECIAL EVENT TRIGGER
The Special Event Trigger of the CCPx/ECCPX module allows periodic ADC measurements without software intervention. When this trigger occurs, the GO/DONE bit is set by hardware and the Timer1 counter resets to zero.

**TABLE 16-2: SPECIAL EVENT TRIGGER**

<table>
<thead>
<tr>
<th>Device</th>
<th>ECCP1</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC12LF1840T39A</td>
<td>ECCP1</td>
</tr>
</tbody>
</table>

Using the Special Event Trigger does not assure proper ADC timing. It is the user’s responsibility to ensure that the ADC timing requirements are met.

Refer to Section 24.0 “Capture/Compare/PWM Modules” for more information.
16.2.6 A/D CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an Analog-to-Digital conversion:

1. Configure Port:
   • Disable pin output driver (Refer to the TRIS register)
   • Configure pin as analog (Refer to the ANSEL register)

2. Configure the ADC module:
   • Select ADC conversion clock
   • Configure voltage reference
   • Select ADC input channel
   • Turn on ADC module

3. Configure ADC interrupt (optional):
   • Clear ADC interrupt flag
   • Enable ADC interrupt
   • Enable peripheral interrupt
   • Enable global interrupt(1)

4. Wait the required acquisition time(2).

5. Start conversion by setting the GO/DONE bit.

6. Wait for ADC conversion to complete by one of the following:
   • Polling the GO/DONE bit
   • Waiting for the ADC interrupt (interrupts enabled)

7. Read ADC Result.

8. Clear the ADC interrupt flag (required if interrupt is enabled).

Note 1: The global interrupt can be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.

2: Refer to Section 16.3 “A/D Acquisition Requirements”.

EXAMPLE 16-1: A/D CONVERSION

```
;This code block configures the ADC
;for polling, Vdd and Vss references, Frc
;clock and AN0 input.
;
;Conversion start & polling for completion
; are included.
;
BANKSEL ADCON1 ;
MOVLW B'11110000' ;Right justify, Frc
MOVWF ADCON1 ;clock
MOVF ADCON1 ;Vdd and Vss Vref
BANKSEL TRISA ;
BSF TRISA,0 ;Set RA0 to input
BANKSEL ANSEL ;
BSF ANSEL,0 ;Set RA0 to analog
BANKSEL ADCON0 ;
MOVLW B'00000001' ;Select channel AN0
MOVF ADCON0 ;Turn ADC On
CALL SampleTime ;Acquisiton delay
BSF ADCON0,ADGO ;Start conversion
BTFSC ADCON0,ADGO ;Is conversion done?
GOTO $-1 ;No, test again
BANKSEL ADRESH ;
MOVF ADRESH,W ;Read upper 2 bits
MOVF RESULTHI ;store in GPR space
BANKSEL ADRESL ;
MOVF ADRESL,W ;Read lower 8 bits
MOVF RESULTLO ;Store in GPR space
```
16.2.7 ADC REGISTER DEFINITIONS

The following registers are used to control the operation of the ADC.

REGISTER 16-1: ADCON0: A/D CONTROL REGISTER 0

<table>
<thead>
<tr>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>CHS&lt;4:0&gt;</td>
<td>GO/DONE</td>
<td>ADON</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

bit 7 Unimplemented: Read as '0'

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

bit 6-2 CHS<4:0>: Analog Channel Select bits

- 00000 = AN0
- 00001 = AN1
- 00010 = AN2
- 00011 = AN3
- 00100 = Reserved. No channel connected.
  -
  -
- 11100 = Reserved. No channel connected.
- 11101 = Temperature Indicator (3)
- 11110 = DAC output (1)
- 11111 = FVR (Fixed Voltage Reference) Buffer 1 Output (2)

bit 1 GO/DONE: A/D Conversion Status bit

- 1 = A/D conversion cycle in progress. Setting this bit starts an A/D conversion cycle.
- 0 = A/D conversion completed/not in progress

bit 0 ADON: ADC Enable bit

- 1 = ADC is enabled
- 0 = ADC is disabled and consumes no operating current

Note 1: See Section 17.0 “Digital-to-Analog Converter (DAC) Module” for more information.

Note 2: See Section 14.0 “Fixed Voltage Reference (FVR)” for more information.

Note 3: See Section 15.0 “Temperature Indicator Module” for more information.
### REGISTER 16-2: ADCON1: A/D CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADFM</td>
<td>ADCS&lt;2:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td>ADPREF&lt;1:0&gt;</td>
<td></td>
</tr>
</tbody>
</table>

#### Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

**bit 7**

**ADFM**: A/D Result Format Select bit

<table>
<thead>
<tr>
<th>Value</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Right justified. Six Most Significant bits of ADRESH are set to ‘0’ when the conversion result is loaded.</td>
</tr>
<tr>
<td>0</td>
<td>Left justified. Six Least Significant bits of ADRESL are set to ‘0’ when the conversion result is loaded.</td>
</tr>
</tbody>
</table>

**bit 6-4**

**ADCS<2:0>**: A/D Conversion Clock Select bits

<table>
<thead>
<tr>
<th>Value</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>Fosc/2</td>
</tr>
<tr>
<td>001</td>
<td>Fosc/8</td>
</tr>
<tr>
<td>010</td>
<td>Fosc/32</td>
</tr>
<tr>
<td>011</td>
<td>FRC (clock supplied from a dedicated RC oscillator)</td>
</tr>
<tr>
<td>100</td>
<td>Fosc/4</td>
</tr>
<tr>
<td>101</td>
<td>Fosc/16</td>
</tr>
<tr>
<td>110</td>
<td>Fosc/64</td>
</tr>
<tr>
<td>111</td>
<td>FRC (clock supplied from a dedicated RC oscillator)</td>
</tr>
</tbody>
</table>

**bit 3-2**

**Unimplemented**: Read as ‘0’

**bit 1-0**

**ADPREF<1:0>**: A/D Positive Voltage Reference Configuration bits

<table>
<thead>
<tr>
<th>Value</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>VREF is connected to VDD</td>
</tr>
<tr>
<td>01</td>
<td>Reserved</td>
</tr>
<tr>
<td>10</td>
<td>VREF is connected to external VREF pin(1)</td>
</tr>
<tr>
<td>11</td>
<td>VREF is connected to internal Fixed Voltage Reference (FVR) module(1)</td>
</tr>
</tbody>
</table>

#### Note 1:
When selecting the FVR or the VREF+ pin as the source of the positive reference, be aware that a minimum voltage specification exists. See Section 31.0 “Electrical Specifications” for details.
**REGISTER 16-3: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 0**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>Upper eight bits of 10-bit conversion result</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- '-n/n' = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

**REGISTER 16-4: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 0**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7-0</td>
<td>Lower two bits of 10-bit conversion result</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- '-n/n' = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

**Reserved: Do not use.**
### REGISTER 16-5: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 1

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>ADRES&lt;9:8&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

- **bit 7-2** **Reserved**: Do not use.
- **bit 1-0** **ADRES<9:8>:** ADC Result Register bits
  - Upper two bits of 10-bit conversion result

### REGISTER 16-6: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 1

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>ADRES&lt;7:0&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

- **bit 7-0** **ADRES<7:0>:** ADC Result Register bits
  - Lower eight bits of 10-bit conversion result
16.3 A/D Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 16-4. The source impedance (Rs) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), refer to Figure 16-4. The maximum recommended impedance for analog sources is 10 kΩ. As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an A/D acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 16-1 may be used. This equation assumes that 1/2 LSb error is used (1,024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

**EQUATION 16-1: ACQUISITION TIME EXAMPLE**

Assumptions:  
Temperature = 50°C and external impedance of 10kΩ 5.0V VDD  

\[ T_{ACQ} = \text{Amplifier Settling Time} + \text{Hold Capacitor Charging Time} + \text{Temperature Coefficient} \]
\[ = T_{AMP} + T_c + T_{COFF} \]
\[ = 2\mu s + T_c + [(\text{Temperature} - 25°C)(0.05\mu s/°C)] \]

The value for \( T_c \) can be approximated with the following equations:

\[ V_{APPLIED}\left(1 - \frac{1}{2^n + 1}\right) = V_{CHOLD} \quad ;[1] \text{CHOLD charged to within 1/2 lsb} \]

\[ V_{APPLIED}\left(1 - e^{-\frac{T_c}{RC}}\right) = V_{CHOLD} \quad ;[2] \text{CHOLD charge response to } V_{APPLIED} \]

\[ V_{APPLIED}\left(1 - e^{-\frac{T_c}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{2^n + 1}\right) \quad ;\text{combining } [1] \text{ and } [2] \]

Note: Where \( n = \text{number of bits of the ADC}. \)

Solving for \( T_c \):

\[ T_c = -CHOLD(R_{IC} + RSS + RS) \ln(1/2047) \]
\[ = -13.5pF(1kΩ + 7kΩ + 10kΩ) \ln(0.0004885) \]
\[ = 1.85\mu s \]

Therefore:

\[ T_{ACQ} = 2\mu s + 1.85\mu s + [(50°C - 25°C)(0.05\mu s/°C)] \]
\[ = 5.1\mu s \]

**Note 1:** The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

**2:** The charge holding capacitor (CHOLD) is not discharged after each conversion.

**3:** The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.
FIGURE 16-4: ANALOG INPUT MODEL

Legend:
- **CHOLD** = Sample/Hold Capacitance
- **CPIN** = Input Capacitance
- **IL LEAKAGE** = Leakage current at the pin due to various junctions
- **Ric** = Interconnect Resistance
- **RSS** = Resistance of Sampling Switch
- **SS** = Sampling Switch
- **VT** = Threshold Voltage

Note 1: Refer to Section 31.0 “Electrical Specifications”.

FIGURE 16-5: ADC TRANSFER FUNCTION
**TABLE 16-3: SUMMARY OF REGISTERS ASSOCIATED WITH ADC**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADCON0</td>
<td>—</td>
<td>CHS4</td>
<td>CHS3</td>
<td>CHS2</td>
<td>CHS1</td>
<td>CHS0</td>
<td>GO/DONE</td>
<td>ADON</td>
<td>121</td>
</tr>
<tr>
<td>ADCON1</td>
<td>ADFM</td>
<td>ADCS2</td>
<td>ADCS1</td>
<td>ADCS0</td>
<td>—</td>
<td>—</td>
<td>ADPREF1</td>
<td>ADPREF0</td>
<td>122</td>
</tr>
<tr>
<td>ADRESH</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>123, 124</td>
</tr>
<tr>
<td>ADRESL</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>123, 124</td>
</tr>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA4</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
</tr>
<tr>
<td>CCP1CON</td>
<td>P1M1</td>
<td>P1M0</td>
<td>DC1B1</td>
<td>DC1B0</td>
<td>CCP1M3</td>
<td>CCP1M2</td>
<td>CCP1M1</td>
<td>CCP1M0</td>
<td>192</td>
</tr>
<tr>
<td>DACCON0</td>
<td>DACEN</td>
<td>DACLPS</td>
<td>DACOE</td>
<td>—</td>
<td>DACPSS1</td>
<td>DACPSS0</td>
<td>—</td>
<td>DACNSS</td>
<td>131</td>
</tr>
<tr>
<td>DACCON1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>DACR4</td>
<td>DACR3</td>
<td>DACR2</td>
<td>DACR1</td>
<td>DACR0</td>
</tr>
<tr>
<td>FVRCON</td>
<td>FVREN</td>
<td>FVRRDY</td>
<td>TSEN</td>
<td>TSRNG</td>
<td>CDAFVR1</td>
<td>CDAFVR0</td>
<td>ADFVR1</td>
<td>ADFVR0</td>
<td>113</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented read as ‘0’. Shaded cells are not used for ADC module.
17.0 DIGITAL-TO-ANALOG CONVERTER (DAC) MODULE

The Digital-to-Analog Converter supplies a variable voltage reference, ratiometric with the input source, with 32 selectable output levels.

The input of the DAC can be connected to:

- External VREF pins
- VDD supply voltage
- FVR (Fixed Voltage Reference)

The output of the DAC can be configured to supply a reference voltage to the following:

- Comparator positive input
- ADC input channel
- DACOUT pin
- Capacitive Sensing (CPS) module

The Digital-to-Analog Converter (DAC) can be enabled by setting the DACEN bit of the DACCON0 register.

17.1 Output Voltage Selection

The DAC has 32 voltage level ranges. The 32 levels are set with the DACR<4:0> bits of the DACCON1 register.

The DAC output voltage is determined by the following equations:

**EQUATION 17-1: DAC OUTPUT VOLTAGE**

\[
\text{IF } \text{DACEN} = 1 \\
V_{\text{OUT}} = (V_{\text{SOURCE}+} - V_{\text{SOURCE}-}) \times \frac{\text{DACR}[4:0]}{2^5} + V_{\text{SOURCE}-}
\]

\[
\text{IF } \text{DACEN} = 0 \text{ & DACLPS = 1 & DACR}[4:0] = 11111 \\
V_{\text{OUT}} = V_{\text{SOURCE}+}
\]

\[
\text{IF } \text{DACEN} = 0 \text{ & DACLPS = 0 & DACR}[4:0] = 00000 \\
V_{\text{OUT}} = V_{\text{SOURCE}-}
\]

\[V_{\text{SOURCE}+} = \text{VD}	ext{D, VREF, or FVR BUFFER 2}\]

\[V_{\text{SOURCE}-} = \text{VSS}\]

17.2 Ratiometric Output Level

The DAC output value is derived using a resistor ladder with each end of the ladder tied to a positive and negative voltage reference input source. If the voltage of either input source fluctuates, a similar fluctuation will result in the DAC output value.

The value of the individual resistors within the ladder can be found in Section 31.0 “Electrical Specifications”.

17.3 DAC Voltage Reference Output

The DAC can be output to the DACOUT pin by setting the DACOE bit of the DACCON0 register to ‘1’. Selecting the DAC reference voltage for output on the DACOUT pin automatically overrides the digital output buffer and digital input threshold detector functions of that pin. Reading the DACOUT pin when it has been configured for DAC reference voltage output will always return a ‘0’.

Due to the limited current drive capability, a buffer must be used on the DAC voltage reference output for external connections to DACOUT. Figure 17-2 shows an example buffering technique.
FIGURE 17-1: DIGITAL-TO-ANALOG CONVERTER BLOCK DIAGRAM

Digital-to-Analog Converter (DAC)

FVR BUFFER2
VDD
VREF
DACPSS<1:0>
DACEN
DACLPS

VSS

VSOURCE+

DACR<4:0>

32-to-1 MUX

Steps

(VDD, CPS and ADC Modules)

DACPSS<1:0>

DACLPS

DACEN

VDD

VREF

VSOURCE−

VSS

FIGURE 17-2: VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE

PIC® MCU

DAC Module

R

Voltage Reference Output Impedance

DACOUT

Buffered DAC Output
17.4 Low-Power Voltage State

In order for the DAC module to consume the least amount of power, one of the two voltage reference input sources to the resistor ladder must be disconnected. Either the positive voltage source, \((V_{SOURCE}^+)\), or the negative voltage source, \((V_{SOURCE}^-)\), can be disabled. The negative voltage source is disabled by setting the DACLPS bit in the DACCON0 register. Clearing the DACLPS bit in the DACCON0 register disables the positive voltage source.

17.4.1 OUTPUT CLAMPED TO POSITIVE VOLTAGE SOURCE

The DAC output voltage can be set to \(V_{SOURCE}^+\) with the least amount of power consumption by performing the following:

- Clearing the DACEN bit in the DACCON0 register.
- Setting the DACLPS bit in the DACCON0 register.
- Configuring the DACPS bits to the proper positive source.
- Configuring the Dacr<4:0> bits to ‘11111’ in the DACCON1 register.

This is also the method used to output the voltage level from the FVR to an output pin. See Section 17.5 “Operation During Sleep” for more information. Reference Figure 17-3 for output clamping examples.

17.4.2 OUTPUT CLAMPED TO NEGATIVE VOLTAGE SOURCE

The DAC output voltage can be set to \(V_{SOURCE}^-\) with the least amount of power consumption by performing the following:

- Clearing the DACEN bit in the DACCON0 register.
- Clearing the DACLPS bit in the DACCON0 register.
- Configuring the Dacr<4:0> bits to ‘00000’ in the DACCON1 register.

This allows the comparator to detect a zero-crossing while not consuming additional current through the DAC module. Reference Figure 17-3 for output clamping examples.

FIGURE 17-3: OUTPUT VOLTAGE CLAMPING EXAMPLES

17.5 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the DACCON0 register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

17.6 Effects of a Reset

A device Reset affects the following:

- DAC is disabled.
- DAC output voltage is removed from the DACOUT pin.
- The Dacr<4:0> range select bits are cleared.
### REGISTER 17-1: DACCON0: VOLTAGE REFERENCE CONTROL REGISTER 0

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>DACEN</td>
<td>DACLPS</td>
<td>DACOE</td>
<td>—</td>
<td>—</td>
<td>DACPSS&lt;1:0&gt;</td>
<td>—</td>
<td>DACNSS</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>U-0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>U-0</td>
<td>R/W-0/0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

<table>
<thead>
<tr>
<th>Bit 7</th>
<th><strong>DACEN</strong>: DAC Enable bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>DAC is enabled</td>
</tr>
<tr>
<td>0</td>
<td>DAC is disabled</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 6</th>
<th><strong>DACLPS</strong>: DAC Low-Power Voltage State Select bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>DAC Positive reference source selected</td>
</tr>
<tr>
<td>0</td>
<td>DAC Negative reference source selected</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 5</th>
<th><strong>DACOE</strong>: DAC Voltage Output Enable bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>DAC voltage level is also an output on the DACOUT pin</td>
</tr>
<tr>
<td>0</td>
<td>DAC voltage level is disconnected from the DACOUT pin</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 4</th>
<th><strong>Unimplemented</strong>: Read as '0'</th>
</tr>
</thead>
</table>

<table>
<thead>
<tr>
<th>Bit 3-2</th>
<th><strong>DACPSS&lt;1:0&gt;</strong>: DAC Positive Source Select bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>VDD</td>
</tr>
<tr>
<td>01</td>
<td>VREF pin</td>
</tr>
<tr>
<td>10</td>
<td>FVR Buffer2 output</td>
</tr>
<tr>
<td>11</td>
<td>Reserved, do not use</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit 1</th>
<th><strong>Unimplemented</strong>: Read as '0'</th>
</tr>
</thead>
</table>

<table>
<thead>
<tr>
<th>Bit 0</th>
<th><strong>DACNSS</strong>: DAC Negative Source Select bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>VREF-</td>
</tr>
<tr>
<td>0</td>
<td>VSS</td>
</tr>
</tbody>
</table>

### REGISTER 17-2: DACCON1: VOLTAGE REFERENCE CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>—</th>
<th>—</th>
<th>—</th>
<th>DACR&lt;4:0&gt;</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>R/W-0/0</th>
</tr>
</thead>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared

<table>
<thead>
<tr>
<th>Bit 7-5</th>
<th><strong>Unimplemented</strong>: Read as '0'</th>
</tr>
</thead>
</table>

| Bit 4-0 | **DACR<4:0>**: DAC Voltage Output Select bits |

### TABLE 17-1: SUMMARY OF REGISTERS ASSOCIATED WITH THE DAC MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on page</th>
</tr>
</thead>
<tbody>
<tr>
<td>FVRCON</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>113</td>
</tr>
<tr>
<td>DACCON0</td>
<td>DACEN</td>
<td>DACLPS</td>
<td>DACOE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>131</td>
</tr>
<tr>
<td>DACCON1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>DACR4</td>
<td>DACR3</td>
<td>DACR2</td>
<td>DACR1</td>
<td>DACR0</td>
</tr>
</tbody>
</table>

**Legend:**
- **—** = unimplemented, read as '0'. Shaded cells are unused by the DAC module.
18.0 SR LATCH
The module consists of a single SR Latch with multiple Set and Reset inputs as well as separate latch outputs. The SR Latch module includes the following features:

- Programmable input selection
- SR Latch output is available externally
- Separate Q and Q̅ outputs
- Firmware Set and Reset

The SR Latch can be used in a variety of analog applications, including oscillator circuits, one-shot circuit, hysteretic controllers, and analog timing applications.

18.1 Latch Operation
The latch is a Set-Reset Latch that does not depend on a clock source. Each of the Set and Reset inputs are active-high. The latch can be set or reset by:

- Software control (SRPS and SRPR bits)
- Comparator C1 output (SYNCC1OUT)
- SRI pin
- Programmable clock (SRCLK)

The SRPS and the SRPR bits of the SRCON0 register may be used to set or reset the SR Latch, respectively. The latch is Reset-dominant. Therefore, if both Set and Reset inputs are high, the latch will go to the Reset state. Both the SRPS and SRPR bits are self resetting which means that a single write to either of the bits is all that is necessary to complete a latch Set or Reset operation.

The output from Comparator C1 can be used as the Set or Reset inputs of the SR Latch. The output of the Comparator can be synchronized to the Timer1 clock source. See Section 19.0 “Comparator Module” and Section 21.0 “Timer1 Module with Gate Control” for more information.

An external source on the SRI pin can be used as the Set or Reset inputs of the SR Latch.

An internal clock source is available that can periodically set or reset the SR Latch. The SRCLK<2:0> bits in the SRCON0 register are used to select the clock source period. The SRSCKE and SRRCKE bits of the SRCON1 register enable the clock source to set or reset the SR Latch, respectively.

18.2 Latch Output
The SRQEN and SRNQEN bits of the SRCON0 register control the Q and Q̅ latch outputs. Both of the SR Latch outputs may be directly output to an I/O pin at the same time.

The applicable TRIS bit of the corresponding port must be cleared to enable the port pin output driver.
18.3 Effects of a Reset

Upon any device Reset, the SR Latch output is not initialized to a known state. The user’s firmware is responsible for initializing the latch output before enabling the output pins.

**FIGURE 18-1: SR LATCH SIMPLIFIED BLOCK DIAGRAM**

### Note
1. If R = 1 and S = 1 simultaneously, Q = 0, Q̅ = 1
2. Pulse generator causes a 1 Q-state pulse width.
3. Name denotes the connection point at the comparator output.
TABLE 18-1:  SRCLK FREQUENCY TABLE

<table>
<thead>
<tr>
<th>SRCLK</th>
<th>Divider</th>
<th>Fosc = 32 MHz</th>
<th>Fosc = 20 MHz</th>
<th>Fosc = 16 MHz</th>
<th>Fosc = 4 MHz</th>
<th>Fosc = 1 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>111</td>
<td>512</td>
<td>62.5 kHz</td>
<td>39.0 kHz</td>
<td>31.3 kHz</td>
<td>7.81 kHz</td>
<td>1.95 kHz</td>
</tr>
<tr>
<td>110</td>
<td>256</td>
<td>125 kHz</td>
<td>78.1 kHz</td>
<td>62.5 kHz</td>
<td>15.6 kHz</td>
<td>3.90 kHz</td>
</tr>
<tr>
<td>101</td>
<td>128</td>
<td>250 kHz</td>
<td>156 kHz</td>
<td>125 kHz</td>
<td>31.25 kHz</td>
<td>7.81 kHz</td>
</tr>
<tr>
<td>100</td>
<td>64</td>
<td>500 kHz</td>
<td>313 kHz</td>
<td>250 kHz</td>
<td>62.5 kHz</td>
<td>15.6 kHz</td>
</tr>
<tr>
<td>011</td>
<td>32</td>
<td>1 MHz</td>
<td>625 kHz</td>
<td>500 kHz</td>
<td>125 kHz</td>
<td>31.3 kHz</td>
</tr>
<tr>
<td>010</td>
<td>16</td>
<td>2 MHz</td>
<td>1.25 MHz</td>
<td>1 MHz</td>
<td>250 kHz</td>
<td>62.5 kHz</td>
</tr>
<tr>
<td>001</td>
<td>8</td>
<td>4 MHz</td>
<td>2.5 MHz</td>
<td>2 MHz</td>
<td>500 kHz</td>
<td>125 kHz</td>
</tr>
<tr>
<td>000</td>
<td>4</td>
<td>8 MHz</td>
<td>5 MHz</td>
<td>4 MHz</td>
<td>1 MHz</td>
<td>250 kHz</td>
</tr>
</tbody>
</table>

REGISTER 18-1:  SRCON0: SR LATCH CONTROL 0 REGISTER

R/W-0/0  R/W-0/0  R/W-0/0  R/W-0/0  R/W-0/0  R/S-0/0  R/S-0/0

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 6</th>
<th>bit 5</th>
<th>bit 4</th>
<th>bit 3</th>
<th>bit 2</th>
<th>bit 1</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>SRLEN</td>
<td>SRCLK2</td>
<td>SRCLK1</td>
<td>SRCLK0</td>
<td>SRQEN</td>
<td>SRNQEN</td>
<td>SRPS</td>
<td>SRPR</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- S = Bit is set only

bit 7  
| SRLEN: SR Latch Enable bit |
| 1 = SR Latch is enabled  |
| 0 = SR Latch is disabled |

bit 6-4  
| SRCLK<2:0>: SR Latch Clock Divider bits |
| 000 = Generates a 1 Fosc wide pulse every 4\textsuperscript{th} Fosc cycle clock |
| 001 = Generates a 1 Fosc wide pulse every 8\textsuperscript{th} Fosc cycle clock |
| 010 = Generates a 1 Fosc wide pulse every 16\textsuperscript{th} Fosc cycle clock |
| 011 = Generates a 1 Fosc wide pulse every 32\textsuperscript{nd} Fosc cycle clock |
| 100 = Generates a 1 Fosc wide pulse every 64\textsuperscript{th} Fosc cycle clock |
| 101 = Generates a 1 Fosc wide pulse every 128\textsuperscript{th} Fosc cycle clock |
| 110 = Generates a 1 Fosc wide pulse every 256\textsuperscript{th} Fosc cycle clock |
| 111 = Generates a 1 Fosc wide pulse every 512\textsuperscript{th} Fosc cycle clock |

bit 3  
| SRQEN: SR Latch Q Output Enable bit |
| If SRLEN = 1: |
| 1 = Q is present on the SRQ pin |
| 0 = External Q output is disabled |
| If SRLEN = 0: |
| SR Latch is disabled |

bit 2  
| SRNQEN: SR Latch \(\overline{Q}\) Output Enable bit |
| If SRLEN = 1: |
| 1 = \(\overline{Q}\) is present on the SRnQ pin |
| 0 = External \(\overline{Q}\) output is disabled |
| If SRLEN = 0: |
| SR Latch is disabled |

bit 1  
| SRPS: Pulse Set Input of the SR Latch bit\textsuperscript{(1)} |
| 1 = Pulse set input for 1 Q-clock period |
| 0 = No effect on set input. |

bit 0  
| SRPR: Pulse Reset Input of the SR Latch bit\textsuperscript{(1)} |
| 1 = Pulse reset input for 1 Q-clock period |
| 0 = No effect on Reset input. |

Note 1: Set only, always reads back ‘0’.
## REGISTER 18-2: SRCON1: SR LATCH CONTROL 1 REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value 1</th>
<th>Value 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>SRSPE: SR Latch Peripheral Set Enable bit</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>6</td>
<td>SRSCKE: SR Latch Set Clock Enable bit</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>5</td>
<td>Reserved: Read as ‘0’. Maintain this bit clear.</td>
<td></td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>SRSC1E: SR Latch C1 Set Enable bit</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>SRRPE: SR Latch Peripheral Reset Enable bit</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>2</td>
<td>SRRC1E: SR Latch C1 Reset Enable bit</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>Reserved: Read as ‘0’. Maintain this bit clear.</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0</td>
<td>SRRC1E: SR Latch C1 Reset Enable bit</td>
<td>1</td>
<td>0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **‘1’** = Bit is set
- **‘0’** = Bit is cleared
- **-n/n** = Value at POR and BOR/Value at all other Resets

- **SRSC1E**: SR Latch C1 Set Enable bit
  - 1 = SR Latch is set when the C1 Comparator output is high
  - 0 = C1 Comparator output has no effect on the set input of the SR Latch

- **SRRPE**: SR Latch Peripheral Reset Enable bit
  - 1 = SR Latch is reset when the SRI pin is high
  - 0 = SRI pin has no effect on the Reset input of the SR Latch
### TABLE 18-2: SUMMARY OF REGISTERS ASSOCIATED WITH SR LATCH MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>SRCON0</td>
<td>SRLEN</td>
<td>SRCLK2</td>
<td>SRCLK1</td>
<td>SRCLK0</td>
<td>SROEN</td>
<td>SRNQEN</td>
<td>SRPS</td>
<td>SRPR</td>
<td>134</td>
</tr>
<tr>
<td>SRCON1</td>
<td>SRSPE</td>
<td>SRSCKE</td>
<td>Reserved</td>
<td>SRSC1E</td>
<td>SRRPE</td>
<td>SRRCKE</td>
<td>Reserved</td>
<td>SRRC1E</td>
<td>135</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented, read as '0'. Shaded cells are unused by the SR Latch module.
19.0 COMPARATOR MODULE

Comparators are used to interface analog circuits to a
digital circuit by comparing two analog voltages and
providing a digital indication of their relative magnitudes. Comparators are very useful mixed signal building
blocks because they provide analog functionality independent of program execution. The analog
comparator module includes the following features:

• Independent comparator control
• Programmable input selection
• Comparator output is available internally/externally
• Programmable output polarity
• Interrupt-on-change
• Wake-up from Sleep
• Programmable Speed/Power optimization
• PWM shutdown
• Programmable and fixed voltage reference

19.1 Comparator Overview

A single comparator is shown in Figure 19-1 along with
the relationship between the analog input levels and
the digital output. When the analog voltage at VIN+ is
less than the analog voltage at VIN-, the output of the
comparator is a digital low level. When the analog
voltage at VIN+ is greater than the analog voltage at
VIN-, the output of the comparator is a digital high level.

Note: The black areas of the output of the comparator represents the uncertainty due to input offsets and response time.
FIGURE 19-2: COMPARATOR 1 MODULE SIMPLIFIED BLOCK DIAGRAM

Note 1: When C1ON = 0, the Comparator will produce a '0' at the output.
2: When C1ON = 0, all multiplexer inputs are disconnected.
3: Output of comparator can be frozen during debugging.
19.2 Comparator Control

The comparator has two control registers: CM1CON0 and CM1CON1.

The CM1CON0 register (see Register 19-1) contains Control and Status bits for the following:

- Enable
- Output selection
- Output polarity
- Speed/Power selection
- Hysteresis enable
- Output synchronization

The CM1CON1 register (see Register 19-2) contains Control bits for the following:

- Interrupt enable
- Interrupt edge polarity
- Positive input channel selection
- Negative input channel selection

19.2.1 COMPARATOR ENABLE

Setting the C1ON bit of the CM1CON0 register enables the comparator for operation. Clearing the C1ON bit disables the comparator resulting in minimum current consumption.

19.2.2 COMPARATOR OUTPUT SELECTION

The output of the comparator can be monitored by reading either the C1OUT bit of the CM1CON0 register or the MC1OUT bit of the CMOUT register. In order to make the output available for an external connection, the following conditions must be true:

- C1OE bit of the CM1CON0 register must be set
- Corresponding TRIS bit must be cleared
- C1ON bit of the CM1CON0 register must be set

19.2.3 COMPARATOR OUTPUT POLARITY

Inverting the output of the comparator is functionally equivalent to swapping the comparator inputs. The polarity of the comparator output can be inverted by setting the C1POL bit of the CM1CON0 register. Clearing the C1POL bit results in a non-inverted output.

Table 19-1 shows the output state versus input conditions, including polarity control.

<table>
<thead>
<tr>
<th>Input Condition</th>
<th>C1POL</th>
<th>C1OUT</th>
</tr>
</thead>
<tbody>
<tr>
<td>C1VN &gt; C1VP</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>C1VN &lt; C1VP</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>C1VN &gt; C1VP</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>C1VN &lt; C1VP</td>
<td>1</td>
<td>0</td>
</tr>
</tbody>
</table>

19.2.4 COMPARATOR SPEED/POWER SELECTION

The trade-off between speed or power can be optimized during program execution with the C1SP control bit. The default state for this bit is ‘1’ which selects the normal speed mode. Device power consumption can be optimized at the cost of slower comparator propagation delay by clearing the C1SP bit to ‘0’.

Note 1: The C1OE bit of the CM1CON0 register overrides the PORT data latch. Setting the C1ON bit of the CM1CON0 register has no impact on the port override.

2: The internal output of the comparator is latched with each instruction cycle. Unless otherwise specified, external outputs are not latched.
19.3 Comparator Hysteresis
A selectable amount of separation voltage can be added to the input pins of each comparator to provide a hysteresis function to the overall operation. Hysteresis is enabled by setting the C1HYS bit of the CM1CON0 register.
See Section 31.0 “Electrical Specifications” for more information.

19.4 Timer1 Gate Operation
The output resulting from a comparator operation can be used as a source for gate control of Timer1. See Section 21.6 “Timer1 Gate” for more information.
This feature is useful for timing the duration or interval of an analog event.
It is recommended that the comparator output be synchronized to Timer1. This ensures that Timer1 does not increment while a change in the comparator is occurring.

19.4.1 COMPARATOR OUTPUT SYNCHRONIZATION
The output from comparator C1 can be synchronized with Timer1 by setting the C1SYNC bit of the CM1CON0 register.
Once enabled, the comparator output is latched on the falling edge of the Timer1 source clock. If a prescaler is used with Timer1, the comparator output is latched after the prescaling function. To prevent a race condition, the comparator output is latched on the falling edge of the Timer1 clock source and Timer1 increments on the rising edge of its clock source. See the Comparator Block Diagram (Figure 19-2) and the Timer1 Block Diagram (Figure 21-1) for more information.

19.5 Comparator Interrupt
An interrupt can be generated upon a change in the output value of the comparator for each comparator, a rising edge detector and a Falling edge detector are present.
When either edge detector is triggered and its associated enable bit is set (C1INTP and/or C1INTN bits of the CM1CON1 register), the Corresponding Interrupt Flag bit (C1IF bit of the PIR2 register) will be set.
To enable the interrupt, you must set the following bits:
• C1ON, C1POL and C1SP bits of the CM1CON0 register
• C1IE bit of the PIE2 register
• C1INTP bit of the CM1CON1 register (for a rising edge detection)
• C1INTN bit of the CM1CON1 register (for a falling edge detection)
• PEIE and GIE bits of the INTCON register
The associated interrupt flag bit, C1IF bit of the PIR2 register, must be cleared in software. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

Note: Although a comparator is disabled, an interrupt can be generated by changing the output polarity with the C1POL bit of the CM1CON0 register, or by switching the comparator on or off with the C1ON bit of the CM1CON0 register.

19.6 Comparator Positive Input Selection
Configuring the C1PCH<1:0> bits of the CM1CON1 register directs an internal voltage reference or an analog pin to the non-inverting input of the comparator:
• C1IN+ analog pin
• DAC
• FVR (Fixed Voltage Reference)
• Vss (Ground)
See Section 14.0 “Fixed Voltage Reference (FVR)” for more information on the Fixed Voltage Reference module.
See Section 17.0 “Digital-to-Analog Converter (DAC) Module” for more information on the DAC input signal.
Any time the comparator is disabled (C1ON = 0), all comparator inputs are disabled.
19.7 Comparator Negative Input Selection

The C1NCH bit of the CM1CON1 register directs one of two analog pins to the comparator inverting input.

**Note:** To use C1IN+ and C1INX- pins as analog input, the appropriate bits must be set in the ANSEL register and the corresponding TRIS bits must also be set to disable the output drivers.

19.8 Comparator Response Time

The comparator output is indeterminate for a period of time after the change of an input source or the selection of a new reference voltage. This period is referred to as the response time. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response time to a comparator input change. See the Comparator and Voltage Reference Specifications in Section 31.0 “Electrical Specifications” for more details.

19.9 Interaction with ECCP Logic

The C1 comparator can be used as a general purpose comparator. The output can be brought out to the C1OUT pin. When the ECCP auto-shutdown is active it can use the comparator signal. If auto-restart is also enabled, the comparator can be configured as a closed loop analog feedback to the ECCP, thereby, creating an analog controlled PWM.

**Note:** When the comparator module is first initialized, the output state is unknown. Upon initialization, the user should verify the output state of the comparator prior to relying on the result, primarily when using the result in connection with other peripheral features, such as the ECCP Auto-Shutdown mode.

19.10 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 19-3. Since the analog input pins share their connection with a digital input, they have reverse biased ESD protection diodes to VDD and VSS. The analog input, therefore, must be between VSS and VDD.

If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur.

A maximum source impedance of 10 kΩ is recommended for the analog sources. Also, any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current to minimize inaccuracies introduced.

**Note 1:** When reading a PORT register, all pins configured as analog inputs will read as a ‘0’. Pins configured as digital inputs will convert as an analog input, according to the input specification.

2: Analog levels on any pin defined as a digital input, may cause the input buffer to consume more current than is specified.
FIGURE 19-3: ANALOG INPUT MODEL

Legend:  
CPIN = Input Capacitance  
ILEAKAGE = Leakage Current at the pin due to various junctions  
RIC = Interconnect Resistance  
RS = Source Impedance  
VA = Analog Voltage  
VT = Threshold Voltage

Note 1: See Section 31.0 "Electrical Specifications".
REGISTER 19-1:  CM1CON0: COMPARATOR C1 CONTROL REGISTER 0

| bit 7 | C1ON: Comparator Enable bit  
| bit 6 | C1OUT: Comparator Output bit  
| bit 5 | C1OE: Comparator Output Enable bit  
| bit 4 | C1POL: Comparator Output Polarity Select bit  
| bit 3 | Unimplemented: Read as ‘0’  
| bit 2 | C1SP: Comparator Speed/Power Select bit  
| bit 1 | C1HYS: Comparator Hysteresis Enable bit  
| bit 0 | C1SYNC: Comparator Output Synchronous Mode bit  

Legend:
R = Readable bit  
W = Writable bit  
U = Unimplemented bit, read as ‘0’  
u = Bit is unchanged  
x = Bit is unknown  
-n/n = Value at POR and BOR/Value at all other Resets  
‘1’ = Bit is set  
‘0’ = Bit is cleared

bit 7  
C1ON: Comparator Enable bit  
1 = Comparator is enabled  
0 = Comparator is disabled and consumes no active power

bit 6  
C1OUT: Comparator Output bit  
If C1POL = 1 (inverted polarity):  
1 = C1VP < C1VN  
0 = C1VP > C1VN  
If C1POL = 0 (non-inverted polarity):  
1 = C1VP > C1VN  
0 = C1VP < C1VN

bit 5  
C1OE: Comparator Output Enable bit  
1 = C1OUT is present on the C1OUT pin. Requires that the associated TRIS bit be cleared to actually drive the pin. Not affected by C1ON.  
0 = C1OUT is internal only

bit 4  
C1POL: Comparator Output Polarity Select bit  
1 = Comparator output is inverted  
0 = Comparator output is not inverted

bit 3  
Unimplemented: Read as ‘0’

bit 2  
C1SP: Comparator Speed/Power Select bit  
1 = Comparator operates in normal power, higher-speed mode  
0 = Comparator operates in low-power, low-speed mode

bit 1  
C1HYS: Comparator Hysteresis Enable bit  
1 = Comparator hysteresis enabled  
0 = Comparator hysteresis disabled

bit 0  
C1SYNC: Comparator Output Synchronous Mode bit  
1 = Comparator output to Timer1 and I/O pin is synchronous to changes on Timer1 clock source. Output updated on the falling edge of Timer1 clock source.  
0 = Comparator output to Timer1 and I/O pin is asynchronous.
TABLE 19-2: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARATOR MODULE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
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<td>—</td>
<td>—</td>
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<td>—</td>
<td>107</td>
</tr>
<tr>
<td>CM1CON0</td>
<td>C1ON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>C1OE</td>
<td>C1POL</td>
<td>—</td>
<td>—</td>
<td>143</td>
</tr>
<tr>
<td>CM1CON1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>C1PCH1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>C1NCH</td>
<td>144</td>
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<td>—</td>
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<td>—</td>
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<td>—</td>
<td>MC1OUT</td>
<td>144</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
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<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>—</td>
<td>C1IE</td>
<td>EEIE</td>
<td>BCL1IE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>79</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>—</td>
<td>C1IF</td>
<td>EEIF</td>
<td>BCL1IF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>81</td>
</tr>
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<td>TRISA</td>
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<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented, read as ‘0’. Shaded cells are unused by the comparator module.
### 20.0 TIMER0 MODULE

The Timer0 module is an 8-bit timer/counter with the following features:
- 8-bit timer/counter register (TMR0)
- 8-bit prescaler (independent of Watchdog Timer)
- Programmable internal or external clock source
- Programmable external clock edge selection
- Interrupt on overflow
- TMR0 can be used to gate Timer1

Figure 20-1 is a block diagram of the Timer0 module.

### 20.1 Timer0 Operation

The Timer0 module can be used as either an 8-bit timer or an 8-bit counter.

#### 20.1.1 8-BIT TIMER MODE

The Timer0 module will increment every instruction cycle, if used without a prescaler. 8-Bit Timer mode is selected by clearing the TMR0CS bit of the OPTION register.

When TMR0 is written, the increment is inhibited for two instruction cycles immediately following the write.

**Note:** The value written to the TMR0 register can be adjusted, in order to account for the two instruction cycle delay when TMR0 is written.

#### 20.1.2 8-BIT COUNTER MODE

In 8-Bit Counter mode, the Timer0 module will increment on every rising or falling edge of the T0CKI pin or the Capacitive Sensing Oscillator (CPSCLK) signal.

8-Bit Counter mode using the T0CKI pin is selected by setting the TMR0CS bit in the OPTION register to ‘1’ and resetting the T0XCS bit in the CPSCON0 register to ‘0’.

8-Bit Counter mode using the Capacitive Sensing Oscillator (CPSCLK) signal is selected by setting the TMR0CS bit in the OPTION register to ‘1’ and setting the T0XCS bit in the CPSCON0 register to ‘1’.

The rising or falling transition of the incrementing edge for either input source is determined by the TMR0SE bit in the OPTION register.

---

**FIGURE 20-1: BLOCK DIAGRAM OF THE TIMER0**
20.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A software programmable prescaler is available for exclusive use with Timer0. The prescaler is enabled by clearing the PSA bit of the OPTION register.

| Note: | The Watchdog Timer (WDT) uses its own independent prescaler. |

There are eight prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be disabled by setting the PSA bit of the OPTION register.

The prescaler is not readable or writable. All instructions writing to the TMR0 register will clear the prescaler.

20.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The TMR0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The TMR0IF bit can only be cleared in software. The Timer0 interrupt enable is the TMR0IE bit of the INTCON register.

| Note: | The Timer0 interrupt cannot wake the processor from Sleep since the timer is frozen during Sleep. |

20.1.5 8-BIT COUNTER MODE SYNCHRONIZATION

When in 8-Bit Counter mode, the incrementing edge on the T0CKI pin must be synchronized to the instruction clock. Synchronization can be accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the instruction clock. The high and low periods of the external clocking source must meet the timing requirements as shown in Section 31.0 “Electrical Specifications”.

20.1.6 OPERATION DURING SLEEP

Timer0 cannot operate while the processor is in Sleep mode. The contents of the TMR0 register will remain unchanged while the processor is in Sleep mode.
### REGISTER 20-1: OPTION_REG: OPTION REGISTER

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>WPUEN</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS&lt;2:0&gt;</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

#### bit 7 WPUEN: Weak Pull-up Enable bit
- 1 = All weak pull-ups are disabled (except MCLR, if it is enabled)
- 0 = Weak pull-ups are enabled by individual WPUA latch values

#### bit 6 INTEDG: Interrupt Edge Select bit
- 1 = Interrupt on rising edge of RA2/INT pin
- 0 = Interrupt on falling edge of RA2/INT pin

#### bit 5 TMR0CS: Timer0 Clock Source Select bit
- 1 = Transition on RA2/T0CKI pin
- 0 = Internal instruction cycle clock (Fosc/4)

#### bit 4 TMR0SE: Timer0 Source Edge Select bit
- 1 = Increment on high-to-low transition on RA2/T0CKI pin
- 0 = Increment on low-to-high transition on RA2/T0CKI pin

#### bit 3 PSA: Prescaler Assignment bit
- 1 = Prescaler is not assigned to the Timer0 module
- 0 = Prescaler is assigned to the Timer0 module

#### bit 2-0 PS<2:0>: Prescaler Rate Select bits

<table>
<thead>
<tr>
<th>Bit Value</th>
<th>Timer0 Rate</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>1 : 2</td>
</tr>
<tr>
<td>001</td>
<td>1 : 4</td>
</tr>
<tr>
<td>010</td>
<td>1 : 8</td>
</tr>
<tr>
<td>011</td>
<td>1 : 16</td>
</tr>
<tr>
<td>100</td>
<td>1 : 32</td>
</tr>
<tr>
<td>011</td>
<td>1 : 64</td>
</tr>
<tr>
<td>110</td>
<td>1 : 128</td>
</tr>
<tr>
<td>111</td>
<td>1 : 256</td>
</tr>
</tbody>
</table>

### TABLE 20-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CPSCON0</td>
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<td></td>
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<td>284</td>
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<td>CPSON</td>
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<td>CPSRM</td>
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<td>WPUEN</td>
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<td>INTEDG</td>
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<td>TMR0SE</td>
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<tr>
<td>PSA</td>
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<td>Timer0 Module Register</td>
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<td></td>
</tr>
</tbody>
</table>

**Legend:**
- — = Unimplemented locations, read as ‘0’.
- Shaded cells are not used by the Timer0 module.

* Page provides register information.
21.0 TIMER1 MODULE WITH GATE CONTROL

The Timer1 module is a 16-bit timer/counter with the following features:

- 16-bit timer/counter register pair (TMR1H:TMR1L)
- Programmable internal or external clock source
- 2-bit prescaler
- Dedicated 32 kHz oscillator circuit
- Optionally synchronized comparator out
- Multiple Timer1 gate (count enable) sources
- Interrupt on overflow
- Wake-up on overflow (external clock, Asynchronous mode only)
- Time base for the Capture/Compare function
- Special Event Trigger (with ECCP)
- Selectable Gate Source Polarity

Figure 21-1 is a block diagram of the Timer1 module.

![Timer1 Block Diagram](image-url)

**FIGURE 21-1: TIMER1 BLOCK DIAGRAM**

- **Note 1:** ST Buffer is high speed type when using T1CKI.
- **Note 2:** Timer1 register increments on rising edge.
- **Note 3:** Synchronize does not operate while in Sleep.
21.1 Timer1 Operation

The Timer1 module is a 16-bit incrementing counter which is accessed through the TMR1H:TMR1L register pair. Writes to TMR1H or TMR1L directly update the counter.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or counter and increments on every selected edge of the external source.

Timer1 is enabled by configuring the TMR1ON and TMR1GE bits in the T1CON and T1GCON registers, respectively. Table 21-1 displays the Timer1 enable selections.

<table>
<thead>
<tr>
<th>TMR1ON</th>
<th>TMR1GE</th>
<th>Timer1 Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>Off</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>Off</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>Always On</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>Count Enabled</td>
</tr>
</tbody>
</table>

21.2 Clock Source Selection

The TMR1CS<1:0> and T1OSCEN bits of the T1CON register are used to select the clock source for Timer1. Table 21-2 displays the clock source selections.

21.2.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected the TMR1H:TMR1L register pair will increment on multiples of Fosc as determined by the Timer1 prescaler.

When the Fosc internal clock source is selected, the Timer1 register value will increment by four counts every instruction clock cycle. Due to this condition, a 2 LSB error in resolution will occur when reading the Timer1 value. To utilize the full resolution of Timer1, an asynchronous input signal must be used to gate the Timer1 clock input.

The following asynchronous sources may be used:
- Asynchronous event on the T1G pin to Timer1 gate
- C1 comparator input to Timer1 gate

21.2.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When enabled to count, Timer1 is incremented on the rising edge of the external clock input T1CKI or the capacitive sensing oscillator signal. Either of these external clock sources can be synchronized to the microcontroller system clock or they can run asynchronously.

When used as a timer with a clock oscillator, an external 32.768 kHz crystal can be used in conjunction with the dedicated internal oscillator circuit.

Note: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge after any one or more of the following conditions:
- Timer1 enabled after POR
- Write to TMR1H or TMR1L
- Timer1 is disabled
- Timer1 is disabled (TMR1ON = 0) when T1CKI is high then Timer1 is enabled (TMR1ON=1) when T1CKI is low.

<table>
<thead>
<tr>
<th>TMR1CS1</th>
<th>TMR1CS0</th>
<th>T1OSCEN</th>
<th>Clock Source</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>1</td>
<td>x</td>
<td>System Clock (Fosc)</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>x</td>
<td>Instruction Clock (Fosc/4)</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>x</td>
<td>Capacitive Sensing Oscillator</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>0</td>
<td>External Clocking on T1CKI Pin</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>1</td>
<td>Osc.Circuit On T1OSI/T1OSO Pins</td>
</tr>
</tbody>
</table>

TABLE 21-2: CLOCK SOURCE SELECTIONS
21.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The T1CKPS bits of the T1CON register control the prescale counter. The prescaler counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

21.4 Timer1 Oscillator

A dedicated low-power 32.768 kHz oscillator circuit is built-in between pins T1OSI (input) and T1OSO (amplifier output). This internal circuit is to be used in conjunction with an external 32.768 kHz crystal.

The oscillator circuit is enabled by setting the T1OSCEN bit of the T1CON register. The oscillator will continue to run during Sleep.

Note: The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to enabling Timer1.

21.5 Timer1 Operation in Asynchronous Counter Mode

If control bit T1SYNC of the T1CON register is set, the external clock input is not synchronized. The timer increments asynchronously to the internal phase clocks. If the external clock source is selected then the timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see Section 21.5.1 “Reading and Writing Timer1 in Asynchronous Counter Mode”).

Note: When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.

21.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TMR1L register pair.

21.6 Timer1 Gate

Timer1 can be configured to count freely or the count can be enabled and disabled using Timer1 gate circuitry. This is also referred to as Timer1 Gate Enable. Timer1 gate can also be driven by multiple selectable sources.

21.6.1 TIMER1 GATE ENABLE

The Timer1 Gate Enable mode is enabled by setting the TMR1GE bit of the T1CON register. The polarity of the Timer1 Gate Enable mode is configured using the T1GPOL bit of the T1GCON register.

When Timer1 Gate Enable mode is enabled, Timer1 will increment on the rising edge of the Timer1 clock source. When Timer1 Gate Enable mode is disabled, no incrementing will occur and Timer1 will hold the current count. See Figure 21-3 for timing details.

### TABLE 21-3: TIMER1 GATE ENABLE SELECTIONS

<table>
<thead>
<tr>
<th>T1CLK</th>
<th>T1GPOL</th>
<th>T1G</th>
<th>Timer1 Operation</th>
</tr>
</thead>
<tbody>
<tr>
<td>↑</td>
<td>0</td>
<td>0</td>
<td>Counts</td>
</tr>
<tr>
<td>↑</td>
<td>0</td>
<td>1</td>
<td>Holds Count</td>
</tr>
<tr>
<td>↑</td>
<td>1</td>
<td>0</td>
<td>Holds Count</td>
</tr>
<tr>
<td>↑</td>
<td>1</td>
<td>1</td>
<td>Counts</td>
</tr>
</tbody>
</table>

21.6.2 TIMER1 GATE SOURCE SELECTION

The Timer1 gate source can be selected from one of four different sources. Source selection is controlled by the T1GSS bits of the T1GCON register. The polarity for each available source is also selectable. Polarity selection is controlled by the T1GPOL bit of the T1GCON register.

### TABLE 21-4: TIMER1 GATE SOURCES

<table>
<thead>
<tr>
<th>T1GSS</th>
<th>Timer1 Gate Source</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Timer1 Gate Pin</td>
</tr>
<tr>
<td>01</td>
<td>Overflow of Timer0 (TMR0 increments from FFh to 00h)</td>
</tr>
<tr>
<td>10</td>
<td>Comparator 1 Output SYNCC1OUT (optionally Timer1 synchronized output)</td>
</tr>
<tr>
<td>11</td>
<td>Reserved</td>
</tr>
</tbody>
</table>

Note: When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.

Note: The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to enabling Timer1.

When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.
21.6.2.1 T1G Pin Gate Operation
The T1G pin is one source for Timer1 gate control. It can be used to supply an external source to the Timer1 gate circuitry.

21.6.2.2 Timer0 Overflow Gate Operation
When Timer0 increments from FFh to 00h, a low-to-high pulse will automatically be generated and internally supplied to the Timer1 gate circuitry.

21.6.2.3 Comparator C1 Gate Operation
The output resulting from a Comparator 1 operation can be selected as a source for Timer1 gate control. The Comparator 1 output (SYNCC1OUT) can be synchronized to the Timer1 clock or left asynchronous. For more information see Section 19.4.1 “Comparator Output Synchronization”.

21.6.3 TIMER1 GATE TOGGLE MODE
When Timer1 Gate Toggle mode is enabled, it is possible to measure the full-cycle length of a Timer1 gate signal, as opposed to the duration of a single level pulse.

The Timer1 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. See Figure 21-4 for timing details.

Timer1 Gate Toggle mode is enabled by setting the T1GTM bit of the T1GCON register. When the T1GTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

Note: Enabling Toggle mode at the same time as changing the gate polarity may result in indeterminate operation.

21.6.4 TIMER1 GATE SINGLE-PULSE MODE
When Timer1 Gate Single-Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer1 Gate Single-Pulse mode is first enabled by setting the T1GSPM bit in the T1GCON register. Next, the T1GGO/DONE bit in the T1GCON register must be set. The Timer1 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the T1GGO/DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1 until the T1GGO/DONE bit is once again set in software. See Figure 21-5 for timing details.

If the Single Pulse Gate mode is disabled by clearing the T1GSPM bit in the T1GCON register, the T1GGO/DONE bit should also be cleared.

Enabling the Toggle mode and the Single-Pulse mode simultaneously will permit both sections to work together. This allows the cycle times on the Timer1 gate source to be measured. See Figure 21-6 for timing details.

21.6.5 TIMER1 GATE VALUE STATUS
When Timer1 gate value status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the T1GVAL bit in the T1GCON register. The T1GVAL bit is valid even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

21.6.6 TIMER1 GATE EVENT INTERRUPT
When Timer1 gate event interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of T1GVAL occurs, the TMR1GIF flag bit in the PIR1 register will be set. If the TMR1GIE bit in the PIE1 register is set, then an interrupt will be recognized.

The TMR1GIF flag bit operates even when the Timer1 gate is not enabled (TMR1GE bit is cleared).
21.7 Timer1 Interrupt

The Timer1 register pair (TMR1H:TMR1L) increments to FFFFh and rolls over to 0000h. When Timer1 rolls over, the Timer1 interrupt flag bit of the PIR1 register is set. To enable the interrupt on rollover, you must set these bits:

- TMR1ON bit of the T1CON register
- TMR1IE bit of the PIE1 register
- PEIE bit of the INTCON register
- GIE bit of the INTCON register

The interrupt is cleared by clearing the TMR1IF bit in the Interrupt Service Routine.

Note: The TMR1H:TMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.

21.8 Timer1 Operation During Sleep

Timer1 can only operate during Sleep when setup in Asynchronous Counter mode. In this mode, an external crystal or clock source can be used to increment the counter. To set up the timer to wake the device:

- TMR1ON bit of the T1CON register must be set
- TMR1IE bit of the PIE1 register must be set
- PEIE bit of the INTCON register must be set
- T1SYNC bit of the T1CON register must be set
- TMR1CS bits of the T1CON register must be configured
- T1OSCEN bit of the T1CON register must be configured

The device will wake-up on an overflow and execute the next instructions. If the GIE bit of the INTCON register is set, the device will call the Interrupt Service Routine.

Timer1 oscillator will continue to operate in Sleep regardless of the T1SYNC bit setting.

FIGURE 21-2: TIMER1 INCREMENTING EDGE

21.9 ECCP/CCP Capture/Compare Time Base

The CCP module uses the TMR1H:TMR1L register pair as the time base when operating in Capture or Compare mode.

In Capture mode, the value in the TMR1H:TMR1L register pair is copied into the CCPR1H:CCPR1L register pair on a configured event.

In Compare mode, an event is triggered when the value in the CCPR1H:CCPR1L register pair matches the value in the TMR1H:TMR1L register pair. This event can be a Special Event Trigger.

For more information, see Section 24.0 “Capture/Compare/PWM Modules”.

21.10 ECCP/CCP Special Event Trigger

When the CCP is configured to trigger a special event, the trigger will clear the TMR1H:TMR1L register pair. This special event does not cause a Timer1 interrupt. The CCP module may still be configured to generate a CCP interrupt.

In this mode of operation, the CCPR1H:CCPR1L register pair becomes the period register for Timer1.

Timer1 should be synchronized and Fosc/4 should be selected as the clock source in order to utilize the Special Event Trigger. Asynchronous operation of Timer1 can cause a Special Event Trigger to be missed.

In the event that a write to TMR1H or TMR1L coincides with a Special Event Trigger from the CCP, the write will take precedence.

For more information, see Section 16.2.5 “Special Event Trigger”.

Note:
The TMR1H:TMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.

T1CKI = 1 when TMR1 Enabled
T1CKI = 0 when TMR1 Enabled

Note 1: Arrows indicate counter increments.

2: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge of the clock.
FIGURE 21-3: TIMER1 GATE ENABLE MODE

FIGURE 21-4: TIMER1 GATE TOGGLE MODE
FIGURE 21-5: TIMER1 GATE SINGLE-PULSE MODE

- TMR1GE
- T1GPOL
- T1GSPM
- T1GGO/
  - Set by software
  - Cleared by hardware on falling edge of T1GVAL
- DONE
  - Counting enabled on rising edge of T1G
- T1G_IN
- T1CKI
- T1GVAL
- Timer1
  - N
  - N + 1
  - N + 2
- TMR1GIF
  - Cleared by software
  - Set by hardware on falling edge of T1GVAL
  - Cleared by software
FIGURE 21-6: TIMER1 GATE SINGLE-PULSE AND TOGGLE COMBINED MODE

- **TMR1GE**
  - Set by software
- **T1G_POL**
  - Counting enabled on rising edge of T1G
- **T1G_IN**
- **T1CKI**
- **T1GVAL**
- **Timer1**
  - N
  - N + 1
  - N + 2
  - N + 3
  - N + 4
- **TMR1GIF**
  - Cleared by software
  - Set by hardware on falling edge of T1GVAL
  - Cleared by software

- **T1G_GOE**
  - Cleared by hardware on falling edge of T1GVAL
- **DONE**
  - Cleared by software
21.11 Timer1 Control Register

The Timer1 Control register (T1CON), shown in Register 21-1, is used to control Timer1 and select the various features of the Timer1 module.

REGISTER 21-1: T1CON: TIMER1 CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th>TMR1CS&lt;1:0&gt;: Timer1 Clock Source Select bits</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>11 = Timer1 clock source is Capacitive Sensing Oscillator (CAPOSC)</td>
</tr>
<tr>
<td></td>
<td>10 = Timer1 clock source is pin or oscillator:</td>
</tr>
<tr>
<td></td>
<td>if T1OSCN = 0:</td>
</tr>
<tr>
<td></td>
<td>External clock from T1CKI pin (on the rising edge)</td>
</tr>
<tr>
<td></td>
<td>if T1OSCN = 1:</td>
</tr>
<tr>
<td></td>
<td>Crystal oscillator on T1OSI/T1OSO pins</td>
</tr>
<tr>
<td></td>
<td>01 = Timer1 clock source is system clock (Fosc)</td>
</tr>
<tr>
<td></td>
<td>00 = Timer1 clock source is instruction clock (Fosc/4)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 5-4</th>
<th>T1CKPS&lt;1:0&gt;: Timer1 Input Clock Prescale Select bits</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>11 = 1:8 Prescale value</td>
</tr>
<tr>
<td></td>
<td>10 = 1:4 Prescale value</td>
</tr>
<tr>
<td></td>
<td>01 = 1:2 Prescale value</td>
</tr>
<tr>
<td></td>
<td>00 = 1:1 Prescale value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 3</th>
<th>T1OSCN: LP Oscillator Enable Control bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1 = Dedicated Timer1 oscillator circuit enabled</td>
</tr>
<tr>
<td></td>
<td>0 = Dedicated Timer1 oscillator circuit disabled</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 2</th>
<th>T1SYNC: Timer1 External Clock Input Synchronization Control bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>TMR1CS&lt;1:0&gt; = 1x</td>
</tr>
<tr>
<td></td>
<td>1 = Do not synchronize external clock input</td>
</tr>
<tr>
<td></td>
<td>0 = Synchronize external clock input with system clock (Fosc)</td>
</tr>
</tbody>
</table>

TMR1CS<1:0> = 0x
This bit is ignored. Timer1 uses the internal clock when TMR1CS<1:0> = 1x.

| bit 1  | Unimplemented: Read as '0' |

<table>
<thead>
<tr>
<th>bit 0</th>
<th>TMR1ON: Timer1 On bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1 = Enables Timer1</td>
</tr>
<tr>
<td></td>
<td>0 = Stops Timer1</td>
</tr>
<tr>
<td></td>
<td>Clears Timer1 gate flip-flop</td>
</tr>
</tbody>
</table>
### 21.12 Timer1 Gate Control Register

The Timer1 Gate Control register (\textit{T1GCON}), shown in \textbf{Register 21-2}, is used to control Timer1 Gate.

#### REGISTER 21-2: \textit{T1GCON: TIMER1 GATE CONTROL REGISTER}

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>\texttt{TMR1GE}: Timer1 Gate Enable bit</td>
</tr>
<tr>
<td>0</td>
<td>\texttt{T1GPOL}: Timer1 Gate Polarity bit</td>
</tr>
<tr>
<td>5</td>
<td>\texttt{T1GTM}: Timer1 Gate Toggle Mode bit</td>
</tr>
<tr>
<td>4</td>
<td>\texttt{T1GSPM}: Timer1 Gate Single-Pulse Mode bit</td>
</tr>
<tr>
<td>3</td>
<td>\texttt{T1GGO/DONE}: Timer1 Gate Single-Pulse Acquisition Status bit</td>
</tr>
<tr>
<td>2</td>
<td>\texttt{T1GVAL}: Timer1 Gate Current State bit</td>
</tr>
<tr>
<td>1-0</td>
<td>\texttt{T1GSS&lt;1:0&gt;}: Timer1 Gate Source Select bits</td>
</tr>
</tbody>
</table>

**Legend:**

- \texttt{R} = Readable bit
- \texttt{W} = Writable bit
- \texttt{U} = Unimplemented bit, read as ‘0’
- \texttt{u} = Bit is unchanged
- \texttt{x} = Bit is unknown
- \texttt{-n/n} = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- \texttt{HC} = Bit is cleared by hardware

- **bit 7:** \texttt{TMR1GE}: Timer1 Gate Enable bit
  - If \texttt{TMR1ON} = 0: This bit is ignored
  - If \texttt{TMR1ON} = 1:
    - 1 = Timer1 counting is controlled by the Timer1 gate function
    - 0 = Timer1 counts regardless of Timer1 gate function

- **bit 6:** \texttt{T1GPOL}: Timer1 Gate Polarity bit
  - 1 = Timer1 gate is active-high (Timer1 counts when gate is high)
  - 0 = Timer1 gate is active-low (Timer1 counts when gate is low)

- **bit 5:** \texttt{T1GTM}: Timer1 Gate Toggle Mode bit
  - 1 = Timer1 Gate Toggle mode is enabled
  - 0 = Timer1 Gate Toggle mode is disabled and toggle flip-flop is cleared
  - Timer1 gate flip-flop toggles on every rising edge.

- **bit 4:** \texttt{T1GSPM}: Timer1 Gate Single-Pulse Mode bit
  - 1 = Timer1 gate Single-Pulse mode is enabled and is controlling Timer1 gate
  - 0 = Timer1 gate Single-Pulse mode is disabled

- **bit 3:** \texttt{T1GGO/DONE}: Timer1 Gate Single-Pulse Acquisition Status bit
  - 1 = Timer1 gate single-pulse acquisition is ready, waiting for an edge
  - 0 = Timer1 gate single-pulse acquisition has completed or has not been started

- **bit 2:** \texttt{T1GVAL}: Timer1 Gate Current State bit
  - Indicates the current state of the Timer1 gate that could be provided to TMR1H:TMR1L. Unaffected by Timer1 Gate Enable (\texttt{TMR1GE}).

- **bit 1-0:** \texttt{T1GSS<1:0>}: Timer1 Gate Source Select bits
  - 00 = Timer1 gate pin
  - 01 = Timer0 overflow output
  - 10 = Comparator 1 optionally synchronized output (SYNCC1OUT)
  - 11 = Reserved
### TABLE 21-5: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER1

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA4</td>
<td>—</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>107</td>
</tr>
<tr>
<td>CCP1CON</td>
<td>P1M1</td>
<td>P1M0</td>
<td>DC1B1</td>
<td>DC1B0</td>
<td>CCP1M3</td>
<td>CCP1M2</td>
<td>CCP1M1</td>
<td>CCP1M0</td>
<td>192</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>TMR1H</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153*</td>
</tr>
<tr>
<td>TMR1L</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>153*</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
<tr>
<td>T1CON</td>
<td>TMR1CS1</td>
<td>TMR1CS0</td>
<td>T1CKPS1</td>
<td>T1CKPS0</td>
<td>T1OSEN</td>
<td>T1SYNC</td>
<td>—</td>
<td>TMR1ON</td>
<td>157</td>
</tr>
<tr>
<td>T1GCON</td>
<td>TMR1GE</td>
<td>T1GPOLO</td>
<td>T1GTM</td>
<td>T1GSPM</td>
<td>T1GGODONE</td>
<td>T1GVAL</td>
<td>T1GSS1</td>
<td>T1GSS0</td>
<td>158</td>
</tr>
</tbody>
</table>

**Legend:**  
— = unimplemented, read as ‘0’. Shaded cells are not used by the Timer1 module.  
*Page provides register information.*
22.0 TIMER2 MODULE

The Timer2 module incorporates the following features:
- 8-bit Timer and Period registers (TMR2 and PR2, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4, 1:16, and 1:64)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMR2 match with PR2, respectively
- Optional use as the shift clock for the MSSP1 modules

See Figure 22-1 for a block diagram of Timer2.

FIGURE 22-1: TIMER2 BLOCK DIAGRAM
22.1 Timer2 Operation

The clock input to the Timer2 modules is the system instruction clock (Fosc/4).

TMR2 increments from 00h on each clock edge.

A 4-bit counter/prescaler on the clock input allows direct input, divide-by-4 and divide-by-16 prescale options. These options are selected by the prescaler control bits, T2CKPS<1:0> of the T2CON register. The value of TMR2 is compared to that of the Period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/postscaler (see Section 22.2 “Timer2 Interrupt”).

The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, whereas the PR2 register initializes to FFh. Both the prescaler and postscaler counters are cleared on the following events:

- a write to the TMR2 register
- a write to the T2CON register
- Power-on Reset (POR)
- Brown-out Reset (BOR)
- MCLR Reset
- Watchdog Timer (WDT) Reset
- Stack Overflow Reset
- Stack Underflow Reset
- RESET Instruction

Note: TMR2 is not cleared when T2CON is written.

22.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2-to-PR2 match) provides the input for the 4-bit counter/postscaler. This counter generates the TMR2 match interrupt flag which is latched in TMR2IF of the PIR1 register. The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE of the PIE1 register.

A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS<3:0>, of the T2CON register.

22.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the CCP1 module, where it is used as a time base for operations in PWM mode.

Timer2 can be optionally used as the shift clock source for the MSSP1 module operating in SPI mode. Additional information is provided in Section 25.1 “Master SSP (MSSP1) Module Overview”

22.4 Timer2 Operation During Sleep

Timer2 cannot be operated while the processor is in Sleep mode. The contents of the TMR2 and PR2 registers will remain unchanged while the processor is in Sleep mode.
REGISTER 22-1:  T2CON: TIMER2 CONTROL REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>T2OUTPS&lt;3:0&gt;</td>
<td>TMR2ON</td>
<td>T2CKPS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:

R = Readable bit  W = Writable bit  U = Unimplemented bit, read as ‘0’
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets

bit 7  **Unimplemented**: Read as ‘0’

bit 6-3  **T2OUTPS<3:0>**: Timer Output Postscaler Select bits

<table>
<thead>
<tr>
<th>Value</th>
<th>Postscaler</th>
</tr>
</thead>
<tbody>
<tr>
<td>0000</td>
<td>1:1 Postscaler</td>
</tr>
<tr>
<td>0001</td>
<td>1:2 Postscaler</td>
</tr>
<tr>
<td>0010</td>
<td>1:3 Postscaler</td>
</tr>
<tr>
<td>0011</td>
<td>1:4 Postscaler</td>
</tr>
<tr>
<td>0100</td>
<td>1:5 Postscaler</td>
</tr>
<tr>
<td>0101</td>
<td>1:6 Postscaler</td>
</tr>
<tr>
<td>0110</td>
<td>1:7 Postscaler</td>
</tr>
<tr>
<td>0111</td>
<td>1:8 Postscaler</td>
</tr>
<tr>
<td>1000</td>
<td>1:9 Postscaler</td>
</tr>
<tr>
<td>1001</td>
<td>1:10 Postscaler</td>
</tr>
<tr>
<td>1010</td>
<td>1:11 Postscaler</td>
</tr>
<tr>
<td>1011</td>
<td>1:12 Postscaler</td>
</tr>
<tr>
<td>1100</td>
<td>1:13 Postscaler</td>
</tr>
<tr>
<td>1101</td>
<td>1:14 Postscaler</td>
</tr>
<tr>
<td>1110</td>
<td>1:15 Postscaler</td>
</tr>
<tr>
<td>1111</td>
<td>1:16 Postscaler</td>
</tr>
</tbody>
</table>

bit 2  **TMR2ON**: Timer2 On bit

<table>
<thead>
<tr>
<th>Value</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Timer2 is on</td>
</tr>
<tr>
<td>0</td>
<td>Timer2 is off</td>
</tr>
</tbody>
</table>

bit 1-0  **T2CKPS<1:0>**: Timer2 Clock Prescale Select bits

<table>
<thead>
<tr>
<th>Value</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>Prescaler is 1</td>
</tr>
<tr>
<td>01</td>
<td>Prescaler is 4</td>
</tr>
<tr>
<td>10</td>
<td>Prescaler is 16</td>
</tr>
<tr>
<td>11</td>
<td>Prescaler is 64</td>
</tr>
</tbody>
</table>
TABLE 22-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER2

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>CCP1CON</td>
<td>P1&lt;0&gt;</td>
<td></td>
<td>DC1B&lt;1:0&gt;</td>
<td></td>
<td>CCP1M&lt;3&gt;</td>
<td></td>
<td></td>
<td></td>
<td>192</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1IE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1E</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1IF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1F</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>PR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>161*</td>
</tr>
<tr>
<td>T2CON</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>T2OUTPS&lt;3&gt;</td>
<td></td>
<td>TMR2ON</td>
<td>T2CKPS1</td>
<td>T2CKPS0</td>
</tr>
<tr>
<td>TMR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>161*</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as "0". Shaded cells are not used for Timer2 module.
* Page provides register information.
23.0 DATA SIGNAL MODULATOR

The Data Signal Modulator (DSM) is a peripheral which allows the user to mix a data stream, also known as a modulator signal, with a carrier signal to produce a modulated output.

Both the carrier and the modulator signals are supplied to the DSM module either internally, from the output of a peripheral, or externally through an input pin.

The modulated output signal is generated by performing a logical “AND” operation of both the carrier and modulator signals and then provided to the MDOUT pin.

The carrier signal is comprised of two distinct and separate signals. A carrier high (CARH) signal and a carrier low (CARL) signal. During the time in which the modulator (MOD) signal is in a logic high state, the DSM mixes the carrier high signal with the modulator signal. When the modulator signal is in a logic low state, the DSM mixes the carrier low signal with the modulator signal.

Using this method, the DSM can generate the following types of Key Modulation schemes:
- Frequency-Shift Keying (FSK)
- Phase-Shift Keying (PSK)
- On-Off Keying (OOK)

Additionally, the following features are provided within the DSM module:
- Carrier Synchronization
- Carrier Source Polarity Select
- Carrier Source Pin Disable
- Programmable Modulator Data
- Modulator Source Pin Disable
- Modulated Output Polarity Select
- Slew Rate Control

Figure 23-1 shows a Simplified Block Diagram of the Data Signal Modulator peripheral.

FIGURE 23-1: SIMPLIFIED BLOCK DIAGRAM OF THE DATA SIGNAL MODULATOR
23.1 DSM Operation

The DSM module can be enabled by setting the MDEN bit in the MDCON register. Clearing the MDEN bit in the MDCON register, disables the DSM module by automatically switching the carrier high and carrier low signals to the Vss signal source. The modulator signal source is also switched to the MDBIT in the MDCON register. This not only assures that the DSM module is inactive, but that it is also consuming the least amount of current.

The values used to select the carrier high, carrier low, and modulator sources held by the Modulation Source, Modulation High Carrier, and Modulation Low Carrier control registers are not affected when the MDEN bit is cleared and the DSM module is disabled. The values inside these registers remain unchanged while the DSM is inactive. The sources for the carrier high, carrier low and modulator signals will once again be selected when the MDEN bit is set and the DSM module is again enabled and active.

The modulated output signal can be disabled without shutting down the DSM module. The DSM module will remain active and continue to mix signals, but the output value will not be sent to the MDOUT pin. During the time that the output is disabled, the MDOUT pin will remain low. The modulated output can be disabled by clearing the MDOE bit in the MDCON register.

23.2 Modulator Signal Sources

The modulator signal can be supplied from the following sources:

- CCP1 Signal
- MSSP1 SD01 Signal (SPI mode only)
- Comparator C1 Signal
- EUSART TX Signal
- External Signal on MDMIN pin
- MDBIT bit in the MDCON register

The modulator signal is selected by configuring the MDMS <3:0> bits in the MDSRC register.

23.3 Carrier Signal Sources

The carrier high signal and carrier low signal can be supplied from the following sources:

- CCP1 Signal
- Reference Clock Module Signal
- External Signal on MDCIN1 pin
- External Signal on MDCIN2 pin
- Vss

The carrier high signal is selected by configuring the MDCH <3:0> bits in the MDCARH register. The carrier low signal is selected by configuring the MDCL <3:0> bits in the MDCARL register.

23.4 Carrier Synchronization

During the time when the DSM switches between carrier high and carrier low signal sources, the carrier data in the modulated output signal can become truncated. To prevent this, the carrier signal can be synchronized to the modulator signal. When synchronization is enabled, the carrier pulse that is being mixed at the time of the transition is allowed to transition low before the DSM switches over to the next carrier source.

Synchronization is enabled separately for the carrier high and carrier low signal sources. Synchronization for the carrier high signal can be enabled by setting the MDCHSYNC bit in the MDCARH register. Synchronization for the carrier low signal can be enabled by setting the MDCLS SYNC bit in the MDCARL register.

Figure 23-1 through Figure 23-5 show timing diagrams of using various synchronization methods.
FIGURE 23-2: ON OFF KEYING (OOK) SYNCHRONIZATION

<table>
<thead>
<tr>
<th>Carrier Low (CARL)</th>
<th>Carrier High (CARH)</th>
<th>Modulator (MOD)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

MDCHSYNC = 1
MDCLSYNC = 0

MDCHSYNC = 1
MDCLSYNC = 1

MDCHSYNC = 0
MDCLSYNC = 0

MDCHSYNC = 0
MDCLSYNC = 1

EXAMPLE 23-1: NO SYNCHRONIZATION (MDHSYNC = 0, MDCLSYNC = 0)

<table>
<thead>
<tr>
<th>Carrier High (CARH)</th>
<th>Carrier Low (CARL)</th>
<th>Modulator (MOD)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Active Carrier State

<table>
<thead>
<tr>
<th>CARH</th>
<th>CARL</th>
<th>CARH</th>
<th>CARL</th>
</tr>
</thead>
<tbody>
<tr>
<td>both</td>
<td>both</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

FIGURE 23-3: CARRIER HIGH SYNCHRONIZATION (MDHSYNC = 1, MDCLSYNC = 0)

<table>
<thead>
<tr>
<th>Carrier High (CARH)</th>
<th>Carrier Low (CARL)</th>
<th>Modulator (MOD)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

MDCHSYNC = 1
MDCLSYNC = 0

Active Carrier State

<table>
<thead>
<tr>
<th>CARH</th>
<th>CARL</th>
<th>CARH</th>
<th>CARL</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>both</td>
<td>CARH</td>
<td>both</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>CARL</td>
</tr>
</tbody>
</table>
**FIGURE 23-4: CARRIER LOW SYNCHRONIZATION (MDSHSYNC = 0, MDCLSYNC = 1)**

<table>
<thead>
<tr>
<th>Carrier High (CARH)</th>
<th>Carrier Low (CARL)</th>
<th>Modulator (MOD)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

MDCHSYNC = 0
MDCLSYNC = 1

Active Carrier State: CARH - CARL - CARH - CARL

**FIGURE 23-5: FULL SYNCHRONIZATION (MDSHSYNC = 1, MDCLSYNC = 1)**

<table>
<thead>
<tr>
<th>Carrier High (CARH)</th>
<th>Carrier Low (CARL)</th>
<th>Modulator (MOD)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

MDCHSYNC = 1
MDCLSYNC = 1

Falling edges used to sync

Active Carrier State: CARH - CARL - CARH - CARL
23.5 Carrier Source Polarity Select

The signal provided from any selected input source for the carrier high and carrier low signals can be inverted. Inverting the signal for the carrier high source is enabled by setting the MDCHPOL bit of the MDCARH register. Inverting the signal for the carrier low source is enabled by setting the MDCLPOL bit of the MDCARL register.

23.6 Carrier Source Pin Disable

Some peripherals assert control over their corresponding output pin when they are enabled. For example, when the CCP1 module is enabled, the output of CCP1 is connected to the CCP1 pin.

This default connection to a pin can be disabled by setting the MDCHODIS bit in the MDCARH register for the carrier high source and the MDCLODIS bit in the MDCARL register for the carrier low source.

23.7 Programmable Modulator Data

The MDBIT of the MDCON register can be selected as the source for the modulator signal. This gives the user the ability to program the value used for modulation.

23.8 Modulator Source Pin Disable

The modulator source default connection to a pin can be disabled by setting the MDMSODIS bit in the MDSRC register.

23.9 Modulated Output Polarity

The modulated output signal provided on the MDOUT pin can also be inverted. Inverting the modulated output signal is enabled by setting the MDOPOL bit of the MDCON register.

23.10 Slew Rate Control

The slew rate limitation on the output port pin can be disabled. The slew rate limitation can be removed by clearing the MDSLRR bit in the MDCON register.

23.11 Operation in Sleep Mode

The Data Signal Modulator (DSM) module is not affected by Sleep mode. The DSM can still operate during Sleep, if the carrier and modulator input sources are also still operable during Sleep.

23.12 Effects of a Reset

Upon any device Reset, the DSM module is disabled. The user’s firmware is responsible for initializing the module before enabling the output. The registers are reset to their default values.
REGISTER 23-1: MDCON: MODULATION CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>MDEN</td>
<td>MDOE</td>
<td>MDSLR</td>
<td>MDOPOL</td>
<td>MDOUT</td>
<td>—</td>
<td>—</td>
<td>MDBIT</td>
</tr>
<tr>
<td>bit 7</td>
<td>bit 6</td>
<td>bit 5</td>
<td>bit 4</td>
<td>bit 3</td>
<td>bit 2-1</td>
<td>bit 0</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit  W = Writable bit  U = Unimplemented bit, read as '0'
u = Bit is unchanged  x = Bit is unknown  -n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set  '0' = Bit is cleared

bit 7    MDEN: Modulator Module Enable bit
         1 = Modulator module is enabled and mixing input signals
         0 = Modulator module is disabled and has no output

bit 6    MDOE: Modulator Module Pin Output Enable bit
         1 = Modulator pin output enabled
         0 = Modulator pin output disabled

bit 5    MDSLR: MDOUT Pin Slew Rate Limiting bit
         1 = MDOUT pin slew rate limiting enabled
         0 = MDOUT pin slew rate limiting disabled

bit 4    MDOPOL: Modulator Output Polarity Select bit
         1 = Modulator output signal is inverted
         0 = Modulator output signal is not inverted

bit 3    MDOUT: Modulator Output bit
         Displays the current output value of the modulator module.\(^{(1)}\)

bit 2-1  Unimplemented: Read as '0'

bit 0    MDBIT: Allows software to manually set modulation source input to module\(^{(1)}\)
         1 = Modulator uses High Carrier source
         0 = Modulator uses Low Carrier source

Note 1: The modulated output frequency can be greater and asynchronous from the clock that updates this register bit, the bit value may not be valid for higher speed modulator or carrier signals.

Note 2: MDBIT must be selected as the modulation source in the MDSRC register for this operation.
### REGISTER 23-2: MDSRC: MODULATION SOURCE CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W- x/u</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>MDMSODIS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MDMS&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

bit 7  **MDMSODIS**: Modulation Source Output Disable bit
1 = Output signal driving the peripheral output pin (selected by MDMS<3:0>) is disabled
0 = Output signal driving the peripheral output pin (selected by MDMS<3:0>) is enabled

bit 6-4  **Unimplemented**: Read as '0'

bit 3-0  **MDMS<3:0>**: Modulation Source Selection bits
1111 = Reserved. No channel connected.
1110 = Reserved. No channel connected.
1101 = Reserved. No channel connected.
1100 = Reserved. No channel connected.
1011 = Reserved. No channel connected.
1010 = EUSART TX output.
1001 = Reserved. No channel connected.
1000 = MSSP1 SDO output
0111 = Reserved. No channel connected.
0110 = Comparator 1 output
0101 = Reserved. No channel connected.
0100 = Reserved. No channel connected.
0011 = Reserved. No channel connected.
0010 = CCP1 output (PWM Output mode only)
0001 = MDMIN port pin
0000 = MDBIT bit of MDCON register is modulation source

**Note 1**: Narrowed carrier pulse widths or spurs may occur in the signal stream if the carrier is not synchronized.
REGISTER 23-3:  MDCARH: MODULATION HIGH CARRIER CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>MDCHODIS: Modulator High Carrier Output Disable bit</td>
<td>1 =</td>
</tr>
<tr>
<td>6</td>
<td>MDCHPOL: Modulator High Carrier Polarity Select bit</td>
<td>1 =</td>
</tr>
<tr>
<td>5</td>
<td>MDCHSYNC: Modulator High Carrier Synchronization Enable bit</td>
<td>1 =</td>
</tr>
<tr>
<td>3-0</td>
<td>MDCH&lt;3:0&gt;: Modulator Data High Carrier Selection bits</td>
<td>1111 =</td>
</tr>
</tbody>
</table>

Legend:

- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets

- '1' = Bit is set
- '0' = Bit is cleared

Note 1: Narrowed carrier pulse widths or spur may occur in the signal stream if the carrier is not synchronized.
REGISTER 23-4: MDCARL: MODULATION LOW CARRIER CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>U-0</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
<th>R/W-x/u</th>
</tr>
</thead>
<tbody>
<tr>
<td>MDCLODIS</td>
<td>MDCLPOL</td>
<td>MDCLSYNC</td>
<td>—</td>
<td>—</td>
<td>MDCL&lt;3:0&gt;</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as '0'
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

<table>
<thead>
<tr>
<th>bit 7</th>
<th>MDCLODIS: Modulator Low Carrier Output Disable bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Output signal driving the peripheral output pin (selected by MDCL&lt;3:0&gt; of the MDCARL register) is disabled</td>
</tr>
<tr>
<td>0</td>
<td>Output signal driving the peripheral output pin (selected by MDCL&lt;3:0&gt; of the MDCARL register) is enabled</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 6</th>
<th>MDCLPOL: Modulator Low Carrier Polarity Select bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Selected low carrier signal is inverted</td>
</tr>
<tr>
<td>0</td>
<td>Selected low carrier signal is not inverted</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 5</th>
<th>MDCLSYNC: Modulator Low Carrier Synchronization Enable bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Modulator waits for a falling edge on the low time carrier signal before allowing a switch to the high time carrier</td>
</tr>
<tr>
<td>0</td>
<td>Modulator Output is not synchronized to the low time carrier signal(1)</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 4</th>
<th>Unimplemented: Read as '0'</th>
</tr>
</thead>
</table>

<table>
<thead>
<tr>
<th>bit 3-0</th>
<th>MDCL&lt;3:0&gt;: Modulator Data High Carrier Selection bits (1)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1111</td>
<td>Reserved. No channel connected.</td>
</tr>
<tr>
<td>0101</td>
<td>Reserved. No channel connected.</td>
</tr>
<tr>
<td>0100</td>
<td>CCP1 output (PWM Output mode only)</td>
</tr>
<tr>
<td>0011</td>
<td>Reference Clock module signal</td>
</tr>
<tr>
<td>0010</td>
<td>Reserved. No channel connected.</td>
</tr>
<tr>
<td>0001</td>
<td>MDCIN1 port pin</td>
</tr>
<tr>
<td>0000</td>
<td>VSS</td>
</tr>
</tbody>
</table>

**Note 1:** Narrowed carrier pulse widths or spurs may occur in the signal stream if the carrier is not synchronized.

TABLE 23-1: SUMMARY OF REGISTERS ASSOCIATED WITH DATA SIGNAL MODULATOR MODE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>MDCARH</td>
<td>MDCLODIS</td>
<td>MDCHPOL</td>
<td>MDCHSYNC</td>
<td>—</td>
<td>MDCH&lt;3:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>MDCARL</td>
<td>MDCLODIS</td>
<td>MDCLPOL</td>
<td>MDCLSYNC</td>
<td>—</td>
<td>MDCL&lt;3:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>MDCON</td>
<td>MDEN</td>
<td>MDOE</td>
<td>MDSLAR</td>
<td>MDOPOL</td>
<td>MDOUT</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MDBIT 170</td>
</tr>
<tr>
<td>MDSRC</td>
<td>MDMOSDIS</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>MDBIT 171</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented, read as '0'. Shaded cells are not used in the Data Signal Modulator mode.
24.0 CAPTURE/COMPARE/PWM MODULES

The Capture/Compare/PWM module is a peripheral which allows the user to time and control different events, and to generate Pulse-Width Modulation (PWM) signals. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate Pulse-Width Modulated signals of varying frequency and duty cycle.

This device contains one Enhanced Capture/Compare/PWM module (ECCP1).

The Half-Bridge ECCP module has two available I/O pins. See Table 24-1.

TABLE 24-1: PWM RESOURCES

<table>
<thead>
<tr>
<th>Device Name</th>
<th>ECCP1</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC12LF1840T39A</td>
<td>Enhanced PWM Half-Bridge</td>
</tr>
</tbody>
</table>
24.1 Capture Mode

Capture mode makes use of the 16-bit Timer1 resource. When an event occurs on the CCP1 pin, the 16-bit CCPR1H:CCPR1L register pair captures and stores the 16-bit value of the TMR1H:TMR1L register pair, respectively. An event is defined as one of the following and is configured by the CCP1M<3:0> bits of the CCP1CON register:

- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

When a capture is made, the Interrupt Request Flag bit CCP1IF of the PIR1 register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPR1H, CCPR1L register pair is read, the old captured value is overwritten by the new captured value.

Figure 24-1 shows a simplified diagram of the capture operation.

24.1.1 CCP1 PIN CONFIGURATION

In Capture mode, the CCP1 pin should be configured as an input by setting the associated TRIS control bit. Also, the CCP1 pin function may be moved to alternative pins using the APFCON register. Refer to Section 12.1 “Alternate Pin Function” for more details.

Note: If the CCP1 pin is configured as an output, a write to the port can cause a capture condition.

FIGURE 24-1: CAPTURE MODE OPERATION BLOCK DIAGRAM

24.1.2 TIMER1 MODE RESOURCE

Timer1 must be running in Timer mode or Synchronized Counter mode for the CCP1 module to use the capture feature. In Asynchronous Counter mode, the capture operation may not work.

See Section 21.0 “Timer1 Module with Gate Control” for more information on configuring Timer1.

24.1.3 SOFTWARE INTERRUPT MODE

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCP1IE interrupt enable bit of the PIE1 register clear to avoid false interrupts. Additionally, the user should clear the CCP1IF interrupt flag bit of the PIR1 register following any change in Operating mode.

Note: Clocking Timer1 from the system clock (Fosc) should not be used in Capture mode. In order for Capture mode to recognize the trigger event on the CCP1 pin, Timer1 must be clocked from the instruction clock (Fosc/4) or from an external clock source.

24.1.4 CCP1 PRESCALER

There are four prescaler settings specified by the CCP1M<3:0> bits of the CCP1CON register. Whenever the CCP1 module is turned off, or the CCP1 module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another does not clear the prescaler and may generate a false interrupt. To avoid this unexpected operation, turn the module off by clearing the CCP1CON register before changing the prescaler. Example 24-1 demonstrates the code to perform this function.

EXAMPLE 24-1: CHANGING BETWEEN CAPTURE PRESCALER

```
BANKSEL CCP1CON ;Set Bank bits to point
;to CCP1CON
CLRF CCP1CON ;Turn CCP1 module off
MOVLM NEW_CAPT_PS ;Load the W reg with
;the new prescaler
MOVF CCP1CON ;Move value and CCP1 ON
```
24.1.5 CAPTURE DURING SLEEP

Capture mode depends upon the Timer1 module for proper operation. There are two options for driving the Timer1 module in Capture mode. It can be driven by the instruction clock (Fosc/4), or by an external clock source.

When Timer1 is clocked by Fosc/4, Timer1 will not increment during Sleep. When the device wakes from Sleep, Timer1 will continue from its previous state.

Capture mode will operate during Sleep when Timer1 is clocked by an external clock source.

24.1.6 ALTERNATE PIN LOCATIONS

This module incorporates I/O pins that can be moved to other locations with the use of the alternate pin function register, APFCON. To determine which pins can be moved and what their default locations are upon a reset, see Section 12.1 “Alternate Pin Function” for more information.

### TABLE 24-2: SUMMARY OF REGISTERS ASSOCIATED WITH CAPTURE

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>APFCON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>T1GSEL</td>
<td>TXCKSEL</td>
<td>P1BSEL</td>
<td>CCP1SEL</td>
<td>103</td>
</tr>
<tr>
<td>CCP1CON</td>
<td>P1M&lt;1:0&gt;</td>
<td>DC1B&lt;1:0&gt;</td>
<td>—</td>
<td>—</td>
<td>CCP1M&lt;3:0&gt;</td>
<td>192</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>CCPR1L</td>
<td>Capture/Compare/PWM Register 1 Low Byte (LSB)</td>
<td>175</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>CCPR1H</td>
<td>Capture/Compare/PWM Register 1 High Byte (MSB)</td>
<td>175</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SPP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>—</td>
<td>C1IE</td>
<td>EEIE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>79</td>
<td></td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>—</td>
<td>C1IF</td>
<td>EEIF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>81</td>
<td></td>
</tr>
<tr>
<td>T1CON</td>
<td>TMR1CS&lt;1:0&gt;</td>
<td>T1CKPS&lt;1:0&gt;</td>
<td>T1OSEN</td>
<td>T1SYNC</td>
<td>—</td>
<td>—</td>
<td>TMR1ON</td>
<td>157</td>
<td></td>
</tr>
<tr>
<td>T1GCON</td>
<td>TMR1GE</td>
<td>T1GPOL</td>
<td>T1GTM</td>
<td>T1GSPM</td>
<td>T1GGO/DONE</td>
<td>T1GVAL</td>
<td>T1GSS&lt;1:0&gt;</td>
<td>158</td>
<td></td>
</tr>
<tr>
<td>TMR1L</td>
<td>Holding Register for the Least Significant Byte of the 16-bit TMR1 Register</td>
<td>153*</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TMR1H</td>
<td>Holding Register for the Most Significant Byte of the 16-bit TMR1 Register</td>
<td>153*</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
</tbody>
</table>

**Legend:** — = Unimplemented location, read as ‘0’. Shaded cells are not used by Capture mode.

* Page provides register information.
24.2 Compare Mode

Compare mode makes use of the 16-bit Timer1 resource. The 16-bit value of the CCPR1H:CCPR1L register pair is constantly compared against the 16-bit value of the TMR1H:TMR1L register pair. When a match occurs, one of the following events can occur:

- Toggle the CCP1 output
- Set the CCP1 output
- Clear the CCP1 output
- Generate a Special Event Trigger
- Generate a Software Interrupt

The action on the pin is based on the value of the CCP1M<3:0> control bits of the CCP1CON register. At the same time, the interrupt flag CCP1IF bit is set.

All Compare modes can generate an interrupt. Figure 24-2 shows a simplified diagram of the Compare operation.

24.2.1 CCP1 PIN CONFIGURATION

The user must configure the CCP1 pin as an output by clearing the associated TRIS bit.

Also, the CCP1 pin function may be moved to alternative pins using the APFCON register. Refer to Section 12.1 “Alternate Pin Function” for more details.

Note: Clearing the CCP1CON register will force the CCP1 compare output latch to the default low level. This is not the PORT I/O data latch.

24.2.2 TIMER1 MODE RESOURCE

In Compare mode, Timer1 must be running in either Timer mode or Synchronized Counter mode. The compare operation may not work in Asynchronous Counter mode.

See Section 21.0 “Timer1 Module with Gate Control” for more information on configuring Timer1.

Note: Clocking Timer1 from the system clock (Fosc) should not be used in Capture mode. In order for Capture mode to recognize the trigger event on the CCP1 pin, Timer1 must be clocked from the instruction clock (Fosc/4) or from an external clock source.

24.2.3 SOFTWARE INTERRUPT MODE

When Generate Software Interrupt mode is chosen (CCP1M<3:0> = 1010), the CCP1 module does not assert control of the CCP1 pin (see the CCP1CON register).

24.2.4 SPECIAL EVENT TRIGGER

When Special Event Trigger mode is chosen (CCP1M<3:0> = 1011), the CCP1 module does the following:

- Resets Timer1
- Starts an ADC conversion if ADC is enabled

The CCP1 module does not assert control of the CCP1 pin in this mode.

The Special Event Trigger output of the CCP1 occurs immediately upon a match between the TMR1H, TMR1L register pair and the CCPR1H, CCPR1L register pair. The TMR1H, TMR1L register pair is not reset until the next rising edge of the Timer1 clock. The Special Event Trigger output starts an A/D conversion (if the A/D module is enabled). This allows the CCPR1H, CCPR1L register pair to effectively provide a 16-bit programmable period register for Timer1.

Note 1: The Special Event Trigger from the CCP1 module does not set interrupt flag bit TMR1IF of the PIR1 register.

2: Removing the match condition by changing the contents of the CCPR1H and CCPR1L register pair, between the clock edge that generates the Special Event Trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.

24.2.5 COMPARE DURING SLEEP

The Compare mode is dependent upon the system clock (Fosc) for proper operation. Since Fosc is shut down during Sleep mode, the Compare mode will not function properly during Sleep.
### 24.2.6 ALTERNATE PIN LOCATIONS

This module incorporates I/O pins that can be moved to other locations with the use of the alternate pin function register, APFCON. To determine which pins can be moved and what their default locations are upon a Reset, see Section 12.1 "Alternate Pin Function" for more information.

<table>
<thead>
<tr>
<th>TABLE 24-3: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARE</th>
</tr>
</thead>
<tbody>
<tr>
<td>Name</td>
</tr>
<tr>
<td>------</td>
</tr>
<tr>
<td>APFCON</td>
</tr>
<tr>
<td>CCP1CON</td>
</tr>
<tr>
<td>CCPR1L</td>
</tr>
<tr>
<td>CCPR1H</td>
</tr>
<tr>
<td>INTCON</td>
</tr>
<tr>
<td>PIE1</td>
</tr>
<tr>
<td>PIE2</td>
</tr>
<tr>
<td>PIR1</td>
</tr>
<tr>
<td>PIR2</td>
</tr>
<tr>
<td>T1CON</td>
</tr>
<tr>
<td>T1GCON</td>
</tr>
<tr>
<td>TMR1L</td>
</tr>
<tr>
<td>TMR1H</td>
</tr>
<tr>
<td>TRISA</td>
</tr>
</tbody>
</table>

**Legend:**
- — = Unimplemented location, read as ‘0’. Shaded cells are not used by Compare mode.
- * Page provides register information.
24.3 PWM Overview

Pulse-Width Modulation (PWM) is a scheme that provides power to a load by switching quickly between fully on and fully off states. The PWM signal resembles a square wave where the high portion of the signal is considered the on state and the low portion of the signal is considered the off state. The high portion, also known as the pulse width, can vary in time and is defined in steps. A larger number of steps applied, which lengthens the pulse width, also supplies more power to the load. Lowering the number of steps applied, which shortens the pulse width, supplies less power. The PWM period is defined as the duration of one complete cycle or the total amount of on and off time combined.

PWM resolution defines the maximum number of steps that can be present in a single PWM period. A higher resolution allows for more precise control of the pulse width time and in turn the power that is applied to the load.

The term duty cycle describes the proportion of the on time to the off time and is expressed in percentages, where 0% is fully off and 100% is fully on. A lower duty cycle corresponds to less power applied and a higher duty cycle corresponds to more power applied.

Figure 24-3 shows a typical waveform of the PWM signal.

24.3.1 STANDARD PWM OPERATION

The standard PWM mode generates a Pulse-Width modulation (PWM) signal on the CCP1 pin with up to ten bits of resolution. The period, duty cycle, and resolution are controlled by the following registers:

- PR2 registers
- T2CON registers
- CCPR1L registers
- CCP1CON registers

Figure 24-4 shows a simplified block diagram of PWM operation.

Note 1: The corresponding TRIS bit must be cleared to enable the PWM output on the CCP1 pin.
2: Clearing the CCP1CON register will relinquish control of the CCP1 pin.

Note 1: The 8-bit timer TMR2 register is concatenated with the 2-bit internal system clock (Fosc), or 2 bits of the prescaler, to create the 10-bit time base.
2: In PWM mode, CCPR1H is a read-only register.
24.3.2 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP1 module for standard PWM operation:

1. Disable the CCP1 pin output driver by setting the associated TRIS bit.
2. Load the PR2 register with the PWM period value.
3. Configure the CCP1 module for the PWM mode by loading the CCP1CON register with the appropriate values.
4. Load the CCPR1L register and the DC1B<1:0> bits of the CCP1CON register, with the PWM duty cycle value.
5. Configure and start Timer2:
   - Clear the TMR2IF interrupt flag bit of the PIR1 register. See Note below.
   - Configure the T2CKPS bits of the T2CON register with the Timer prescale value.
   - Enable the Timer by setting the TMR2ON bit of the T2CON register.
6. Enable PWM output pin:
   - Wait until the Timer overflows and the TMR2IF bit of the PIR1 register is set. See Note below.
   - Enable the CCP1 pin output driver by clearing the associated TRIS bit.

**Note:** In order to send a complete duty cycle and period on the first PWM output, the above steps must be included in the setup sequence. If it is not critical to start with a complete PWM signal on the first output, then step 6 may be ignored.

24.3.3 PWM PERIOD

The PWM period is specified by the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 24-1.

**EQUATION 24-1: PWM PERIOD**

\[
\text{PWM Period} = [(PR2) + 1] \cdot \frac{T_{osc}}{(TMR2 \text{ Prescale Value})}
\]

**Note 1:** \(T_{osc} = \frac{1}{F_{OSC}}\)

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP1 pin is set. (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM duty cycle is latched from CCPR1L into CCPR1H.

**Note:** The Timer postscaler (see Section 22.1 “Timer2 Operation”) is not used in the determination of the PWM frequency.

24.3.4 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to multiple registers: CCPR1L register and DC1B<1:0> bits of the CCP1CON register. The CCPR1L contains the eight MSbs and the DC1B<1:0> bits of the CCP1CON register contain the two LSbs. CCPR1L and DC1B<1:0> bits of the CCP1CON register can be written to at any time. The duty cycle value is not latched into CCPR1H until after the period completes (i.e., a match between PR2 and TMR2 registers occurs). While using the PWM, the CCPR1H register is read-only.

**Equation 24-2** is used to calculate the PWM pulse width.

**Equation 24-3** is used to calculate the PWM duty cycle ratio.

**EQUATION 24-2: PULSE WIDTH**

\[
\text{Pulse Width} = (\text{CCPR1L}:\text{CCP1CON}<5:4>) \cdot \frac{T_{osc}}{(TMR2 \text{ Prescale Value})}
\]

**EQUATION 24-3: DUTY CYCLE RATIO**

\[
\text{Duty Cycle Ratio} = \frac{(\text{CCPR1L}:\text{CCP1CON}<5:4>)}{4(\text{PR2} + 1)}
\]

The CCPR1H register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

The 8-bit timer TMR2 register is concatenated with either the 2-bit internal system clock (FOSC), or two bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

When the 10-bit time base matches the CCPR1H and 2-bit latch, then the CCP1 pin is cleared (see Figure 24-4).
24.3.5 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is 10 bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 24-4.

\[
Resolution = \frac{\log[4(PR2 + 1)]}{\log(2)} \text{ bits}
\]

Note: If the pulse width value is greater than the period the assigned PWM pin(s) will remain unchanged.

**TABLE 24-4: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 32 MHz)**

<table>
<thead>
<tr>
<th>PWM Frequency</th>
<th>1.95 kHz</th>
<th>7.81 kHz</th>
<th>31.25 kHz</th>
<th>125 kHz</th>
<th>250 kHz</th>
<th>333.3 kHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>Timer Prescale (1, 4, 16)</td>
<td>16</td>
<td>4</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>PR2 Value</td>
<td>0xFF</td>
<td>0xFF</td>
<td>0xFF</td>
<td>0x3F</td>
<td>0x1F</td>
<td>0x17</td>
</tr>
<tr>
<td>Maximum Resolution (bits)</td>
<td>10</td>
<td>10</td>
<td>10</td>
<td>8</td>
<td>7</td>
<td>6.6</td>
</tr>
</tbody>
</table>

**TABLE 24-5: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz)**

<table>
<thead>
<tr>
<th>PWM Frequency</th>
<th>1.22 kHz</th>
<th>4.88 kHz</th>
<th>19.53 kHz</th>
<th>78.12 kHz</th>
<th>156.3 kHz</th>
<th>208.3 kHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>Timer Prescale (1, 4, 16)</td>
<td>16</td>
<td>4</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>PR2 Value</td>
<td>0xFF</td>
<td>0xFF</td>
<td>0xFF</td>
<td>0x3F</td>
<td>0x1F</td>
<td>0x17</td>
</tr>
<tr>
<td>Maximum Resolution (bits)</td>
<td>10</td>
<td>10</td>
<td>10</td>
<td>8</td>
<td>7</td>
<td>6.6</td>
</tr>
</tbody>
</table>

**TABLE 24-6: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz)**

<table>
<thead>
<tr>
<th>PWM Frequency</th>
<th>1.22 kHz</th>
<th>4.90 kHz</th>
<th>19.61 kHz</th>
<th>76.92 kHz</th>
<th>153.85 kHz</th>
<th>200.0 kHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>Timer Prescale (1, 4, 16)</td>
<td>16</td>
<td>4</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>PR2 Value</td>
<td>0x65</td>
<td>0x65</td>
<td>0x65</td>
<td>0x19</td>
<td>0x0C</td>
<td>0x09</td>
</tr>
<tr>
<td>Maximum Resolution (bits)</td>
<td>8</td>
<td>8</td>
<td>8</td>
<td>6</td>
<td>5</td>
<td>5</td>
</tr>
</tbody>
</table>
24.3.6 OPERATION IN SLEEP MODE
In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the CCP1 pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

24.3.7 CHANGES IN SYSTEM CLOCK FREQUENCY
The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See Section 5.0 “Oscillator Module (With Fail-Safe Clock Monitor)” for additional details.

24.3.8 EFFECTS OF RESET
Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

24.3.9 ALTERNATE PIN LOCATIONS
This module incorporates I/O pins that can be moved to other locations with the use of the alternate pin function register, APFCON. To determine which pins can be moved and what their default locations are upon a reset, see Section 12.1 “Alternate Pin Function” for more information.

### TABLE 24-7: SUMMARY OF REGISTERS ASSOCIATED WITH STANDARD PWM

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>APFCON</td>
<td>—</td>
<td></td>
<td></td>
<td></td>
<td>T1GSEL</td>
<td>TXCKSEL</td>
<td>P1BSEL</td>
<td>CCP1SEL</td>
<td>103</td>
</tr>
<tr>
<td>CCP1CON</td>
<td>P1M&lt;1:0&gt;</td>
<td>DC1B&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>192</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>PR2</td>
<td></td>
<td></td>
<td>Timer2 Period Register</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>161*</td>
</tr>
<tr>
<td>T2CON</td>
<td>—</td>
<td></td>
<td>T2OUTPS&lt;3:0&gt;</td>
<td></td>
<td>TMR2ON</td>
<td>T2CKPS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td>163</td>
</tr>
<tr>
<td>TMR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>161</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
</tbody>
</table>

Legend: — = Unimplemented location, read as ‘0’. Shaded cells are not used by the PWM.

* Page provides register information.
24.4 PWM (Enhanced Mode)

The enhanced PWM mode generates a Pulse-Width Modulation (PWM) signal on up to two different output pins with up to ten bits of resolution. The period, duty cycle, and resolution are controlled by the following registers:

- PR2 registers
- T2CON registers
- CCPRI1 registers
- CCP1CON registers

The ECCP modules have the following additional PWM registers which control Auto-shutdown, Auto-restart, Dead-band Delay and PWM Steering modes:

- CCP1AS registers
- PSTR1CON registers
- PWM1CON registers

The enhanced PWM module can generate the following three PWM Output modes:

- Single PWM
- Half-Bridge PWM
- Single PWM with PWM Steering Mode

To select an Enhanced PWM Output mode, the P1M bits of the CCP1CON register must be configured appropriately.

The PWM outputs are multiplexed with I/O pins and are designated P1A and P1B. The polarity of the PWM pins is configurable and is selected by setting the bits CCP1M<3:0> in the CCP1CON register appropriately.

Figure 24-5 shows an example of a simplified block diagram of the Enhanced PWM module.

Table 24-8 shows the pin assignments for various Enhanced PWM modes.

Note 1: The corresponding TRIS bit must be cleared to enable the PWM output on the CCP1 pin.

2: Clearing the CCP1CON register will relinquish control of the CCP1 pin.

3: Any pin not used in the enhanced PWM mode is available for alternate pin functions, if applicable.

4: To prevent the generation of an incomplete waveform when the PWM is first enabled, the ECCP module waits until the start of a new PWM period before generating a PWM signal.

FIGURE 24-5: EXAMPLE SIMPLIFIED BLOCK DIAGRAM OF THE ENHANCED PWM MODE

Note 1: The 8-bit timer TMR1 register is concatenated with the 2-bit internal Q clock, or two bits of the prescaler to create the 10-bit time base.
TABLE 24-8: EXAMPLE PIN ASSIGNMENTS FOR VARIOUS PWM ENHANCED MODES

<table>
<thead>
<tr>
<th>ECCP Mode</th>
<th>P1M&lt;1:0&gt;</th>
<th>CCP1/P1A</th>
<th>P1B</th>
</tr>
</thead>
<tbody>
<tr>
<td>Single</td>
<td>00</td>
<td>Yes(1)</td>
<td>Yes(1)</td>
</tr>
<tr>
<td>Half-Bridge</td>
<td>10</td>
<td>Yes</td>
<td>Yes</td>
</tr>
</tbody>
</table>

Note 1: PWM Steering enables outputs in Single mode.

FIGURE 24-6: EXAMPLE PWM (ENHANCED MODE) OUTPUT RELATIONSHIPS (ACTIVE-HIGH STATE)

FIGURE 24-7: EXAMPLE ENHANCED PWM OUTPUT RELATIONSHIPS (ACTIVE-LOW STATE)
24.4.1 HALF-BRIDGE MODE

In Half-Bridge mode, two pins are used as outputs to drive push-pull loads. The PWM output signal is output on the CCP1/P1A pin, while the complementary PWM output signal is output on the P1B pin (see Figure 24-9). This mode can be used for half-bridge applications, as shown in Figure 24-9, or for full-bridge applications, where four power switches are being modulated with two PWM signals.

In Half-Bridge mode, the programmable dead-band delay can be used to prevent shoot-through current in half-bridge power devices. The value of the P1DC<6:0> bits of the PWM1CON register sets the number of instruction cycles before the output is driven active. If the value is greater than the duty cycle, the corresponding output remains inactive during the entire cycle. See Section 24.4.4 “Programmable Dead-Band Delay Mode” for more details of the dead-band delay operations.

Since the P1A and P1B outputs are multiplexed with the PORT data latches, the associated TRIS bits must be cleared to configure P1A and P1B as outputs.

**FIGURE 24-8: EXAMPLE OF HALF-BRIDGE PWM OUTPUT**

**FIGURE 24-9: EXAMPLE OF HALF-BRIDGE APPLICATIONS**
24.4.2 ENHANCED PWM AUTO-SHUTDOWN MODE

The PWM mode supports an Auto-Shutdown mode that will disable the PWM outputs when an external shutdown event occurs. Auto-Shutdown mode places the PWM output pins into a predetermined state. This mode is used to help prevent the PWM from damaging the application.

The auto-shutdown sources are selected using the CCP1AS<2:0> bits of the CCP1AS register. A shutdown event may be generated by:

- A logic ‘0’ on the INT pin
- Comparator C1
- Setting the CCP1ASE bit in firmware

A shutdown condition is indicated by the CCP1ASE (Auto-Shutdown Event Status) bit of the CCP1AS register. If the bit is a ‘0’, the PWM pins are operating normally. If the bit is a ‘1’, the PWM outputs are in the shutdown state.

When a shutdown event occurs, two things happen:

The CCP1ASE bit is set to ‘1’. The CCP1ASE will remain set until cleared in firmware or an auto-restart occurs (see Section 24.4.3 “Auto-Restart Mode”).

The enabled PWM pins are asynchronously placed in their shutdown states. The PWM output pins are grouped into pairs [P1A] and [P1B]. The state of each pin pair is determined by the PSS1AC and PSS1BD bits of the CCP1AS register. Each pin pair may be placed into one of three states:

- Drive logic ‘1’
- Drive logic ‘0’
- Tri-state (high-impedance)

Note 1: The auto-shutdown condition is a level-based signal, not an edge-based signal. As long as the level is present, the auto-shutdown will persist.

Note 2: Writing to the CCP1ASE bit is disabled while an auto-shutdown condition persists.

Note 3: Once the auto-shutdown condition has been removed and the PWM restarted (either through firmware or auto-restart) the PWM signal will always restart at the beginning of the next PWM period.

FIGURE 24-10: PWM AUTO-SHUTDOWN WITH FIRMWARE RESTART (P1RSEN = 0)
24.4.3 AUTO-RESTART MODE

The Enhanced PWM can be configured to automatically restart the PWM signal once the auto-shutdown condition has been removed. Auto-restart is enabled by setting the P1RSEN bit in the PWM1CON register.

If auto-restart is enabled, the CCP1ASE bit will remain set as long as the auto-shutdown condition is active. When the auto-shutdown condition is removed, the CCP1ASE bit will be cleared via hardware and normal operation will resume.

FIGURE 24-11: PWM AUTO-SHUTDOWN WITH AUTO-RESTART (P1RSEN = 1)
24.4.4 PROGRAMMABLE DEAD-BAND DELAY MODE

In Half-Bridge applications where all power switches are modulated at the PWM frequency, the power switches normally require more time to turn off than to turn on. If both the upper and lower power switches are switched at the same time (one turned on, and the other turned off), both switches may be on for a short period of time until one switch completely turns off. During this brief interval, a very high current (shoot-through current) will flow through both power switches, shorting the bridge supply. To avoid this potentially destructive shoot-through current from flowing during switching, turning on either of the power switches is normally delayed to allow the other switch to completely turn off.

In Half-Bridge mode, a digitally programmable dead-band delay is available to avoid shoot-through current from destroying the bridge power switches. The delay occurs at the signal transition from the non-active state to the active state. See Figure 24-12 for illustration. The lower seven bits of the associated PWM1CON register (Register 24-3) sets the delay period in terms of microcontroller instruction cycles (TcY or 4 Tosc).

FIGURE 24-12: EXAMPLE OF HALF-BRIDGE PWM OUTPUT

FIGURE 24-13: EXAMPLE OF HALF-BRIDGE APPLICATIONS
24.4.5 PWM STEERING MODE

In Single Output mode, PWM steering allows any of the PWM pins to be the modulated signal. Additionally, the same PWM signal can be simultaneously available on multiple pins.

Once the Single Output mode is selected (CCPM<3:2> = 11 and P1M<1:0> = 00 of the CCP1CON register), the user firmware can bring out the same PWM signal to one or two output pins by setting the appropriate STR1 bits of the PSTR1CON register, as shown in Table 24-8.

Note: The associated TRIS bits must be set to output (‘0’) to enable the pin output driver in order to see the PWM signal on the pin.

While the PWM Steering mode is active, the CCP1M<1:0> bits of the CCP1CON register determine the polarity of the output pins.

The PWM auto-shutdown operation also applies to PWM Steering mode as described in Section 24.4.2 “Enhanced PWM Auto-shutdown mode”. An auto-shutdown event will only affect pins that have PWM outputs enabled.

24.4.5.1 Steering Synchronization

The STR1SYNC bit of the PSTR1CON register gives the user two selections of when the steering event will happen. When the STR1SYNC bit is ‘0’, the steering event will happen at the end of the instruction that writes to the PSTR1CON register. In this case, the output signal at the output pins may be an incomplete PWM waveform. This operation is useful when the user firmware needs to immediately remove a PWM signal from the pin.

When the STR1SYNC bit is ‘1’, the effective steering update will happen at the beginning of the next PWM period. In this case, steering on/off the PWM output will always produce a complete PWM waveform.

Figures 24-15 and 24-16 illustrate the timing diagrams of the PWM steering depending on the STR1SYNC setting.

24.4.6 START-UP CONSIDERATIONS

When any PWM mode is used, the application hardware must use the proper external pull-up and/or pull-down resistors on the PWM output pins.

The CCP1M<1:0> bits of the CCP1CON register allow the user to choose whether the PWM output signals are active-high or active-low for each of the PWM output pins (P1A and P1B). The PWM output polarities must be selected before the PWM pin output drivers are enabled. Changing the polarity configuration while the PWM pin output drivers are enable is not recommended since it may result in damage to the application circuits.

The P1A and P1B output latches may not be in the proper states when the PWM module is initialized. Enabling the PWM pin output drivers at the same time as the Enhanced PWM modes may cause damage to the application circuit. The Enhanced PWM modes must be enabled in the proper Output mode and complete a full PWM cycle before enabling the PWM pin output drivers. The completion of a full PWM cycle is indicated by the TMR2IF bit of the PIR1 register being set as the second PWM period begins.

Note: When the microcontroller is released from Reset, all of the I/O pins are in the high-impedance state. The external circuits must keep the power switch devices in the Off state until the microcontroller drives the I/O pins with the proper signal levels or activates the PWM output(s).
FIGURE 24-15: EXAMPLE OF STEERING EVENT AT END OF INSTRUCTION (STR1SYNC = 0)

FIGURE 24-16: EXAMPLE OF STEERING EVENT AT BEGINNING OF INSTRUCTION (STR1SYNC = 1)
24.4.7 ALTERNATE PIN LOCATIONS
This module incorporates I/O pins that can be moved to other locations with the use of the alternate pin function register, APFCON. To determine which pins can be moved and what their default locations are upon a Reset, see Section 12.1 “Alternate Pin Function” for more information.

### TABLE 24-9: SUMMARY OF REGISTERS ASSOCIATED WITH ENHANCED PWM

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>APFCON</td>
<td></td>
<td>RXDTSEL</td>
<td>SDOSEL</td>
<td>SSSEL</td>
<td>—</td>
<td>T1GSEL</td>
<td>TXCKSEL</td>
<td>P1BSEL</td>
<td>CCP1SEL</td>
</tr>
<tr>
<td>CCP1CON</td>
<td>P1M&lt;1:0&gt;</td>
<td>DC1B&lt;1:0&gt;</td>
<td>CCP1M&lt;3:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>CCP1AS</td>
<td>CCP1ASE</td>
<td>CCP1AS&lt;2:0&gt;</td>
<td>PSS1AC&lt;1:0&gt;</td>
<td>PSS1BD&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IO CIF</td>
<td></td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td></td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td></td>
</tr>
<tr>
<td>PR2</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>PSTR1SYNC</td>
<td>Reserved</td>
<td>STR1B</td>
</tr>
<tr>
<td>PSTR1CON</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>STR1SYNC</td>
<td>Reserved</td>
<td>Reserved</td>
<td>STR1B</td>
<td>STR1A</td>
</tr>
<tr>
<td>PWM1CON</td>
<td>P1RSEN</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>P1DC&lt;6:0&gt;</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>T2CON</td>
<td>—</td>
<td>T2OUTPS&lt;3:0&gt;</td>
<td>TMR2ON</td>
<td>T2CKPS&lt;1:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TMR2</td>
<td>Timer2 Module Register</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td></td>
</tr>
</tbody>
</table>

**Legend:** — = Unimplemented location, read as ‘0’. Shaded cells are not used by the PWM.

* Page provides register information.
## REGISTER 24-1: CCP1CON: CCP1 CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit 7-6</th>
<th>P1M&lt;1:0&gt;: Enhanced PWM Output Configuration bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>Capture mode: Unused</td>
<td></td>
</tr>
<tr>
<td>Compare mode: Unused</td>
<td></td>
</tr>
<tr>
<td>PWM mode:</td>
<td></td>
</tr>
<tr>
<td>If CCP1M&lt;3:2&gt; = 00, 01, 10:</td>
<td></td>
</tr>
<tr>
<td>xx = P1A assigned as Capture/Compare input; P1B assigned as port pins(1)</td>
<td></td>
</tr>
<tr>
<td>If CCP1M&lt;3:2&gt; = 11:</td>
<td></td>
</tr>
<tr>
<td>00 = Single output; P1A modulated; P1B assigned as port pins</td>
<td></td>
</tr>
<tr>
<td>01 = Reserved</td>
<td></td>
</tr>
<tr>
<td>10 = Half-Bridge output; P1A, P1B modulated with dead-band control</td>
<td></td>
</tr>
<tr>
<td>11 = Reserved</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 5-4</th>
<th>DC1B&lt;1:0&gt;: PWM Duty Cycle Least Significant bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>Capture mode: Unused</td>
<td></td>
</tr>
<tr>
<td>Compare mode: Unused</td>
<td></td>
</tr>
<tr>
<td>PWM mode:</td>
<td></td>
</tr>
<tr>
<td>These bits are the two LSbs of the PWM duty cycle. The eight MSbs are found in CCPR1L.</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 3-0</th>
<th>CCP1M&lt;3:0&gt;: ECCP1 Mode Select bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>0000 = Capture/Compare/PWM off (resets ECCP1 module)</td>
<td></td>
</tr>
<tr>
<td>0001 = Reserved</td>
<td></td>
</tr>
<tr>
<td>0010 = Compare mode: toggle output on match</td>
<td></td>
</tr>
<tr>
<td>0011 = Reserved</td>
<td></td>
</tr>
<tr>
<td>0100 = Capture mode: every falling edge</td>
<td></td>
</tr>
<tr>
<td>0101 = Capture mode: every rising edge</td>
<td></td>
</tr>
<tr>
<td>0110 = Capture mode: every 4th rising edge</td>
<td></td>
</tr>
<tr>
<td>0111 = Capture mode: every 16th rising edge</td>
<td></td>
</tr>
<tr>
<td>1000 = Compare mode: initialize ECCP1 pin low; set output on compare match (set CCP1IF)</td>
<td></td>
</tr>
<tr>
<td>1001 = Compare mode: initialize ECCP1 pin high; clear output on compare match (set CCP1IF)</td>
<td></td>
</tr>
<tr>
<td>1010 = Compare mode: generate software interrupt only; ECCP1 pin reverts to I/O state</td>
<td></td>
</tr>
<tr>
<td>1011 = Compare mode: Special Event Trigger (CCP1 resets Timer, sets CCP1IF bit, and starts A/D conversion if A/D module is enabled)</td>
<td></td>
</tr>
</tbody>
</table>

### PWM mode:
- 1100 = PWM mode: P1A active-high; P1B active-high
- 1101 = PWM mode: P1A active-high; P1B active-low
- 1110 = PWM mode: P1A active-low; P1B active-high
- 1111 = PWM mode: P1A active-low; P1B active-low

---

The table above illustrates the different configurations for CCP1CON register, detailing the functionality for P1M<1:0>, DC1B<1:0>, and CCP1M<3:0> bits. Each configuration is described in terms of Capture, Compare, and PWM modes, with specific conditions for each mode and the respective states of the bits in question.
### REGISTER 24-2: CCP1AS: CCP1 AUTO-SHUTDOWN CONTROL REGISTER

<table>
<thead>
<tr>
<th>bit</th>
<th>CCP1ASE</th>
<th>CCP1AS&lt;2:0&gt;</th>
<th>PSS1AC&lt;1:0&gt;</th>
<th>PSS1BD&lt;1:0&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
</tr>
<tr>
<td>6-4</td>
<td></td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
</tr>
<tr>
<td>3-2</td>
<td></td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
</tr>
<tr>
<td>1-0</td>
<td></td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
<td>R/W-0/0</td>
</tr>
</tbody>
</table>

**Legend:**
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

**bit 7**  
**CCP1ASE:** CCP1 Auto-Shutdown Event Status bit  
1 = A shutdown event has occurred; CCP1 outputs are in shutdown state  
0 = CCP1 outputs are operating

**bit 6-4**  
**CCP1AS<2:0>:** CCP1 Auto-Shutdown Source Select bits  
000 = Auto-shutdown is disabled  
001 = Comparator C1 output low(1)  
010 = Reserved  
011 = Either Comparator C1 output low(1)  
100 = VIL on INT pin  
101 = VIL on INT pin or Comparator C1 low(1)  
110 = Reserved  
111 = VIL on INT pin or Comparator C1 low(1)

**bit 3-2**  
**PSS1AC<1:0>:** Pin P1A Shutdown State Control bits  
00 = Drive pin P1A to ‘0’  
01 = Drive pin P1A to ‘1’  
1x = Pin P1A tri-state

**bit 1-0**  
**PSS1BD<1:0>:** Pin P1B Shutdown State Control bits  
00 = Drive pin P1B to ‘0’  
01 = Drive pin P1B to ‘1’  
1x = Pin P1B tri-state

**Note 1:** If C1SYNC is enabled, the shutdown will be delayed by Timer1.
## REGISTER 24-3: PWM1CON: ENHANCED PWM CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>P1RSEN</td>
<td></td>
<td></td>
<td></td>
<td>P1DC&lt;6:0&gt;</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W =Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

**bit 7**

<table>
<thead>
<tr>
<th>P1RSEN: PWM Restart Enable bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1 = Upon auto-shutdown, the CCP1ASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically</td>
</tr>
<tr>
<td>0 = Upon auto-shutdown, CCP1ASE must be cleared in software to restart the PWM</td>
</tr>
</tbody>
</table>

**bit 6-0**

<table>
<thead>
<tr>
<th>P1DC&lt;6:0&gt;: PWM Delay Count bits</th>
</tr>
</thead>
<tbody>
<tr>
<td>P1DC1 = Number of Fosc/4 (4 * Tosc) cycles between the scheduled time when a PWM signal should transition active and the actual time it transitions active</td>
</tr>
</tbody>
</table>

**Note 1:** Bit resets to ‘0’ with Two-Speed Start-up and LP, XT or HS selected as the Oscillator mode or Fail-Safe mode is enabled.

## REGISTER 24-4: PSTR1CON: PWM STEERING CONTROL REGISTER

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td>—</td>
<td>—</td>
<td>—</td>
<td>STR1SYNC</td>
<td>Reserved</td>
<td>Reserved</td>
<td>STR1B</td>
<td>STR1A</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets

**bit 7**

### Unimplemented: Read as ‘0’

**bit 4**

<table>
<thead>
<tr>
<th>STR1SYNC: Steering Sync bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>1 = Output steering update occurs on next PWM period</td>
</tr>
<tr>
<td>0 = Output steering update occurs at the beginning of the instruction cycle boundary</td>
</tr>
</tbody>
</table>

**bit 3-2**

### Reserved: Read as ‘0’. Maintain these bits clear.

**bit 1**

<table>
<thead>
<tr>
<th>STR1B: Steering Enable bit B</th>
</tr>
</thead>
<tbody>
<tr>
<td>1 = P1B pin has the PWM waveform with polarity control from CCP1M&lt;1:0&gt;</td>
</tr>
<tr>
<td>0 = P1B pin is assigned to port pin</td>
</tr>
</tbody>
</table>

**bit 0**

<table>
<thead>
<tr>
<th>STR1A: Steering Enable bit A</th>
</tr>
</thead>
<tbody>
<tr>
<td>1 = P1A pin has the PWM waveform with polarity control from CCP1M&lt;1:0&gt;</td>
</tr>
<tr>
<td>0 = P1A pin is assigned to port pin</td>
</tr>
</tbody>
</table>

**Note 1:** The PWM Steering mode is available only when the CCP1CON register bits CCP1M<3:2> = 11 and P1M<1:0> = 00.
25.0 MASTER SYNCHRONOUS SERIAL PORT MODULE

25.1 Master SSP (MSSP1) Module

Overview

The Master Synchronous Serial Port (MSSP1) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be Serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP1 module can operate in one of two modes:

• Serial Peripheral Interface (SPI)
• Inter-Integrated Circuit (I^2C™)

The SPI interface supports the following modes and features:

• Master mode
• Slave mode
• Clock Parity
• Slave Select Synchronization (Slave mode only)
• Daisy-chain connection of slave devices

Figure 25-1 is a block diagram of the SPI interface module.

**FIGURE 25-1: MSSP1 BLOCK DIAGRAM (SPI MODE)**
The I²C interface supports the following modes and features:

- Master mode
- Slave mode
- Byte NACKing (Slave mode)
- Limited Multi-master support
- 7-bit and 10-bit addressing
- Start and Stop interrupts
- Interrupt masking
- Clock stretching
- Bus collision detection
- General call address matching
- Address masking
- Address Hold and Data Hold modes
- Selectable SDA hold times

Figure 25-2 is a block diagram of the I²C interface module in Master mode. Figure 25-3 is a diagram of the I²C interface module in Slave mode.

**FIGURE 25-2: MSSP1 BLOCK DIAGRAM (I²C™ MASTER MODE)**
FIGURE 25-3: MSSP1 BLOCK DIAGRAM (I²C™ SLAVE MODE)
25.2 SPI Mode Overview

The Serial Peripheral Interface (SPI) bus is a synchronous serial data communication bus that operates in Full-Duplex mode. Devices communicate in a master/slave environment where the master device initiates the communication. A slave device is controlled through a chip select known as Slave Select. The SPI bus specifies four signal connections:

- Serial Clock (SCK)
- Serial Data Out (SDO)
- Serial Data In (SDI)
- Slave Select (SS)

Figure 25-1 shows the block diagram of the MSSP1 module when operating in SPI Mode.

The SPI bus operates with a single master device and one or more slave devices. When multiple slave devices are used, an independent Slave Select connection is required from the master device to each slave device.

Figure 25-4 shows a typical connection between a master device and multiple slave devices.

The master selects only one slave at a time. Most slave devices have tri-state outputs so their output signal appears disconnected from the bus when they are not selected.

Transmissions involve two shift registers, eight bits in size, one in the master and one in the slave. With either the master or the slave device, data is always shifted out one bit at a time, with the Most Significant bit (MSb) shifted out first. At the same time, a new Least Significant bit (LSb) is shifted into the same register.

Figure 25-5 shows a typical connection between two processors configured as master and slave devices.

Data is shifted out of both shift registers on the programmed clock edge and latched on the opposite edge of the clock.

The master device transmits information out on its SDO output pin which is connected to, and received by, the slave’s SDI input pin. The slave device transmits information out on its SDO output pin, which is connected to, and received by, the master’s SDI input pin.

To begin communication, the master device first sends out the clock signal. Both the master and the slave devices should be configured for the same clock polarity.

The master device starts a transmission by sending out the MSb from its shift register. The slave device reads this bit from that same line and saves it into the LSb position of its shift register.

During each SPI clock cycle, a full-duplex data transmission occurs. This means that while the master device is sending out the MSb from its shift register (on its SDO pin) and the slave device is reading this bit and saving it as the LSb of its shift register, that the slave device is also sending out the MSb from its shift register (on its SDO pin) and the master device is reading this bit and saving it as the LSb of its shift register.

After eight bits have been shifted out, the master and slave have exchanged register values.

If there is more data to exchange, the shift registers are loaded with new data and the process repeats itself.

Whether the data is meaningful or not (dummy data), depends on the application software. This leads to three scenarios for data transmission:

- Master sends useful data and slave sends dummy data
- Master sends useful data and slave sends useful data
- Master sends dummy data and slave sends useful data

Transmissions may involve any number of clock cycles. When there is no more data to be transmitted, the master stops sending the clock signal and it deselects the slave.

Every slave device connected to the bus that has not been selected through its slave select line must disregard the clock and transmission signals and must not transmit out any data of its own.
25.2.1 SPI MODE REGISTERS

The MSSP1 module has five registers for SPI mode operation. These are:

- MSSP1 STATUS register (SSP1STAT)
- MSSP1 Control Register 1 (SSP1CON1)
- MSSP1 Control Register 3 (SSP1CON3)
- MSSP1 Data Buffer register (SSP1BUF)
- MSSP1 Address register (SSP1ADD)
- MSSP1 Shift register (SSP1SR)
  (Not directly accessible)

SSP1CON1 and SSP1STAT are the control and STATUS registers in SPI mode operation. The SSP1CON1 register is readable and writable. The lower six bits of the SSP1STAT are read-only. The upper two bits of the SSP1STAT are read/write.

In one SPI master mode, SSP1ADD can be loaded with a value used in the Baud Rate Generator. More information on the Baud Rate Generator is available in Section 25.7 “Baud Rate Generator”.

SSP1SR is the shift register used for shifting data in and out. SSP1BUF provides indirect access to the SSP1SR register. SSP1BUF is the buffer register to which data bytes are written, and from which data bytes are read.

In receive operations, SSP1SR and SSP1BUF together create a buffered receiver. When SSP1SR receives a complete byte, it is transferred to SSP1BUF and the SSP1IF interrupt is set.

During transmission, the SSP1BUF is not buffered. A write to SSP1BUF will write to both SSP1BUF and SSP1SR.
25.2.2 SPI MODE OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSP1CON1<5:0> and SSP1STAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCK1 is the clock output)
- Slave mode (SCK1 is the clock input)
- Clock Polarity (Idle state of SCK1)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCK1)
- Clock Rate (Master mode only)
- Slave Select mode (Slave mode only)

To enable the serial port, SSP1 Enable bit, SSP1EN of the SSP1CON1 register must be set. To reset or reconfigure SPI mode, clear the SSP1EN bit, re-initialize the SSP1CONx registers and then set the SSP1EN bit. This configures the SDI, SDO, SCK and SS pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- SDI must have corresponding TRIS bit set
- SDO must have corresponding TRIS bit cleared
- SCK (Master mode) must have corresponding TRIS bit cleared
- SCK (Slave mode) must have corresponding TRIS bit set
- SS must have corresponding TRIS bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.

The MSSP1 consists of a transmit/receive shift register (SSP1SR) and a buffer register (SSP1BUF). The SSP1SR shifts the data in and out of the device, MSb first. The SSP1BUF holds the data that was written to the SSP1SR until the received data is ready. Once the 8 bits of data have been received, that byte is moved to the SSP1BUF register. Then, the Buffer Full detect bit, BF of the SSP1STAT register, and the interrupt flag bit, SSP1IF, are set. This double-buffering of the received data (SSP1BUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSP1BUF register during transmission/reception of data will be ignored and the Write Collision detect bit, WCOL, of the SSP1CON1 register, will be set. User software must clear the WCOL bit to allow the following write(s) to the SSP1BUF register to complete successfully.

When the application software is expecting to receive valid data, the SSP1BUF should be read before the next byte of data to transfer is written to the SSP1BUF. The Buffer Full bit, BF of the SSP1STAT register, indicates when SSP1BUF has been loaded with the received data (transmission is complete). When the SSP1BUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP1 interrupt is used to determine when the transmission/reception has completed. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur.

The SSP1SR is not directly readable or writable and can only be accessed by addressing the SSP1BUF register. Additionally, the SSP1STAT register indicates the various Status conditions.
FIGURE 25-5: SPI MASTER/SLAVE CONNECTION

SPI Master SSP1M<3:0> = 00xx
= 1010

SPI Slave SSP1M<3:0> = 010x

Serial Input Buffer (SSP1BUF)
Shift Register (SSP1SR)

Serial Input Buffer (SSP1BUF)
Shift Register (SSP1SR)

Processor 1
Processor 2

SDI
SDO
SCK
SS
General I/O (optional)

Serial Clock
Slave Select

MSb
LSb

MSb
LSb
25.2.3 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK line. The master determines when the slave (Processor 2, Figure 25-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSP1BUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSP1SR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSP1BUF register as if a normal received byte (interrupts and Status bits appropriately set).

The clock polarity is selected by appropriately programming the CKP bit of the SSP1CON1 register and the CKE bit of the SSP1STAT register. This then, would give waveforms for SPI communication as shown in Figure 25-6, Figure 25-8 and Figure 25-9, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- \( \text{Fosc}/4 \) (or TCY)
- \( \text{Fosc}/16 \) (or \( 4 \times \text{TCY} \))
- \( \text{Fosc}/64 \) (or \( 16 \times \text{TCY} \))
- Timer2 output/2
- \( \text{Fosc}/(4 \times (\text{SSP1ADD} + 1)) \)

Figure 25-6 shows the waveforms for Master mode.

When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSP1BUF is loaded with the received data is shown.

**FIGURE 25-6: SPI MODE WAVEFORM (MASTER MODE)**
25.2.4  SPI SLAVE MODE

In Slave mode, the data is transmitted and received as external clock pulses appear on SCK. When the last bit is latched, the SSP1IF interrupt flag bit is set.

Before enabling the module in SPI Slave mode, the clock line must match the proper Idle state. The clock line can be observed by reading the SCK pin. The Idle state is determined by the CKP bit of the SSP1CON1 register.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. The shift register is clocked from the SCK pin input and when a byte is received, the device will generate an interrupt. If enabled, the device will wake-up from Sleep.

25.2.4.1  Daisy-Chain Configuration

The SPI bus can sometimes be connected in a daisy-chain configuration. The first slave output is connected to the second slave input, the second slave output is connected to the third slave input, and so on. The final slave output is connected to the master input. Each slave sends out, during a second group of clock pulses, an exact copy of what was received during the first group of clock pulses. The whole chain acts as one large communication shift register. The daisy-chain feature only requires a single Slave Select line from the master device.

Figure 25-7 shows the block diagram of a typical daisy-chain connection when operating in SPI Mode.

In a daisy-chain configuration, only the most recent byte on the bus is required by the slave. Setting the BOEN bit of the SSP1CON3 register will enable writes to the SSP1BUF register, even if the previous byte has not been read. This allows the software to ignore data that may not apply to it.

25.2.5  SLAVE SELECT SYNCHRONIZATION

The Slave Select can also be used to synchronize communication. The Slave Select line is held high until the master device is ready to communicate. When the Slave Select line is pulled low, the slave knows that a new transmission is starting.

If the slave fails to receive the communication properly, it will be reset at the end of the transmission, when the Slave Select line returns to a high state. The slave is then ready to receive a new transmission when the Slave Select line is pulled low again. If the Slave Select line is not used, there is a risk that the slave will eventually become out of sync with the master. If the slave misses a bit, it will always be one bit off in future transmissions. Use of the Slave Select line allows the slave and master to align themselves at the beginning of each transmission.

The S5 pin allows a Synchronous Slave mode. The SPI must be in Slave mode with S5 pin control enabled (SS1CON<3:0> = 0100).

When the S5 pin is low, transmission and reception are enabled and the SDO pin is driven.

When the S5 pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

Note 1: When the SPI is in Slave mode with S5 pin control enabled (SS1CON<3:0> = 0100), the SPI module will reset if the S5 pin is set to VDD.

2: When the SPI is used in Slave mode with CKE set; the user must enable S5 pin control.

3: While operated in SPI Slave mode the SMP bit of the SSP1STAT register must remain clear.

When the SPI module resets, the bit counter is forced to ‘0’. This can be done by either forcing the S5 pin to a high level or clearing the SSP1EN bit.
FIGURE 25-7: SPI DAISY-CHAIN CONNECTION

FIGURE 25-8: SLAVE SELECT SYNCHRONOUS WAVEFORM
FIGURE 25-9: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)

FIGURE 25-10: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)
25.2.6 SPI OPERATION IN SLEEP MODE

In SPI Master mode, module clocks may be operating at a different speed than when in full power mode; in the case of the Sleep mode, all clocks are halted.

Special care must be taken by the user when the MSSP1 clock is much faster than the system clock.

In Slave mode, when MSSP1 interrupts are enabled, after the master completes sending data, an MSSP1 interrupt will wake the controller from Sleep.

If an exit from Sleep mode is not desired, MSSP1 interrupts should be disabled.

In SPI Master mode, when the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the device wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in Sleep mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP1 interrupt flag bit will be set and if enabled, will wake the device.

### TABLE 25-1: SUMMARY OF REGISTERS ASSOCIATED WITH SPI OPERATION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA4</td>
<td>—</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
<td>107</td>
</tr>
<tr>
<td>APFCON</td>
<td>RXDTSEL</td>
<td>SDOSEL</td>
<td>SSSEL</td>
<td>—</td>
<td>T1GSEL</td>
<td>TXCKSEL</td>
<td>P1BSEL</td>
<td>CCP1SEL</td>
<td>103</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>SSP1BUF</td>
<td>Synchronous Serial Port Receive Buffer/Transmit Register</td>
<td>199*</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SSP1CON1</td>
<td>WCOL</td>
<td>SSP1OV</td>
<td>SSP1EN</td>
<td>CKP</td>
<td>SSP1M&lt;3:0&gt;</td>
<td>246</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SSP1CON3</td>
<td>ACKTIM</td>
<td>PCIE</td>
<td>SCIE</td>
<td>BOEN</td>
<td>SDAHT</td>
<td>SBCDE</td>
<td>AHEN</td>
<td>DHEN</td>
<td>248</td>
</tr>
<tr>
<td>SSP1STAT</td>
<td>SMP</td>
<td>CKE</td>
<td>D/A</td>
<td>P</td>
<td>S</td>
<td>R/W</td>
<td>UA</td>
<td>BF</td>
<td>245</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
</tbody>
</table>

**Legend:**
- = Unimplemented location, read as ‘0’. Shaded cells are not used by the MSSP1 in SPI mode.
- Page provides register information.
25.3  \( \text{i}^2\text{C} \) MODE OVERVIEW

The Inter-Integrated Circuit Bus (\( \text{i}^2\text{C} \)) is a multi-master serial data communication bus. Devices communicate in a master/slave environment where the master devices initiate the communication. A Slave device is controlled through addressing.

The \( \text{i}^2\text{C} \) bus specifies two signal connections:
- Serial Clock (SCL)
- Serial Data (SDA)

Figure 25-11 shows the block diagram of the MSSP1 module when operating in \( \text{i}^2\text{C} \) mode.

Both the SCL and SDA connections are bidirectional open-drain lines, each requiring pull-up resistors for the supply voltage. Pulling the line to ground is considered a logical zero and letting the line float is considered a logical one.

Figure 25-11 shows a typical connection between two processors configured as master and slave devices.

The \( \text{i}^2\text{C} \) bus can operate with one or more master devices and one or more slave devices.

There are four potential modes of operation for a given device:
- Master Transmit mode (master is transmitting data to a slave)
- Master Receive mode (master is receiving data from a slave)
- Slave Transmit mode (slave is transmitting data to a master)
- Slave Receive mode (slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device.

If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an \( \text{ACK} \). The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

The \( \text{i}^2\text{C} \) bus specifies three message protocols;
- Single message where a master writes data to a slave.
- Single message where a master reads data from a slave.
- Combined message where a master initiates a minimum of two writes, or two reads, or a combination of writes and reads, to one or more slaves.
When one device is transmitting a logical one, or letting the line float, and a second device is transmitting a logical zero, or holding the line low, the first device can detect that the line is not a logical one. This detection, when used on the SCL line, is called clock stretching. Clock stretching gives slave devices a mechanism to control the flow of data. When this detection is used on the SDA line, it is called arbitration. Arbitration ensures that there is only one master device communicating at any single time.

### 25.3.1 CLOCK STRETCHING

When a slave device has not completed processing data, it can delay the transfer of more data through the process of clock stretching. An addressed slave device may hold the SCL clock line low after receiving or sending a bit, indicating that it is not yet ready to continue. The master that is communicating with the slave will attempt to raise the SCL line in order to transfer the next bit, but will detect that the clock line has not yet been released. Because the SCL connection is open-drain, the slave has the ability to hold that line low until it is ready to continue communicating.

Clock stretching allows receivers that cannot keep up with a transmitter to control the flow of incoming data.

### 25.3.2 ARBITRATION

Each master device must monitor the bus for Start and Stop bits. If the device detects that the bus is busy, it cannot begin a new message until the bus returns to an Idle state.

However, two master devices may try to initiate a transmission on or about the same time. When this occurs, the process of arbitration begins. Each transmitter checks the level of the SDA data line and compares it to the level that it expects to find. The first transmitter to observe that the two levels do not match, loses arbitration, and must stop transmitting on the SDA line.

For example, if one transmitter holds the SDA line to a logical one (lets it float) and a second transmitter holds it to a logical zero (pulls it low), the result is that the SDA line will be low. The first transmitter then observes that the level of the line is different than expected and concludes that another transmitter is communicating.

The first transmitter to notice this difference is the one that loses arbitration and must stop driving the SDA line. If this transmitter is also a master device, it also must stop driving the SCL line. It then can monitor the lines for a Stop condition before trying to reissue its transmission. In the meantime, the other device that has not noticed any difference between the expected and actual levels on the SDA line continues with its original transmission. It can do so without any complications, because so far, the transmission appears exactly as expected with no other transmitter disturbing the message.

Slave Transmit mode can also be arbitrated, when a master addresses multiple slaves, but this is less common.

If two master devices are sending a message to two different slave devices at the address stage, the master sending the lower slave address always wins arbitration. When two master devices send messages to the same slave address, and addresses can sometimes refer to multiple slaves, the arbitration process must continue into the data stage.

Arbitration usually occurs very rarely, but it is a necessary process for proper multi-master support.

### 25.4 I2C Mode Operation

All MSSP1 I2C communication is byte oriented and shifted out MSb first. Six SFR registers and two interrupt flags interface the module with the PIC® microcontroller and user software. Two pins, SDA and SCL, are exercised by the module to communicate with other external I2C devices.

#### 25.4.1 BYTE FORMAT

All communication in I2C is done in 9-bit segments. A byte is sent from a master to a slave or vice-versa, followed by an Acknowledge bit sent back. After the 8th falling edge of the SCL line, the device outputting data on the SDA changes that pin to an input and reads in an acknowledge value on the next clock pulse.

The clock signal, SCL, is provided by the master. Data is valid to change while the SCL signal is low, and sampled on the rising edge of the clock. Changes on the SDA line while the SCL line is high define special conditions on the bus, explained below.

#### 25.4.2 DEFINITION OF I2C TERMINOLOGY

There is language and terminology in the description of I2C communication that have definitions specific to I2C. That word usage is defined below and may be used in the rest of this document without explanation. This table was adapted from the Phillips I2C™ specification.

#### 25.4.3 SDA AND SCL PINS

Selection of any I2C mode with the SSP1EN bit set, forces the SCL and SDA pins to be open-drain. These pins should be set by the user to inputs by setting the appropriate TRIS bits.

**Note:** Data is tied to output zero when an I2C mode is enabled.
25.4.4 SDA HOLD TIME

The hold time of the SDA pin is selected by the SDAHT bit of the SSP1CON3 register. Hold time is the time SDA is held valid after the falling edge of SCL. Setting the SDAHT bit selects a longer 300 ns minimum hold time and may help on buses with large capacitance.

### TABLE 25-2: I2C BUS TERMS

<table>
<thead>
<tr>
<th>TERM</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>Transmitter</td>
<td>The device which shifts data out onto the bus.</td>
</tr>
<tr>
<td>Receiver</td>
<td>The device which shifts data in from the bus.</td>
</tr>
<tr>
<td>Master</td>
<td>The device that initiates a transfer, generates clock signals and terminates a transfer.</td>
</tr>
<tr>
<td>Slave</td>
<td>The device addressed by the master.</td>
</tr>
<tr>
<td>Multi-master</td>
<td>A bus with more than one device that can initiate data transfers.</td>
</tr>
<tr>
<td>Arbitration</td>
<td>Procedure to ensure that only one master at a time controls the bus. Winning arbitration ensures that the message is not corrupted.</td>
</tr>
<tr>
<td>Synchronization</td>
<td>Procedure to synchronize the clocks of two or more devices on the bus.</td>
</tr>
<tr>
<td>Idle</td>
<td>No master is controlling the bus, and both SDA and SCL lines are high.</td>
</tr>
<tr>
<td>Active</td>
<td>Any time one or more master devices are controlling the bus.</td>
</tr>
<tr>
<td>Addressed Slave</td>
<td>Slave device that has received a matching address and is actively being clocked by a master.</td>
</tr>
<tr>
<td>Matching Address</td>
<td>Address byte that is clocked into a slave that matches the value stored in SSP1ADD.</td>
</tr>
<tr>
<td>Write Request</td>
<td>Slave receives a matching address with R/W bit clear, and is ready to clock in data.</td>
</tr>
<tr>
<td>Read Request</td>
<td>Master sends an address byte with the R/W bit set, indicating that it wishes to clock data out of the Slave. This data is the next and all following bytes until a Restart or Stop.</td>
</tr>
<tr>
<td>Clock Stretching</td>
<td>When a device on the bus holds SCL low to stall communication.</td>
</tr>
<tr>
<td>Bus Collision</td>
<td>Any time the SDA line is sampled low by the module while it is outputting and expected high state.</td>
</tr>
</tbody>
</table>

25.4.5 START CONDITION

The I2C specification defines a Start condition as a transition of SDA from a high to a low state while SCL line is high. A Start condition is always generated by the master and signifies the transition of the bus from an Idle to an Active state. Figure 25-12 shows wave forms for Start and Stop conditions.

A bus collision can occur on a Start condition if the module samples the SDA line low before asserting it low. This does not conform to the I2C specification that states no bus collision can occur on a Start.

25.4.6 STOP CONDITION

A Stop condition is a transition of the SDA line from low-to-high state while the SCL line is high.

**Note:** At least one SCL low time must appear before a Stop is valid, therefore, if the SDA line goes low then high again while the SCL line stays high, only the Start condition is detected.

25.4.7 RESTART CONDITION

A Restart is valid any time that a Stop would be valid. A master can issue a Restart if it wishes to hold the bus after terminating the current transfer. A Restart has the same effect on the slave as a Start would, resetting all slave logic and preparing it to clock in an address. The master may want to address the same or another slave. Figure 25-13 shows wave forms for a RESTART condition.

In 10-bit Addressing Slave mode a Restart is required for the master to clock data out of the addressed slave. Once a slave has been fully addressed, matching both high and low address bytes, the master can issue a Restart and the high address byte with the R/W bit set. The slave logic will then hold the clock and prepare to clock out data.

After a full match with R/W clear in 10-bit mode, a prior match flag is set and maintained. Until a Stop condition, a high address with R/W clear, or high address match fails.

25.4.8 START/STOP CONDITION INTERRUPT MASKING

The SCIE and PCIE bits of the SSP1CON3 register can enable the generation of an interrupt in Slave modes that do not typically support this function. Slave modes where interrupt on Start and Stop detect are already enabled, these bits will have no effect.
FIGURE 25-12: \( \text{I}^2\text{C} \) START AND STOP CONDITIONS

FIGURE 25-13: \( \text{I}^2\text{C} \) RESTART CONDITION
25.4.9 ACKNOWLEDGE SEQUENCE

The 9th SCL pulse for any transferred byte in I²C is dedicated as an Acknowledge. It allows receiving devices to respond back to the transmitter by pulling the SDA line low. The transmitter must release control of the line during this time to shift in the response. The Acknowledge (ACK) is an active-low signal, pulling the SDA line low indicated to the transmitter that the device has received the transmitted data and is ready to receive more.

The result of an ACK is placed in the ACKSTAT bit of the SSP1CON2 register.

Slave software, when the AHEN and DHEN bits are set, allow the user to set the ACK value sent back to the transmitter. The ACKDT bit of the register is set/cleared to determine the response.

Slave hardware will generate an ACK response if the AHEN and DHEN bits of the SSP1CON3 register are clear.

There are certain conditions where an ACK will not be sent by the slave. If the BF bit of the SSP1STAT register or the SSP1OV bit of the SSP1CON1 register are set when a byte is received.

When the module is addressed, after the 8th falling edge of SCL on the bus, the ACKTIM bit of the SSP1CON3 register is set. The ACKTIM bit indicates the acknowledge time of the active bus. The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is enabled.

25.5 I²C SLAVE MODE OPERATION

The MSSP1 Slave mode operates in one of four modes selected in the SSP1M bits of the SSP1CON1 register. The modes can be divided into 7-bit and 10-bit Addressing mode. 10-bit Addressing modes operate the same as 7-bit with some additional overhead for handling the larger addresses.

Modes with Start and Stop bit interrupts operated the same as the other modes with SSP1IF additionally getting set upon detection of a Start, Restart or Stop condition.

25.5.1 SLAVE MODE ADDRESSES

The SSP1ADD register (Register 25-6) contains the Slave mode address. The first byte received after a Start or Restart condition is compared against the value stored in this register. If the byte matches, the value is loaded into the SSP1BUF register and an interrupt is generated. If the value does not match, the module goes Idle and no indication is given to the software that anything happened.

The SSP Mask register (Register 25-5) affects the address matching process. See Section 25.5.9 “SSP1 Mask Register” for more information.

25.5.1.1 I²C Slave 7-bit Addressing Mode

In 7-bit Addressing mode, the LSb of the received data byte is ignored when determining if there is an address match.

25.5.1.2 I²C Slave 10-bit Addressing Mode

In 10-bit Addressing mode, the first received byte is compared to the binary value of "1 1 1 1 0 A9 A8 0". A9 and A8 are the two MSb’s of the 10-bit address and stored in bits 2 and 1 of the SSP1ADD register.

After the acknowledge of the high byte the UA bit is set and SCL is held low until the user updates SSP1ADD with the low address. The low address byte is clocked in and all 8 bits are compared to the low address value in SSP1ADD. Even if there is not an address match; SSP1IF and UA are set, and SCL is held low until SSP1ADD is updated to receive a high byte again. When SSP1ADD is updated the UA bit is cleared. This ensures the module is ready to receive the high address byte on the next communication.

A high and low address match as a write request is required at the start of all 10-bit addressing communication. A transmission can be initiated by issuing a Restart once the slave is addressed, and clocking in the high address with the R/W bit set. The slave hardware will then acknowledge the read request and prepare to clock out data. This is only valid for a slave after it has received a complete high and low address byte match.
25.5.2 SLAVE RECEPTION

When the R/W bit of a matching received address byte is clear, the R/W bit of the SSP1STAT register is cleared. The received address is loaded into the SSP1BUF register and acknowledged.

When the overflow condition exists for a received address, then not Acknowledge is given. An overflow condition is defined as either bit BF of the SSP1STAT register is set, or bit SSP1OV of the SSP1CON1 register is set. The BOEN bit of the SSP1CON3 register modifies this operation. For more information see Register 25-4.

An MSSP1 interrupt is generated for each transferred data byte. Flag bit, SSP1IF, must be cleared by software.

When the SEN bit of the SSP1CON2 register is set, SCL will be held low (clock stretch) following each received byte. The clock must be released by setting the CKP bit of the SSP1CON1 register, except sometimes in 10-bit mode. See Section 25.2.3 “SPI Master Mode” for more detail.

25.5.2.1 7-bit Addressing Reception

This section describes a standard sequence of events for the MSSP1 module configured as an I2C Slave in 7-bit Addressing mode. All decisions made by hardware or software and their effect on reception. Figure 25-13 and Figure 25-14 is used as a visual reference for this description.

This is a step by step process of what typically must be done to accomplish I2C communication.

1. Start bit detected.
2. S bit of SSP1STAT is set; SSP1IF is set if interrupt on Start detect is enabled.
3. Matching address with R/W bit clear is received.
4. The slave pulls SDA low sending an ACK to the master, and sets SSP1IF bit.
5. Software clears the SSP1IF bit.
6. Software reads received address from SSP1BUF clearing the BF flag.
7. If SEN = 1; Slave software sets CKP bit to release the SCL line.
8. The master clocks out a data byte.
9. Slave drives SDA low sending an ACK to the master, and sets SSP1IF bit.
10. Software clears SSP1IF.
11. Software reads the received byte from SSP1BUF clearing BF.
12. Steps 8-12 are repeated for all received bytes from the Master.

25.5.2.2 7-bit Reception with AHEN and DHEN

Slave device reception with AHEN and DHEN set operate the same as without these options with extra interrupts and clock stretching added after the 8th falling edge of SCL. These additional interrupts allow the slave software to decide whether it wants to ACK the receive address or data byte, rather than the hardware. This functionality adds support for PMBus™ that was not present on previous versions of this module.

This list describes the steps that need to be taken by slave software to use these options for I2C communication. Figure 25-15 displays a module using both address and data holding. Figure 25-16 includes the operation with the SEN bit of the SSP1CON2 register set.

1. S bit of SSP1STAT is set; SSP1IF is set if interrupt on Start detect is enabled.
2. Matching address with R/W bit clear is clocked in. SSP1IF is set and CKP cleared after the 8th falling edge of SCL.
3. Slave clears the SSP1IF.
4. Slave can look at the ACKTIM bit of the SSP1CON3 register to determine if the SSP1IF was after or before the ACK.
5. Slave reads the address value from SSP1BUF, clearing the BF flag.
6. Slave sets ACK value clocked out to the master by setting ACKDT.
7. Slave releases the clock by setting CKP.
8. SSP1IF is set after an ACK, not after a NACK.
9. If SEN = 1 the slave hardware will stretch the clock after the ACK.
10. Slave clears SSP1IF.

Note: SSP1IF is still set after the 9th falling edge of SCL, even if there is no clock stretching and BF has been cleared. Only if NACK is sent to Master is SSP1IF not set.

11. SSP1IF set and CKP cleared after 8th falling edge of SCL for a received data byte.
12. Slave looks at ACKTIM bit of SSP1CON3 to determine the source of the interrupt.
13. Slave reads the received data from SSP1BUF clearing BF.
14. Steps 7-14 are the same for each received data byte.
15. Communication is ended by either the slave sending an ACK = 1, or the master sending a Stop condition. If a Stop is sent and Interrupt-on-Stop Detect is disabled, the slave will only know by polling the P bit of the SSTSTAT register.
FIGURE 25-14:  \textit{I}^2\textit{C} SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 0, DHEN = 0)

Bus Master sends
Stop condition

Receiving Address

\begin{align*}
\text{A7} & \quad \text{A6} & \quad \text{A5} & \quad \text{A4} & \quad \text{A3} & \quad \text{A2} & \quad \text{A1} \\
\text{D7} & \quad \text{D6} & \quad \text{D5} & \quad \text{D4} & \quad \text{D3} & \quad \text{D2} & \quad \text{D1} & \quad \text{D0}
\end{align*}

From Slave to Master

SCL

SSP1IF

BF

SSP1OV

ACK is not sent.

SSP1OV set because
SSP1BUF is still full.

Cleared by software

First byte of data available
in SSP1BUF

Cleared by software

SSP1BUF is read

P

Bus Master sends
Stop condition

Received Data

ACK

SSP1IF set on 9th
falling edge of SCL

Cleared by software

Received Data ACK

From Slave to Master
FIGURE 25-15: I²C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 0, DHEN = 0)
FIGURE 25-16: I₂C SLAVE, 7-BIT ADDRESS, RECEIPTION (SEN = 0, AHEN = 1, DHEN = 1)
FIGURE 25-17: I2C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 1, DHEN = 1)

- Master sends Stop condition
- Master releases SDA to slave for ACK sequence
- ACK
- Slave software clears ACKDT to ACK
- ACKTIM is cleared by software
- SSP1BUF can be read any time before next byte is loaded
- Release SCL on 9th rising edge of SCL
- Received address is loaded into SSP1BUF
- Received data available on SSP1BUF
- Slave sends not ACK
- CKP is not cleared if not ACK
- Master sends Stop condition
- No interrupt after Stop
- ACK
- ACKTIM is cleared by hardware
- SCL p edge
- Stop condition
- SCL p edge
25.5.3 SLAVE TRANSMISSION

When the R/W bit of the incoming address byte is set and an address match occurs, the R/W bit of the SSP1STAT register is set. The received address is loaded into the SSP1BUF register, and an ACK pulse is sent by the slave on the ninth bit.

Following the ACK, slave hardware clears the CKP bit and the SCL pin is held low (see Section 25.5.6 "Clock Stretching" for more detail). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data.

The transmit data must be loaded into the SSP1BUF register which also loads the SSP1SR register. Then the SCL pin should be released by setting the CKP bit of the SSP1CON1 register. The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time.

The ACK pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. This ACK value is copied to the ACKSTAT bit of the SSP1CON2 register. If ACKSTAT is set (not ACK), then the data transfer is complete. In this case, when the not ACK is latched by the slave, the slave goes Idle and waits for another occurrence of the Start bit. If the SDA line was low (ACK), the next transmit data must be loaded into the SSP1BUF register. Again, the SCL pin must be released by setting bit CKP.

An MSSP1 interrupt is generated for each data transfer byte. The SSP1IF bit must be cleared by software and the SSP1STAT register is used to determine the status of the byte. The SSP1IF bit is set on the falling edge of the ninth clock pulse.

25.5.3.1 Slave Mode Bus Collision

A slave receives a read request and begins shifting data out on the SDA line. If a bus collision is detected and the SBCDE bit of the SSP1CON3 register is set, the BCL1IF bit of the PIRx register is set. Once a bus collision is detected, the slave goes Idle and waits to be addressed again. User software can use the BCL1IF bit to handle a slave bus collision.

25.5.3.2 7-bit Transmission

A master device can transmit a read request to a slave, and then clock data out of the slave. The list below outlines what software for a slave will need to do to accomplish a standard transmission. Figure 25-17 can be used as a reference to this list.

1. Master sends a Start condition on SDA and SCL.
2. S bit of SSP1STAT is set; SSP1IF is set if interrupt-on-Start detect is enabled.
3. Matching address with R/W bit set is received by the Slave setting SSP1IF bit.
4. Slave hardware generates an ACK and sets SSP1IF.
5. SSP1IF bit is cleared by user.
6. Software reads the received address from SSP1BUF, clearing BF.
7. R/W is set so CKP was automatically cleared after the ACK.
8. The slave software loads the transmit data into SSP1BUF.
9. CKP bit is set releasing SCL, allowing the master to clock the data out of the slave.
10. SSP1IF is set after the ACK response from the master is loaded into the ACKSTAT register.
11. SSP1IF bit is cleared.
12. The slave software checks the ACKSTAT bit to see if the master wants to clock out more data.

13. Steps 9-13 are repeated for each transmitted byte.
14. If the master sends a not ACK; the clock is not held, but SSP1IF is still set.
15. The master sends a Restart condition or a Stop.
16. The slave is no longer addressed.

Note 1: If the master ACKs the clock will be stretched.

2: ACKSTAT is the only bit updated on the rising edge of SCL (9th) rather than the falling.
FIGURE 25-18: \( \text{i}^2\text{C SLAVE, 7-BIT ADDRESS, TRANSMISSION (AHEN = 0)} \)

- Master sends Stop condition
- Ack
- Transmitting Data: Automatic
- Receiving Address: Automatic

- Data to transmit is loaded into SSP1BUF
- BF is automatically cleared after 8th falling edge of SCL
- CP is not held for not ACK
- R/W is copied from the received address
- Indicates an address has been received
- Masters not ACK

- SDA
- SCL
- SSP1IF
- BF
- ACKP
- ACKSTAT
- R/W
- D/A
- S
- P

- Received address is read from SSP1BUF
- When R/W is set, this is always after 9th SCL falling edge
- ACK is set by software
- Master not ACK
- R/W is copied from the received address
- ACK is copied by ACKSTAT
- Masters not ACK

- Receiving Address: Automatic
- Transmitting Data: Automatic
- Received address

- R/W = 1

- Master sends Stop condition
25.5.3.3 7-bit Transmission with Address Hold Enabled

Setting the AHEN bit of the SSP1CON3 register enables additional clock stretching and interrupt generation after the 8th falling edge of a received matching address. Once a matching address has been clocked in, CKP is cleared and the SSP1IF interrupt is set.

Figure 25-18 displays a standard waveform of a 7-bit Address Slave Transmission with AHEN enabled.

1. Bus starts idle.
2. Master sends Start condition; the S bit of SSP1STAT is set; SSP1IF is set if interrupt-on-Start detect is enabled.
3. Master sends matching address with R/W bit set. After the 8th falling edge of the SCL line the CKP bit is cleared and SSP1IF interrupt is generated.
4. Slave software clears SSP1IF.
5. Slave software reads ACKTIM bit of SSP1CON3 register, and R/W and D/A of the SSP1STAT register to determine the source of the interrupt.
6. Slave reads the address value from the SSP1BUF register clearing the BF bit.
7. Slave software decides from this information if it wishes to ACK or not ACK and sets the ACKDT bit of the SSP1CON2 register accordingly.
8. Slave sets the CKP bit releasing SCL.
9. Master clocks in the ACK value from the slave.
10. Slave hardware automatically clears the CKP bit and sets SSP1IF after the ACK if the R/W bit is set.
11. Slave software clears SSP1IF.
12. Slave loads value to transmit to the master into SSP1BUF setting the BF bit.

**Note:** SSP1BUF cannot be loaded until after the ACK.

13. Slave sets CKP bit releasing the clock.
14. Master clocks out the data from the slave and sends an ACK value on the 9th SCL pulse.
15. Slave hardware copies the ACK value into the ACKSTAT bit of the SSP1CON2 register.
16. Steps 10-15 are repeated for each byte transmitted to the master from the slave.
17. If the master sends a not ACK the slave releases the bus, allowing the master to send a Stop and end the communication.

**Note:** Master must send a not ACK on the last byte to ensure that the slave releases the SCL line to receive a Stop.
FIGURE 25-19: \(^2\)C SLAVE, 7-BIT ADDRESS, TRANSMISSION (AHEN = 1)

- Master sends Stop condition
- Master releases SDA to start for \(\text{ACK}\) sequence
- Data to transmit is loaded into SSP1BUF
- ACKDT is cleared by software
- Master clears ACKDT to ACK address
- Master's ACK response is copied to SSP1STAT
- CKP is cleared by hardware after receiving matching address.
- When \(\text{R/W} = 1\); CKP is always cleared after ACK.
- ACKTM is cleared on 8th falling edge of SCL.
25.5.4 SLAVE MODE 10-BIT ADDRESS RECEPTION

This section describes a standard sequence of events for the MSSP1 module configured as an I2C Slave in 10-bit Addressing mode.

Figure 25-19 is used as a visual reference for this description.

This is a step by step process of what must be done by slave software to accomplish I2C communication.

1. Bus starts Idle.
2. Master sends Start condition; S bit of SSP1STAT is set; SSP1IF is set if interrupt-on-Start detect is enabled.
3. Master sends matching high address with R/W bit clear; UA bit of the SSP1STAT register is set.
4. Slave sends ACK and SSP1IF is set.
5. Software clears the SSP1IF bit.
6. Software reads received address from SSP1BUF clearing the BF flag.
7. Slave loads low address into SSP1ADD, releasing SCL.
8. Master sends matching low address byte to the Slave; UA bit is set.

9. Slave sends ACK and SSP1IF is set.

**Note:** Updates to the SSP1ADD register are not allowed until after the ACK sequence.

10. Slave clears SSP1IF.
11. Slave reads the received matching address from SSP1BUF clearing BF.
12. Slave loads high address into SSP1ADD.
13. Master clocks a data byte to the slave and clocks out the slaves ACK on the 9th SCL pulse; SSP1IF is set.
14. If SEN bit of SSP1CON2 is set, CKP is cleared by hardware and the clock is stretched.
15. Slave clears SSP1IF.
16. Slave reads the received byte from SSP1BUF clearing BF.
17. If SEN is set the slave sets CKP to release the SCL.
18. Steps 13-17 repeat for each received byte.
19. Master sends Stop to end the transmission.

25.5.5 10-BIT ADDRESSING WITH ADDRESS OR DATA HOLD

Reception using 10-bit addressing with AHEN or DHEN set is the same as with 7-bit modes. The only difference is the need to update the SSP1ADD register using the UA bit. All functionality, specifically when the CKP bit is cleared and SCL line is held low are the same. Figure 25-20 can be used as a reference of a slave in 10-bit addressing with AHEN set.

Figure 25-21 shows a standard waveform for a slave transmitter in 10-bit Addressing mode.

**Note:** Updates to the SSP1ADD register are not allowed until after the ACK sequence.

**Note:** If the low address does not match, SSP1IF and UA are still set so that the slave software can set SSP1ADD back to the high address. BF is not set because there is no match. CKP is unaffected.

10. Slave clears SSP1IF.
11. Slave reads the received matching address from SSP1BUF clearing BF.
12. Slave loads high address into SSP1ADD.
13. Master clocks a data byte to the slave and clocks out the slaves ACK on the 9th SCL pulse; SSP1IF is set.
14. If SEN bit of SSP1CON2 is set, CKP is cleared by hardware and the clock is stretched.
15. Slave clears SSP1IF.
16. Slave reads the received byte from SSP1BUF clearing BF.
17. If SEN is set the slave sets CKP to release the SCL.
18. Steps 13-17 repeat for each received byte.
19. Master sends Stop to end the transmission.
FIGURE 25-20: \(^2\text{C SLAVE, 10-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 0, DHEN = 0)}\)
FIGURE 25-21:  \( \text{I}^{2}\text{C} \text{ SLAVE, 10-BIT ADDRESS, RECEPTION} \) (\( \text{SEN} = 0, \text{AHEN} = 1, \text{DHEN} = 0 \))

- Receive First Address Byte
- Receive Second Address Byte
- Receive Data
- ACK

SSP1BUF is read from
- Received data
- SSP1BUF can be read anytime before the next received byte
- SSP1BUF is cleared by software
- Update to SSP1ADD is not allowed until 9th falling edge of SCL
- Slave software clears ACKDT to ACK
- If when AHEN = 1; on the 8th falling edge of SCL of an address byte, CKP is cleared
- ACKTIM is set by hardware on 8th falling edge of SCL
- ACKTIM is cleared by software on 8th falling edge of SCL
FIGURE 25-22: \( \text{I}^2\text{C SLAVE, 10-BIT ADDRESS, TRANSMISSION (SEN = 0, AHEN = 0, DHEN = 0)} \)
25.5.6 CLOCK STRETCHING
Clock stretching occurs when a device on the bus holds the SCL line low, effectively pausing communication. The slave may stretch the clock to allow more time to handle data or prepare a response for the master device. A master device is not concerned with stretching as anytime it is active on the bus and not transferring data it is stretching. Any stretching done by a slave is invisible to the master software and handled by the hardware that generates SCL.

The CKP bit of the SSP1CON1 register is used to control stretching in software. Any time the CKP bit is cleared, the module will wait for the SCL line to go low and then hold it. Setting CKP will release SCL and allow more communication.

25.5.6.1 Normal Clock Stretching
Following an ACK if the R/W bit of SSP1STAT is set, a read request, the slave hardware will clear CKP. This allows the slave time to update SSP1BUF with data to transfer to the master. If the SEN bit of SSP1CON2 is set, the slave hardware will always stretch the clock after the ACK sequence. Once the slave is ready; CKP is set by software and communication resumes.

Note 1: The BF bit has no effect on if the clock will be stretched or not. This is different than previous versions of the module that would not stretch the clock, clear CKP, if SSP1BUF was read before the 9th falling edge of SCL.

2: Previous versions of the module did not stretch the clock for a transmission if SSP1BUF was loaded before the 9th falling edge of SCL. It is now always cleared for read requests.

25.5.6.2 10-bit Addressing Mode
In 10-bit Addressing mode, when the UA bit is set, the clock is always stretched. This is the only time the SCL is stretched without CKP being cleared. SCL is released immediately after a write to SSP1ADD.

Note: Previous versions of the module did not stretch the clock if the second address byte did not match.

25.5.6.3 Byte NACKing
When the AHEN bit of SSP1CON3 is set; CKP is cleared by hardware after the 8th falling edge of SCL for a received matching address byte. When the DHEN bit of SSP1CON3 is set; CKP is cleared after the 8th falling edge of SCL for received data.

Stretching after the 8th falling edge of SCL allows the slave to look at the received address or data and decide if it wants to ACK the received data.
25.5.7 CLOCK SYNCHRONIZATION AND THE CKP BIT

Any time the CKP bit is cleared, the module will wait for the SCL line to go low and then hold it. However, clearing the CKP bit will not assert the SCL output low until the SCL output is already sampled low. Therefore, the CKP bit will not assert the SCL line until an external I²C master device has already asserted the SCL line. The SCL output will remain low until the CKP bit is set and all other devices on the I²C bus have released SCL. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCL (see Figure 25-22).

FIGURE 25-23: CLOCK SYNCHRONIZATION TIMING

![CLOCK SYNCHRONIZATION TIMING Diagram]
25.5.8 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the \( \text{I}^2\text{C} \) bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master device. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an acknowledge.

The general call address is a reserved address in the \( \text{I}^2\text{C} \) protocol, defined as address 0x00. When the GCEN bit of the SSP1CON2 register is set, the slave module will automatically ACK the reception of this address, regardless of the value stored in SSP1ADD. After the slave clocks in an address of all zeros with the R/W bit clear, an interrupt is generated and slave software can read SSP1BUF and respond. Figure 25-23 shows a general call reception sequence.

**FIGURE 25-23: SLAVE MODE GENERAL CALL ADDRESS SEQUENCE**

In 10-bit Address mode, the UA bit will not be set on the reception of the general call address. The slave will prepare to receive the second byte as data, just as it would in 7-bit mode.

If the AHEN bit of the SSP1CON3 register is set, just as with any other address reception, the slave hardware will stretch the clock after the 8th falling edge of SCL. The slave must then set its ACKDT value and release the clock with communication progressing as it would normally.

25.5.9 SSP1 MASK REGISTER

An SSP1 Mask (SSP1MSK) register (Register 25-5) is available in \( \text{I}^2\text{C} \) Slave mode as a mask for the value held in the SSP1SR register during an address comparison operation. A zero (‘0’) bit in the SSP1MSK register has the effect of making the corresponding bit of the received address a “don’t care.”

This register is reset to all ‘1’s upon any Reset condition and, therefore, has no effect on standard SSP1 operation until written with a mask value.

The SSP1 Mask register is active during:

- 7-bit Address mode: address compare of A<7:1>.
- 10-bit Address mode: address compare of A<7:0> only. The SSP1 mask has no effect during the reception of the first (high) byte of the address.
25.6 I\textsuperscript{2}C MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSP1M bits in the SSP1CON1 register and by setting the SSP1EN bit. In Master mode, the SCL and SDA lines are set as inputs and are manipulated by the MSSP1 hardware.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP1 module is disabled. Control of the I\textsuperscript{2}C bus may be taken when the P bit is set, or the bus is idle.

In Firmware Controlled Master mode, user code conducts all I\textsuperscript{2}C bus operations based on Start and Stop bit condition detection. Start and Stop condition detection is the only active circuitry in this mode. All other communication is done by the user software directly manipulating the SDA and SCL lines.

The following events will cause the SSP1 Interrupt Flag bit, SSP1IF, to be set (SSP1 interrupt, if enabled):

- Start condition detected
- Stop condition detected
- Data transfer byte transmitted/received
- Acknowledge transmitted/received
- Repeated Start generated

**Note 1:** The MSSP1 module, when configured in \textsuperscript{2}C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSP1BUF register to initiate transmission before the Start condition is complete. In this case, the SSP1BUF will not be written to and the WCOL bit will be set, indicating that a write to the SSP1BUF did not occur.

**Note 2:** When in Master mode, Start/Stop detection is masked and an interrupt is generated when the SEN/PEN bit is cleared and the generation is complete.

25.6.1 I\textsuperscript{2}C MASTER MODE OPERATION

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I\textsuperscript{2}C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic ‘0’. Serial data is transmitted 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic ‘1’. Thus, the first byte transmitted is a 7-bit slave address followed by a ‘1’ to indicate the receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

A Baud Rate Generator is used to set the clock frequency output on SCL. See Section 25.7 “Baud Rate Generator” for more detail.
25.6.2 CLOCK ARBITRATION

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, releases the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 25-25).

FIGURE 25-25: BAUD RATE GENERATOR TIMING WITH CLOCK ARBITRATION

25.6.3 WCOL STATUS FLAG

If the user writes the SSP1BUF when a Start, Restart, Stop, Receive or Transmit sequence is in progress, the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur). Any time the WCOL bit is set it indicates that an action on SSP1BUF was attempted while the module was not Idle.

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSP1CON2 is disabled until the Start condition is complete.

25.6.4 I²C MASTER MODE START CONDITION TIMING

To initiate a Start condition, the user sets the Start Enable bit, SEN bit of the SSP1CON2 register. If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (Tpwrt), the SDA pin is driven low. The action of the SDA being driven low while SCL is high is the Start condition and causes the S bit of the SSP1STAT1 register to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSP1ADD<7:0> and resumes its count. When the Baud Rate Generator times out (Tpwrt), the SEN bit of the SSP1CON2 register will be automatically cleared by hardware; the Baud Rate Generator is suspended, leaving the SDA line held low and the Start condition is complete.

Note 1: If at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition, the SCL line is sampled low before the SDA line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCL1IF, is set, the Start condition is aborted and the I²C module is reset into its Idle state.

2: The Philips I²C Specification states that a bus collision cannot occur on a Start.
FIGURE 25-26: FIRST START BIT TIMING

Write to SEN bit occurs here
SDA = 1, SCL = 1
At completion of Start bit, hardware clears SEN bit and sets SSP1IF bit

Set S bit (SSP1STAT<3>)
Write to SSP1BUF occurs here
1st bit
2nd bit

SDA
SCL
S
TBRG
TBRG
TBRG
TBRG
25.6.5  \textit{I}^2\text{C} MASTEr MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit of the SSP1CON2 register is programmed high and the Master state machine is no longer active. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (Tpwrt). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be deasserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded and begins counting. SDA and SCL must be sampled high for one Tpwrt. This action is then followed by assertion of the SDA pin (SDA = 0) for one Tpwrt while SCL is high. SCL is asserted low. Following this, the RSEN bit of the SSP1CON2 register will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit of the SSP1STAT register will be set. The SSP1IF bit will not be set until the Baud Rate Generator has timed out.

\begin{itemize}
  \item \textbf{Note 1:} If RSEN is programmed while any other event is in progress, it will not take effect.
  \item \textbf{2:} A bus collision during the Repeated Start condition occurs if:
    \begin{itemize}
      \item SDA is sampled low when SCL goes from low-to-high.
      \item SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data ‘1’.
    \end{itemize}
\end{itemize}

\begin{figure}[h]
\centering
\includegraphics[width=\textwidth]{repeat-start-condition-waveform}
\caption{Repeat Start Condition Waveform}
\end{figure}
25.6.6  I2C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSP1BUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted. SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high. When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an ACK bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKSTAT bit on the rising edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared. If not, the bit is set. After the ninth clock, the SSP1IF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSP1BUF, leaving SCL low and SDA unchanged (Figure 25-27).

After the write to the SSP1BUF, each bit of the address will be shifted out on the falling edge of SCL until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will release the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT Status bit of the SSP1CON2 register. Following the falling edge of the ninth clock transmission of the address, the SSP1IF bit is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSP1BUF takes place, holding SCL low and allowing SDA to float.

25.6.6.1  BF Status Flag

In Transmit mode, the BF bit of the SSP1STAT register is set when the CPU writes to SSP1BUF and is cleared when all eight bits are shifted out.

25.6.6.2  WCOL Status Flag

If the user writes the SSP1BUF when a transmit is already in progress (i.e., SSP1SR is still shifting out a data byte), the WCOL is set and the contents of the buffer are unchanged (the write does not occur).

WCOL must be cleared by software before the next transmission.

25.6.6.3  ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit of the SSP1CON2 register is cleared when the slave has sent an Acknowledge (ACK = 0) and is set when the slave does not Acknowledge (ACK = 1). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

25.6.6.4  Typical transmit sequence:

1. The user generates a Start condition by setting the SEN bit of the SSP1CON2 register.
2. SSP1IF is set by hardware on completion of the Start.
3. SSP1IF is cleared by software.
4. The MSSP1 module will wait the required start time before any other operation takes place.
5. The user loads the SSP1BUF with the slave address to transmit.
6. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSP1BUF is written to.
7. The MSSP1 module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
8. The MSSP1 module generates an interrupt at the end of the ninth clock cycle by setting the SSP1IF bit.
9. The user loads the SSP1BUF with eight bits of data.
10. Data is shifted out the SDA pin until all 8 bits are transmitted.
11. The MSSP1 module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
12. Steps 8-11 are repeated for all transmitted data bytes.
13. The user generates a Stop or Restart condition by setting the PEN or RSEN bits of the SSP1CON2 register. Interrupt is generated once the Stop/Restart condition is complete.
FIGURE 25-28: I^2C MASTER MODE WAVEFORM (TRANSMISSION, 7 OR 10-BIT ADDRESS)
25.6.7  I2C MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN bit of the SSP1CON2 register.

**Note:** The MSSP1 module must be in an Idle state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting and on each rollover, the state of the SCL pin changes (high-to-low/low-to-high) and data is shifted into the SSP1SR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSP1SR are loaded into the SSP1BUF, the BF flag bit is set, the SSP1IF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP1 is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable, ACKEN bit of the SSP1CON2 register.

25.6.7.1  BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSP1BUF from SSP1SR. It is cleared when the SSP1BUF register is read.

25.6.7.2  SSP1OV Status Flag

In receive operation, the SSP1OV bit is set when eight bits are received into the SSP1SR and the BF flag bit is already set from a previous reception.

25.6.7.3  WCOL Status Flag

If the user writes the SSP1BUF when a receive is already in progress (i.e., SSP1SR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

25.6.7.4  Typical Receive Sequence:

1. The user generates a Start condition by setting the SEN bit of the SSP1CON2 register.
2. SSP1IF is set by hardware on completion of the Start.
3. SSP1IF is cleared by software.
4. User writes SSP1BUF with the slave address to transmit and the R/W bit set.
5. Address is shifted out the SDA pin until all 8 bits are transmitted. Transmission begins as soon as SSP1BUF is written to.
6. The MSSP1 module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSP1CON2 register.
7. The MSSP1 module generates an interrupt at the end of the ninth clock cycle by setting the SSP1IF bit.
8. User sets the RCEN bit of the SSP1CON2 register and the master clocks in a byte from the slave.
9. After the 8th falling edge of SCL, SSP1IF and BF are set.
10. Master clears SSP1IF and reads the received byte from SSP1UF, clears BF.
11. Master sets ACK value sent to slave in ACKDT bit of the SSP1CON2 register and initiates the ACK by setting the ACKEN bit.
12. Masters ACK is clocked out to the slave and SSP1IF is set.
13. User clears SSP1IF.
14. Steps 8-13 are repeated for each received byte from the slave.
15. Master sends a not ACK or Stop to end communication.
FIGURE 25-29:  I²C MASTER MODE WAVEFORM (RECEPTION, 7-BIT ADDRESS)
25.6.8 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit, ACKEN bit of the SSP1CON2 register. When this bit is set, the SCL pin is pulled low and the contents of the Acknowledge data bit are presented on the SDA pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCL pin is deasserted (pulled high). When the SCL pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG. The SCL pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP1 module then goes into Idle mode (Figure 25-29).

25.6.8.1 WCOL Status Flag

If the user writes the SSP1BUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write does not occur).

25.6.9 STOP CONDITION TIMING

A Stop bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN bit of the SSP1CON2 register. At the end of a receive/transmit, the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the Baud Rate Generator is reloaded and counts down to 0. When the Baud Rate Generator times out, the SCL pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDA pin will be deasserted. When the SDA pin is sampled high while SCL is high, the P bit of the SSP1STAT register is set. A TBRG later, the PEN bit is cleared and the SSP1IF bit is set (Figure 25-30).

25.6.9.1 WCOL Status Flag

If the user writes the SSP1BUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

FIGURE 25-30: ACKNOWLEDGE SEQUENCE WAVEFORM

![Acknowledge Sequence Waveform](image)

**Note:** TBRG = one Baud Rate Generator period.
25.6.10 SLEEP OPERATION
While in Sleep mode, the \( \text{I}^2\text{C} \) slave module can receive addresses or data and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP1 interrupt is enabled).

25.6.11 EFFECTS OF A RESET
A Reset disables the MSSP1 module and terminates the current transfer.

25.6.12 MULTI-MASTER MODE
In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP1 module is disabled. Control of the \( \text{I}^2\text{C} \) bus may be taken when the P bit of the SSP1STAT register is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the SSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDA line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed by hardware with the result placed in the BCL1IF bit.

The states where arbitration can be lost are:
- Address Transfer
- Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

25.6.13 MULTI-MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION
Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a ‘1’ on SDA, by letting SDA float high and another master asserts a ‘0’. When the SCL pin floats high, data should be stable. If the expected data on SDA is a ‘1’ and the data sampled on the SDA pin is ‘0’, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCL1IF, and reset the \( \text{I}^2\text{C} \) port to its Idle state (Figure 25-31).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDA and SCL lines are deasserted and the SSP1BUF can be written to. When the user services the bus collision Interrupt Service Routine and if the \( \text{I}^2\text{C} \) bus is free, the user can resume communication by asserting a Start condition.

If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are deasserted and the respective control bits in the SSP1CON2 register are cleared. When the user services the bus collision Interrupt Service Routine and if the \( \text{I}^2\text{C} \) bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDA and SCL pins. If a Stop condition occurs, the SSP1IF bit will be set.

A write to the SSP1BUF will start the transmission of data at the first data bit, regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the \( \text{I}^2\text{C} \) bus can be taken when the P bit is set in the SSP1STAT register, or the bus is Idle and the S and P bits are cleared.
FIGURE 25-32: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE

- Data changes while SCL = 0
- SDA line pulled low by another source
- SDA released by master
- Sample SDA. While SCL is high, data does not match what is driven by the master. Bus collision has occurred.
- Set bus collision interrupt (BCL1IF)
25.6.13.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 25-32).
b) SCL is sampled low before SDA is asserted low (Figure 25-33).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- the Start condition is aborted,
- the BCL1IF flag is set and
- the MSSP1 module is reset to its Idle state (Figure 25-32).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded and counts down. If the SCL pin is sampled low while SDA is high, a bus collision occurs because it is assumed that another master is attempting to drive a data ‘1’ during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 25-34). If, however, a ‘1’ is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to zero; if the SCL pin is sampled as ‘0’ during this time, a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

**Note:** The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.
FIGURE 25-34: BUS COLLISION DURING START CONDITION (SCL = 0)

SDA = 0, SCL = 1

SDA

SCL

SEN

BCL1IF

S

SSP1IF

SDA pulled low by other master. Reset BRG and assert SDA.

S

SCL

SEN

BCL1IF

S

SSP1IF

FIGURE 25-35: BRG RESET DUE TO SDA ARBITRATION DURING START CONDITION

SDA = 0, SCL = 1

Less than TBRG

Set S

Set S

Set S

Set S

Set SEN, enable Start sequence if SDA = 1, SCL = 1

S

SCL pulled low after BRG time-out

Set SEN, enable Start sequence if SDA = 1, SCL = 1

S

SDA = 0, SCL = 1, set SSP1IF

Interrupts cleared by software

Interrupts cleared by software
25.6.13.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

a) A low level is sampled on SDA when SCL goes from low level to high level.

b) SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data ’1’.

When the user releases SDA and the pin is allowed to float high, the BRG is loaded with SSP1ADD and counts down to zero. The SCL pin is then deasserted and when sampled high, the SDA pin is sampled.

If SDA is low, a bus collision has occurred (i.e., another master is attempting to transmit a data ’0’, Figure 25-35). If SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high-to-low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data ’1’ during the Repeated Start condition, see Figure 25-36.

If, at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.

**FIGURE 25-36: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)**

<table>
<thead>
<tr>
<th>SDA</th>
<th>SCL</th>
<th>RSEN</th>
<th>BCL1IF</th>
<th>S</th>
<th>SSP1IF</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Sample SDA when SCL goes high.
If SDA = 0, set BCL1IF and release SDA and SCL.

Cleared by software

**FIGURE 25-37: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)**

<table>
<thead>
<tr>
<th>SDA</th>
<th>SCL</th>
<th>BCL1IF</th>
<th>RSEN</th>
<th>S</th>
<th>SSP1IF</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

SCL goes low before SDA, set BCL1IF. Release SDA and SCL.
Interrupt cleared by software

S = ’0’
25.6.13.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

a) After the SDA pin has been deasserted and allowed to float high, SDA is sampled low after the BRG has timed out.

b) After the SCL pin is deasserted, SCL is sampled low before SDA goes high.

The Stop condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSP1ADD and counts down to ‘0’. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data ‘0’ (Figure 25-37). If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data ‘0’ (Figure 25-38).

**FIGURE 25-38: BUS COLLISION DURING A STOP CONDITION (CASE 1)**

<table>
<thead>
<tr>
<th>TBRG</th>
<th>TBRG</th>
<th>TBRG</th>
</tr>
</thead>
<tbody>
<tr>
<td>SDA</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SCL</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>PEN</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BCL1IF</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>P</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SSP1IF</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**FIGURE 25-39: BUS COLLISION DURING A STOP CONDITION (CASE 2)**

<table>
<thead>
<tr>
<th>TBRG</th>
<th>TBRG</th>
<th>TBRG</th>
</tr>
</thead>
<tbody>
<tr>
<td>SDA</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SCL</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>PEN</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BCL1IF</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>P</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SSP1IF</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>
# TABLE 25-3: SUMMARY OF REGISTERS ASSOCIATED WITH I²C™ OPERATION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Reset Values on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIE2</td>
<td>OSFIE</td>
<td>—</td>
<td>C1IE</td>
<td>EEIE</td>
<td>BCL1IE</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>79</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>PIR2</td>
<td>OSFIF</td>
<td>—</td>
<td>C1IF</td>
<td>EEIF</td>
<td>BCL1IF</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>81</td>
</tr>
<tr>
<td>SSP1ADD</td>
<td>ADD7</td>
<td>ADD6</td>
<td>ADD5</td>
<td>ADD4</td>
<td>ADD3</td>
<td>ADD2</td>
<td>ADD1</td>
<td>ADD0</td>
<td>249</td>
</tr>
<tr>
<td>SSP1BUF</td>
<td>Synchronous Serial Port Receive Buffer/Transmit Register</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>199*</td>
</tr>
<tr>
<td>SSP1CON1</td>
<td>WCOL</td>
<td>SSP1OV</td>
<td>SSP1EN</td>
<td>CKP</td>
<td>SSP1M&lt;3:0&gt;</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>246</td>
</tr>
<tr>
<td>SSP1CON2</td>
<td>GCEN</td>
<td>ACKSTAT</td>
<td>ACKDT</td>
<td>ACKEN</td>
<td>RCEN</td>
<td>PEN</td>
<td>RSEN</td>
<td>SEN</td>
<td>247</td>
</tr>
<tr>
<td>SSP1CON3</td>
<td>ACKTIM</td>
<td>PCIE</td>
<td>SCIE</td>
<td>BOEN</td>
<td>SDAHT</td>
<td>SBCDE</td>
<td>AHEN</td>
<td>DHEN</td>
<td>248</td>
</tr>
<tr>
<td>SSP1MSK</td>
<td>MSK7</td>
<td>MSK6</td>
<td>MSK5</td>
<td>MSK4</td>
<td>MSK3</td>
<td>MSK2</td>
<td>MSK1</td>
<td>MSK0</td>
<td>249</td>
</tr>
<tr>
<td>SSP1STAT</td>
<td>SMP</td>
<td>CKE</td>
<td>D/A</td>
<td>P</td>
<td>S</td>
<td>R/W</td>
<td>UA</td>
<td>BF</td>
<td>245</td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td>106</td>
</tr>
</tbody>
</table>

**Legend:**
- — = unimplemented location, read as ‘0’.
- Shaded cells are not used by the MSSP module in I²C™ mode.
- * Page provides register information.
25.7 BAUD RATE GENERATOR

The MSSP1 module has a Baud Rate Generator available for clock generation in both I²C and SPI Master modes. The Baud Rate Generator (BRG) reload value is placed in the SSP1ADD register (Register 25-6). When a write occurs to SSP1BUF, the Baud Rate Generator will automatically begin counting down.

Once the given operation is complete, the internal clock will automatically stop counting and the clock pin will remain in its last state.

An internal signal “Reload” in Figure 25-39 triggers the value from SSP1ADD to be loaded into the BRG counter. This occurs twice for each oscillation of the module clock line. The logic dictating when the reload signal is asserted depends on the mode the MSSP1 is being operated in. Table 25-4 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSP1ADD.

**EQUATION 25-1:**

\[
F_{\text{clock}} = \frac{F_{\text{osc}}}{(SSP_{xADD} + 1)(4)}
\]

**FIGURE 25-40: BAUD RATE GENERATOR BLOCK DIAGRAM**

**TABLE 25-4: MSSP1 CLOCK RATE W/BRG**

<table>
<thead>
<tr>
<th>Fosc</th>
<th>FCy</th>
<th>BRG Value</th>
<th>FClock (2 Rollovers of BRG)</th>
</tr>
</thead>
<tbody>
<tr>
<td>32 MHz</td>
<td>8 MHz</td>
<td>13h</td>
<td>400 kHz(^{(1)})</td>
</tr>
<tr>
<td>32 MHz</td>
<td>8 MHz</td>
<td>19h</td>
<td>308 kHz</td>
</tr>
<tr>
<td>32 MHz</td>
<td>8 MHz</td>
<td>4Fh</td>
<td>100 kHz</td>
</tr>
<tr>
<td>16 MHz</td>
<td>4 MHz</td>
<td>09h</td>
<td>400 kHz(^{(1)})</td>
</tr>
<tr>
<td>16 MHz</td>
<td>4 MHz</td>
<td>0Ch</td>
<td>308 kHz</td>
</tr>
<tr>
<td>16 MHz</td>
<td>4 MHz</td>
<td>27h</td>
<td>100 kHz</td>
</tr>
<tr>
<td>4 MHz</td>
<td>1 MHz</td>
<td>09h</td>
<td>100 kHz</td>
</tr>
</tbody>
</table>

**Note:** Values of 0x00, 0x01 and 0x02 are not valid for SSP1ADD when used as a Baud Rate Generator for I²C. This is an implementation limitation.

**Note 1:** Refer to the I/O port electrical and timing specifications in Table 31-18 and Figure 31-20 to ensure the system is designed to support the I/O timing requirements.
### REGISTER 25-1: **SSP1STAT: SSP1 STATUS REGISTER**

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>SMP</td>
<td>CKE</td>
<td>D/A</td>
<td>P</td>
<td>S</td>
<td>R/W</td>
<td>UA</td>
<td>BF</td>
</tr>
</tbody>
</table>

#### bit 7    SMP: SPI Data Input Sample bit

**SPI Master mode:**
- 1 = Input data sampled at end of data output time
- 0 = Input data sampled at middle of data output time

**SPI Slave mode:**
- SMP must be cleared when SPI is used in Slave mode

**I^2^C Master or Slave mode:**
- 1 = Slew rate control disabled for standard speed mode (100 kHz and 1 MHz)
- 0 = Slew rate control enabled for high speed mode (400 kHz)

#### bit 6    CKE: SPI Clock Edge Select bit (SPI mode only)

**In SPI Master or Slave mode:**
- 1 = Transmit occurs on transition from active to Idle clock state
- 0 = Transmit occurs on transition from Idle to active clock state

**In I^2^C Master or Slave mode:**
- 1 = Enable input logic so that thresholds are compliant with SMBus specification
- 0 = Disable SMBus specific inputs

#### bit 5    D/A: Data/Address bit (I^2^C mode only)

- 1 = Indicates that the last byte received or transmitted was data
- 0 = Indicates that the last byte received or transmitted was address

#### bit 4    P: Stop bit

(I^2^C mode only. This bit is cleared when the MSSP1 module is disabled, SSP1EN is cleared.)
- 1 = Indicates that a Stop bit has been detected last (this bit is ‘0’ on Reset)
- 0 = Stop bit was not detected last

#### bit 3    S: Start bit

(I^2^C mode only. This bit is cleared when the MSSP1 module is disabled, SSP1EN is cleared.)
- 1 = Indicates that a Start bit has been detected last (this bit is ‘0’ on Reset)
- 0 = Start bit was not detected last

#### bit 2    R/W: Read/Write bit information (I^2^C mode only)

This bit holds the R/W bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit, or not ACK bit.

**In I^2^C Slave mode:**
- 1 = Read
- 0 = Write

**In I^2^C Master mode:**
- 1 = Transmit is in progress
- 0 = Transmit is not in progress

OR-ing this bit with SEN, RSEN, PEN, RCEN or ACKEN will indicate if the MSSP1 is in Idle mode.

#### bit 1    UA: Update Address bit (10-bit I^2^C mode only)

- 1 = Indicates that the user needs to update the address in the SSP1ADD register
- 0 = Address does not need to be updated

#### bit 0    BF: Buffer Full Status bit

**Receive (SPI and I^2^C modes):**
- 1 = Receive complete, SSP1BUF is full
- 0 = Receive not complete, SSP1BUF is empty

**Transmit (I^2^C mode only):**
- 1 = Data transmit in progress (does not include the ACK and Stop bits), SSP1BUF is full
- 0 = Data transmit complete (does not include the ACK and Stop bits), SSP1BUF is empty

---

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

---

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REGISTER 25-2: SSP1CON1: SSP1 CONTROL REGISTER 1

<table>
<thead>
<tr>
<th>R/C/HS-0/0</th>
<th>R/C/HS-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>bit 6</td>
<td>bit 5</td>
<td>bit 4</td>
<td>bit 3</td>
<td>bit 2</td>
<td>bit 1</td>
</tr>
</tbody>
</table>

**Legend:**

- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- **'1'** = Bit is set
- **'0'** = Bit is cleared
- **HS** = Bit is set by hardware
- **C** = User cleared

**Legend:**

- **WCOL**: Write Collision Detect bit
  - **Master mode**: A write to the SSP1BUF register was attempted while the I2C conditions were not valid for a transmission to be started
  - **Slave mode**: The SSP1BUF register is written while it is still transmitting the previous word (must be cleared in software)
  - **0** = No collision
  - **1** = A write to the SSP1BUF register was attempted while the I2C conditions were not valid for a transmission to be started
  - **0** = No collision

**Legend:**

- **SSP1OV**: Receive Overflow Indicator bit
  - **In SPI mode**: A new byte is received while the SSP1BUF register is still holding the previous data. In case of overflow, the data in SSP1SR is lost. Overflow can only occur in Slave mode. In Slave mode, the user must read the SSP1BUF, even if only transmitting data, to avoid setting overflow. In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSP1BUF register (must be cleared in software).
  - **0** = No overflow
  - **1** = A new byte is received while the SSP1BUF register is still holding the previous byte. SSP1OV is a “don’t care” in Transmit mode (must be cleared in software).

**Legend:**

- **SSP1EN**: Synchronous Serial Port Enable bit
  - **In SPI mode**: Enables serial port and configures SCK, SDO, SDI and SS as the source of the serial port pins
  - **0** = Disables serial port and configures these pins as I/O port pins

**Legend:**

- **CKP**: Clock Polarity Select bit
  - **In SPI mode**: Idle state for clock is a high level
  - **0** = Idle state for clock is a low level
  - **SCL release control**
  - **1** = Enable clock
  - **0** = Holds clock low (clock stretch). (Used to ensure data setup time.)

**Legend:**

- **SSP1M<3:0>**: Synchronous Serial Port Mode Select bits
  - **Unspecified**
  - **In I2C mode**: Enables the serial port and configures SDA and SCL pins as the source of the serial port pins
  - **0** = Enables serial port and configures these pins as I/O port pins
  - **1** = A byte is received while the SSP1BUF register is still holding the previous byte. SSP1OV is a “don’t care” in Transmit mode (must be cleared in software).

**Legend:**

- **Note**: In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSP1BUF register.
- **2**: When enabled, these pins must be properly configured as input or output.
- **3**: When enabled, the SDA and SCL pins must be configured as inputs.
- **4**: SSP1ADD values of 0, 1 or 2 are not supported for I2C mode.
- **5**: SSP1ADD value of ‘1’ is not supported. Use SSP1M = 0000 instead.
REGISTER 25-3:  : SSP1 CONTROL REGISTER 2

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>R/W-0/0</th>
<th>R/S-0/0</th>
<th>R/S-0/0</th>
<th>R/S-0/0</th>
<th>R/S-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GCEN</td>
<td>ACKSTAT</td>
<td>ACKDT</td>
<td>ACKEN</td>
<td>RCEN</td>
<td>PEN</td>
<td>RSEN</td>
<td>SEN</td>
</tr>
</tbody>
</table>

Legend:
- **R** = Readable bit
- **W** = Writable bit
- **U** = Unimplemented bit, read as ‘0’
- **u** = Bit is unchanged
- **x** = Bit is unknown
- **-n/n** = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared
- **HC** = Cleared by hardware
- **S** = User set

**bit 7**  
**GCEN:** General Call Enable bit (in **I²C** Slave mode only)
- **1** = Enable interrupt when a general call address (0x00 or 00h) is received in the SSP1SR
- **0** = General call address disabled

**bit 6**  
**ACKSTAT:** Acknowledge Status bit (in **I²C** mode only)
- **1** = Acknowledge was not received
- **0** = Acknowledge was received

**bit 5**  
**ACKDT:** Acknowledge Data bit (in **I²C** mode only)

  - In **Receive** mode:
    - Value transmitted when the user initiates an Acknowledge sequence at the end of a receive
    - **1** = Not Acknowledge
    - **0** = Acknowledge

**bit 4**  
**ACKEN:** Acknowledge Sequence Enable bit (in **I²C** Master mode only)

  - In **Master Receive** mode:
    - **1** = Initiate Acknowledge sequence on SDA and SCL pins, and transmit ACKDT data bit. Automatically cleared by hardware.
    - **0** = Acknowledge sequence Idle

**bit 3**  
**RCEN:** Receive Enable bit (in **I²C** Master mode only)
- **1** = Enables Receive mode for **I²C**
- **0** = Receive Idle

**bit 2**  
**PEN:** Stop Condition Enable bit (in **I²C** Master mode only)

  - **SCK Release Control**:
    - **1** = Initiate Stop condition on SDA and SCL pins. Automatically cleared by hardware.
    - **0** = Stop condition Idle

**bit 1**  
**RSEN:** Repeated Start Condition Enabled bit (in **I²C** Master mode only)
- **1** = Initiate Repeated Start condition on SDA and SCL pins. Automatically cleared by hardware.
- **0** = Repeated Start condition Idle

**bit 0**  
**SEN:** Start Condition Enabled bit (in **I²C** Master mode only)

  - **In Master mode**:
    - **1** = Initiate Start condition on SDA and SCL pins. Automatically cleared by hardware.
    - **0** = Start condition Idle

  - **In Slave mode**:
    - **1** = Clock stretching is enabled for both slave transmit and slave receive (stretch enabled)
    - **0** = Clock stretching is disabled

**Note 1:** For bits ACKEN, RCEN, PEN, RSEN, SEN: If the **I²C** module is not in the Idle mode, this bit may not be set (no spooling) and the SSP1BUF may not be written (or writes to the SSP1BUF are disabled).
## REGISTER 25-4:  SSP1CON3: SSP1 CONTROL REGISTER 3

<table>
<thead>
<tr>
<th>Bit 7</th>
<th>Bit 0</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>ACKTIM</td>
<td>PCIE</td>
<td>SCIE</td>
<td>BOEN</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as '0'
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- '1' = Bit is set
- '0' = Bit is cleared

**bit 7**  
**ACKTIM:** Acknowledge Time Status bit (I²C mode only)  
1 = Indicates the I²C bus is in an Acknowledge sequence, set on 8TH falling edge of SCL clock  
0 = Not an Acknowledge sequence, cleared on 9TH rising edge of SCL clock

**bit 6**  
**PCIE:** Stop Condition Interrupt Enable bit (I²C mode only)  
1 = Enable interrupt on detection of Stop condition  
0 = Stop detection interrupts are disabled

**bit 5**  
**SCIE:** Start Condition Interrupt Enable bit (I²C mode only)  
1 = Enable interrupt on detection of Start or Restart conditions  
0 = Start detection interrupts are disabled

**bit 4**  
**BOEN:** Buffer Overwrite Enable bit  
In SPI Slave mode:
1 = SSP1BUF updates every time that a new data byte is shifted in ignoring the BF bit  
0 = If new byte is received with BF bit of the SSP1STAT register already set, SSP1OV bit of the SSP1CON1 register is set, and the buffer is not updated  
In I²C Master mode and SPI Master mode:
This bit is ignored.  
In I²C Slave mode:
1 = SSP1BUF is updated and ACK is generated for a received address/data byte, ignoring the state of the SSP1OV bit only if the BF bit = 0.  
0 = SSP1BUF is only updated when SSP1OV is clear

**bit 3**  
**SDAHT:** SDA Hold Time Selection bit (I²C mode only)  
1 = Minimum of 300 ns hold time on SDA after the falling edge of SCL  
0 = Minimum of 100 ns hold time on SDA after the falling edge of SCL

**bit 2**  
**SBCDE:** Slave Mode Bus Collision Detect Enable bit (I²C Slave mode only)  
If on the rising edge of SCL, SDA is sampled low when the module is outputting a high state, the BCL1IF bit of the PIR2 register is set, and bus goes Idle  
1 = Enable slave bus collision interrupts  
0 = Slave bus collision interrupts are disabled

**bit 1**  
**AHEN:** Address Hold Enable bit (I²C Slave mode only)  
1 = Following the 8th falling edge of SCL for a matching received address byte; CKP bit of the SSP1CON1 register will be cleared and the SCL will be held low.  
0 = Address holding is disabled

**bit 0**  
**DHEN:** Data Hold Enable bit (I²C Slave mode only)  
1 = Following the 8th falling edge of SCL for a received data byte; slave hardware clears the CKP bit of the SSP1CON1 register and SCL is held low.  
0 = Data holding is disabled

**Note 1:** For daisy-chained SPI operation; allows the user to ignore all but the last received byte. SSP1OV is still set when a new byte is received and BF = 1, but hardware continues to write the most recent byte to SSP1BUF.

**Note 2:** This bit has no effect in Slave modes that Start and Stop condition detection is explicitly listed as enabled.

**Note 3:** The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is set.
# REGISTER 25-5: SSP1MSK: SSP1 MASK REGISTER

<table>
<thead>
<tr>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
<th>R/W-1/1</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 7-0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R**: Readable bit
- **W**: Writable bit
- **U**: Unimplemented bit, read as ‘0’
- **u**: Bit is unchanged
- **x**: Bit is unknown
- **-n/n**: Value at POR and BOR/Value at all other Resets
- **’1’**: Bit is set
- **’0’**: Bit is cleared

**bit 7-1**

- **MSK<7:1>**: Mask bits
  - 1: The received address bit n is compared to SSP1ADD<n> to detect I²C address match
  - 0: The received address bit n is not used to detect I²C address match

**bit 0**

- **MSK<0>**: Mask bit for I²C Slave mode, 10-bit Address
  - I²C Slave mode, 10-bit address (SSP1M<3:0> = 0111 or 1111):
    - 1: The received address bit 0 is compared to SSP1ADD<0> to detect I²C address match
    - 0: The received address bit 0 is not used to detect I²C address match
  - I²C Slave mode, 7-bit address, the bit is ignored

# REGISTER 25-6: SSP1ADD: MSSP1 ADDRESS AND BAUD RATE REGISTER (I²C MODE)

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>bit 7-0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Legend:**

- **R**: Readable bit
- **W**: Writable bit
- **U**: Unimplemented bit, read as ‘0’
- **u**: Bit is unchanged
- **x**: Bit is unknown
- **-n/n**: Value at POR and BOR/Value at all other Resets
- **’1’**: Bit is set
- **’0’**: Bit is cleared

**Master mode:**

**bit 7-0**

- **ADD<7:0>**: Baud Rate Clock Divider bits
  - SCL pin clock period = ((ADD<7:0> + 1) * 4)/FOSC

**10-Bit Slave mode — Most Significant Address Byte:**

**bit 7-3**

- **Not used**: Unused for Most Significant Address Byte. Bit state of this register is a "don't care". Bit pattern sent by master is fixed by I²C specification and must be equal to '11110'. However, those bits are compared by hardware and are not affected by the value in this register.

**bit 2-1**

- **ADD<2:1>**: Two Most Significant bits of 10-bit address

**bit 0**

- **Not used**: Unused in this mode. Bit state is a "don't care."

**10-Bit Slave mode — Least Significant Address Byte:**

**bit 7-0**

- **ADD<7:0>**: Eight Least Significant bits of 10-bit address

**7-Bit Slave mode:**

**bit 7-1**

- **ADD<7:1>**: 7-bit address

**bit 0**

- **Not used**: Unused in this mode. Bit state is a "don't care."
26.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer independent of device program execution. The EUSART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or half-duplex synchronous system. Full-Duplex mode is useful for communications with peripheral systems, such as CRT terminals and personal computers. Half-Duplex Synchronous mode is intended for communications with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs or other microcontrollers. These devices typically do not have internal clocks for baud rate generation and require the external clock signal provided by a master synchronous device.

The EUSART module includes the following capabilities:
- Full-duplex asynchronous transmit and receive
- Two-character input buffer
- One-character output buffer
- Programmable 8-bit or 9-bit character length
- Address detection in 9-bit mode
- Input buffer overrun error detection
- Received character framing error detection
- Half-duplex synchronous master
- Half-duplex synchronous slave
- Programmable clock polarity in synchronous modes
- Sleep operation

The EUSART module implements the following additional features, making it ideally suited for use in Local Interconnect Network (LIN) bus systems:
- Automatic detection and calibration of the baud rate
- Wake-up on Break reception
- 13-bit Break character transmit

Block diagrams of the EUSART transmitter and receiver are shown in Figure 26-1 and Figure 26-2.
The operation of the EUSART module is controlled through three registers:

- Transmit Status and Control (TXSTA)
- Receive Status and Control (RCSTA)
- Baud Rate Control (BAUDCON)

These registers are detailed in Register 26-1, Register 26-2 and Register 26-3, respectively.

When the receiver or transmitter section is not enabled then the corresponding RX or TX pin may be used for general purpose input and output.
26.1 EUSART Asynchronous Mode

The EUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH mark state which represents a ‘1’ data bit, and a VOL space state which represents a ‘0’ data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the Mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is 8 bits. Each transmitted bit persists for a period of \( \frac{1}{\text{Baud Rate}} \). An on-chip dedicated 8-bit/16-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See Table 26-5 for examples of baud rate configurations.

The EUSART transmits and receives the LSB first. The EUSART’s transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

26.1.1 EUSART Asynchronous Transmitter

The EUSART transmitter block diagram is shown in Figure 26-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TXREG register.

26.1.1.1 Enabling the Transmitter

The EUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

- TXEN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TXSTA register enables the transmitter circuitry of the EUSART. Clearing the SYNC bit of the TXSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the EUSART and automatically configures the TX/CK I/O pin as an output. If the TX/CK pin is shared with an analog peripheral, the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

**Note 1:** The TXIF Transmitter Interrupt flag is set when the TXEN enable bit is set.
26.1.1.4 TSR Status

The TRMT bit of the TXSTA register indicates the status of the TSR register. This is a read-only bit. The TRMT bit is set when the TSR register is empty and is cleared when a character is transferred to the TSR register from the TXREG. The TRMT bit remains clear until all bits have been shifted out of the TSR register. No interrupt logic is tied to this bit, so the user has to poll this bit to determine the TSR status.

Note: The TSR register is not mapped in data memory, so it is not available to the user.

26.1.1.5 Transmitting 9-Bit Characters

The EUSART supports 9-bit character transmissions. When the TX9 bit of the TXSTA register is set, the EUSART will shift nine bits out for each character transmitted. The TX9D bit of the TXSTA register is the ninth, and Most Significant data bit. When transmitting 9-bit data, the TX9D data bit must be written before writing the eight Least Significant data bits into the TXREG. All nine bits of data will be transferred to the TSR shift register immediately after the TXREG is written.

A special 9-bit Address mode is available for use with multiple receivers. See Section 26.1.2.7 “Address Detection” for more information on the address mode.

26.1.1.6 Asynchronous Transmission Set-up:

1. Initialize the SPBRGH, SPBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 26.3 “EUSART Baud Rate Generator (BRG)”).
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If 9-bit transmission is desired, set the TX9 control bit. A set ninth data bit will indicate that the eight Least Significant data bits are an address when the receiver is set for address detection.
4. Enable the transmission by setting the TXEN control bit. This will cause the TXIF interrupt bit to be set.
5. If interrupts are desired, set the TXIE interrupt enable bit of the PIE1 register. An interrupt will occur immediately provided that the GIE and PEIE bits of the INTCON register are also set.
6. If 9-bit transmission is selected, the ninth bit should be loaded into the TX9D data bit.
7. Load 8-bit data into the TXREG register. This will start the transmission.

FIGURE 26-3: ASYNCHRONOUS TRANSMISSION
FIGURE 26-4: ASYNCHRONOUS TRANSMISSION (BACK-TO-BACK)

TABLE 26-1: SUMMARY OF REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUDCON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>ABDEN</td>
<td>261</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>RCSTA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>260</td>
</tr>
<tr>
<td>SPBRGL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SPBRGH</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TXREG</td>
<td>EUSART Transmit Data Register</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TXSTA</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>259</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used for asynchronous transmission.

* Page provides register information.
26.1.2 EUSART ASYNCHRONOUS RECEIVER

The Asynchronous mode is typically used in RS-232 systems. The receiver block diagram is shown in Figure 26-2. The data is received on the RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at 16 times the baud rate, whereas the serial Receive Shift Register (RSR) operates at the bit rate. When all eight or nine bits of the character have been shifted in, they are immediately transferred to a two character First-In-First-Out (FIFO) memory. The FIFO buffering allows reception of two complete characters and the start of a third character before software must start servicing the EUSART receiver. The FIFO and RSR registers are not directly accessible by software. Access to the received data is via the RCREG register.

26.1.2.1 Enabling the Receiver

The EUSART receiver is enabled for asynchronous operation by configuring the following three control bits:

- CREN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the CREN bit of the RCSTA register enables the receiver circuitry of the EUSART. Clearing the SYNC bit of the TXSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the EUSART. The programmer must set the corresponding TRIS bit to configure the RX/DT I/O pin as an input.

**Note 1:** If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.

26.1.2.2 Receiving Data

The receiver data recovery circuit initiates character reception on the falling edge of the first bit. The first bit, also known as the Start bit, is always a zero. The data recovery circuit counts one-half bit time to the center of the Start bit and verifies that the bit is still a zero. If it is not a zero then the data recovery circuit aborts character reception, without generating an error, and resumes looking for the falling edge of the Start bit. If the Start bit zero verification succeeds then the data recovery circuit counts a full bit time to the center of the next bit. The bit is then sampled by a majority detect circuit and the resulting ‘0’ or ‘1’ is shifted into the RSR. This repeats until all data bits have been sampled and shifted into the RSR. One final bit time is measured and the level sampled. This is the Stop bit, which is always a ‘1’. If the data recovery circuit samples a ‘0’ in the Stop bit position then a framing error is set for this character, otherwise the framing error is cleared for this character. See **Section 26.1.2.4 “Receive Framing Error”** for more information on framing errors.

Immediately after all data bits and the Stop bit have been received, the character in the RSR is transferred to the EUSART receive FIFO and the RCIF interrupt flag bit of the PIR1 register is set. The top character in the FIFO is transferred out of the FIFO by reading the RCREG register.

**Note:** If the receive FIFO is overrun, no additional characters will be received until the overrun condition is cleared. See **Section 26.1.2.5 “Receive Overrun Error”** for more information on overrun errors.

26.1.2.3 Receive Interrupts

The RCIF interrupt flag bit of the PIR1 register is set whenever the EUSART receiver is enabled and there is an unread character in the receive FIFO. The RCIF interrupt flag bit is read-only, it cannot be set or cleared by software.

RCIF interrupts are enabled by setting all of the following bits:

- RCIE interrupt enable bit of the PIE1 register
- Peripheral Interrupt Enable (PEIE) bit of the INTCON register
- Global Interrupt Enable (GIE) bit of the INTCON register

The RCIF interrupt flag bit will be set when there is an unread character in the FIFO, regardless of the state of interrupt enable bits.
26.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCSTA register which resets the EUSART. Clearing the CREN bit of the RCSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

**Note:** If all receive characters in the receive FIFO have framing errors, repeated reads of the RCREG will not clear the FERR bit.

26.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCSTA register or by resetting the EUSART by clearing the SPEN bit of the RCSTA register.

26.1.2.6 Receiving 9-bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCREG.

26.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.
26.1.2.8 Asynchronous Reception Set-up:

1. Initialize the SPBRGH, SPBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 26.3 “EUSART Baud Rate Generator (BRG)”).
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
4. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
5. If 9-bit reception is desired, set the RX9 bit.
6. Enable reception by setting the CREN bit.
7. The RCIF interrupt flag bit will be set when a character is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
8. Read the RCSTA register to get the error flags and, if 9-bit data reception is enabled, the ninth data bit.
9. Get the received 8 Least Significant data bits from the receive buffer by reading the RCREG register.
10. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.

26.1.2.9 9-bit Address Detection Mode Set-up

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

1. Initialize the SPBRGH, SPBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 26.3 “EUSART Baud Rate Generator (BRG)”).
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
4. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
5. Enable 9-bit reception by setting the RX9 bit.
6. Enable address detection by setting the ADDEN bit.
7. Enable reception by setting the CREN bit.
8. The RCIF interrupt flag bit will be set when a character with the ninth bit set is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
9. Read the RCSTA register to get the error flags. The ninth data bit will always be set.
10. Get the received eight Least Significant data bits from the receive buffer by reading the RCREG register. Software determines if this is the device’s address.
11. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.
12. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and generate interrupts.

**FIGURE 26-5: ASYNCHRONOUS RECEPTION**

- RX/DT pin
- Start bit
- bit 0
- bit 1
- bit 7
- bit 8
- Stop bit
- Word 1
- RCREG
- Word 2
- RCREG
- RCIDL
- Read Rcv Buffer Reg.
- Rcv Shift Reg
- Rcv Buffer Reg.
- RCIF (Interrupt Flag)
- OERR bit
- CREN

**Note:** This timing diagram shows three words appearing on the RX input. The RCREG (receive buffer) is read after the third word, causing the OERR (overrun) bit to be set.
**TABLE 26-2: SUMMARY OF REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION**

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUDCON</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>261</td>
</tr>
<tr>
<td>INTCON</td>
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<td></td>
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<td></td>
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<td></td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
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<td></td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
</tr>
<tr>
<td>PIR1</td>
<td></td>
<td></td>
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<td>80</td>
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<tr>
<td>RCREG</td>
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<td></td>
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<td></td>
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<td></td>
<td></td>
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<td>255*</td>
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<tr>
<td>RCSTA</td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>260</td>
</tr>
<tr>
<td>SPBRGL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>262*</td>
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<tr>
<td>SPBRGH</td>
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<td></td>
<td></td>
<td></td>
<td>262*</td>
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<tr>
<td>TXSTA</td>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>259</td>
</tr>
</tbody>
</table>

*Page provides register information.*
26.2 Clock Accuracy with Asynchronous Operation

The factory calibrates the internal oscillator block output (INTOSC). However, the INTOSC frequency may drift as VDD or temperature changes, and this directly affects the asynchronous baud rate. Two methods may be used to adjust the baud rate clock, but both require a reference clock source of some kind.

The first (preferred) method uses the OSCTUNE register to adjust the INTOSC output. Adjusting the value in the OSCTUNE register allows for fine resolution changes to the system clock source. See Section 5.2.2 “Internal Clock Sources” for more information.

The other method adjusts the value in the Baud Rate Generator. This can be done automatically with the Auto-Baud Detect feature (see Section 26.3.1 “Auto-Baud Detect”). There may not be fine enough resolution when adjusting the Baud Rate Generator to compensate for a gradual change in the peripheral clock frequency.

REGISTER 26-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER

<table>
<thead>
<tr>
<th>R/W-0</th>
<th>R/W-0/0</th>
<th>R/W-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-1/1</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN(?1)</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- ‘-n/n’ = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

<table>
<thead>
<tr>
<th>bit 7</th>
<th>CSRC: Clock Source Select bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Asynchronous mode:</td>
<td></td>
</tr>
<tr>
<td>Don’t care</td>
<td></td>
</tr>
<tr>
<td>Synchronous mode:</td>
<td></td>
</tr>
<tr>
<td>1 = Master mode (clock generated internally from BRG)</td>
<td></td>
</tr>
<tr>
<td>0 = Slave mode (clock from external source)</td>
<td></td>
</tr>
<tr>
<td>bit 6</td>
<td>TX9: 9-bit Transmit Enable bit</td>
</tr>
<tr>
<td>1 = Selects 9-bit transmission</td>
<td></td>
</tr>
<tr>
<td>0 = Selects 8-bit transmission</td>
<td></td>
</tr>
<tr>
<td>bit 5</td>
<td>TXEN: Transmit Enable bit(?1)</td>
</tr>
<tr>
<td>1 = Transmit enabled</td>
<td></td>
</tr>
<tr>
<td>0 = Transmit disabled</td>
<td></td>
</tr>
<tr>
<td>bit 4</td>
<td>SYNC: EUSART Mode Select bit</td>
</tr>
<tr>
<td>1 = Synchronous mode</td>
<td></td>
</tr>
<tr>
<td>0 = Asynchronous mode</td>
<td></td>
</tr>
<tr>
<td>bit 3</td>
<td>SENDB: Send Break Character bit</td>
</tr>
<tr>
<td>Asynchronous mode:</td>
<td></td>
</tr>
<tr>
<td>1 = Send Sync Break on next transmission (cleared by hardware upon completion)</td>
<td></td>
</tr>
<tr>
<td>0 = Sync Break transmission completed</td>
<td></td>
</tr>
<tr>
<td>Synchronous mode:</td>
<td></td>
</tr>
<tr>
<td>Don’t care</td>
<td></td>
</tr>
<tr>
<td>bit 2</td>
<td>BRGH: High Baud Rate Select bit</td>
</tr>
<tr>
<td>Asynchronous mode:</td>
<td></td>
</tr>
<tr>
<td>1 = High speed</td>
<td></td>
</tr>
<tr>
<td>0 = Low speed</td>
<td></td>
</tr>
<tr>
<td>Synchronous mode:</td>
<td></td>
</tr>
<tr>
<td>Unused in this mode</td>
<td></td>
</tr>
<tr>
<td>bit 1</td>
<td>TRMT: Transmit Shift Register Status bit</td>
</tr>
<tr>
<td>1 = TSR empty</td>
<td></td>
</tr>
<tr>
<td>0 = TSR full</td>
<td></td>
</tr>
<tr>
<td>bit 0</td>
<td>TX9D: Ninth bit of Transmit Data</td>
</tr>
<tr>
<td>Can be address/data bit or a parity bit.</td>
<td></td>
</tr>
</tbody>
</table>

Note 1: SREN/CREN overrides TXEN in Sync mode.
REGISTER 26-2: RCSTA: RECEIVE STATUS AND CONTROL REGISTER(1)

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>R-0/0</th>
<th>R-x/x</th>
</tr>
</thead>
<tbody>
<tr>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
</tr>
</tbody>
</table>

Legend:
- R = Readable bit
- W = Writable bit
- U = Unimplemented bit, read as ‘0’
- u = Bit is unchanged
- x = Bit is unknown
- -n/n = Value at POR and BOR/Value at all other Resets
- ‘1’ = Bit is set
- ‘0’ = Bit is cleared

bit 7  
SPEN: Serial Port Enable bit
1 = Serial port enabled (configures RX/DT and TX/CK pins as serial port pins)
0 = Serial port disabled (held in Reset)

bit 6  
RX9: 9-bit Receive Enable bit
1 = Selects 9-bit reception
0 = Selects 8-bit reception

bit 5  
SREN: Single Receive Enable bit
- Asynchronous mode:
  Don’t care
- Synchronous mode – Master:
  1 = Enables single receive
  0 = Disables single receive
  This bit is cleared after reception is complete.
- Synchronous mode – Slave
  Don’t care

bit 4  
CREN: Continuous Receive Enable bit
- Asynchronous mode:
  1 = Enables receiver
  0 = Disables receiver
- Synchronous mode:
  1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN)
  0 = Disables continuous receive

bit 3  
ADDEN: Address Detect Enable bit
- Asynchronous mode 9-bit (RX9 = 1):
  1 = Enables address detection, enable interrupt and load the receive buffer when RSR<8> is set
  0 = Disables address detection, all bytes are received and ninth bit can be used as parity bit
- Asynchronous mode 8-bit (RX9 = 0):
  Don’t care

bit 2  
FERR: Framing Error bit
1 = Framing error (can be updated by reading RCREG register and receive next valid byte)
0 = No framing error

bit 1  
OERR: Overrun Error bit
1 = Overrun error (can be cleared by clearing bit CREN)
0 = No overrun error

bit 0  
RX9D: Ninth bit of Received Data
This can be address/data bit or a parity bit and must be calculated by user firmware.
## REGISTER 26-3: BAUDCON: BAUD RATE CONTROL REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Asynchronous mode</th>
<th>Synchronous mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>7</td>
<td>ABDOVF</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Auto-Baud Detect Overflow bit</td>
<td>1 = Auto-baud timer overflowed</td>
<td>0 = Auto-baud timer did not overflow</td>
</tr>
<tr>
<td>6</td>
<td>RCIDL</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Receive Idle Flag bit</td>
<td>1 = Receiver is Idle</td>
<td>0 = Start bit has been received and the receiver is receiving</td>
</tr>
<tr>
<td>5</td>
<td>Unimplemented</td>
<td>Read as ’0’</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>SCKP</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Synchronous Clock Polarity Select bit</td>
<td>1 = Transmit inverted data to the TX/CK pin</td>
<td>0 = Transmit non-inverted data to the TX/CK pin</td>
</tr>
<tr>
<td>3</td>
<td>BRG16</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>16-bit Baud Rate Generator bit</td>
<td>1 = 16-bit Baud Rate Generator is used</td>
<td>0 = 8-bit Baud Rate Generator is used</td>
</tr>
<tr>
<td>2</td>
<td>Unimplemented</td>
<td>Read as ’0’</td>
<td></td>
</tr>
<tr>
<td>1</td>
<td>WUE</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Wake-up Enable bit</td>
<td>1 = Receiver is waiting for a falling edge. No character will be received, byte RCIF will be set. WUE will automatically clear after RCIF is set.</td>
<td>0 = Receiver is operating normally</td>
</tr>
<tr>
<td>0</td>
<td>ABDEN</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Auto-Baud Detect Enable bit</td>
<td>1 = Auto-Baud Detect mode is enabled (clears when auto-baud is complete)</td>
<td>0 = Auto-Baud Detect mode is disabled</td>
</tr>
</tbody>
</table>

**Legend:**
- R = Readable bit
- W =Writable bit
- U = Unimplemented bit, read as ’0’
- u = Bit is unchanged
- x = Bit is unknown
- ’1’ = Bit is set
- ’0’ = Bit is cleared
- -n/n = Value at POR and BOR/Value at all other Resets
26.3 EUSART Baud Rate Generator (BRG)

The Baud Rate Generator (BRG) is an 8-bit or 16-bit timer that is dedicated to the support of both the asynchronous and synchronous EUSART operation. By default, the BRG operates in 8-bit mode. Setting the BRG16 bit of the BAUDCON register selects 16-bit mode.

The SPBRGH, SPBRGL register pair determines the period of the free running baud rate timer. In Asynchronous mode the multiplier of the baud rate period is determined by both the BRGH bit of the TXSTA register and the BRG16 bit of the BAUDCON register. In Synchronous mode, the BRGH bit is ignored.

Table 26-3 contains the formulas for determining the baud rate. Example 26-1 provides a sample calculation for determining the baud rate and baud rate error.

Typical baud rates and error values for various asynchronous modes have been computed for your convenience and are shown in Table 26-3. It may be advantageous to use the high baud rate (BRGH = 1), or the 16-bit BRG (BRG16 = 1) to reduce the baud rate error. The 16-bit BRG mode is used to achieve slow baud rates for fast oscillator frequencies.

Writing a new value to the SPBRGH, SPBRGL register pair causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

If the system clock is changed during an active receive operation, a receive error or data loss may result. To avoid this problem, check the status of the RCIDL bit to make sure that the receive operation is Idle before changing the system clock.

---

**EXAMPLE 26-1: CALCULATING BAUD RATE ERROR**

For a device with Fosc of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:

\[
\text{Desired Baud Rate} = \frac{Fosc}{64(\text{SPBRGH}\text{;}\text{SPBRGL}) + 1}
\]

Solving for SPBRGH:SPBRGL:

\[
X = \frac{\text{Desired Baud Rate}}{64} - 1
\]

\[
= \frac{16000000}{64} - 1
\]

\[
= 9600 - 1
\]

\[
= [25.042] = 25
\]

**Calculated Baud Rate**

\[
= \frac{16000000}{64(25 + 1)}
\]

\[
= 9615
\]

**Error**

\[
= \frac{\text{Calc. Baud Rate} - \text{Desired Baud Rate}}{\text{Desired Baud Rate}}
\]

\[
= \frac{(9615 - 9600)}{9600} = 0.16\%
\]
### TABLE 26-3: BAUD RATE FORMULAS

<table>
<thead>
<tr>
<th>Configuration Bits</th>
<th>BRG/EUSART Mode</th>
<th>Baud Rate Formula</th>
</tr>
</thead>
<tbody>
<tr>
<td>SYNC</td>
<td>BRG16</td>
<td>BRGH</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>x</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>x</td>
</tr>
</tbody>
</table>

**Legend:**  
- *x* = Don’t care,  
- *n* = value of SPBRGH, SPBRGL register pair.

### TABLE 26-4: SUMMARY OF REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUDCON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>ABDEN</td>
<td>261</td>
</tr>
<tr>
<td>RCSTA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>260</td>
</tr>
<tr>
<td>SPBRGL</td>
<td>BRG&lt;7:0&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>262*</td>
</tr>
<tr>
<td>SPBRGH</td>
<td>BRG&lt;15:8&gt;</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>262*</td>
</tr>
<tr>
<td>TXSTA</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRG16</td>
<td>TRMT</td>
<td>TX9D</td>
<td>259</td>
</tr>
</tbody>
</table>

**Legend:**  
- — = unimplemented location, read as ‘0’. Shaded cells are not used for the Baud Rate Generator.  
- *Page provides register information.*
### TABLE 26-5: BAUD RATES FOR ASYNCHRONOUS MODES

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Fosc = 32.000 MHz</th>
<th>Fosc = 20.000 MHz</th>
<th>Fosc = 18.432 MHz</th>
<th>Fosc = 11.0592 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>Actual Rate</td>
<td>% Error</td>
</tr>
<tr>
<td></td>
<td>SPBRG value</td>
<td>(decimal)</td>
<td>SPBRG value</td>
<td>(decimal)</td>
</tr>
<tr>
<td>300</td>
<td>300</td>
<td>—</td>
<td>300</td>
<td>—</td>
</tr>
<tr>
<td>1200</td>
<td>1200</td>
<td>—</td>
<td>1200</td>
<td>—</td>
</tr>
<tr>
<td>2400</td>
<td>2404</td>
<td>0.16</td>
<td>2404</td>
<td>0.16</td>
</tr>
<tr>
<td>9600</td>
<td>9615</td>
<td>0.16</td>
<td>9615</td>
<td>0.16</td>
</tr>
<tr>
<td>10417</td>
<td>10417</td>
<td>0.00</td>
<td>10417</td>
<td>0.00</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.23k</td>
<td>0.16</td>
<td>19.53k</td>
<td>1.73</td>
</tr>
<tr>
<td>57.6k</td>
<td>55.55k</td>
<td>-3.55</td>
<td>57.60k</td>
<td>0.00</td>
</tr>
<tr>
<td>115.2k</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Fosc = 8.000 MHz</th>
<th>Fosc = 4.000 MHz</th>
<th>Fosc = 3.6864 MHz</th>
<th>Fosc = 1.000 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>Actual Rate</td>
<td>% Error</td>
</tr>
<tr>
<td></td>
<td>SPBRG value</td>
<td>(decimal)</td>
<td>SPBRG value</td>
<td>(decimal)</td>
</tr>
<tr>
<td>300</td>
<td>300</td>
<td>0.16</td>
<td>300</td>
<td>0.16</td>
</tr>
<tr>
<td>1200</td>
<td>1202</td>
<td>0.16</td>
<td>1202</td>
<td>0.16</td>
</tr>
<tr>
<td>2400</td>
<td>2404</td>
<td>0.16</td>
<td>2404</td>
<td>0.16</td>
</tr>
<tr>
<td>9600</td>
<td>9615</td>
<td>0.16</td>
<td>9615</td>
<td>0.16</td>
</tr>
<tr>
<td>10417</td>
<td>10417</td>
<td>0.00</td>
<td>10417</td>
<td>0.00</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.23k</td>
<td>0.16</td>
<td>19.53k</td>
<td>1.73</td>
</tr>
<tr>
<td>57.6k</td>
<td>55.55k</td>
<td>-3.55</td>
<td>57.60k</td>
<td>0.00</td>
</tr>
<tr>
<td>115.2k</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Fosc = 32.000 MHz</th>
<th>Fosc = 20.000 MHz</th>
<th>Fosc = 18.432 MHz</th>
<th>Fosc = 11.0592 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>Actual Rate</td>
<td>% Error</td>
</tr>
<tr>
<td></td>
<td>SPBRG value</td>
<td>(decimal)</td>
<td>SPBRG value</td>
<td>(decimal)</td>
</tr>
<tr>
<td>300</td>
<td>300</td>
<td>—</td>
<td>300</td>
<td>—</td>
</tr>
<tr>
<td>1200</td>
<td>1200</td>
<td>—</td>
<td>1200</td>
<td>—</td>
</tr>
<tr>
<td>2400</td>
<td>2404</td>
<td>0.16</td>
<td>2404</td>
<td>0.16</td>
</tr>
<tr>
<td>9600</td>
<td>9615</td>
<td>0.16</td>
<td>9615</td>
<td>0.16</td>
</tr>
<tr>
<td>10417</td>
<td>10417</td>
<td>0.00</td>
<td>10417</td>
<td>0.00</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.23k</td>
<td>0.16</td>
<td>19.53k</td>
<td>1.73</td>
</tr>
<tr>
<td>57.6k</td>
<td>55.55k</td>
<td>-3.55</td>
<td>57.60k</td>
<td>0.00</td>
</tr>
<tr>
<td>115.2k</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Fosc = 32.000 MHz</th>
<th>Fosc = 20.000 MHz</th>
<th>Fosc = 18.432 MHz</th>
<th>Fosc = 11.0592 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Actual Rate</td>
<td>% Error</td>
<td>Actual Rate</td>
<td>% Error</td>
</tr>
<tr>
<td></td>
<td>SPBRG value</td>
<td>(decimal)</td>
<td>SPBRG value</td>
<td>(decimal)</td>
</tr>
<tr>
<td>300</td>
<td>300</td>
<td>—</td>
<td>300</td>
<td>—</td>
</tr>
<tr>
<td>1200</td>
<td>1200</td>
<td>—</td>
<td>1200</td>
<td>—</td>
</tr>
<tr>
<td>2400</td>
<td>2404</td>
<td>0.16</td>
<td>2404</td>
<td>0.16</td>
</tr>
<tr>
<td>9600</td>
<td>9615</td>
<td>0.16</td>
<td>9615</td>
<td>0.16</td>
</tr>
<tr>
<td>10417</td>
<td>10417</td>
<td>0.00</td>
<td>10417</td>
<td>0.00</td>
</tr>
<tr>
<td>19.2k</td>
<td>19.23k</td>
<td>0.16</td>
<td>19.53k</td>
<td>1.73</td>
</tr>
<tr>
<td>57.6k</td>
<td>55.55k</td>
<td>-3.55</td>
<td>57.60k</td>
<td>0.00</td>
</tr>
<tr>
<td>115.2k</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>
### TABLE 26-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

#### SYNC = 0, BRGH = 1, BRG16 = 0

<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>Actual Rate</th>
<th>% Error</th>
<th>FOSC = 8.000 MHz</th>
<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
<th>FOSC = 4.000 MHz</th>
<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
<th>FOSC = 3.6864 MHz</th>
<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
<th>FOSC = 1.000 MHz</th>
<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
</tr>
</thead>
<tbody>
<tr>
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#### SYNC = 0, BRGH = 0, BRG16 = 1

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<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
<th>FOSC = 20.000 MHz</th>
<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
<th>FOSC = 18.432 MHz</th>
<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
<th>FOSC = 11.0592 MHz</th>
<th>Actual Rate</th>
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<th>SPBRG value (decimal)</th>
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#### SYNC = 0, BRGH = 0, BRG16 = 1

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<th>Actual Rate</th>
<th>% Error</th>
<th>SPBRG value (decimal)</th>
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<th>Actual Rate</th>
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<th>SPBRG value (decimal)</th>
<th>FOSC = 18.432 MHz</th>
<th>Actual Rate</th>
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### TABLE 26-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

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<th>BAUD RATE</th>
<th>( F_{OSC} = 32.000 \text{ MHz} )</th>
<th>( F_{OSC} = 20.000 \text{ MHz} )</th>
<th>( F_{OSC} = 18.432 \text{ MHz} )</th>
<th>( F_{OSC} = 11.0592 \text{ MHz} )</th>
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<td>Actual Rate % Error SPBRG value (decimal)</td>
<td>Actual Rate % Error SPBRG value (decimal)</td>
<td>Actual Rate % Error SPBRG value (decimal)</td>
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<td>1200.0 0.00 3839</td>
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<td>2400.0 0.02 2082</td>
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<td>57.47k -0.22 86</td>
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<td>57.60k 0.00 47</td>
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<td>115.2k 0.00 39</td>
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<table>
<thead>
<tr>
<th>BAUD RATE</th>
<th>( F_{OSC} = 8.000 \text{ MHz} )</th>
<th>( F_{OSC} = 4.000 \text{ MHz} )</th>
<th>( F_{OSC} = 3.6864 \text{ MHz} )</th>
<th>( F_{OSC} = 1.000 \text{ MHz} )</th>
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<tr>
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<td>Actual Rate % Error SPBRG value (decimal)</td>
<td>Actual Rate % Error SPBRG value (decimal)</td>
<td>Actual Rate % Error SPBRG value (decimal)</td>
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<td>1200.0 0.00 767</td>
<td>1202.0 0.16 207</td>
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<td>2398.0 0.08 416</td>
<td>2400.0 0.00 383</td>
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<td>115.2k 0.00 7</td>
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**SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1**
26.3.1 AUTO-BAUD DETECT

The EUSART module supports automatic detection and calibration of the baud rate.

In the Auto-Baud Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RX signal, the RX signal is timing the BRG. The Baud Rate Generator is used to time the period of a received 55h (ASCII "U") which is the Sync character for the LIN bus. The unique feature of this character is that it has five rising edges including the Stop bit edge.

Setting the ABDEN bit of the BAUDCON register starts the auto-baud calibration sequence (Figure 26-6). While the ABD sequence takes place, the EUSART state machine is held in Idle. On the first rising edge of the receive line, after the Start bit, the SPBRG begins counting up using the BRG counter clock as shown in Table 26-6. The fifth rising edge will occur on the RX pin at the end of the eighth bit period. At that time, an accumulated value totaling the proper BRG period is left in the SPBRGH, SPBRGL register pair, the ABDEN bit is automatically cleared and the RCIF interrupt flag is set. The value in the RCREG needs to be read to clear the RCIF interrupt. RCREG content should be discarded. When calibrating for modes that do not use the SPBRGH register the user can verify that the SPBRGL register did not overflow by checking for 00h in the SPBRG register.

The BRG auto-baud clock is determined by the BRG16 and BRGH bits as shown in Table 26-6. During ABD, both the SPBRGH and SPBRGL registers are used as a 16-bit counter, independent of the BRG16 bit setting. While calibrating the baud rate period, the SPBRGH and SPBRGL registers are clocked at 1/8th the BRG base clock rate. The resulting byte measurement is the average bit time when clocked at full speed.

Note 1: If the WUE bit is set with the ABDEN bit, auto-baud detection will occur on the byte following the Break character (see Section 26.3 “EUSART Baud Rate Generator (BRG)”.

2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible.

3: During the auto-baud process, the auto-baud counter starts counting at ‘1’. Upon completion of the auto-baud sequence, to achieve maximum accuracy, subtract 1 from the SPBRGH:SPBRGL register pair.

TABLE 26-6: BRG COUNTER CLOCK RATES

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<tr>
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<th>BRG Base Clock</th>
<th>BRG ABD Clock(1)</th>
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<td>Fosc/512</td>
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<tr>
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<td>1</td>
<td>1</td>
<td>Fosc/4</td>
<td>Fosc/32</td>
</tr>
</tbody>
</table>

Note: During the ABD sequence, SPBRGL and SPBRGH registers are both used as a 16-bit counter, independent of BRG16 setting.

![FIGURE 26-6: AUTOMATIC BAUD RATE CALIBRATION](image-url)

Note 1: The ABD sequence requires the EUSART module to be configured in Asynchronous mode.
26.3.2 AUTO-BAUD OVERFLOW

During the course of automatic baud detection, the ABDOVF bit of the BAUDCON register will be set if the baud rate counter overflows before the fifth rising edge is detected on the RX pin. The ABDOVF bit indicates that the counter has exceeded the maximum count that can fit in the 16 bits of the SPBRGH:SPBRGL register pair. After the ABDOVF has been set, the counter continues to count until the fifth rising edge is detected on the RX pin. Upon detecting the fifth RX edge, the hardware will set the RCIF interrupt flag and clear the ABDEN bit of the BAUDCON register. The RCIF flag can be subsequently cleared by reading the RCREG register. The ABDOVF flag of the BAUDCON register can be cleared by software directly.

To terminate the auto-baud process before the RCIF flag is set, clear the ABDEN bit then clear the ABDOVF bit of the BAUDCON register. The ABDOVF bit will remain set if the ABDEN bit is not cleared first.

26.3.3 AUTO-WAKE-UP ON BREAK

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper character reception cannot be performed. The Auto-Wake-up feature allows the controller to wake-up due to activity on the RX/DT line. This feature is available only in Asynchronous mode.

The Auto-Wake-up feature is enabled by setting the WUE bit of the BAUDCON register. Once set, the normal receive sequence on RX/DT is disabled, and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX/DT line. (This coincides with the start of a Sync Break or a wake-up signal character for the LIN protocol.)

The EUSART module generates an RCIF interrupt coincident with the wake-up event. The interrupt is generated synchronously to the Q clocks in normal CPU operating modes (Figure 26-7), and asynchronously if the device is in Sleep mode (Figure 26-8). The interrupt condition is cleared by reading the RCREG register.

The WUE bit is automatically cleared by the low-to-high transition on the RX line at the end of the Break. This signals to the user that the Break event is over. At this point, the EUSART module is in Idle mode waiting to receive the next character.

26.3.3.1 Special Considerations

Break Character

To avoid character errors or character fragments during a wake-up event, the wake-up character must be all zeros.

When the wake-up is enabled the function works independent of the low time on the data stream. If the WUE bit is set and a valid non-zero character is received, the low time from the Start bit to the first rising edge will be interpreted as the wake-up event. The remaining bits in the character will be received as a fragmented character and subsequent characters can result in framing or overrun errors.

Therefore, the initial character in the transmission must be all ‘0’s. This must be 10 or more bit times, 13-bit times recommended for LIN bus, or any number of bit times for standard RS-232 devices.

Oscillator Start-up Time

Oscillator start-up time must be considered, especially in applications using oscillators with longer start-up intervals (i.e., LP, XT or HS/PLL mode). The Sync Break (or wake-up signal) character must be of sufficient length, and be followed by a sufficient interval, to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

WUE Bit

The wake-up event causes a receive interrupt by setting the RCIF bit. The WUE bit is cleared in hardware by a rising edge on RX/DT. The interrupt condition is then cleared in software by reading the RCREG register and discarding its contents.

To ensure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process before setting the WUE bit. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.
FIGURE 26-7: AUTO-WAKE-UP BIT (WUE) TIMING DURING NORMAL OPERATION

Note 1: The EUSART remains in Idle while the WUE bit is set.

FIGURE 26-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP

Note 1: If the wake-up event requires long oscillator warm-up time, the automatic clearing of the WUE bit can occur while the stposc signal is still active. This sequence should not depend on the presence of Q clocks.

2: The EUSART remains in Idle while the WUE bit is set.
26.3.4 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. A Break character consists of a Start bit, followed by 12 ‘0’ bits and a Stop bit.

To send a Break character, set the SENDB and TXEN bits of the TXSTA register. The Break character transmission is then initiated by a write to the TXREG. The value of data written to TXREG will be ignored and all ‘0’s will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

The TRMT bit of the TXSTA register indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 26-9 for the timing of the Break character sequence.

26.3.4.1 Break and Sync Transmit Sequence

The following sequence will start a message frame header made up of a Break, followed by an auto-baud Sync byte. This sequence is typical of a LIN bus master.

1. Configure the EUSART for the desired mode.
2. Set the TXEN and SENDB bits to enable the Break sequence.
3. Load the TXREG with a dummy character to initiate transmission (the value is ignored).
4. Write ‘55h’ to TXREG to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TXREG becomes empty, as indicated by the TXIF, the next data byte can be written to TXREG.

26.3.5 RECEIVING A BREAK CHARACTER

The Enhanced EUSART module can receive a Break character in two ways.

The first method to detect a Break character uses the FERR bit of the RCSTA register and the received data as indicated by RCREG. The Baud Rate Generator is assumed to have been initialized to the expected baud rate.

A Break character has been received when;
- RCIF bit is set
- FERR bit is set
- RCREG = 00h

The second method uses the Auto-Wake-up feature described in Section 26.3.3 “Auto-Wake-up on Break”. By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCIF interrupt, and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Detect feature. For both methods, the user can set the ABDEN bit of the BAUDCON register before placing the EUSART in Sleep mode.

FIGURE 26-9: SEND BREAK CHARACTER SEQUENCE
26.4 EUSART Synchronous Mode

Synchronous serial communications are typically used in systems with a single master and one or more slaves. The master device contains the necessary circuitry for baud rate generation and supplies the clock for all devices in the system. Slave devices can take advantage of the master clock by eliminating the internal clock generation circuitry.

There are two signal lines in Synchronous mode: a bidirectional data line and a clock line. Slaves use the external clock supplied by the master to shift the serial data into and out of their respective receive and transmit shift registers. Since the data line is bidirectional, synchronous operation is half-duplex only. Half-duplex refers to the fact that master and slave devices can receive and transmit data but not both simultaneously. The EUSART can operate as either a master or slave device.

Start and Stop bits are not used in synchronous transmissions.

26.4.1 SYNCHRONOUS MASTER MODE

The following bits are used to configure the EUSART for Synchronous Master operation:

- SYNC = 1
- CSRC = 1
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXSTA register configures the device for synchronous operation. Setting the CSRC bit of the TXSTA register configures the device as a master. Clearing the SREN and CREN bits of the RCSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTA register enables the EUSART.

26.4.1.1 Master Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a master transmits the clock on the TX/CK line. The TX/CK pin output driver is automatically enabled when the EUSART is configured for synchronous transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One clock cycle is generated for each data bit. Only as many clock cycles are generated as there are data bits.

26.4.1.2 Clock Polarity

A clock polarity option is provided for Microwire compatibility. Clock polarity is selected with the SCKP bit of the BAUDCON register. Setting the SCKP bit sets the clock Idle state as high. When the SCKP bit is set, the data changes on the falling edge of each clock. Clearing the SCKP bit sets the Idle state as low. When the SCKP bit is cleared, the data changes on the rising edge of each clock.

26.4.1.3 Synchronous Master Transmission

Data is transferred out of the device on the RX/DT pin. The RX/DT and TX/CK pin output drivers are automatically enabled when the EUSART is configured for synchronous master transmit operation.

A transmission is initiated by writing a character to the TXREG register. If the TSR still contains all or part of a previous character the new character data is held in the TXREG until the last bit of the previous character has been transmitted. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXREG is immediately transferred to the TSR. The transmission of the character commences immediately following the transfer of the data to the TSR from the TXREG.

Each data bit changes on the leading edge of the master clock and remains valid until the subsequent leading clock edge.

Note: The TSR register is not mapped in data memory, so it is not available to the user.

26.4.1.4 Synchronous Master Transmission Set-up:

1. Initialize the SPBRGH, SPBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 26.3 “EUSART Baud Rate Generator (BRG)”).
2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
3. Disable Receive mode by clearing bits SREN and CREN.
4. Enable Transmit mode by clearing bits SREN and CREN.
5. If 9-bit transmission is desired, set the TX9 bit.
6. If interrupts are desired, set the TXIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
7. If 9-bit transmission is selected, the ninth bit should be loaded in the TX9D bit.
8. Start transmission by loading data to the TXREG register.
FIGURE 26-10: SYNCHRONOUS TRANSMISSION

- RX/DT pin
- TX/CK pin (SCKP = 0)
- TX/CK pin (SCKP = 1)
- Word 1
- Word 2
- Write to TXREG reg
- TXIF bit (Interrupt Flag)
- TRMT bit
- TXEN bit

Note: Sync Master mode, SPBRGL = 0, continuous transmission of two 8-bit words.

FIGURE 26-11: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)

- RX/DT pin
- TX/CK pin
- Write to TXREG reg
- TXIF bit
- TRMT bit
- TXEN bit

TABLE 26-7: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUDCON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>ABDEN</td>
<td>261</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOCIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>RCSTA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>260</td>
</tr>
<tr>
<td>SPBRGL</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>262*</td>
</tr>
<tr>
<td>SPBRGH</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>262*</td>
</tr>
<tr>
<td>TXREG</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>259</td>
</tr>
</tbody>
</table>

Legend:
- = unimplemented location, read as '0'. Shaded cells are not used for synchronous master transmission.
- * Page provides register information.
26.4.1.5 Synchronous Master Reception

Data is received at the RX/DT pin. The RX/DT pin output driver is automatically disabled when the EUSART is configured for synchronous master receive operation.

In Synchronous mode, reception is enabled by setting either the Single Receive Enable bit (SREN of the RCSTA register) or the Continuous Receive Enable bit (CREN of the RCSTA register).

When SREN is set and CREN is clear, only as many clock cycles are generated as there are data bits in a single character. The SREN bit is automatically cleared at the completion of one character. When CREN is set, clocks are continuously generated until CREN is cleared. If CREN is cleared in the middle of a character the CK clock stops immediately and the partial character is discarded. If SREN and CREN are both set, then SREN is cleared at the completion of the first character and CREN takes precedence.

To initiate reception, set either SREN or CREN. Data is sampled at the RX/DT pin on the trailing edge of the TX/CK clock pin and is shifted into the Receive Shift Register (RSR). When a complete character is received into the RSR, the RCIF bit is set and the character is automatically transferred to the two character receive FIFO. The Least Significant eight bits of the top character in the receive FIFO are available in RCREG. The RCIF bit remains set as long as there are unread characters in the receive FIFO.

Note: If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.

26.4.1.6 Slave Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a slave receives the clock on the TX/CK line. The TX/CK pin output driver is automatically disabled when the device is configured for synchronous slave transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One data bit is transferred for each clock cycle. Only as many clock cycles should be received as there are data bits.

Note: If the device is configured as a slave and the TX/CK function is on an analog pin, the corresponding ANSEL bit must be cleared.

26.4.1.7 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before RCREG is read to access the FIFO. When this happens the OERR bit of the RCSTA register is set. Previous data in the FIFO will not be overwritten. The two characters in the FIFO buffer can be read, however, no additional characters will be received until the error is cleared. The OERR bit can only be cleared by clearing the overrun condition. If the overrun error occurred when the SREN bit is set and CREN is clear then the error is cleared by reading RCREG. If the overrun occurred when the CREN bit is set then the error condition is cleared by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

26.4.1.8 Receiving 9-bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the EUSART will shift 9-bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth, and Most Significant, data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCREG.

26.4.1.9 Synchronous Master Reception

Set-up:
1. Initialize the SPBRG, SPBRGL register pair for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
4. Ensure bits CREN and SREN are clear.
5. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
6. If 9-bit reception is desired, set bit RX9.
7. Start reception by setting the SREN bit or for continuous reception, set the CREN bit.
8. Interrupt flag bit RCIF will be set when reception of a character is complete. An interrupt will be generated if the enable bit RCIE was set.
9. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
10. Read the 8-bit received data by reading the RCREG register.
11. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

Note: If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.

Note: If the device is configured as a slave and the TX/CK function is on an analog pin, the corresponding ANSEL bit must be cleared.
FIGURE 26-12: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

TABLE 26-8: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUDCON</td>
<td>ABOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
<td>ABDEN</td>
<td>261</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IO CIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
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</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
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</tr>
<tr>
<td>RCREG</td>
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<td>—</td>
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<td>—</td>
<td>—</td>
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<td>255*</td>
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<tr>
<td>RCSTA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>260</td>
</tr>
<tr>
<td>SPBRGL</td>
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<td></td>
<td></td>
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<td>SPBRGH</td>
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<td></td>
<td></td>
<td></td>
<td>262*</td>
</tr>
<tr>
<td>TXSTA</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>259</td>
</tr>
</tbody>
</table>

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for synchronous master reception.

* Page provides register information.

Note: Timing diagram demonstrates Sync Master mode with bit SREN = 1 and bit BRGH = 0.
26.4.2 SYMMONOUS SLAVE MODE

The following bits are used to configure the EUSART for Synchronous slave operation:

- **SYNC = 1**
- **CSRC = 0**
- **SREN = 0** (for transmit); **SREN = 1** (for receive)
- **CREN = 0** (for transmit); **CREN = 1** (for receive)
- **SPEN = 1**

Setting the SYNC bit of the TXSTA register configures the device for synchronous operation. Clearing the CSRC bit of the TXSTA register configures the device as a slave. Clearing the SREN and CREN bits of the RCSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTA register enables the EUSART.

26.4.2.1 EUSART Synchronous Slave Transmit

The operation of the Synchronous Master and Slave modes are identical (see Section 26.4.1.3 “Synchronous Master Transmission”), except in the case of the Sleep mode.

If two words are written to the TXREG and then the **SLEEP** instruction is executed, the following will occur:

1. The first character will immediately transfer to the TSR register and transmit.
2. The second word will remain in the TXREG register.
3. The TXIF bit will not be set.
4. After the first character has been shifted out of TSR, the TXREG register will transfer the second character to the TSR and the TXIF bit will now be set.
5. If the PEIE and TXIE bits are set, the interrupt will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will call the Interrupt Service Routine.

26.4.2.2 Synchronous Slave Transmission

Set-up:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Clear the ANSEL bit for the CK pin (if applicable).
3. Clear the CREN and SREN bits.
4. If interrupts are desired, set the TXIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
5. If 9-bit transmission is desired, set the TX9 bit.
6. Enable transmission by setting the TXEN bit.
7. If 9-bit transmission is selected, insert the Most Significant bit into the TX9D bit.
8. Start transmission by writing the Least Significant eight bits to the TXREG register.

### TABLE 26-9: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUDCON</td>
<td>ABDOVF</td>
<td>RCIDL</td>
<td>—</td>
<td>SCKP</td>
<td>BRG16</td>
<td>—</td>
<td>WUE</td>
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<td>INTCON</td>
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<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IO CIF</td>
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<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
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<td>ADDEN</td>
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<td>OERR</td>
<td>RX9D</td>
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<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>259</td>
</tr>
</tbody>
</table>

**Legend:** — = unimplemented location, read as ’0’. Shaded cells are not used for synchronous slave transmission. *Page provides register information.
26.4.2.3 EUSART Synchronous Slave Reception

The operation of the Synchronous Master and Slave modes is identical (Section 26.4.1.5 “Synchronous Master Reception”), with the following exceptions:

- Sleep
- CREN bit is always set, therefore the receiver is never idle
- SREN bit, which is a “don’t care” in Slave mode

A character may be received while in Sleep mode by setting the CREN bit prior to entering Sleep. Once the word is received, the RSR register will transfer the data to the RCREG register. If the RCIE enable bit is set, the interrupt generated will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will branch to the interrupt vector.

26.4.2.4 Synchronous Slave Reception

Set-up:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Clear the ANSEL bit for both the CK and DT pins (if applicable).
3. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
4. If 9-bit reception is desired, set the RX9 bit.
5. Set the CREN bit to enable reception.
6. The RCIF bit will be set when reception is complete. An interrupt will be generated if the RCIE bit was set.
7. If 9-bit mode is enabled, retrieve the Most Significant bit from the RX9D bit of the RCSTA register.
8. Retrieve the eight Least Significant bits from the receive FIFO by reading the RCREG register.
9. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

### TABLE 26-10: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>BAUDCON</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>261</td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOCIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IO CIF</td>
<td>77</td>
</tr>
<tr>
<td>PIE1</td>
<td>TMR1GIE</td>
<td>ADIE</td>
<td>RCIE</td>
<td>TXIE</td>
<td>SSP1IE</td>
<td>CCP1IE</td>
<td>TMR2IE</td>
<td>TMR1IE</td>
<td>78</td>
</tr>
<tr>
<td>PIR1</td>
<td>TMR1GIF</td>
<td>ADIF</td>
<td>RCIF</td>
<td>TXIF</td>
<td>SSP1IF</td>
<td>CCP1IF</td>
<td>TMR2IF</td>
<td>TMR1IF</td>
<td>80</td>
</tr>
<tr>
<td>RCREG</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>255*</td>
</tr>
<tr>
<td>RCSTA</td>
<td>SPEN</td>
<td>RX9</td>
<td>SREN</td>
<td>CREN</td>
<td>ADDEN</td>
<td>FERR</td>
<td>OERR</td>
<td>RX9D</td>
<td>260</td>
</tr>
<tr>
<td>TXSTA</td>
<td>CSRC</td>
<td>TX9</td>
<td>TXEN</td>
<td>SYNC</td>
<td>SENDB</td>
<td>BRGH</td>
<td>TRMT</td>
<td>TX9D</td>
<td>259</td>
</tr>
</tbody>
</table>

*Page provides register information.

Legend: — = unimplemented location, read as ‘0’. Shaded cells are not used for synchronous slave reception.
26.5 EUSART Operation During Sleep

The EUSART will remain active during Sleep only in the Synchronous Slave mode. All other modes require the system clock and therefore cannot generate the necessary signals to run the Transmit or Receive Shift registers during Sleep.

Synchronous Slave mode uses an externally generated clock to run the Transmit and Receive Shift registers.

26.5.1 SYNCHRONOUS RECEIVE DURING SLEEP

To receive during Sleep, all the following conditions must be met before entering Sleep mode:

- The RCSTA and TXSTA Control registers must be configured for Synchronous Slave Reception (see Section 26.4.2.4 “Synchronous Slave Reception Set-up:”).
- If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- The RCIF interrupt flag must be cleared by reading RCREG to unload any pending characters in the receive buffer.

Upon entering Sleep mode, the device will be ready to accept data and clocks on the RX/DT and TX/CK pins, respectively. When the data word has been completely clocked in by the external device, the RCIF interrupt flag bit of the PIR1 register will be set. Thereby, waking the processor from Sleep.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit is also set then the Interrupt Service Routine at address 0004h will be called.

26.5.2 SYNCHRONOUS TRANSMIT DURING SLEEP

To transmit during Sleep, all the following conditions must be met before entering Sleep mode:

- RCSTA and TXSTA Control registers must be configured for Synchronous Slave Transmission (see Section 26.4.2.2 “Synchronous Slave Transmission Set-up:”).
- The TXIF interrupt flag must be cleared by writing the output data to the TXREG, thereby filling the TSR and transmit buffer.
- If interrupts are desired, set the TXIE bit of the PIE1 register and the PEIE bit of the INTCON register.
- Interrupt enable bits TXIE of the PIE1 register and PEIE of the INTCON register must set.

Upon entering Sleep mode, the device will be ready to accept clocks on the TX/CK pin and transmit data on the RX/DT pin. When the data word in the TSR has been completely clocked out by the external device, the pending byte in the TXREG will transfer to the TSR and the TXIF flag will be set. Thereby, waking the processor from Sleep. At this point, the TXREG is available to accept another character for transmission, which will clear the TXIF flag.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit is also set then the Interrupt Service Routine at address 0004h will be called.

26.5.3 ALTERNATE PIN LOCATIONS

This module incorporates I/O pins that can be moved to other locations with the use of the alternate pin function register, APFCON. To determine which pins can be moved and what their default locations are upon a Reset, see Section 12.1 “Alternate Pin Function” for more information.
27.0 CAPACITIVE SENSING (CPS) MODULE

The Capacitive Sensing (CPS) module allows for an interaction with an end user without a mechanical interface. In a typical application, the CPS is attached to a pad on a Printed Circuit Board (PCB), which is electrically isolated from the end user. When the end user places their finger over the PCB pad, a capacitive load is added, causing a frequency shift in the CPS module. The CPS module requires software and at least one timer resource to determine the change in frequency. Key features of this module include:

- Analog MUX for monitoring multiple inputs
- Capacitive sensing oscillator
- Multiple Power modes
- High power range with variable voltage references
- Multiple timer resources
- Software control
- Operation during Sleep

**FIGURE 27-1: CAPACITIVE SENSING BLOCK DIAGRAM**

*Note 1: If CPSON = 0, disabling capacitive sensing, no channel is selected.*
FIGURE 27-2: CAPACITIVE SENSING OSCILLATOR BLOCK DIAGRAM

Note 1: Module Enable and Power mode selections are not shown.
2: Comparator remains active in Noise Detection mode.
27.1 Analog MUX

The CPS can monitor up to four inputs. See Register 27-2 for details. The capacitive sensing inputs are defined as CPS<7:0>, as applicable to device. To determine if a frequency change has occurred the user must:

- Select the appropriate CPS pin by setting the appropriate CPSCH bits of the CPSCON1 register.
- Set the corresponding ANSEL bit.
- Set the corresponding TRIS bit.
- Run the software algorithm.

Selection of the CPSx pin while the module is enabled will cause the capacitive sensing oscillator to be on the CPSx pin. Failure to set the corresponding ANSEL and TRIS bits can cause the capacitive sensing oscillator to stop, leading to false frequency readings.

27.2 Capacitive Sensing Oscillator

The capacitive sensing oscillator consists of a constant current source and a constant current sink, to produce a triangle waveform. The CPSOUT bit of the CPSCON0 register shows the status of the capacitive sensing oscillator, whether it is a sinking or sourcing current. The oscillator is designed to drive a capacitive load (single PCB pad) and at the same time, be a clock source to either Timer0 or Timer1. The oscillator has three different current settings as defined by CPSRNG<1:0> of the CPSCON0 register. The different current settings for the oscillator serve two purposes:

- Maximize the number of counts in a timer for a fixed time base.
- Maximize the count differential in the timer during a change in frequency.

27.3 Voltage References

The capacitive sensing oscillator uses voltage references to provide two voltage thresholds for oscillation. The upper voltage threshold is referred to as Ref+ and the lower voltage threshold is referred to as Ref-.

The user can elect to use fixed voltage references, which are internal to the capacitive sensing oscillator, or variable voltage references, which are supplied by the Fixed Voltage Reference (FVR) module and the Digital-to-Analog Converter (DAC) module.

When the fixed voltage references are used, the VSS voltage determines the lower threshold level (Ref-) and the VDD voltage determines the upper threshold level (Ref+).

When the variable voltage references are used, the DAC voltage determines the lower threshold level (Ref-) and the FVR voltage determines the upper threshold level (Ref+). An advantage of using these reference sources is that oscillation frequency remains constant with changes in VDD.

Different oscillation frequencies can be obtained through the use of these variable voltage references. The more the upper voltage reference level is lowered and the more the lower voltage reference level is raised, the higher the capacitive sensing oscillator frequency becomes.

Selection between the voltage references is controlled by the CPSRM bit of the CPSCON0 register. Setting this bit selects the variable voltage references and clearing this bit selects the fixed voltage references.

Please see Section 14.0 “Fixed Voltage Reference (FVR)” and Section 17.0 “Digital-to-Analog Converter (DAC) Module” for more information on configuring the variable voltage levels.
### 27.4 Power Modes

The capacitive sensing oscillator can operate in one of seven different power modes. The power modes are separated into two ranges; the low range and the high range.

When the oscillator’s low range is selected, the fixed internal voltage references of the capacitive sensing oscillator are being used. When the oscillator’s high range is selected, the variable voltage references supplied by the FVR and DAC modules are being used. Selection between the voltage references is controlled by the CPSRM bit of the CPSCON0 register. See Section 27.3 “Voltage References” for more information.

Within each range there are three distinct Power modes; low, medium and high. Current consumption is dependent upon the range and mode selected. Selecting Power modes within each range is accomplished by configuring the CPSRNG <1:0> bits in the CPSCON0 register. See Table 27-1 for proper Power mode selection.

The remaining mode is a Noise Detection mode that resides within the high range. The Noise Detection mode is unique in that it disables the sinking and sourcing of current on the analog pin but leaves the rest of the oscillator circuitry active. This reduces the oscillation frequency on the analog pin to zero and also greatly reduces the current consumed by the oscillator module.

When noise is introduced onto the pin, the oscillator is driven at the frequency determined by the noise. This produces a detectable signal at the comparator output, indicating the presence of activity on the pin.

Figure 27-2 shows a more detailed drawing of the current sources and comparators associated with the oscillator.

<table>
<thead>
<tr>
<th>CPSRM</th>
<th>Range</th>
<th>CPSRNG&lt;1:0&gt;</th>
<th>Mode(1)</th>
<th>Current Range(1)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>Low</td>
<td>00</td>
<td>Off</td>
<td>Noise Detection</td>
</tr>
<tr>
<td></td>
<td></td>
<td>01</td>
<td>Low</td>
<td>Low</td>
</tr>
<tr>
<td></td>
<td></td>
<td>10</td>
<td>Medium</td>
<td>Medium</td>
</tr>
<tr>
<td></td>
<td></td>
<td>11</td>
<td>High</td>
<td>High</td>
</tr>
<tr>
<td>1</td>
<td>High</td>
<td>00</td>
<td>Noise Detection</td>
<td>Off</td>
</tr>
<tr>
<td></td>
<td></td>
<td>01</td>
<td>Low</td>
<td>Low</td>
</tr>
<tr>
<td></td>
<td></td>
<td>10</td>
<td>Medium</td>
<td>Medium</td>
</tr>
<tr>
<td></td>
<td></td>
<td>11</td>
<td>High</td>
<td>High</td>
</tr>
</tbody>
</table>

**Note 1:** See Power-Down Currents (IPD) in Section 31.0 “Electrical Specifications” for more information.
27.5 Timer Resources
To measure the change in frequency of the capacitive sensing oscillator, a fixed time base is required. For the period of the fixed time base, the capacitive sensing oscillator is used to clock either Timer0 or Timer1. The frequency of the capacitive sensing oscillator is equal to the number of counts in the timer, divided by the period of the fixed time base.

27.6 Fixed Time Base
To measure the frequency of the capacitive sensing oscillator, a fixed time base is required. Any timer resource or software loop can be used to establish the fixed time base. It is up to the end user to determine the method in which the fixed time base is generated.

**Note:** The fixed time base can not be generated by the timer resource that the capacitive sensing oscillator is clocking.

27.6.1 TIMER0
To select Timer0 as the timer resource for the CPS:
- Set the T0XCS bit of the CPSCON0 register.
- Clear the TMROCS bit of the OPTION register.

When Timer0 is chosen as the timer resource, the capacitive sensing oscillator will be the clock source for Timer0. Refer to Section 20.0 “Timer0 Module” for additional information.

27.6.2 TIMER1
To select Timer1 as the timer resource for the CPS module, set the TMR1CS<1:0> of the T1CON register to ‘11’. When Timer1 is chosen as the timer resource, the capacitive sensing oscillator will be the clock source for Timer1. Because the Timer1 module has a gate control, developing a time base for the frequency measurement can be simplified by using the Timer0 overflow flag.

It is recommend that the Timer0 overflow flag, in conjunction with the Toggle mode of the Timer1 Gate, be used to develop the fixed time base required by the software portion of the CPS module. Refer to Section 21.12 “Timer1 Gate Control Register” for additional information.

<table>
<thead>
<tr>
<th>TABLE 27-2: TIMER1 ENABLE FUNCTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>TMR1ON</td>
</tr>
<tr>
<td>--------</td>
</tr>
<tr>
<td>0</td>
</tr>
<tr>
<td>0</td>
</tr>
<tr>
<td>1</td>
</tr>
<tr>
<td>1</td>
</tr>
</tbody>
</table>

27.7 Software Control
The software portion of the CPS module is required to determine the change in frequency of the capacitive sensing oscillator. This is accomplished by the following:
- Setting a fixed time base to acquire counts on Timer0 or Timer1.
- Establishing the nominal frequency for the capacitive sensing oscillator.
- Establishing the reduced frequency for the capacitive sensing oscillator due to an additional capacitive load.
- Set the frequency threshold.

27.7.1 NOMINAL FREQUENCY (NO CAPACITIVE LOAD)
To determine the nominal frequency of the capacitive sensing oscillator:
- Remove any extra capacitive load on the selected CPSx pin.
- At the start of the fixed time base, clear the timer resource.
- At the end of the fixed time base save the value in the timer resource.

The value of the timer resource is the number of oscillations of the capacitive sensing oscillator for the given time base. The frequency of the capacitive sensing oscillator is equal to the number of counts on the timer, divided by the period of the fixed time base.

27.7.2 REDUCED FREQUENCY (ADDITIONAL CAPACITIVE LOAD)
The extra capacitive load will cause the frequency of the capacitive sensing oscillator to decrease. To determine the reduced frequency of the capacitive sensing oscillator:
- Add a typical capacitive load on the selected CPSx pin.
- Use the same fixed time base as the nominal frequency measurement.
- At the start of the fixed time base, clear the timer resource.
- At the end of the fixed time base, save the value in the timer resource.

The value of the timer resource is the number of oscillations of the capacitive sensing oscillator with an additional capacitive load. The frequency of the capacitive sensing oscillator is equal to the number of counts on in the timer, divided by the period of the fixed time base. This frequency should be less than the value obtained during the nominal frequency measurement.
27.7.3 FREQUENCY THRESHOLD

The frequency threshold should be placed midway between the value of nominal frequency and the reduced frequency of the capacitive sensing oscillator. Refer to Application Note AN1103, “Software Handling for Capacitive Sensing” (DS01103) for more detailed information on the software required for CPS module.

Note: For more information on general capacitive sensing refer to Application Notes:

- AN1101, “Introduction to Capacitive Sensing” (DS01101)
- AN1102, “Layout and Physical Design Guidelines for Capacitive Sensing” (DS01102)

27.8 Operation during Sleep

The capacitive sensing oscillator will continue to run as long as the module is enabled, independent of the part being in Sleep. In order for the software to determine if a frequency change has occurred, the part must be awake. However, the part does not have to be awake when the timer resource is acquiring counts.

Note: Timer0 does not operate when in Sleep, and therefore, cannot be used for capacitive sense measurements in Sleep.
REGISTER 27-1:  CPSCON0: CAPACITIVE SENSING CONTROL REGISTER 0

<table>
<thead>
<tr>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/W-0/0</th>
<th>R/W-0/0</th>
<th>R-0/0</th>
<th>R/W-0/0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CPSON</td>
<td>CPSRM</td>
<td>CPSRNG&lt;1:0&gt;</td>
<td>CPSOUT</td>
<td>T0XCS</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>bit 7</th>
<th>bit 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>CPSON: CPS Module Enable bit</td>
<td></td>
</tr>
<tr>
<td>1 = CPS module is enabled</td>
<td></td>
</tr>
<tr>
<td>0 = CPS module is disabled</td>
<td></td>
</tr>
<tr>
<td>CPSRM: Capacitive Sensing Reference Mode bit</td>
<td></td>
</tr>
<tr>
<td>1 = CPS module is in high range. DAC and FVR provide oscillator voltage references.</td>
<td></td>
</tr>
<tr>
<td>0 = CPS module is in the low range. Internal oscillator voltage references are used.</td>
<td></td>
</tr>
<tr>
<td>CPSRNG&lt;1:0&gt;: Capacitive Sensing Current Range bit</td>
<td></td>
</tr>
<tr>
<td>If CPSRM = 0 (low range):</td>
<td></td>
</tr>
<tr>
<td>00 = Oscillator is off</td>
<td></td>
</tr>
<tr>
<td>01 = Oscillator is in Low Range. Charge/Discharge Current is nominally 0.1 µA</td>
<td></td>
</tr>
<tr>
<td>10 = Oscillator is in Medium Range. Charge/Discharge Current is nominally 1.2 µA</td>
<td></td>
</tr>
<tr>
<td>11 = Oscillator is in High Range. Charge/Discharge Current is nominally 18 µA</td>
<td></td>
</tr>
<tr>
<td>If CPSRM = 1 (high range):</td>
<td></td>
</tr>
<tr>
<td>00 = Oscillator is on. Noise Detection mode. No Charge/Discharge current is supplied.</td>
<td></td>
</tr>
<tr>
<td>01 = Oscillator is in Low Range. Charge/Discharge Current is nominally 9 µA</td>
<td></td>
</tr>
<tr>
<td>10 = Oscillator is in Medium Range. Charge/Discharge Current is nominally 30 µA</td>
<td></td>
</tr>
<tr>
<td>11 = Oscillator is in High Range. Charge/Discharge Current is nominally 100 µA</td>
<td></td>
</tr>
<tr>
<td>CPSOUT: Capacitive Sensing Oscillator Status bit</td>
<td></td>
</tr>
<tr>
<td>1 = Oscillator is sourcing current (Current flowing out of the pin)</td>
<td></td>
</tr>
<tr>
<td>0 = Oscillator is sinking current (Current flowing into the pin)</td>
<td></td>
</tr>
<tr>
<td>T0XCS: Timer0 External Clock Source Select bit</td>
<td></td>
</tr>
<tr>
<td>If TMR0CS = 1:</td>
<td></td>
</tr>
<tr>
<td>The T0XCS bit controls which clock external to the core/Timer0 module supplies Timer0:</td>
<td></td>
</tr>
<tr>
<td>1 = Timer0 clock source is the capacitive sensing oscillator</td>
<td></td>
</tr>
<tr>
<td>0 = Timer0 clock source is the T0CKI pin</td>
<td></td>
</tr>
<tr>
<td>If TMR0CS = 0:</td>
<td></td>
</tr>
<tr>
<td>Timer0 clock source is controlled by the core/Timer0 module and is Fosc/4</td>
<td></td>
</tr>
</tbody>
</table>

Legend:
R = Readable bit
W = Writable bit
U = Unimplemented bit, read as ‘0’
u = Bit is unchanged
x = Bit is unknown
- n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set
‘0’ = Bit is cleared
TABLE 27-3: SUMMARY OF REGISTERS ASSOCIATED WITH CAPACITIVE SENSING

<table>
<thead>
<tr>
<th>Name</th>
<th>Bit 7</th>
<th>Bit 6</th>
<th>Bit 5</th>
<th>Bit 4</th>
<th>Bit 3</th>
<th>Bit 2</th>
<th>Bit 1</th>
<th>Bit 0</th>
<th>Register on Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>ANSELA</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>ANSA4</td>
<td>ANSA2</td>
<td>ANSA1</td>
<td>ANSA0</td>
</tr>
<tr>
<td>CPSCON0</td>
<td>CPSON</td>
<td>CPSRM</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>CPSRNG1</td>
<td>CPSRNG0</td>
<td>CPSOUT</td>
<td>TOXCS</td>
</tr>
<tr>
<td>CPSCON1</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>CPSCH1</td>
<td>CPSCH0</td>
<td></td>
</tr>
<tr>
<td>INTCON</td>
<td>GIE</td>
<td>PEIE</td>
<td>TMR0IE</td>
<td>INTE</td>
<td>IOIE</td>
<td>TMR0IF</td>
<td>INTF</td>
<td>IOIF</td>
<td></td>
</tr>
<tr>
<td>OPTION_REG</td>
<td>WPUEN</td>
<td>INTEDG</td>
<td>TMR0CS</td>
<td>TMR0SE</td>
<td>PSA</td>
<td>PS2</td>
<td>PS1</td>
<td>PS0</td>
<td></td>
</tr>
<tr>
<td>T1CON</td>
<td>TMR1CS0</td>
<td>TMR1CS0</td>
<td>T1CKPS0</td>
<td>T1CKPS0</td>
<td>T1OSCEN</td>
<td>T1SYNC</td>
<td>—</td>
<td>TMR1ON</td>
<td></td>
</tr>
<tr>
<td>TRISA</td>
<td>—</td>
<td>—</td>
<td>TRISA5</td>
<td>TRISA4</td>
<td>TRISA3</td>
<td>TRISA2</td>
<td>TRISA1</td>
<td>TRISA0</td>
<td></td>
</tr>
</tbody>
</table>

Legend: — = Unimplemented locations, read as ‘0’. Shaded cells are not used by the CPS module.
28.0 RF TRANSMITTER

The RF transmitter is an ultra low-power, integrated multi-band Sub-GHz transmitter. It is capable of operating in the 310, 433, 868, and 915 MHz license-free frequency bands using Frequency Shift Keying (FSK) or On-Off Keying (OOK) modulation of an input data stream.

28.1 Circuit Description

The RF transmitter block diagram is shown in Figure 28-1 and the I/O pin definitions are shown in Table 28-1.
The RF transmitter contains a sigma-delta fractional-N Phase-Locked Loop (PLL) frequency synthesizer. Frequency Shift Keying (FSK) modulation is made inside the PLL bandwidth. On-Off Keying (OOK) modulation is made by turning on and off the Power Amplifier (PA).

The reference frequency is generated by an internal crystal oscillator. An external quartz crystal resonator is connected to the XTAL pin and Ground (VSSRF). The choice of crystal frequency depends on the frequency band of choice.

The RF transmitter can deliver 0 dBm or +10 dBm into a 50Ω load via the RF OUT pin. An external matching network is required for each power setting and frequency band for the best efficiency to the antenna.

### 28.2 Configuring the RF Transmitter

The CTRL and DATA pins are used to configure the RF transmitter for transmit frequency, output power, modulation, FSK frequency deviation, and Sleep time. Once configured, the DATA pin is used to encode transmit data.

#### 28.2.1 POWER-ON RESET (POR)

At power-on, the CTRL pin is sampled as shown in Figure 28-2 and depending on the CTRL pin logic level, the RF transmitter will enter one of two Power-on Reset (POR) values as shown in Table 28-3 and Table 28-4. To continue using the RF transmitter with these POR values, maintain the CTRL pin stable and at the powered-on logic level. With the DATA pin at logic ‘0’, the RF transmitter will enter Sleep mode.

#### 28.2.2 RF TRANSMITTER REGISTERS

RF transmitter has three registers: Application, Frequency, and Status. These are used to write and read configuration parameters related to transmit frequency, output power, modulation, FSK frequency deviation, and Sleep time. A summary of register values are shown in Table 28-2. A detailed explanation of Application register is shown in Table 28-3, Frequency register values in Table 28-4, and STATUS register in Table 28-5.

To access the registers, the DATA line is sampled at each low-to-high transition on the CTRL pin. A total of 24 transitions are required on the CTRL pin to successfully write or read a value in the registers. Register write and read operations are shown in Figure 28-3.

Writing and reading the RF transmitter registers should be done when the device is in Sleep mode. See Section 28.2.4 “Sleep Mode”.
In the event that spurious activity (for example MCU interrupt or Reset) or less than 24 clock cycles on the CTRL pin, a special sequence over the CTRL and DATA pins can be used to recover serial communications with the RF transmitter. The recover sequence is shown in Figure 28-4.

**FIGURE 28-3: REGISTER WRITE AND READ OPERATIONS**

![Figure 28-3: Register Write and Read Operations](image)

**FIGURE 28-4: RECOVERY SEQUENCE TIMING**

![Figure 28-4: Recovery Sequence Timing](image)

Note 1: Refer to Section 31.7 “RF Transmitter Electrical Specifications”.

Note 2: Exactly 24 clock cycles are required for proper configuration.
### TABLE 28-2: RF TRANSMITTER REGISTER SUMMARY

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Value</th>
<th>Setting</th>
<th>Power-on Reset</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>CTRL = 0</td>
<td>CTRL = 1</td>
</tr>
<tr>
<td>23</td>
<td></td>
<td>0</td>
<td>Automatic</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>22</td>
<td></td>
<td>0</td>
<td>Manual</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>21</td>
<td></td>
<td>0</td>
<td>FSK</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>20</td>
<td></td>
<td>0</td>
<td>OOK</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>19</td>
<td></td>
<td>0</td>
<td>310 - 450 MHz</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>18</td>
<td></td>
<td>0</td>
<td>860 - 870 MHz</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>17</td>
<td></td>
<td>0</td>
<td>—</td>
<td>—</td>
<td>0x06(1)</td>
</tr>
<tr>
<td>16</td>
<td></td>
<td>0</td>
<td>0 dBm</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>15</td>
<td></td>
<td>0</td>
<td>10 dBm</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>14</td>
<td></td>
<td>0</td>
<td>2 ms</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>13</td>
<td></td>
<td>0</td>
<td>20 ms</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>12</td>
<td>Reserved</td>
<td>100(2)</td>
<td></td>
<td>100</td>
<td>100</td>
</tr>
</tbody>
</table>

**Note 1:** Actual frequency deviation value dependant on crystal frequency.

**Note 2:** When writing to the Application register, DA<2:0> must be 0b100.

### TABLE 28-3: APPLICATION REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Value</th>
<th>Setting</th>
<th>Power-on Reset</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>CTRL = 0</td>
<td>CTRL = 1</td>
</tr>
<tr>
<td>DA15</td>
<td>Mode</td>
<td>0</td>
<td>Automatic</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>Manual</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>DA14</td>
<td>Modulation</td>
<td>0</td>
<td>FSK</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>OOK</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>DA13</td>
<td>Band</td>
<td>0</td>
<td>310 - 450 MHz</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>860 - 870 MHz</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>DA&lt;12:5&gt;</td>
<td>Frequency Deviation (fDEV)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>0x06(1)</td>
</tr>
<tr>
<td>DA4</td>
<td>Output Power</td>
<td>0</td>
<td>0 dBm</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>10 dBm</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>DA3</td>
<td>Transmitter Off Time (tOFFT)</td>
<td>0</td>
<td>2 ms</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>20 ms</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>DA&lt;2:0&gt;</td>
<td>Reserved</td>
<td>100(2)</td>
<td></td>
<td>100</td>
<td>100</td>
</tr>
</tbody>
</table>

**Note 1:** Actual frequency deviation value dependant on crystal frequency.

**Note 2:** When writing to the Application register, DA<2:0> must be 0b100.

### TABLE 28-4: FREQUENCY REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Value</th>
<th>Setting</th>
<th>Power-on Reset</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>CTRL = 0</td>
<td>CTRL = 1</td>
</tr>
<tr>
<td>DF&lt;18:0&gt;</td>
<td>Transmit Frequency (fRF)</td>
<td>—</td>
<td>—</td>
<td>0x42C1C(1)</td>
<td>Refer to Section 28.4 “Frequency Selection and Configuration”. When reading frequency, the Most Significant three bits are read from the STATUS register (see Table 28-5)</td>
</tr>
</tbody>
</table>

**Note 1:** Actual frequency value dependant on crystal frequency.
28.2.3 DATA TRANSMISSION

TABLE 28-5: STATUS REGISTER

<table>
<thead>
<tr>
<th>Bit</th>
<th>Name</th>
<th>Value</th>
<th>Setting</th>
<th>Power-on Reset</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>DV&lt;7:0&gt;</td>
<td>Chip Version</td>
<td>0x11</td>
<td>—</td>
<td>0x11 0x11 0x11</td>
<td>0x11 = Version A1</td>
</tr>
<tr>
<td>DS&lt;4:2&gt;</td>
<td>Reserved</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>DS1</td>
<td>TX Ready</td>
<td>0</td>
<td>Sleep</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>Transmitting</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>DS0</td>
<td>Reserved</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>DF&lt;18:16&gt;</td>
<td>Transmit Frequency (fRF)</td>
<td>—</td>
<td>—</td>
<td>0b100 0b100</td>
<td>Refer to Section 28.4 “Frequency Selection and Configuration”. When reading frequency, the Most Significant three bits are read from the STATUS register (see Table 28-5)</td>
</tr>
</tbody>
</table>

RF data is transmitted when the DATA pin is at a logic ‘1’ for greater than tWAKE as shown in Figure 28-5. The CTRL pin must remain stable (either logic ‘0’ or ‘1’). If the modulation mode is OOK, the transmitted signal is turned on and off by the DATA pin. If the modulation mode is FSK, the transmitted signal is frequency shifted by the DATA pin. The encoding of the transmitted signal is determined by the length of time the DATA pin is held logic ‘0’ or ‘1’.
28.2.4 SLEEP MODE

The RF transmitter will automatically enter Sleep mode when the DATA pin is a logic ‘0’ for greater than $t_{OFFT}$, as shown in Figure 28-5. $t_{OFFT}$ can be configured for 2 or 20 ms in the Application register (see Table 28-3).

**FIGURE 28-5: DATA PIN TRANSMIT TIMING DIAGRAM**

![Data Pin Transmit Timing Diagram](image)

*Note 1:* The CTRL pin must remain stable (logic ‘0’ or ‘1’).
28.2.5 MANUAL TRANSMIT MODE
The RF transmitter can continuously transmit by setting the Mode bit (DA15) to a logic ‘1’ in the Applications register (see Table 28-3). It will continuously transmit RF data presented on the DATA pin without automatically entering Sleep mode. To cease transmission the Mode bit is must be cleared (DA15 = ‘0’). Figure 28-6 shows the Manual Transmit mode timing.

FIGURE 28-6: MANUAL TRANSMIT MODE TIMING

28.3 Modulation Selection

28.3.1 ON-OFF KEYING (OOK)
OOK modulation can be configured by setting the modulation DA14 bit in the Application register (Table 28-3). Data is transmitted as stated in Section 28.2.3 “Data Transmission”.

28.3.2 FREQUENCY SHIFT KEYING (FSK)
FSK modulation can be configured by clearing the modulation DA14 bit in the Application register. Frequency Deviation ($f_{DEV}$) is configured by setting the DA<12:5> bits in the Application register. Data is transmitted as stated in Section 28.2.3 “Data Transmission”.

28.3.3 DIGITAL TRANSMISSION SYSTEM (DTS)
In the United States and Canada, digital modulation techniques are permitted (FCC Part 15.247 and RSS-210, respectively). The RF transmitter can be configured for DTS mode by selecting FSK and $f_{DEV} = 200$ kHz. Data encoding techniques, such as data whitening, may be needed to ensure that the minimum 6 dB bandwidth is at least 500 kHz.
28.4 Frequency Selection and Configuration

The RF transmitter is capable of generating many of the popular RF frequencies that are permitted within the Radio Regulations of the country the finished product will be sold. The RF frequency configuration is performed by determining which frequency band, selecting the crystal frequency, and setting the frequency value in the Frequency register DF<18:0>. If FSK modulation is used, the frequency deviation is set in the Application register DA<12:5>. See Section 28.2.2 “RF Transmitter Registers” for information on Configuration register settings.

### 28.4.1 BAND SELECTION

The Band bit DA13 in the Application register configures the RF transmitter for a range of frequencies for a given crystal frequency as shown in Table 28-6.

### 28.4.2 CRYSTAL SELECTION

Once the frequency band has been selected, the choice of crystal frequency is flexible provided the crystal meets the specifications summarized in Table 28-7, the boundaries of the Frequency register DF<18:0> are followed as shown in Figure 28-7, and RF transmit frequency error is acceptable (see Section 28.4.3 “Frequency Calculation”).

### TABLE 28-6: FREQUENCY BAND SELECTION

<table>
<thead>
<tr>
<th>Band Setting DA&lt;13&gt;</th>
<th>Frequency Band (fRF)</th>
<th>Crystal Frequency (fXTAL)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>310 - 450 MHz</td>
<td>22 MHz</td>
</tr>
<tr>
<td></td>
<td>312 - 450 MHz</td>
<td>24 MHz</td>
</tr>
<tr>
<td></td>
<td>338 - 450 MHz</td>
<td>26 MHz</td>
</tr>
<tr>
<td>1</td>
<td>863 - 870 MHz</td>
<td>22 MHz</td>
</tr>
<tr>
<td></td>
<td>902 - 924 MHz</td>
<td>26 MHz</td>
</tr>
<tr>
<td></td>
<td>863 - 870 MHz</td>
<td>24 MHz</td>
</tr>
<tr>
<td></td>
<td>902 - 928 MHz</td>
<td>26 MHz</td>
</tr>
</tbody>
</table>

### TABLE 28-7: CRYSTAL RESONATOR SPECIFICATIONS

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Description</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>FXTAL</td>
<td>Crystal Frequency</td>
<td>22</td>
<td>—</td>
<td>26</td>
<td>MHz</td>
</tr>
<tr>
<td>CL</td>
<td>Load Capacitance</td>
<td>—</td>
<td>15</td>
<td>—</td>
<td>pF</td>
</tr>
<tr>
<td>ESR</td>
<td>Equivalent Series Resistance</td>
<td>—</td>
<td>—</td>
<td>100</td>
<td>Ohms</td>
</tr>
</tbody>
</table>

The crystal frequency tolerance and frequency stability over the operating temperature range depends on the system frequency budget. Typically, the receiver crystal frequency tolerance, stability, and receiver bandwidth will have the greatest influence. For OOK modulation, the transmitted RF signal (fRF) should remain inside the receiver bandwidth, otherwise signal degradation will occur. For FSK modulation, fRF should remain inside the receiver bandwidth and within 0.5 * fDEV.

As a general practice, do not choose a RF transmit signal (fRF) with an integer or near integer multiple of fXTAL. This will result in higher noise and spurious emissions.

### 28.4.3 FREQUENCY CALCULATION

Once the frequency band and crystal frequency are selected, the RF transmit signal (fRF) is calculated by setting the Frequency register DF(18:0) bits according to the formula shown in Figure 28-7. If the calculated value for DF(18:0) is not an integer, there will be an associated transmit frequency error. Ensure that this error is within the acceptable system frequency budget. Similarly, the frequency deviation is calculated as shown in Figure 28-7.
FIGURE 28-7: FREQUENCY CALCULATION

<table>
<thead>
<tr>
<th>Band 0</th>
<th>Band 1</th>
</tr>
</thead>
<tbody>
<tr>
<td>DF(18:0) = ( \frac{f_{RF} \times 16384}{f_{XTAL}} )</td>
<td>DF(18:0) = ( \frac{f_{RF} \times 8192}{f_{XTAL}} )</td>
</tr>
<tr>
<td>212992 &lt; DF(18:0) &lt; 344064</td>
<td>212992 &lt; DF(18:0) &lt; 344064</td>
</tr>
</tbody>
</table>

**Note:** Check \( f_{RF} \) frequency error by calculating \( f_{RF} \) with integer value of DF(18:0).

| DA(12:5) = \( \frac{f_{DEV} \times 16384}{f_{XTAL}} \) | DA(12:5) = \( \frac{f_{DEV} \times 8192}{f_{XTAL}} \) |
| 10 kHz \( \leq f_{DEV} \leq 200 \text{ kHz} \) | 10 kHz \( \leq f_{DEV} \leq 200 \text{ kHz} \) |

**Note:** Check \( f_{DEV} \) frequency error by calculating \( f_{DEV} \) with integer value of DA(12:5).

\( f_{RF} \) and \( f_{XTAL} \) values in the range shown in Table 9-6.
28.5 Applications

28.5.1 SOFTWARE INITIALIZATION

EXAMPLE 28-1: SAMPLE INITIALIZATION CODE

```c
#define APP_REG_PREFIX 0
#define FREQ_REG_PREFIX 0x18

void sendTxCommand(unsigned char cmd)
{
    // The 'T39A samples data on the rising edge of clock. Clock is idle low.
    unsigned char i;
    for (i=0; i<8; i++)
    {
        if (cmd & 0x80)
            DATA_OUT = 1;
        else
            DATA_OUT = 0;

        CTRL_OUT = 1;
        NOP();
        NOP();
        CTRL_OUT = 0;
        cmd = cmd << 1;
    }
}

void TX_Init(void)
{
    unsigned char app_high = (T39A_APP_CONFIG & 0x00FF00) >> 8;
    unsigned char app_low  = (T39A_APP_CONFIG & 0x0000FF);
    unsigned char f_upper  = (T39A_FREQ_CONFIG & 0x70000) >> 16;
    unsigned char f_high   = (T39A_FREQ_CONFIG & 0xFF000) >> 8;
    unsigned char f_low    = (T39A_FREQ_CONFIG & 0x000FF);

    sendTxCommand(APP_REG_PREFIX);
    sendTxCommand(app_high);
    sendTxCommand(app_low);
    sendTxCommand(FREQ_REG_PREFIX | f_upper);
    sendTxCommand(f_high);
    sendTxCommand(f_low);
    return;
}
```

28.5.2 APPLICATION CIRCUIT

Figure 28-8 describes a sample four-button remote transmitter application schematic. Table 28-8 contains its bill of materials. This schematic and bill of materials is a design suggestion only. Actual component values will be dependent on implementation parameters.
### TABLE 28-8: BILL OF MATERIALS

<table>
<thead>
<tr>
<th>Designator</th>
<th>Value</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>Common</td>
<td></td>
<td></td>
</tr>
<tr>
<td>U1</td>
<td>PIC18FXXXX</td>
<td>Microcontroller with integrated UHF transmitter</td>
</tr>
<tr>
<td>C6, C7</td>
<td>0.1 µF</td>
<td>Decoupling</td>
</tr>
<tr>
<td>R6</td>
<td>470 Ω</td>
<td>Current limiting</td>
</tr>
<tr>
<td>DS1</td>
<td>RED</td>
<td>LED</td>
</tr>
<tr>
<td>R3</td>
<td>10 kΩ</td>
<td>Weak pull-down for RF configuration</td>
</tr>
<tr>
<td>R4</td>
<td>100 Ω</td>
<td></td>
</tr>
<tr>
<td>R1, R5</td>
<td>47 kΩ</td>
<td>Voltage divider</td>
</tr>
<tr>
<td>C4</td>
<td>1000 pF</td>
<td></td>
</tr>
<tr>
<td>L5</td>
<td>120 nH</td>
<td></td>
</tr>
<tr>
<td>C5</td>
<td>100 pF</td>
<td></td>
</tr>
<tr>
<td>C3, L1, L3</td>
<td>0Ω</td>
<td>Matching to 50 Ω</td>
</tr>
<tr>
<td>L4</td>
<td>39 nH</td>
<td></td>
</tr>
<tr>
<td>C2</td>
<td>6.8 pF</td>
<td></td>
</tr>
<tr>
<td>L2</td>
<td>2.2 nH</td>
<td></td>
</tr>
<tr>
<td>X1</td>
<td>24 MHz</td>
<td></td>
</tr>
<tr>
<td>434 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td>L5</td>
<td>12 nH</td>
<td></td>
</tr>
<tr>
<td>C5</td>
<td>1 pF</td>
<td></td>
</tr>
<tr>
<td>C3, L4</td>
<td>DNP</td>
<td></td>
</tr>
<tr>
<td>L3</td>
<td>27 nH</td>
<td>Matching to 50 Ω</td>
</tr>
<tr>
<td>C2</td>
<td>2.7 pF</td>
<td></td>
</tr>
<tr>
<td>L1, L2</td>
<td>0Ω</td>
<td></td>
</tr>
<tr>
<td>X1</td>
<td>26 MHz</td>
<td></td>
</tr>
<tr>
<td>868 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td>L5</td>
<td>8.2 nH</td>
<td></td>
</tr>
<tr>
<td>L1, L4, C2</td>
<td>0Ω</td>
<td>Matching to 50 Ω</td>
</tr>
<tr>
<td>C5</td>
<td>4.7 pF</td>
<td></td>
</tr>
<tr>
<td>C3</td>
<td>1.2 pF</td>
<td></td>
</tr>
<tr>
<td>L3</td>
<td>2.4 nH</td>
<td></td>
</tr>
<tr>
<td>L2</td>
<td>10 nH</td>
<td></td>
</tr>
<tr>
<td>X1</td>
<td>26 MHz</td>
<td></td>
</tr>
<tr>
<td>915 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td>L5</td>
<td>8.2 nH</td>
<td></td>
</tr>
<tr>
<td>L1, L4, C2</td>
<td>0Ω</td>
<td></td>
</tr>
<tr>
<td>C5</td>
<td>4.7 pF</td>
<td></td>
</tr>
<tr>
<td>C3</td>
<td>1.2 pF</td>
<td></td>
</tr>
<tr>
<td>L3</td>
<td>2.4 nH</td>
<td></td>
</tr>
<tr>
<td>L2</td>
<td>10 nH</td>
<td></td>
</tr>
<tr>
<td>X1</td>
<td>26 MHz</td>
<td></td>
</tr>
</tbody>
</table>
29.0  IN-CIRCUIT SERIAL PROGRAMMING™ (ICSP™)

ICSP™ programming allows customers to manufacture circuit boards with unprogrammed devices. Programming can be done after the assembly process allowing the device to be programmed with the most recent firmware or a custom firmware. Five pins are needed for ICSP™ programming:

- ICSPCLK
- ICSPDAT
- MCLR/VPP
- VDD
- VSS

In Program/Verify mode the program memory, user IDs and the Configuration Words are programmed through serial communications. The ICSPDAT pin is a bidirectional I/O used for transferring the serial data and the ICSPCLK pin is the clock input. For more information on ICSP™ refer to the "PIC12LF1840T48A/T39A Programming Specification" (DS41595).

29.1  High-Voltage Programming Entry Mode

The device is placed into High-Voltage Programming Entry mode by holding the ICSPCLK and ICSPDAT pins low then raising the voltage on MCLR/VPP to Vihh. Some programmers produce VPP greater than Vihh (9.0V), an external circuit is required to limit the VPP voltage. See Figure 29-1 for example circuit.

FIGURE 29-1:  VPP LIMITER EXAMPLE CIRCUIT

![VPP Limiter Example Circuit](image)

Note:  The MPLAB® ICD 2 produces a VPP voltage greater than the maximum VPP specification of the PIC12LF1840T39A.
29.2 Low-Voltage Programming Entry Mode

The Low-Voltage Programming Entry mode allows the PIC12LF1840T39A device to be programmed using VDD only, without high voltage. When the LVP bit of Configuration Word 2 is set to ‘1’, the low-voltage ICSP programming entry is enabled. To disable the Low-Voltage ICSP mode, the LVP bit must be programmed to ‘0’.

Entry into the Low-Voltage Programming Entry mode requires the following steps:
1. MCLR is brought to VIL.
2. A 32-bit key sequence is presented on ICSPDAT, while clocking ICSPCLK.

Once the key sequence is complete, MCLR must be held at VIL for as long as Program/Verify mode is to be maintained.

If low-voltage programming is enabled (LVP = 1), the MCLR Reset function is automatically enabled and cannot be disabled. See Section 7.3 “MCLR” for more information.

The LVP bit can only be reprogrammed to ‘0’ by using the High-Voltage Programming mode.

29.3 Common Programming Interfaces

Connection to a target device is typically done through an ICSP™ header. A commonly found connector on development tools is the RJ-11 in the 6P6C (6 pin, 6 connector) configuration. See Figure 29-2.

Another connector often found in use with the PICkit™ programmers is a standard 6-pin header with 0.1 inch spacing. Refer to Figure 29-3.
For additional interface recommendations, refer to your specific device programmer manual prior to PCB design.

It is recommended that isolation devices be used to separate the programming pins from other circuitry. The type of isolation is highly dependent on the specific application and may include devices such as resistors, diodes, or even jumpers. See Figure 29-4 for more information.

**FIGURE 29-4: TYPICAL CONNECTION FOR ICSP™ PROGRAMMING**

![Diagram](image)

* Isolation devices (as required).
30.0 INSTRUCTION SET SUMMARY

Each PIC16 instruction is a 14-bit word containing the operation code (opcode) and all required operands. The opcodes are broken into three broad categories.

• Byte Oriented
• Bit Oriented
• Literal and Control

The literal and control category contains the most varied instruction word format.

Table 30-3 lists the instructions recognized by the MPASM™ assembler.

All instructions are executed within a single instruction cycle, with the following exceptions, which may take two or three cycles:

• Subroutine takes two cycles (CALL, CALLW)
• Returns from interrupts or subroutines take two cycles (RETURN, RETLW, RETFIE)
• Program branching takes two cycles (GOTO, BRA, BRW, BTFSS, BTFSC, DECFSZ, INCFSZ)
• One additional instruction cycle will be used when any instruction references an indirect file register and the file select register is pointing to program memory.

One instruction cycle consists of four oscillator cycles; for an oscillator frequency of 4 MHz, this gives a nominal instruction execution rate of 1 MHz.

All instruction examples use the format ‘0xhh’ to represent a hexadecimal number, where ‘h’ signifies a hexadecimal digit.

30.1 Read-Modify-Write Operations

Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (R-M-W) operation. The register is read, the data is modified, and the result is stored according to either the instruction, or the destination designator ‘d’. A read operation is performed on a register even if the instruction writes to that register.

TABLE 30-1: OPCODE FIELD DESCRIPTIONS

<table>
<thead>
<tr>
<th>Field</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>f</td>
<td>Register file address (0x00 to 0x7F)</td>
</tr>
<tr>
<td>W</td>
<td>Working register (accumulator)</td>
</tr>
<tr>
<td>b</td>
<td>Bit address within an 8-bit file register</td>
</tr>
<tr>
<td>k</td>
<td>Literal field, constant data or label</td>
</tr>
<tr>
<td>x</td>
<td>Don’t care location (= 0 or 1). The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.</td>
</tr>
<tr>
<td>d</td>
<td>Destination select: d = 0: store result in W, d = 1: store result in file register f. Default is d = 1.</td>
</tr>
<tr>
<td>n</td>
<td>FSR or INDF number. (0-1)</td>
</tr>
<tr>
<td>mm</td>
<td>Pre-post increment-decrement mode selection</td>
</tr>
</tbody>
</table>

TABLE 30-2: ABBREVIATION DESCRIPTIONS

<table>
<thead>
<tr>
<th>Field</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>PC</td>
<td>Program Counter</td>
</tr>
<tr>
<td>TO</td>
<td>Time-out bit</td>
</tr>
<tr>
<td>C</td>
<td>Carry bit</td>
</tr>
<tr>
<td>DC</td>
<td>Digit carry bit</td>
</tr>
<tr>
<td>Z</td>
<td>Zero bit</td>
</tr>
<tr>
<td>PD</td>
<td>Power-down bit</td>
</tr>
</tbody>
</table>
FIGURE 30-1: GENERAL FORMAT FOR INSTRUCTIONS

<table>
<thead>
<tr>
<th>Byte-oriented file register operations</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  8  7  6  0</td>
</tr>
<tr>
<td>OPCODE                  d  f (FILE #)</td>
</tr>
<tr>
<td>d = 0 for destination W</td>
</tr>
<tr>
<td>d = 1 for destination f</td>
</tr>
<tr>
<td>f = 7-bit file register address</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit-oriented file register operations</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  10  9  7  6  0</td>
</tr>
<tr>
<td>OPCODE                  b (BIT #)  f (FILE #)</td>
</tr>
<tr>
<td>b = 3-bit bit address</td>
</tr>
<tr>
<td>f = 7-bit file register address</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Literal and control operations</th>
</tr>
</thead>
<tbody>
<tr>
<td>General</td>
</tr>
<tr>
<td>13  8  7  0</td>
</tr>
<tr>
<td>OPCODE                 k (literal)</td>
</tr>
<tr>
<td>k = 8-bit immediate value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>CALL and GOTO instructions only</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  11  10  0</td>
</tr>
<tr>
<td>OPCODE                 k (literal)</td>
</tr>
<tr>
<td>k = 11-bit immediate value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>MOVLP instruction only</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  7  6  0</td>
</tr>
<tr>
<td>OPCODE                 k (literal)</td>
</tr>
<tr>
<td>k = 7-bit immediate value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>MOVLB instruction only</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  5  4  0</td>
</tr>
<tr>
<td>OPCODE                 k (literal)</td>
</tr>
<tr>
<td>k = 5-bit immediate value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>BRA instruction only</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  9  8  0</td>
</tr>
<tr>
<td>OPCODE                 k (literal)</td>
</tr>
<tr>
<td>k = 9-bit immediate value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>FSR Offset instructions</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  7  6  5  0</td>
</tr>
<tr>
<td>OPCODE                   n  k (literal)</td>
</tr>
<tr>
<td>n = appropriate FSR</td>
</tr>
<tr>
<td>k = 6-bit immediate value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>FSR Increment instructions</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  3  2  1  0</td>
</tr>
<tr>
<td>OPCODE                  n  m (mode)</td>
</tr>
<tr>
<td>n = appropriate FSR</td>
</tr>
<tr>
<td>m = 2-bit mode value</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>OPCODE only</th>
</tr>
</thead>
<tbody>
<tr>
<td>13  0</td>
</tr>
<tr>
<td>OPCODE</td>
</tr>
</tbody>
</table>
### TABLE 30-3: PIC12LF1840T39A ENHANCED INSTRUCTION SET

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>1</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td><strong>BYTE-ORIENTED FILE REGISTER OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>ADDWF f, d</td>
<td>Add W and f</td>
<td>1</td>
<td>00 0111 dfff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>ADDWFC f, d</td>
<td>Add with Carry W and f</td>
<td>1</td>
<td>11 1101 dfff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>ANDWF f, d</td>
<td>AND W with f</td>
<td>1</td>
<td>00 0101 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>ASRF f, d</td>
<td>Arithmetic Right Shift</td>
<td>1</td>
<td>11 0111 dfff ffff</td>
<td>C, Z</td>
<td>2</td>
</tr>
<tr>
<td>LSLF f, d</td>
<td>Logical Left Shift</td>
<td>1</td>
<td>11 0101 dfff ffff</td>
<td>C, Z</td>
<td>2</td>
</tr>
<tr>
<td>LSRF f, d</td>
<td>Logical Right Shift</td>
<td>1</td>
<td>11 0110 dfff ffff</td>
<td>C, Z</td>
<td>2</td>
</tr>
<tr>
<td>CLRF f</td>
<td>Clear f</td>
<td>1</td>
<td>00 0001 1fff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>CLRW –</td>
<td>Clear W</td>
<td>1</td>
<td>00 0000 0000 00xx</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>COMF f, d</td>
<td>Complement f</td>
<td>1</td>
<td>00 1001 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>DECF f, d</td>
<td>Decrement f</td>
<td>1</td>
<td>00 0011 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>INCF f, d</td>
<td>Increment f</td>
<td>1</td>
<td>00 1010 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>IORWF f, d</td>
<td>Inclusive OR W with f</td>
<td>1</td>
<td>00 0100 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>MOVf f, d</td>
<td>Move f</td>
<td>1</td>
<td>00 1000 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>MOVWF f</td>
<td>Move W to f</td>
<td>1</td>
<td>00 0000 1fff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>RLf f</td>
<td>Rotate Left f through Carry</td>
<td>1</td>
<td>00 1100 dfff ffff</td>
<td>C</td>
<td>2</td>
</tr>
<tr>
<td>RRF f, d</td>
<td>Rotate Right f through Carry</td>
<td>1</td>
<td>00 1100 dfff ffff</td>
<td>C</td>
<td>2</td>
</tr>
<tr>
<td>SUBWF f</td>
<td>Subtract W from f</td>
<td>1</td>
<td>00 0010 dfff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>SUBWF f, d</td>
<td>Subtract with Borrow W from f</td>
<td>1</td>
<td>11 1011 dfff ffff</td>
<td>C, DC, Z</td>
<td>2</td>
</tr>
<tr>
<td>SWAPf f, d</td>
<td>Swap nibbles in f</td>
<td>1</td>
<td>00 1110 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>XORWF f, d</td>
<td>Exclusive OR W with f</td>
<td>1</td>
<td>00 0110 dfff ffff</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td><strong>BYTE ORIENTED SKIP OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>DECFSZ f, d</td>
<td>Decrement f, Skip if 0</td>
<td>1(2)</td>
<td>00 1011 dfff ffff</td>
<td>1, 2</td>
<td></td>
</tr>
<tr>
<td>INCFSZ f, d</td>
<td>Increment f, Skip if 0</td>
<td>1(2)</td>
<td>00 1111 dfff ffff</td>
<td>1, 2</td>
<td></td>
</tr>
<tr>
<td><strong>BIT-ORIENTED FILE REGISTER OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BCF f, b</td>
<td>Bit Clear f</td>
<td>1</td>
<td>01 00bb bfff ffff</td>
<td>2</td>
<td></td>
</tr>
<tr>
<td>BSF f, b</td>
<td>Bit Set f</td>
<td>1</td>
<td>01 01bb bfff ffff</td>
<td>2</td>
<td></td>
</tr>
<tr>
<td><strong>BIT-ORIENTED SKIP OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BTFSC f, b</td>
<td>Bit Test f, Skip if Clear</td>
<td>1(2)</td>
<td>01 10bb bfff ffff</td>
<td>1, 2</td>
<td></td>
</tr>
<tr>
<td>BTFSS f, b</td>
<td>Bit Test f, Skip if Set</td>
<td>1(2)</td>
<td>01 11bb bfff ffff</td>
<td>1, 2</td>
<td></td>
</tr>
<tr>
<td><strong>LITERAL OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>ADDLW k</td>
<td>Add literal and W</td>
<td>1</td>
<td>11 1110 kkkk kkkk</td>
<td>C, DC, Z</td>
<td></td>
</tr>
<tr>
<td>ANDLW k</td>
<td>AND literal with W</td>
<td>1</td>
<td>11 1001 kkkk kkkk</td>
<td>Z</td>
<td></td>
</tr>
<tr>
<td>IORLW k</td>
<td>Inclusive OR literal with W</td>
<td>1</td>
<td>11 1000 kkkk kkkk</td>
<td>Z</td>
<td></td>
</tr>
<tr>
<td>MOVLB k</td>
<td>Move literal to BSR</td>
<td>1</td>
<td>00 0000 001k kkkk</td>
<td></td>
<td></td>
</tr>
<tr>
<td>MOVLB k</td>
<td>Move literal to PCLATH</td>
<td>1</td>
<td>11 0001 1kkk kkkk</td>
<td></td>
<td></td>
</tr>
<tr>
<td>MOVLW k</td>
<td>Move literal to W</td>
<td>1</td>
<td>11 0000 kkkk kkkk</td>
<td></td>
<td></td>
</tr>
<tr>
<td>SUBLW k</td>
<td>Subtract W from literal</td>
<td>1</td>
<td>11 1100 kkkk kkkk</td>
<td>C, DC, Z</td>
<td></td>
</tr>
<tr>
<td>XORLW k</td>
<td>Exclusive OR literal with W</td>
<td>1</td>
<td>11 1010 kkkk kkkk</td>
<td>Z</td>
<td></td>
</tr>
</tbody>
</table>

**Note 1:** If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOOP.

**Note 2:** If this instruction addresses an INDF register and the MSb of the corresponding FSR is set, this instruction will require one additional instruction cycle.
### TABLE 30-3: PIC12LF1840T39A ENHANCED INSTRUCTION SET (CONTINUED)

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>14-Bit Opcode</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>MSb</td>
<td>LSb</td>
<td></td>
</tr>
<tr>
<td><strong>CONTROL OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BRA k</td>
<td>Relative Branch</td>
<td>2</td>
<td>11</td>
<td>001k</td>
<td>kkkk</td>
</tr>
<tr>
<td>BRW –</td>
<td>Relative Branch with W</td>
<td>2</td>
<td>00</td>
<td>0000</td>
<td>0000</td>
</tr>
<tr>
<td>CALL k</td>
<td>Call Subroutine</td>
<td>2</td>
<td>10</td>
<td>0kkk</td>
<td>kkkk</td>
</tr>
<tr>
<td>CALLW –</td>
<td>Call Subroutine with W</td>
<td>2</td>
<td>00</td>
<td>0000</td>
<td>0000</td>
</tr>
<tr>
<td>GOTO k</td>
<td>Go to address</td>
<td>2</td>
<td>10</td>
<td>1kkk</td>
<td>kkkk</td>
</tr>
<tr>
<td>RETFIE k</td>
<td>Return from interrupt</td>
<td>2</td>
<td>00</td>
<td>0000</td>
<td>0000</td>
</tr>
<tr>
<td>RETLW k</td>
<td>Return with literal in W</td>
<td>2</td>
<td>11</td>
<td>0100</td>
<td>kkkk</td>
</tr>
<tr>
<td>RETURN –</td>
<td>Return from Subroutine</td>
<td>2</td>
<td>00</td>
<td>0000</td>
<td>0000</td>
</tr>
<tr>
<td><strong>INHERENT OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>CLRWD –</td>
<td>Clear Watchdog Timer</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0110</td>
</tr>
<tr>
<td>NOP –</td>
<td>No Operation</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0000</td>
</tr>
<tr>
<td>OPTION –</td>
<td>Load OPTION_REG register with W</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0110</td>
</tr>
<tr>
<td>RESET –</td>
<td>Software device Reset</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0000</td>
</tr>
<tr>
<td>SLEEP –</td>
<td>Go into Standby mode</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0110</td>
</tr>
<tr>
<td>TRIS f</td>
<td>Load TRIS register with W</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0110</td>
</tr>
<tr>
<td><strong>C-COMpiler OPTIMIZED</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>ADDFSR n, k</td>
<td>Add Literal k to FSRn</td>
<td>1</td>
<td>11</td>
<td>0001</td>
<td>0nkk</td>
</tr>
<tr>
<td>MOVIW n mm</td>
<td>Move Indirect FSRn to W with pre/post inc/dec modifier, mm</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0001</td>
</tr>
<tr>
<td>k[n]</td>
<td>Move INDFn to W, Indexed Indirect.</td>
<td>1</td>
<td>11</td>
<td>1111</td>
<td>0nkk</td>
</tr>
<tr>
<td>MOVWI n mm</td>
<td>Move W to Indirect FSRn with pre/post inc/dec modifier, mm</td>
<td>1</td>
<td>00</td>
<td>0000</td>
<td>0001</td>
</tr>
<tr>
<td>k[n]</td>
<td>Move W to INDFn, Indexed Indirect.</td>
<td>1</td>
<td>11</td>
<td>1111</td>
<td>1nkk</td>
</tr>
</tbody>
</table>

**Note 1:** If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

**Note 2:** If this instruction addresses an INDF register and the MSb of the corresponding FSR is set, this instruction will require one additional instruction cycle.

**Note 3:** See Table in the MOVIW and MOVWI instruction descriptions.
### 30.2 Instruction Descriptions

#### ADDFSR
**Add Literal to FSRn**

**Syntax:** 
\[
\text{[label]} \text{ ADDFSR } \text{ FSRn, k}
\]

**Operands:** 
\[-32 \leq k \leq 31\]
\[n \in \{ 0, 1 \}\]

**Operation:** 
\[\text{FSR}(n) + k \rightarrow \text{FSR}(n)\]

**Status Affected:** None

**Description:**
The signed 6-bit literal \(k\) is added to the contents of the FSR\(nH:FSRnL\) register pair. FSR\(n\) is limited to the range 0000h - FFFFh. Moving beyond these bounds will cause the FSR to wrap-around.

#### ADDLW
**Add literal and W**

**Syntax:** 
\[
\text{[label]} \text{ ADDLW } k
\]

**Operands:** 
\[0 \leq k \leq 255\]

**Operation:** 
\[(W) + k \rightarrow (W)\]

**Status Affected:** C, DC, Z

**Description:**
The contents of the W register are added to the 8-bit literal \(k\) and the result is placed in the W register.

#### ADDWF
**Add W and f**

**Syntax:** 
\[
\text{[label]} \text{ ADDWF } f,d
\]

**Operands:** 
\[0 \leq f \leq 127\]
\[d \in \{ 0, 1 \}\]

**Operation:** 
\[(W) + (f) \rightarrow (destination)\]

**Status Affected:** C, DC, Z

**Description:**
Add the contents of the W register with register \(f\). If \(d\) is \(0\), the result is stored in the W register. If \(d\) is \(1\), the result is stored back in register \(f\).

#### ADDWFC
**ADD W and CARRY bit to f**

**Syntax:** 
\[
\text{[label]} \text{ ADDWFC } f,d
\]

**Operands:** 
\[0 \leq f \leq 127\]
\[d \in \{ 0, 1 \}\]

**Operation:** 
\[(W) + (f) + (C) \rightarrow \text{dest}\]

**Status Affected:** C, DC, Z

**Description:**
Add W, the Carry flag and data memory location \(f\). If \(d\) is \(0\), the result is placed in W. If \(d\) is \(1\), the result is placed in data memory location \(f\).

#### ANDLW
**AND literal with W**

**Syntax:** 
\[
\text{[label]} \text{ ANDLW } k
\]

**Operands:** 
\[0 \leq k \leq 255\]

**Operation:** 
\[(W) \cdot \text{AND} \cdot (k) \rightarrow (W)\]

**Status Affected:** Z

**Description:**
The contents of W register are AND’ed with the 8-bit literal \(k\). The result is placed in the W register.

#### ANDWF
**AND W with f**

**Syntax:** 
\[
\text{[label]} \text{ ANDWF } f,d
\]

**Operands:** 
\[0 \leq f \leq 127\]
\[d \in \{ 0, 1 \}\]

**Operation:** 
\[(W) \cdot \text{AND} \cdot (f) \rightarrow (destination)\]

**Status Affected:** Z

**Description:**
AND the W register with register \(f\). If \(d\) is \(0\), the result is stored in the W register. If \(d\) is \(1\), the result is stored back in register \(f\).

#### ASRF
**Arithmetic Right Shift**

**Syntax:** 
\[
\text{[label]} \text{ ASRF } f,d
\]

**Operands:** 
\[0 \leq f \leq 127\]
\[d \in \{ 0, 1 \}\]

**Operation:** 
\[
\begin{align*}
(f<7>) & \rightarrow \text{dest}<7> \\
(f<7:1>) & \rightarrow \text{dest}<6:0> \\
(f<0>) & \rightarrow C,
\end{align*}
\]

**Status Affected:** C, Z

**Description:**
The contents of register \(f\) are shifted one bit to the right through the Carry flag. The MSb remains unchanged. If \(d\) is \(0\), the result is placed in W. If \(d\) is \(1\), the result is stored back in register \(f\).
### BCF  Bit Clear f

**Syntax:**  
\[
\text{[ label]} \text{ BCF } f,b
\]

**Operands:**  
\[
0 \leq f \leq 127  \\
0 \leq b \leq 7
\]

**Operation:**  
\[
0 \rightarrow (f<b>)
\]

**Status Affected:**  
None

**Description:**  
Bit ‘b’ in register ‘f’ is cleared.

---

### BRA  Relative Branch

**Syntax:**  
\[
\text{[ label]} \text{ BRA } label  \\
\text{[ label]} \text{ BRA } $+k$
\]

**Operands:**  
\[
-256 \leq \text{label} - \text{PC} + 1 \leq 255  \\
-256 \leq k \leq 255
\]

**Operation:**  
\[
(\text{PC}) + 1 + k \rightarrow \text{PC}
\]

**Status Affected:**  
None

**Description:**  
Add the signed 9-bit literal ‘k’ to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + k. This instruction is a 2-cycle instruction. This branch has a limited range.

---

### BRW  Relative Branch with W

**Syntax:**  
\[
\text{[ label]} \text{ BRW}
\]

**Operands:**  
None

**Operation:**  
\[
(\text{PC}) + (W) \rightarrow \text{PC}
\]

**Status Affected:**  
None

**Description:**  
Add the contents of W (unsigned) to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + (W). This instruction is a 2-cycle instruction.

---

### BSF  Bit Set f

**Syntax:**  
\[
\text{[ label]} \text{ BSF } f,b
\]

**Operands:**  
\[
0 \leq f \leq 127  \\
0 \leq b \leq 7
\]

**Operation:**  
\[
1 \rightarrow (f<b>)
\]

**Status Affected:**  
None

**Description:**  
Bit ‘b’ in register ‘f’ is set.

---

### BTFSC  Bit Test f, Skip if Clear

**Syntax:**  
\[
\text{[ label]} \text{ BTFSC } f,b
\]

**Operands:**  
\[
0 \leq f \leq 127  \\
0 \leq b \leq 7
\]

**Operation:**  
\[
\text{skip if } (f < b) = 0
\]

**Status Affected:**  
None

**Description:**  
If bit ‘b’ in register ‘f’ is ‘1’, the next instruction is executed. If bit ‘b’, in register ‘f’, is ‘0’, the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.

---

### BTFSS  Bit Test f, Skip if Set

**Syntax:**  
\[
\text{[ label]} \text{ BTFSS } f,b
\]

**Operands:**  
\[
0 \leq f \leq 127  \\
0 \leq b < 7
\]

**Operation:**  
\[
\text{skip if } (f < b) = 1
\]

**Status Affected:**  
None

**Description:**  
If bit ‘b’ in register ‘f’ is ‘0’, the next instruction is executed. If bit ‘b’ is ‘1’, then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.
CALL Call Subroutine
Syntax: \[
\text{[label]} \text{ CALL } k
\]
Operands: \(0 \leq k \leq 2047\)
Operation: \((\text{PC} + 1) \rightarrow \text{TOS},\)
\(k \rightarrow \text{PC}<10:0>,\)
\((\text{PCLATH}<6:3>) \rightarrow \text{PC}<14:11>\)
Status Affected: None
Description: Call Subroutine. First, return address \((\text{PC} + 1)\) is pushed onto the stack. The 11-bit immediate address is loaded into \(\text{PC}<10:0>\). The upper bits of the \(\text{PC}\) are loaded from \(\text{PCLATH}\). \text{CALL} is a 2-cycle instruction.

CALLW Subroutine Call With W
Syntax: \[
\text{[label]} \text{ CALLW}
\]
Operands: None
Operation: \((\text{PC} + 1) \rightarrow \text{TOS},\)
\((W) \rightarrow \text{PC}<7:0>,\)
\((\text{PCLATH}<6:0>) \rightarrow \text{PC}<14:8>\)
Status Affected: None
Description: Subroutine call with \(W\). First, the return address \((\text{PC} + 1)\) is pushed onto the return stack. Then, the contents of \(W\) is loaded into \(\text{PC}<7:0>\), and the contents of \(\text{PCLATH}\) into \(\text{PC}<14:8>\). \text{CALLW} is a 2-cycle instruction.

CLRF Clear f
Syntax: \[
\text{[label]} \text{ CLRF } f
\]
Operands: \(0 \leq f \leq 127\)
Operation: \(00h \rightarrow (f),\)
\(1 \rightarrow Z\)
Status Affected: \(Z\)
Description: The contents of register ‘\(f\)’ are cleared and the \(Z\) bit is set.

CLRWDT Clear Watchdog Timer
Syntax: \[
\text{[label]} \text{ CLRWDT}
\]
Operands: None
Operation: \(00h \rightarrow \text{WDT}\)
\(0 \rightarrow \text{WDT prescaler},\)
\(1 \rightarrow \text{TO}\)
\(1 \rightarrow \text{PD}\)
Status Affected: \(\text{TO}, \text{PD}\)
Description: \text{CLRWDT} instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. Status bits \(\text{TO}\) and \(\text{PD}\) are set.

COMF Complement f
Syntax: \[
\text{[label]} \text{ COMF } f,d
\]
Operands: \(0 \leq f \leq 127\)
\(d \in [0,1]\)
Operation: \((\text{f}) \rightarrow (\text{destination})\)
Status Affected: \(Z\)
Description: The contents of register ‘\(f\)’ are complemented. If ‘\(d\)’ is ‘0’, the result is stored in \(W\). If ‘\(d\)’ is ‘1’, the result is stored back in register ‘\(f\)’.

DECF Decrement f
Syntax: \[
\text{[label]} \text{ DECF } f,d
\]
Operands: \(0 \leq f \leq 127\)
\(d \in [0,1]\)
Operation: \((f) - 1 \rightarrow (\text{destination})\)
Status Affected: \(Z\)
Description: Decrement register ‘\(f\)’. If ‘\(d\)’ is ‘0’, the result is stored in the \(W\) register. If ‘\(d\)’ is ‘1’, the result is stored back in register ‘\(f\)’.

CLRW Clear W
Syntax: \[
\text{[label]} \text{ CLRW}
\]
Operands: None
Operation: \(00h \rightarrow (W),\)
\(1 \rightarrow Z\)
Status Affected: \(Z\)
Description: \(W\) register is cleared. Zero bit (\(Z\)) is set.
### DECFSZ  Decrement f, Skip if 0

**Syntax:**  
\[ \text{[ label]} \] DECFSZ f,d

**Operands:**  \( 0 \leq f \leq 127 \)  
\( d \in \{0,1\} \)

**Operation:**  
\((f) - 1 \rightarrow (\text{destination}); \)  
skip if result = 0

**Status Affected:** None

**Description:** The contents of register 'f' are decremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle instruction.

### INCFSZ  Increment f, Skip if 0

**Syntax:**  
\[ \text{[ label]} \] INCFSZ f,d

**Operands:**  \( 0 \leq f \leq 127 \)  
\( d \in \{0,1\} \)

**Operation:**  
\((f) + 1 \rightarrow (\text{destination}); \)  
skip if result = 0

**Status Affected:** None

**Description:** The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle instruction.

### GOTO  Unconditional Branch

**Syntax:**  
\[ \text{[ label]} \] GOTO k

**Operands:**  \( 0 \leq k \leq 2047 \)

**Operation:**  
\( k \rightarrow \text{PC<10:0>} \)  
\( \text{PCLATH<6:3>} \rightarrow \text{PC<14:11>} \)

**Status Affected:** None

**Description:** GOTO is an unconditional branch. The 11-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<6:3>. GOTO is a 2-cycle instruction.

### IORLW  Inclusive OR literal with W

**Syntax:**  
\[ \text{[ label]} \] IORLW k

**Operands:**  \( 0 \leq k \leq 255 \)

**Operation:**  
\( (W) \ .OR. k \rightarrow (W) \)

**Status Affected:** Z

**Description:** The contents of the W register are OR'ed with the 11-bit literal 'k'. The result is placed in the W register.

### INCF  Increment f

**Syntax:**  
\[ \text{[ label]} \] INCF f,d

**Operands:**  \( 0 \leq f \leq 127 \)  
\( d \in \{0,1\} \)

**Operation:**  
\((f) + 1 \rightarrow (\text{destination})\)

**Status Affected:** Z

**Description:** The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

### IORWF  Inclusive OR W with f

**Syntax:**  
\[ \text{[ label]} \] IORWF f,d

**Operands:**  \( 0 \leq f \leq 127 \)  
\( d \in \{0,1\} \)

**Operation:**  
\( (W) \ .OR. (f) \rightarrow (\text{destination}) \)

**Status Affected:** Z

**Description:** Inclusive OR the W register with register 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.
### LSLF Logical Left Shift

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>[ label ] LSLF f {,d}</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 127</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>(&lt;7&gt;) → C</td>
</tr>
<tr>
<td></td>
<td>(&lt;6:0&gt;) → dest&lt;7:1&gt;</td>
</tr>
<tr>
<td></td>
<td>0 → dest&lt;0&gt;</td>
</tr>
<tr>
<td>Status Affected</td>
<td>C, Z</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register 'f' are shifted one bit to the left through the Carry flag. A '0' is shifted into the LSb. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.</td>
</tr>
</tbody>
</table>

### LSRF Logical Right Shift

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>[ label ] LSRF f {,d}</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 127</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>0 → dest&lt;7&gt;</td>
</tr>
<tr>
<td></td>
<td>(&lt;7:1&gt;) → dest&lt;6:0&gt;,</td>
</tr>
<tr>
<td></td>
<td>(&lt;0&gt;) → C</td>
</tr>
<tr>
<td>Status Affected</td>
<td>C, Z</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register 'f' are shifted one bit to the right through the Carry flag. A '0' is shifted into the MSb. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.</td>
</tr>
</tbody>
</table>

### MOVF Move f

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>[ label ] MOVF f,d</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 127</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>(f) → (dest)</td>
</tr>
<tr>
<td>Status Affected</td>
<td>Z</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register f is moved to a destination dependent upon the status of d. If d = 0, destination is W register. If d = 1, the destination is file register f itself. d = 1 is useful to test a file register since status flag Z is affected.</td>
</tr>
</tbody>
</table>

Example: 

MOVF FSR, 0  

After Instruction  

W = value in FSR register  

Z = 1
MOVIW  Move INDFn to W

Syntax:  
[ label ] MOVIW ++FSRn  
[ label ] MOVIW --FSRn  
[ label ] MOVIW FSRn++  
[ label ] MOVIW FSRn--  
[ label ] MOVIW k[FSRn]

Operands:  
n ∈ [0, 1]  
mm ∈ [00, 01, 10, 11]  
-32 ≤ k ≤ 31

Operation:  
INDFn → W  
Effective address is determined by  
• FSR + 1 (preincrement)  
• FSR - 1 (predecrement)  
• FSR + k (relative offset)  
After the Move, the FSR value will be  
• FSR + 1 (all increments)  
• FSR - 1 (all decrements)  
• Unchanged

Status Affected:  
Z

Mode  Syntax  mm
Preincrement  ++FSRn  00
Predecrement  --FSRn  01
Postincrement  FSRn++  10
Postdecrement  FSRn--  11

Description:  
This instruction is used to move data  
between W and one of the indirect  
registers (INDFn). Before/after this  
move, the pointer (FSRn) is updated by  
pre/post incrementing/decrementing it.

Note:  The INDFn registers are not  
physical registers. Any instruction that  
accesses an INDFn register actually  
accesses the register at the address  
specified by the FSRn.

FSRn is limited to the range 0000h -  
FFFFFFh. Incrementing/decrementing it  
beyond these bounds will cause it to  
wrap-around.

MOVP  Move literal to PCLATH

Syntax:  
[ label ] MOVP k

Operands:  
0 ≤ k ≤ 127

Operation:  
k → PCLATH

Status Affected:  
None

Description:  
The 7-bit literal ‘k’ is loaded into the  
PCLATH register.

MOVLW  Move literal to W

Syntax:  
[ label ] MOVLW k

Operands:  
0 ≤ k ≤ 255

Operation:  
k → (W)

Status Affected:  
None

Description:  
The 8-bit literal ‘k’ is loaded into W  
register. The “don’t cares” will  
assemble as ‘0’s.

Words:  1
Cycles:  1

Example:  
MOVLW 0x5A

After Instruction  
W = 0x5A

MOVWF  Move W to f

Syntax:  
[ label ] MOVWF f

Operands:  
0 ≤ f ≤ 127

Operation:  
(W) → (f)

Status Affected:  
None

Description:  
Move data from W register to register  
f.

Words:  1
Cycles:  1

Example:  
MOVWF OPTION

Before Instruction  
OPTION = 0xFF  
W = 0x4F

After Instruction  
OPTION = 0x4F  
W = 0x4F

MOVLB  Move literal to BSR

Syntax:  
[ label ] MOVLB k

Operands:  
0 ≤ k ≤ 15

Operation:  
k → BSR

Status Affected:  
None

Description:  
The 5-bit literal ‘k’ is loaded into the  
Bank Select Register (BSR).
MOVWI Move W to INDFn

Syntax:
- \([ label ]\) MOVWI ++FSRn
- \([ label ]\) MOVWI --FSRn
- \([ label ]\) MOVWI FSRn++
- \([ label ]\) MOVWI FSRn--
- \([ label ]\) MOVWI k[FSRn]

Operands:
- \(n \in [0,1]\)
- \(mm \in [00,01,10,11]\)
- \(-32 \leq k \leq 31\)

Operation:
- \(W \rightarrow \text{INDFn}\)
Effector address is determined by
- \(\text{FSR} + 1\) (preincrement)
- \(\text{FSR} - 1\) (predecrement)
- \(\text{FSR} + k\) (relative offset)
After the Move, the FSR value will be either:
- \(\text{FSR} + 1\) (all increments)
- \(\text{FSR} - 1\) (all decrements)
- Unchanged

Status Affected: None

Description:
This instruction is used to move data between W and one of the indirect registers (INDFn). Before/after this move, the pointer (FSRn) is updated by pre/post incrementing/decrementing it.

Note: The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the FSRn.

FSRn is limited to the range 0000h - FFFFh. Incrementing/decrementing it beyond these bounds will cause it to wrap-around.

The increment/decrement operation on FSRn WILL NOT affect any Status bits.

NOP No Operation

Syntax: \([ label ]\) NOP

Operands: None

Operation: No operation

Status Affected: None

Description: No operation.

Words: 1

Cycles: 1

Example:

OPTON Load OPTION_REG Register with W

Syntax: \([ label ]\) OPTION

Operands: None

Operation: \(\text{(W)} \rightarrow \text{OPTION_REG}\)

Status Affected: None

Description: Move data from W register to OPTION_REG register.

RESET Software Reset

Syntax: \([ label ]\) RESET

Operands: None

Operation: Execute a device Reset. Resets the nRI flag of the PCON register.

Status Affected: None

Description: This instruction provides a way to execute a hardware Reset by software.
### RETFIE (Return from Interrupt)

**Syntax:**

\[
\text{[ label ] RETFIE } k
\]

**Operands:** None

**Operation:**

\[
\text{TOS} \rightarrow \text{PC, } 1 \rightarrow \text{GIE}
\]

**Status Affected:** None

**Description:**

Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a 2-cycle instruction.

**Words:** 1

**Cycles:** 2

**Example:**

```
After Interrupt
PC = TOS
GIE = 1
```

### RETLW (Return with literal in W)

**Syntax:**

\[
\text{[ label ] RETLW } k
\]

**Operands:**

\[0 \leq k \leq 255\]

**Operation:**

\[k \rightarrow (W); \text{TOS} \rightarrow \text{PC}\]

**Status Affected:** None

**Description:**

The W register is loaded with the 8-bit literal ‘k’. The program counter is loaded from the top of the stack (the return address). This is a 2-cycle instruction.

**Words:** 1

**Cycles:** 2

**Example:**

```
CALL TABLE;W contains table
;offset value
 ;W now has table value

TABLE

ADDWF PC ;W = offset
RETLW k1 ;Begin table
RETLW k2 ;

RETLW kn ; End of table
```

### RETURN (Return from Subroutine)

**Syntax:**

\[
\text{[ label ] RETURN}
\]

**Operands:** None

**Operation:**

\[\text{TOS} \rightarrow \text{PC}\]

**Status Affected:** None

**Description:**

Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a 2-cycle instruction.

**Words:** 1

**Cycles:** 2

### RLF (Rotate Left f through Carry)

**Syntax:**

\[
\text{[ label ] RLF } f,d
\]

**Operands:**

\[0 \leq f \leq 127\]

\[d \in [0,1]\]

**Operation:**

See description below

**Status Affected:** C

**Description:**

The contents of register ‘f’ are rotated one bit to the left through the Carry flag. If ‘d’ is ‘0’, the result is placed in the W register. If ‘d’ is ‘1’, the result is stored back in register ‘f’.

**Words:** 1

**Cycles:** 1

**Example:**

```
RLF REG1,0
```

Before Instruction

<table>
<thead>
<tr>
<th>REG1</th>
<th>C</th>
</tr>
</thead>
<tbody>
<tr>
<td>1110 0110</td>
<td>0</td>
</tr>
</tbody>
</table>

After Instruction

<table>
<thead>
<tr>
<th>REG1</th>
<th>W</th>
<th>C</th>
</tr>
</thead>
<tbody>
<tr>
<td>1110 0110</td>
<td>1100 1110</td>
<td>1</td>
</tr>
</tbody>
</table>
### RRF
**Rotate Right f through Carry**

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>[ label ] RRF f,d</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 127</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>See description below</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>C</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.</td>
</tr>
</tbody>
</table>

### SLEEP
**Enter Sleep mode**

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>[ label ] SLEEP</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>None</td>
</tr>
<tr>
<td>Operation:</td>
<td>00h → WDT,</td>
</tr>
<tr>
<td></td>
<td>0 → WDT prescaler,</td>
</tr>
<tr>
<td></td>
<td>1 → TO,</td>
</tr>
<tr>
<td></td>
<td>0 → PD</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>TO, PD</td>
</tr>
<tr>
<td>Description:</td>
<td>The power-down Status bit, PD is cleared. Time-out Status bit, TO is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.</td>
</tr>
</tbody>
</table>

### SUBLW
**Subtract W from literal**

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>[ label ] SUBLW k</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ k ≤ 255</td>
</tr>
<tr>
<td>Operation:</td>
<td>k - (W) → (W)</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>C, DC, Z</td>
</tr>
<tr>
<td>Description:</td>
<td>The W register is subtracted (2's complement method) from the 8-bit literal 'k'. The result is placed in the W register.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>C = 0</th>
<th>W &gt; k</th>
</tr>
</thead>
<tbody>
<tr>
<td>C = 1</td>
<td>W ≤ k</td>
</tr>
<tr>
<td>DC = 0</td>
<td>W&lt;3:0&gt; &gt; k&lt;3:0&gt;</td>
</tr>
<tr>
<td>DC = 1</td>
<td>W&lt;3:0&gt; ≤ k&lt;3:0&gt;</td>
</tr>
</tbody>
</table>

### SUBWF
**Subtract W from f**

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>[ label ] SUBWF f,d</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 127</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>(f) - (W) → (destination)</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>C, DC, Z</td>
</tr>
<tr>
<td>Description:</td>
<td>Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>C = 0</th>
<th>W &gt; f</th>
</tr>
</thead>
<tbody>
<tr>
<td>C = 1</td>
<td>W ≤ f</td>
</tr>
<tr>
<td>DC = 0</td>
<td>W&lt;3:0&gt; &gt; f&lt;3:0&gt;</td>
</tr>
<tr>
<td>DC = 1</td>
<td>W&lt;3:0&gt; ≤ f&lt;3:0&gt;</td>
</tr>
</tbody>
</table>

### SUBWFB
**Subtract W from f with Borrow**

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>SUBWFB f {,d}</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 127</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>(f) - (W) - (B) → dest</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>C, DC, Z</td>
</tr>
<tr>
<td>Description:</td>
<td>Subtract W and the BORROW flag (CARRY) from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>C = 0</th>
<th>W &gt; f</th>
</tr>
</thead>
<tbody>
<tr>
<td>C = 1</td>
<td>W ≤ f</td>
</tr>
<tr>
<td>DC = 0</td>
<td>W&lt;3:0&gt; &gt; f&lt;3:0&gt;</td>
</tr>
<tr>
<td>DC = 1</td>
<td>W&lt;3:0&gt; ≤ f&lt;3:0&gt;</td>
</tr>
</tbody>
</table>
### SWAPF

**Swap Nibbles in f**

<table>
<thead>
<tr>
<th>Syntax</th>
<th>SWAPF f,d</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Operands:</strong></td>
<td>0 ≤ f ≤ 127 &lt;br&gt; d ∈ [0, 1]</td>
</tr>
<tr>
<td><strong>Operation:</strong></td>
<td>(f&lt;3:0&gt;) → (destination&lt;7:4&gt;), &lt;br&gt;(f&lt;7:4&gt;) → (destination&lt;3:0&gt;)</td>
</tr>
<tr>
<td><strong>Status Affected:</strong></td>
<td>None</td>
</tr>
<tr>
<td><strong>Description:</strong></td>
<td>The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.</td>
</tr>
</tbody>
</table>

### XORLW

**Exclusive OR literal with W**

<table>
<thead>
<tr>
<th>Syntax</th>
<th>XORLW k</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Operands:</strong></td>
<td>0 ≤ k ≤ 255</td>
</tr>
<tr>
<td><strong>Operation:</strong></td>
<td>(W) .XOR. k → (W)</td>
</tr>
<tr>
<td><strong>Status Affected:</strong></td>
<td>Z</td>
</tr>
<tr>
<td><strong>Description:</strong></td>
<td>The contents of the W register are XOR'ed with the 8-bit literal 'k'. The result is placed in the W register.</td>
</tr>
</tbody>
</table>

### TRIS

**Load TRIS Register with W**

<table>
<thead>
<tr>
<th>Syntax</th>
<th>TRIS f</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Operands:</strong></td>
<td>5 ≤ f ≤ 7</td>
</tr>
<tr>
<td><strong>Operation:</strong></td>
<td>(W) → TRIS register 'f'</td>
</tr>
<tr>
<td><strong>Status Affected:</strong></td>
<td>None</td>
</tr>
<tr>
<td><strong>Description:</strong></td>
<td>Move data from W register to TRIS register. When 'f' = 5, TRISA is loaded. When 'f' = 6, TRISB is loaded. When 'f' = 7, TRISC is loaded.</td>
</tr>
</tbody>
</table>

### XORWF

**Exclusive OR W with f**

<table>
<thead>
<tr>
<th>Syntax</th>
<th>XORWF f,d</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Operands:</strong></td>
<td>0 ≤ f ≤ 127 &lt;br&gt; d ∈ [0, 1]</td>
</tr>
<tr>
<td><strong>Operation:</strong></td>
<td>(W) .XOR. (f) → (destination)</td>
</tr>
<tr>
<td><strong>Status Affected:</strong></td>
<td>Z</td>
</tr>
<tr>
<td><strong>Description:</strong></td>
<td>Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.</td>
</tr>
</tbody>
</table>
31.0 ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings(†)

Ambient temperature under bias....................................................................................................... -40°C to +125°C
Storage temperature ........................................................................................................................ -55°C to +150°C
Voltage on VDD with respect to Vss, PIC12LF1840T39A .................................................................. -0.3V to +4.0V
Voltage on MCLR with respect to Vss ................................................................................................. -0.3V to +9.0V
Voltage on all other pins with respect to VSS ........................................................................... -0.3V to (VDD + 0.3V)
Total power dissipation(†) ...............................................................................................................................800 mW
Maximum current out of VSS pin, -40°C ≤ TA ≤ +85°C for industrial............................................................... 170 mA
Maximum current into VDD pin, -40°C ≤ TA ≤ +85°C for industrial.................................................................. 170 mA
Clamp current, IK (VPIN < 0 or VPIN > VDD)...........................................................................................± 20 mA
Maximum output current sunk by any I/O pin.............................................................................................. 25 mA
Maximum output current sourced by any I/O pin ..........................................................................................25 mA

Note 1: Power dissipation is calculated as follows: PDIS = VDD x (IDD – Σ IOH) + Σ {(VDD – VOH) x IOH} + Σ(VOI x IOL).

† NOTICE: Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the
device. This is a stress rating only and functional operation of the device at those or any other conditions above those
indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for
extended periods may affect device reliability.
FIGURE 31-1: PIC12LF1840T39A VOLTAGE FREQUENCY GRAPH, -40°C ≤ TA ≤ -85°C

- Note 1: The shaded region indicates the permissible combinations of voltage and frequency.
- Note 2: Refer to Table 31-3 for each Oscillator mode’s supported frequencies.

FIGURE 31-2: HFINTOSC FREQUENCY ACCURACY OVER DEVICE VDD AND TEMPERATURE
# 31.1 DC Characteristics: PIC12LF1840T39A-I (Industrial)

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>D001</td>
<td>VDD</td>
<td>Supply Voltage (VDDMin, VDDMax)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>PIC12LF1840T39A</td>
<td>1.8</td>
<td>—</td>
<td>3.6</td>
<td>V</td>
<td>Fosc ≤ 16 MHz:</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>2.5</td>
<td>—</td>
<td>3.6</td>
<td>V</td>
<td>Fosc ≤ 32 MHz (Note 2)</td>
</tr>
<tr>
<td>D002*</td>
<td>VDR</td>
<td>RAM Data Retention Voltage(1)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>PIC12LF1840T39A</td>
<td>1.5</td>
<td>—</td>
<td>—</td>
<td>V</td>
<td>Device in Sleep mode</td>
</tr>
<tr>
<td>VPOR*</td>
<td></td>
<td>Power-on Reset Release Voltage</td>
<td>—</td>
<td>1.6</td>
<td>—</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>VPORR*</td>
<td></td>
<td>Power-on Reset Rearm Voltage</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>PIC12LF1840T39A</td>
<td>—</td>
<td>0.8</td>
<td>—</td>
<td>V</td>
<td>Device in Sleep mode</td>
</tr>
<tr>
<td>D003</td>
<td>VADFVR</td>
<td>Fixed Voltage Reference Voltage for ADC, Initial Accuracy</td>
<td>-8</td>
<td>6</td>
<td>6</td>
<td>%</td>
<td>1.024V, VDD ≥ 2.5V, 85°C (Note 3)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-8</td>
<td>6</td>
<td>6</td>
<td>%</td>
<td>2.048V, VDD ≥ 2.5V, 85°C</td>
</tr>
<tr>
<td>D003A</td>
<td>VCDAFVR</td>
<td>Fixed Voltage Reference Voltage for Comparator and DAC, Initial Accuracy</td>
<td>-11</td>
<td>7</td>
<td>7</td>
<td>%</td>
<td>1.024V, VDD ≥ 2.5V, 85°C</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>-11</td>
<td>7</td>
<td>7</td>
<td>%</td>
<td>2.048V, VDD ≥ 2.5V, 85°C</td>
</tr>
<tr>
<td>D003C*</td>
<td>TCVFVR</td>
<td>Temperature Coefficient, Fixed Voltage Reference</td>
<td>—</td>
<td>-114</td>
<td>—</td>
<td>ppm/°C</td>
<td></td>
</tr>
<tr>
<td>D003D*</td>
<td>ΔVFVR/ΔVIN</td>
<td>Line Regulation, Fixed Voltage Reference</td>
<td>—</td>
<td>0.225</td>
<td>—</td>
<td>%/V</td>
<td></td>
</tr>
<tr>
<td>D004*</td>
<td>SVDD</td>
<td>VDD Rise Rate to ensure internal Power-on Reset signal</td>
<td>0.05</td>
<td>—</td>
<td>—</td>
<td>V/ms</td>
<td>See Section 7.1 “Power-on Reset (POR)” for details.</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note**
1: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.
2: PLL required for 32 MHz operation.
3: For proper operation, the minimum value of the ADC positive voltage reference must be 1.8V or greater. When selecting the FVR or the VREF+ pin as the source of the ADC positive voltage reference, be aware that the voltage must be 1.8V or greater.
FIGURE 31-3: POR AND POR REARM WITH SLOW RISING VDD

Note 1: When NPOR is low, the device is held in Reset.
2: TPOR 1 μs typical.
3: TVLOW 2.7 μs typical.
# 31.2 DC Characteristics: PIC12LF1840T39A-I (Industrial)

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Device Characteristics</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
<th>VDD</th>
<th>Note</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D010</td>
<td>Supply Current (IDD)</td>
<td>5</td>
<td>12</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 32 kHz, -40°C to +85°C</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>7.4</td>
<td>25</td>
<td>µA</td>
<td>3.0</td>
<td>LP Oscillator mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D011</td>
<td></td>
<td>60</td>
<td>95</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 1 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>119</td>
<td>180</td>
<td>µA</td>
<td>3.0</td>
<td>XT Oscillator mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D012</td>
<td></td>
<td>165</td>
<td>240</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 4 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>309</td>
<td>430</td>
<td>µA</td>
<td>3.0</td>
<td>XT Oscillator mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D013</td>
<td></td>
<td>34</td>
<td>120</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 1 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>69</td>
<td>200</td>
<td>µA</td>
<td>3.0</td>
<td>EC Oscillator mode, Medium-Power mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D014</td>
<td></td>
<td>118</td>
<td>210</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 4 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>222</td>
<td>380</td>
<td>µA</td>
<td>3.0</td>
<td>EC Oscillator mode, Medium-Power mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D015</td>
<td>Supply Current (IDD)</td>
<td>6.5</td>
<td>20</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 31 kHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>9</td>
<td>31</td>
<td>µA</td>
<td>3.0</td>
<td>LFINTOSC mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D016</td>
<td></td>
<td>103</td>
<td>190</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 500 kHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>124</td>
<td>220</td>
<td>µA</td>
<td>3.0</td>
<td>MFINTOSC mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D017*</td>
<td></td>
<td>0.5</td>
<td>0.9</td>
<td>mA</td>
<td>1.8</td>
<td>Fosc = 8 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>0.8</td>
<td>1.3</td>
<td>mA</td>
<td>3.0</td>
<td>HFINTOSC mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D018</td>
<td></td>
<td>0.7</td>
<td>1.2</td>
<td>mA</td>
<td>1.8</td>
<td>Fosc = 16 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>1.2</td>
<td>1.8</td>
<td>mA</td>
<td>3.0</td>
<td>HFINTOSC mode</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D019</td>
<td></td>
<td>1.3</td>
<td>3</td>
<td>mA</td>
<td>3.0</td>
<td>Fosc = 32 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>2.3</td>
<td>4</td>
<td>mA</td>
<td>3.6</td>
<td>HFINTOSC mode (Note 3)</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D020</td>
<td>Supply Current (IDD)</td>
<td>1.3</td>
<td>2.5</td>
<td>mA</td>
<td>3.0</td>
<td>Fosc = 32 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>1.7</td>
<td>3.0</td>
<td>mA</td>
<td>3.6</td>
<td>HS Oscillator mode (Note 4)</td>
<td></td>
<td></td>
</tr>
<tr>
<td>D021</td>
<td></td>
<td>185</td>
<td>300</td>
<td>µA</td>
<td>1.8</td>
<td>Fosc = 4 MHz</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>390</td>
<td>480</td>
<td>µA</td>
<td>3.0</td>
<td>EXTRC mode (Note 5)</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Note 1: The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

Note 2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

Note 3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula \( I_{REXT} = VDD/2\times REXT \) (mA) with REXT in kΩ.
### 31.3 DC Characteristics: PIC12LF1840T39A-I (Power-Down)

#### PIC12LF1840T39A

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Device Characteristics</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max. +85°C</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Vdd</td>
</tr>
<tr>
<td>Power-down Base Current (IpD)(2)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D022</td>
<td>—</td>
<td>—</td>
<td>0.02</td>
<td>1</td>
<td>µA</td>
<td>1.8</td>
</tr>
<tr>
<td></td>
<td>—</td>
<td>—</td>
<td>0.03</td>
<td>2</td>
<td>µA</td>
<td>3.0</td>
</tr>
<tr>
<td>D023</td>
<td>—</td>
<td>—</td>
<td>0.3</td>
<td>6</td>
<td>µA</td>
<td>1.8</td>
</tr>
<tr>
<td></td>
<td>—</td>
<td>—</td>
<td>0.5</td>
<td>7</td>
<td>µA</td>
<td>3.0</td>
</tr>
<tr>
<td>D023A</td>
<td>—</td>
<td>—</td>
<td>8.5</td>
<td>23</td>
<td>µA</td>
<td>1.8</td>
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<td>µA</td>
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<td>28</td>
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**Note**: The peripheral current is the sum of the base Idd or IpD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base Idd or IpD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to Vdd.

3: A/D oscillator source is FRC.
### 31.4 DC Characteristics: PIC12LF1840T39A-I

#### DC CHARACTERISTICS

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
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<td>D031</td>
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<td>with Schmitt Trigger buffer</td>
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<td>with I²C™ levels</td>
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<tr>
<td>D033</td>
<td></td>
<td>with SMBus levels</td>
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<td>MCLR, OSC1 (RC mode)(1)</td>
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<td>D043B</td>
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<td>D045</td>
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<td>Input Leakage Current(2)</td>
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<tr>
<td>D060</td>
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<td>I/O ports</td>
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<td>D061A</td>
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<td>MCLR(3)</td>
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<td>D070*</td>
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<td>Weak Pull-up Current</td>
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<td>D080</td>
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<td>D101*</td>
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<td>Capacitive Loading Specs on Output Pins</td>
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<td>D101A*</td>
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</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.

**Note 2:** Negative current is defined as current sourced by the pin.

**Note 3:** The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

**Note 4:** Including OSC2 in CLKOUT mode.
### 31.5 Memory Programming Requirements

#### DC CHARACTERISTICS

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
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<tr>
<td>D110</td>
<td>VIHH</td>
<td>Program Memory Programming Specifications</td>
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<tr>
<td>D111</td>
<td>IDD P</td>
<td>Voltage on MCLR/VPP/RA5 pin</td>
<td>8.0</td>
<td>—</td>
<td>9.0</td>
<td>V</td>
<td>(Note 3, Note 4)</td>
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<tr>
<td>D112</td>
<td>VDD</td>
<td>Supply Current during Programming</td>
<td>2.7</td>
<td>—</td>
<td>VDD</td>
<td>V</td>
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<td>D113</td>
<td>VPEW</td>
<td>VDD for Bulk Erase</td>
<td>VDD</td>
<td>—</td>
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<td>D114</td>
<td>IPPGM</td>
<td>Current on MCLR/VPP during Erase/Write</td>
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<td>1</td>
<td>—</td>
<td>mA</td>
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<td>D115</td>
<td>IDDP GM</td>
<td>Current on VDD during Erase/Write</td>
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<td>5</td>
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<td>mA</td>
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<td>D116</td>
<td>ED</td>
<td>Byte Endurance</td>
<td>100K</td>
<td>—</td>
<td>—</td>
<td>E/W</td>
<td>-40°C to +85°C</td>
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<td>D117</td>
<td>VORW</td>
<td>VDD for Read/Write</td>
<td>VDD</td>
<td>—</td>
<td>VDD</td>
<td>V</td>
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<td>D118</td>
<td>TDEW</td>
<td>Erase/Write Cycle Time</td>
<td>—</td>
<td>4.0</td>
<td>5.0</td>
<td>ms</td>
<td>Provided no other specifications are violated</td>
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<td>D119</td>
<td>TRED</td>
<td>Characteristic Retention</td>
<td>—</td>
<td>40</td>
<td>—</td>
<td>Year</td>
<td>-40°C to +85°C</td>
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<tr>
<td>D120</td>
<td>TREF</td>
<td>Number of Total Erase/Write Cycles before Refresh$^2$</td>
<td>1M</td>
<td>10M</td>
<td>—</td>
<td>E/W</td>
<td>-40°C to +85°C</td>
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<td>D121</td>
<td>EP</td>
<td>Cell Endurance</td>
<td>10K</td>
<td>—</td>
<td>—</td>
<td>E/W</td>
<td>-40°C to +85°C (Note 1)</td>
</tr>
<tr>
<td>D122</td>
<td>VPR</td>
<td>VDD for Read</td>
<td>VDD</td>
<td>—</td>
<td>VDD</td>
<td>V</td>
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<td>D123</td>
<td>TW</td>
<td>Self-timed Write Cycle Time</td>
<td>—</td>
<td>2</td>
<td>2.5</td>
<td>ms</td>
<td>Provided no other specifications are violated</td>
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<tr>
<td>D124</td>
<td>TRWD</td>
<td>Characteristic Retention</td>
<td>—</td>
<td>40</td>
<td>—</td>
<td>Year</td>
<td></td>
</tr>
</tbody>
</table>

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Self-write and Block Erase.

**Note 2:** Refer to Section 11.2 “Using the Data EEPROM” for a more detailed discussion on data EEPROM endurance.

**Note 3:** Required only if single-supply programming is disabled.

**Note 4:** The MPLAB ICD 2 does not support variable VPP output. Circuitry to limit the MPLAB ICD 2 VPP voltage must be placed between the MPLAB ICD 2 and target system when programming or debugging with the MPLAB ICD 2.
31.6 Thermal Considerations

Standard Operating Conditions (unless otherwise stated)
Operating temperature -40°C ≤ TA ≤ +125°C

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Typ.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
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<tbody>
<tr>
<td>TH01</td>
<td>0JA</td>
<td>Thermal Resistance Junction to Ambient</td>
<td>TBD</td>
<td>°C/W</td>
<td>14-pin TSSOP 4x4mm package</td>
</tr>
<tr>
<td>TH02</td>
<td>0JC</td>
<td>Thermal Resistance Junction to Case</td>
<td>TBD</td>
<td>°C/W</td>
<td>14-pin TSSOP 4x4mm package</td>
</tr>
<tr>
<td>TH03</td>
<td>TMJAX</td>
<td>Maximum Junction Temperature</td>
<td>150</td>
<td>°C</td>
<td></td>
</tr>
<tr>
<td>TH04</td>
<td>PD</td>
<td>Power Dissipation</td>
<td>—</td>
<td>W</td>
<td>PD = PINTERNAL + P/I/O</td>
</tr>
<tr>
<td>TH05</td>
<td>PINTERNAL</td>
<td>Internal Power Dissipation</td>
<td>—</td>
<td>W</td>
<td>PINTERNAL = IDD x VDD(1)</td>
</tr>
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<td>TH06</td>
<td>P/I/O</td>
<td>I/O Power Dissipation</td>
<td>—</td>
<td>W</td>
<td>P/I/O = Σ (IOL * VOL) + Σ (IOH * (VDD - VOH))</td>
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<td>TH07</td>
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<td>Derated Power</td>
<td>—</td>
<td>W</td>
<td>PDER = PDMAX (TJ - TA)/θJA(2)</td>
</tr>
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</table>

Legend:  
TBD = To Be Determined

Note 1:  
(1) IDD is current to run the chip alone without driving any load on the output pins.
(2) TA = Ambient Temperature.
(3) TJ = Junction Temperature.

TABLE 31-1: RF Transmitter Electrical Specifications

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Description</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
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<tbody>
<tr>
<td>IDDSL</td>
<td>Supply Current in Sleep mode</td>
<td>—</td>
<td>—</td>
<td>0.5</td>
<td>1</td>
<td>μA</td>
</tr>
<tr>
<td>IDDT_315</td>
<td>Supply Current in Transmit mode at 315 MHz*</td>
<td>RFOP = +10 dBm 50% OOK</td>
<td>—</td>
<td>11</td>
<td>—</td>
<td>mA</td>
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<tr>
<td></td>
<td></td>
<td>RFOP = +10 dBm FSK</td>
<td>—</td>
<td>15</td>
<td>—</td>
<td>mA</td>
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<tr>
<td></td>
<td></td>
<td>RFOP = 0 dBm FSK</td>
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<td>9</td>
<td>—</td>
<td>mA</td>
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<td>IDDT_915</td>
<td>Supply Current in Transmit mode at 915 MHz*</td>
<td>RFOP = +10 dBm FSK</td>
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<td>17.5</td>
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<td>mA</td>
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<td></td>
<td></td>
<td>RFOP = 0 dBm FSK</td>
<td>—</td>
<td>10.5</td>
<td>—</td>
<td>mA</td>
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<td>FBAND</td>
<td>Accessible Frequency Bands</td>
<td>Band 0, with FXOSC = 22 MHz</td>
<td>310</td>
<td>—</td>
<td>450</td>
<td>MHz</td>
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<td></td>
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<td>Band 0, with FXOSC = 24 MHz</td>
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<td>MHz</td>
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<td></td>
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<td>Band 0, with FXOSC = 26 MHz</td>
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<td>—</td>
<td>450</td>
<td>MHz</td>
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<td></td>
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<td>Band 1, with FXOSC = 26 MHz</td>
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<td>MHz</td>
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<td></td>
<td></td>
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<td>902</td>
<td>—</td>
<td>928</td>
<td>MHz</td>
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<td>FDA</td>
<td>Frequency deviation, FSK</td>
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<td>—</td>
<td>200</td>
<td>kHz</td>
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<td>Bit rate, FSK</td>
<td>Permissible Range</td>
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<td>kbps</td>
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<td>OOK Modulation Depth</td>
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<td>dB</td>
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<td>RFOP</td>
<td>RF output power in 50 Ohms in either frequency bands</td>
<td>High-Power Setting</td>
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<td>dBm</td>
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<td>Low-Power Setting*</td>
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<td>dB</td>
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<td>dB</td>
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<td>1.8V to 3.7V</td>
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<td>—</td>
<td>7</td>
<td>dB</td>
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<td>PHN</td>
<td>Transmitter phase noise</td>
<td>At offset:</td>
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<td>-82</td>
<td>-76</td>
<td>dBc/Hz</td>
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<tr>
<td></td>
<td></td>
<td>100 kHz</td>
<td>—</td>
<td>-92</td>
<td>-81</td>
<td>dBc/Hz</td>
</tr>
<tr>
<td></td>
<td></td>
<td>350 kHz</td>
<td>—</td>
<td>-92</td>
<td>-81</td>
<td>dBc/Hz</td>
</tr>
<tr>
<td></td>
<td></td>
<td>550 kHz</td>
<td>—</td>
<td>-96</td>
<td>-91</td>
<td>dBc/Hz</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1.15 MHz</td>
<td>—</td>
<td>-103</td>
<td>-101</td>
<td>dBc/Hz</td>
</tr>
<tr>
<td>STEP_22</td>
<td>RF frequency step</td>
<td>FXOSC = 22 MHz, Band 0</td>
<td>—</td>
<td>1.34277</td>
<td>—</td>
<td>kHz</td>
</tr>
<tr>
<td>STEP_24</td>
<td>RF frequency step</td>
<td>FXOSC = 24 MHz, Band 0</td>
<td>—</td>
<td>1.46484</td>
<td>—</td>
<td>kHz</td>
</tr>
<tr>
<td>STEP_26</td>
<td>RF frequency step</td>
<td>FXOSC = 26 MHz, Band 0</td>
<td>—</td>
<td>1.58691</td>
<td>—</td>
<td>kHz</td>
</tr>
<tr>
<td></td>
<td></td>
<td>FXOSC = 26 MHz, Band 1</td>
<td>—</td>
<td>3.17383</td>
<td>—</td>
<td>kHz</td>
</tr>
</tbody>
</table>
### TABLE 31-1: RF Transmitter Electrical Specifications (CONTINUED)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Description</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>FXOSC</td>
<td>Crystal Oscillator Frequency</td>
<td>XTAL dependant, with spec’d XTAL</td>
<td>22</td>
<td>24</td>
<td>26</td>
<td>MHz</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>22</td>
<td>24</td>
<td>26</td>
<td>MHz</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>22</td>
<td>24</td>
<td>26</td>
<td>MHz</td>
</tr>
</tbody>
</table>

#### Timing Specifications

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Description</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>t_WAKE</td>
<td>Time from Sleep to Tx mode</td>
<td>XTAL</td>
<td>650</td>
<td>2000</td>
<td>us</td>
<td></td>
</tr>
<tr>
<td>t_OFFT</td>
<td>Timer from Tx data activity to Sleep</td>
<td>Programmable</td>
<td>2</td>
<td>20</td>
<td>ms</td>
<td></td>
</tr>
<tr>
<td>t_RAMP</td>
<td>PA Ramp up and down time</td>
<td>20</td>
<td>us</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_START</td>
<td>Time before CTRL pin mode selection.</td>
<td>Time from power on to sampling of CTRL</td>
<td>1</td>
<td>ms</td>
<td></td>
<td></td>
</tr>
<tr>
<td>f_CTRL</td>
<td>CTRL Clock Frequency</td>
<td>10</td>
<td>MHz</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_CH</td>
<td>CTRL Clock High time</td>
<td>45</td>
<td>ns</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_CL</td>
<td>CTRL Clock Low time</td>
<td>45</td>
<td>ns</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_RISE</td>
<td>CTRL Clock Rise time</td>
<td>5</td>
<td>ns</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_FALL</td>
<td>CTRL Clock Fall time</td>
<td>5</td>
<td>ns</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_SETUP</td>
<td>DATA Setup time</td>
<td>From DATA transition to CTRL rising edge</td>
<td>45</td>
<td>ns</td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_HOLD</td>
<td>DATA Hold time</td>
<td>From CTRL rising edge to DATA transition</td>
<td>45</td>
<td>ns</td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_0</td>
<td>Time at ‘1’ on DATA during Recovery Sequence Timing</td>
<td>See Figure 28-4</td>
<td>5</td>
<td>ns</td>
<td></td>
<td></td>
</tr>
<tr>
<td>t_1</td>
<td>Time at ‘0’ on DATA during Recovery Sequence Timing</td>
<td>See Figure 28-4</td>
<td>5</td>
<td>ns</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

### TABLE 31-2: POWER CONSUMPTION IN TX MODE

<table>
<thead>
<tr>
<th>Frequency Band</th>
<th>Conditions</th>
<th>Typical Current Drain</th>
</tr>
</thead>
<tbody>
<tr>
<td>310 to 450 MHz</td>
<td>POUT = +10 dBm, OOK modulation with 50% duty cycle</td>
<td>11 mA</td>
</tr>
<tr>
<td></td>
<td>POUT = +10 dBm, FSK modulation</td>
<td>15 mA</td>
</tr>
<tr>
<td></td>
<td>POUT = 0 dBm, FSK modulation</td>
<td>9 mA</td>
</tr>
<tr>
<td>860 to 870 MHz</td>
<td>POUT = +10 dBm, FSK modulation</td>
<td>16.5 mA</td>
</tr>
<tr>
<td></td>
<td>POUT = 0 dBm, FSK modulation</td>
<td>10 mA</td>
</tr>
<tr>
<td>902 to 928 MHz</td>
<td>POUT = +10 dBm, FSK modulation</td>
<td>17.5 mA</td>
</tr>
<tr>
<td></td>
<td>POUT = 0 dBm, FSK modulation</td>
<td>10.5 mA</td>
</tr>
</tbody>
</table>
31.7 Timing Parameter Symbology

The timing parameter symbols have been created with one of the following formats:

1. TppS2ppS
2. TppS

<table>
<thead>
<tr>
<th>T</th>
<th>F</th>
<th>Frequency</th>
<th>T</th>
<th>Time</th>
</tr>
</thead>
<tbody>
<tr>
<td>pp</td>
<td>cc</td>
<td>CCP1</td>
<td>osc</td>
<td>OSC1</td>
</tr>
<tr>
<td></td>
<td>ck</td>
<td>CLKOUT</td>
<td>rd</td>
<td>RD</td>
</tr>
<tr>
<td></td>
<td>cs</td>
<td>CS</td>
<td>rw</td>
<td>RD or WR</td>
</tr>
<tr>
<td></td>
<td>di</td>
<td>SDIx</td>
<td>sc</td>
<td>SCKx</td>
</tr>
<tr>
<td></td>
<td>do</td>
<td>SDO</td>
<td>ss</td>
<td>SS</td>
</tr>
<tr>
<td></td>
<td>dt</td>
<td>Data in</td>
<td>t0</td>
<td>T0CKI</td>
</tr>
<tr>
<td></td>
<td>io</td>
<td>I/O PORT</td>
<td>t1</td>
<td>T1CKI</td>
</tr>
<tr>
<td></td>
<td>mc</td>
<td>MCLR</td>
<td>wr</td>
<td>WR</td>
</tr>
</tbody>
</table>

Lowercase letters (pp) and their meanings:

Uppercase letters and their meanings:

FIGURE 31-4: LOAD CONDITIONS

Load Condition

Legend: CL = 50 pF for all pins, 15 pF for OSC2 output
### 31.8 AC Characteristics: PIC12LF1840T39A-I

**FIGURE 31-5: CLOCK TIMING**

![Circuit Diagram](image)

**TABLE 31-3: CLOCK OSCILLATOR TIMING REQUIREMENTS**

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>OS01</td>
<td>Fosc</td>
<td>External CLKIN Frequency(1)</td>
<td>DC</td>
<td>—</td>
<td>0.5</td>
<td>MHz</td>
<td>EC Oscillator mode (low)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>DC</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>EC Oscillator mode (medium)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>DC</td>
<td>—</td>
<td>32</td>
<td>MHz</td>
<td>EC Oscillator mode (high)</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Oscillator Frequency(1)</td>
<td>—</td>
<td>32.768</td>
<td>—</td>
<td>kHz</td>
<td>LP Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0.1</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>XT Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>HS Oscillator mode, $V_{DD} \leq 2.7$V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1</td>
<td>—</td>
<td>20</td>
<td>MHz</td>
<td>HS Oscillator mode, $V_{DD} &gt; 2.7$V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>DC</td>
<td>—</td>
<td>4</td>
<td>MHz</td>
<td>RC Oscillator mode</td>
</tr>
<tr>
<td>OS02</td>
<td>Tosc</td>
<td>External CLKIN Period(1)</td>
<td>27</td>
<td>—</td>
<td>$\infty$</td>
<td>μs</td>
<td>LP Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>$\infty$</td>
<td>ns</td>
<td>XT Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>50</td>
<td>—</td>
<td>$\infty$</td>
<td>ns</td>
<td>HS Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>50</td>
<td>—</td>
<td>$\infty$</td>
<td>ns</td>
<td>EC Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Oscillator Period(1)</td>
<td>—</td>
<td>30.5</td>
<td>—</td>
<td>μs</td>
<td>LP Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>10,000</td>
<td>ns</td>
<td>XT Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>50</td>
<td>—</td>
<td>1,000</td>
<td>ns</td>
<td>HS Oscillator mode</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>250</td>
<td>—</td>
<td>$\infty$</td>
<td>ns</td>
<td>RC Oscillator mode</td>
</tr>
<tr>
<td>OS03</td>
<td>TCY</td>
<td>Instruction Cycle Time(1)</td>
<td>200</td>
<td>—</td>
<td>DC</td>
<td>ns</td>
<td>$T_{CY} = \frac{F_{OSC}}{4}$</td>
</tr>
<tr>
<td>OS04*</td>
<td>TosH, ToSL</td>
<td>External CLKIN High, External CLKIN Low</td>
<td>2</td>
<td>—</td>
<td>—</td>
<td>μs</td>
<td>LP oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>100</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>XT oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>HS oscillator</td>
</tr>
<tr>
<td>OS05*</td>
<td>TosR, TosF</td>
<td>External CLKIN Rise, External CLKIN Fall</td>
<td>0</td>
<td>—</td>
<td>$\infty$</td>
<td>ns</td>
<td>LP oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0</td>
<td>—</td>
<td>$\infty$</td>
<td>ns</td>
<td>XT oscillator</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0</td>
<td>—</td>
<td>$\infty$</td>
<td>ns</td>
<td>HS oscillator</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Instruction cycle period ($T_{CY}$) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at “min” values with an external clock applied to OSC1 pin. When an external clock input is used, the “max” cycle time limit is “DC” (no clock) for all devices.
### TABLE 31-4: OSCILLATOR PARAMETERS

**Standard Operating Conditions (unless otherwise stated)**

Operating Temperature: 

-40°C ≤ TA ≤ +125°C

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Freq. Tolerance</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>OS08</td>
<td>HFosc</td>
<td>Internal Calibrated HFINTOSC Frequency(^2)</td>
<td>±2%</td>
<td>16.0</td>
<td>—</td>
<td>MHz</td>
<td>0°C ≤ TA ≤ +60°C, VDD ≥ 2.5V</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>±3%</td>
<td>16.0</td>
<td>—</td>
<td>MHz</td>
<td>60°C ≤ TA ≤ +85°C, VDD ≥ 2.5V</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>±5%</td>
<td>16.0</td>
<td>—</td>
<td>MHz</td>
<td>-40°C ≤ TA ≤ +125°C</td>
<td></td>
</tr>
<tr>
<td>OS08A</td>
<td>MFosc</td>
<td>Internal Calibrated MFINTOSC Frequency(^2)</td>
<td>±2%</td>
<td>500</td>
<td>—</td>
<td>kHz</td>
<td>0°C ≤ TA ≤ +60°C, VDD ≥ 2.5V</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>±3%</td>
<td>500</td>
<td>—</td>
<td>kHz</td>
<td>60°C ≤ TA ≤ +85°C, VDD ≥ 2.5V</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>±5%</td>
<td>500</td>
<td>—</td>
<td>kHz</td>
<td>-40°C ≤ TA ≤ +125°C</td>
<td></td>
</tr>
<tr>
<td>OS09</td>
<td>LFosc</td>
<td>Internal LFINTOSC Frequency</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>kHz</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

| OS10*     | Tiosc| Wake-up from Sleep Start-up Time |
|           |      | —    | 5    | 8    | µs    |

| OS10*     | Tiosc| Wake-up from Sleep Start-up Time |
|           |      | —    | 20   | 30   | µs    |

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at “min” values with an external clock applied to the OSC1 pin. When an external clock input is used, the “max” cycle time limit is “DC” (no clock) for all devices.

2: To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 µF and 0.01 µF values in parallel are recommended.

3: By design.

### TABLE 31-5: PLL CLOCK TIMING SPECIFICATIONS (VDD = 2.7V TO 3.6V)

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>F10</td>
<td>Fosc</td>
<td>Oscillator Frequency Range</td>
<td>4</td>
<td>—</td>
<td>8</td>
<td>MHz</td>
<td>—</td>
</tr>
<tr>
<td>F11</td>
<td>Fsys</td>
<td>On-Chip VCO System Frequency</td>
<td>16</td>
<td>—</td>
<td>32</td>
<td>MHz</td>
<td>—</td>
</tr>
<tr>
<td>F12</td>
<td>Trc</td>
<td>PLL Start-up Time (Lock Time)</td>
<td>—</td>
<td>—</td>
<td>2</td>
<td>ms</td>
<td>—</td>
</tr>
<tr>
<td>F13*</td>
<td>Δclk</td>
<td>CLKOUT Stability (Jitter)</td>
<td>-0.25%</td>
<td>—</td>
<td>+0.25%</td>
<td>%</td>
<td>—</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
FIGURE 31-6: CLKOUT AND I/O TIMING

<table>
<thead>
<tr>
<th>Cycle</th>
<th>Write</th>
<th>Fetch</th>
<th>Read</th>
<th>Execute</th>
</tr>
</thead>
<tbody>
<tr>
<td>Fosc</td>
<td>Q4</td>
<td>Q1</td>
<td>Q2</td>
<td>Q3</td>
</tr>
<tr>
<td>CLKOUT</td>
<td>OS11</td>
<td>OS20</td>
<td>OS16</td>
<td>OS18</td>
</tr>
<tr>
<td>I/O pin (Input)</td>
<td>OS19</td>
<td>OS13</td>
<td>OS17</td>
<td></td>
</tr>
<tr>
<td>I/O pin (Output)</td>
<td>OS15</td>
<td>OS14</td>
<td>OS18, OS19</td>
<td>Old Value</td>
</tr>
</tbody>
</table>
### TABLE 31-6: CLKOUT AND I/O TIMING PARAMETERS

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>OS11</td>
<td>TosH2ckL</td>
<td>Fosc↑ to CLKOUT↓ (1)</td>
<td>—</td>
<td>—</td>
<td>70 ns</td>
<td>ns</td>
<td>VDD = 3.0-3.6V</td>
</tr>
<tr>
<td>OS12</td>
<td>TosH2ckH</td>
<td>Fosc↑ to CLKOUT↑ (1)</td>
<td>—</td>
<td>—</td>
<td>72 ns</td>
<td>ns</td>
<td>VDD = 3.0-3.6V</td>
</tr>
<tr>
<td>OS13</td>
<td>TckL2ioV</td>
<td>CLKOUT↓ to Port out valid (1)</td>
<td>—</td>
<td>—</td>
<td>20 ns</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>OS14</td>
<td>TioV2ckH</td>
<td>Port input valid before CLKOUT↑(1)</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>OS15</td>
<td>TosH2ioV</td>
<td>Fosc↑ (Q1 cycle) to Port out valid</td>
<td>—</td>
<td>50 ns</td>
<td>70* ns</td>
<td>ns</td>
<td>VDD = 3.0-3.6V</td>
</tr>
<tr>
<td>OS16</td>
<td>TosH2iol</td>
<td>Fosc↑ (Q2 cycle) to Port input invalid (I/O in hold time)</td>
<td>50</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>VDD = 3.0-3.6V</td>
</tr>
<tr>
<td>OS17</td>
<td>TioV2osH</td>
<td>Port input valid to Fosc↑ (Q2 cycle) (I/O in setup time)</td>
<td>20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>OS18*</td>
<td>TioR</td>
<td>Port output rise time</td>
<td>—</td>
<td>40 ns</td>
<td>72 ns</td>
<td>ns</td>
<td>VDD = 1.8V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>—</td>
<td>15 ns</td>
<td>32 ns</td>
<td>ns</td>
<td>VDD = 3.0-3.6V</td>
</tr>
<tr>
<td>OS19*</td>
<td>TioF</td>
<td>Port output fall time</td>
<td>—</td>
<td>28 ns</td>
<td>55 ns</td>
<td>ns</td>
<td>VDD = 1.8V</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>—</td>
<td>15 ns</td>
<td>30 ns</td>
<td>ns</td>
<td>VDD = 3.0-3.6V</td>
</tr>
<tr>
<td>OS20*</td>
<td>Tinp</td>
<td>INT pin input high or low time</td>
<td>25</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>OS21*</td>
<td>Tioc</td>
<td>Interrupt-on-change new input level time</td>
<td>25</td>
<td>—</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated.

**Note 1:** Measurements are taken in RC mode where CLKOUT output is 4 x Tosc.

### FIGURE 31-7: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

[Diagram showing the timing of various events such as VDD, MCLR, Internal POR, PWRT, OSC start-up time, internal reset, watchdog timer reset, and I/O pins.]
FIGURE 31-8: BROWN-OUT RESET TIMING AND CHARACTERISTICS

Note 1: 64 ms delay only if PWRTE bit in the Configuration Word 1 is programmed to '0'.
2 ms delay if PWRTE = 0 and VREGEN = 1.
### TABLE 31-7: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET PARAMETERS

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>30</td>
<td>TMCL</td>
<td>MCLR Pulse Width (low)</td>
<td>2</td>
<td></td>
<td>5</td>
<td>μs</td>
<td>VDD = 3.3-3.6V, -40°C to +85°C</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>31</td>
<td>TWDTLP</td>
<td>Watchdog Timer Time-out Period</td>
<td>10</td>
<td>16</td>
<td>27</td>
<td>ms</td>
<td>VDD = 3.3V-5V, 1:16 Prescaler used</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>32</td>
<td>TOST</td>
<td>Oscillator Start-up Timer Period(1), (2)</td>
<td></td>
<td>1024</td>
<td></td>
<td></td>
<td>VDD = 3.3V-5V, Tosc (Note 3)</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>33*</td>
<td>TPWRT</td>
<td>Power-up Timer Period, PWRTE = 0</td>
<td>40</td>
<td>65</td>
<td>140</td>
<td>ms</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>34*</td>
<td>TIOZ</td>
<td>I/O high-impedance from MCLR Low or Watchdog Timer Reset</td>
<td></td>
<td></td>
<td></td>
<td>μs</td>
<td></td>
</tr>
<tr>
<td>35</td>
<td>Vbor</td>
<td>Brown-out Reset Voltage BORV = 0</td>
<td>2.55</td>
<td>2.70</td>
<td>2.85</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>BORV = 1</td>
<td>1.80</td>
<td>1.9</td>
<td>2.05</td>
<td></td>
<td></td>
</tr>
<tr>
<td>37*</td>
<td>VHYST</td>
<td>Brown-out Reset Hysteresis</td>
<td>0</td>
<td>25</td>
<td>60</td>
<td>mV</td>
<td>-40°C to +85°C</td>
</tr>
<tr>
<td>38*</td>
<td>TBORDC</td>
<td>Brown-out Reset DC Response Time</td>
<td>1</td>
<td>3</td>
<td>35</td>
<td>μs</td>
<td>VDD ≤ VBOR</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at “min” values with an external clock applied to the OSC1 pin. When an external clock input is used, the “max” cycle time limit is “DC” (no clock) for all devices.

2: By design.
3: Period of the slower clock.
4: To ensure these voltage tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.


### FIGURE 31-9: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

![Diagram showing waveforms for T0CKI, T1CKI, TMR0 or TMR1]

### TABLE 31-8: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>40*</td>
<td>T0H</td>
<td>T0CKI High Pulse Width</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>No Prescaler</td>
<td>0.5 TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>With Prescaler</td>
<td>10</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>41*</td>
<td>T0L</td>
<td>T0CKI Low Pulse Width</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>No Prescaler</td>
<td>0.5 TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>With Prescaler</td>
<td>10</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>42*</td>
<td>T0P</td>
<td>T0CKI Period</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Greater of:</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td>N = prescale value (2, 4, ..., 256)</td>
</tr>
<tr>
<td></td>
<td></td>
<td>20 or TCY + 40</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>45*</td>
<td>T1H</td>
<td>T1CKI High Time</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Synchronous, No Prescaler</td>
<td>0.5 TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Synchronous, with Prescaler</td>
<td>15</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Asynchronous</td>
<td>30</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>46*</td>
<td>T1L</td>
<td>T1CKI Low Time</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Synchronous, No Prescaler</td>
<td>0.5 TCY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Synchronous, with Prescaler</td>
<td>15</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Asynchronous</td>
<td>30</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>47*</td>
<td>T1P</td>
<td>T1CKI Input Period</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Synchronous</td>
<td>Greater of:</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>N = prescale value (1, 2, 4, 8)</td>
</tr>
<tr>
<td></td>
<td></td>
<td>30 or TCY + 40</td>
<td></td>
<td></td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Asynchronous</td>
<td>360</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>48</td>
<td>F1</td>
<td>Timer1 Oscillator Input Frequency Range</td>
<td>32.4</td>
<td>32.768</td>
<td>33.1</td>
<td>kHz</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>(oscillator enabled by setting bit T1OSCEN)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>49*</td>
<td>TCEZmr1</td>
<td>Delay from External Clock Edge to Timer Increment</td>
<td>2 Tosc</td>
<td>—</td>
<td>7 Tosc</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

---

**Standard Operating Conditions (unless otherwise stated)**

Operating Temperature -40°C ≤ TA ≤ +125°C
FIGURE 31-10: CAPTURE/COMPARE/PWM TIMINGS (CCP)

![Capture/Compare/PWM Timings Diagram]

Note: Refer to Figure 31-4 for load conditions.

TABLE 31-9: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP)

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>CC01*</td>
<td>TccL</td>
<td>CCP Input Low Time</td>
<td>No Prescaler</td>
<td>0.5T CY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>With Prescaler</td>
<td>20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td>CC02*</td>
<td>TccH</td>
<td>CCP Input High Time</td>
<td>No Prescaler</td>
<td>0.5T CY + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>With Prescaler</td>
<td>20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td>CC03*</td>
<td>TccP</td>
<td>CCP Input Period</td>
<td>3T CY + 40 N</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td>N = prescale value (1, 4 or 16)</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 31-10: PIC12LF1840T39A A/D CONVERTER (ADC) CHARACTERISTICS:

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>AD01</td>
<td>NR</td>
<td>Resolution</td>
<td>—</td>
<td>—</td>
<td>10</td>
<td>bit</td>
<td></td>
</tr>
<tr>
<td>AD02</td>
<td>EIL</td>
<td>Integral Error</td>
<td>—</td>
<td>±1</td>
<td>±1.7</td>
<td>LSB</td>
<td>VREF = 3.0V</td>
</tr>
<tr>
<td>AD03</td>
<td>EDL</td>
<td>Differential Error</td>
<td>—</td>
<td>±1</td>
<td>±1</td>
<td>LSB</td>
<td>No missing codes VREF = 3.0V</td>
</tr>
<tr>
<td>AD04</td>
<td>EOFF</td>
<td>Offset Error</td>
<td>—</td>
<td>±1</td>
<td>±2.5</td>
<td>LSB</td>
<td>VREF = 3.0V</td>
</tr>
<tr>
<td>AD05</td>
<td>EG</td>
<td>Gain Error</td>
<td>—</td>
<td>±1</td>
<td>±2</td>
<td>LSB</td>
<td>VREF = 3.0V</td>
</tr>
<tr>
<td>AD06</td>
<td>VREF</td>
<td>Reference Voltage(3)</td>
<td>1.8</td>
<td>—</td>
<td>VDD</td>
<td>V</td>
<td>VREF = (VREF+ minus VREF-) (Note 5)</td>
</tr>
<tr>
<td>AD07</td>
<td>VAIN</td>
<td>Full-Scale Range</td>
<td>VSS</td>
<td>—</td>
<td>VREF</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>AD08</td>
<td>ZAIN</td>
<td>Recommended Impedance of Analog Voltage Source</td>
<td>—</td>
<td>—</td>
<td>10</td>
<td>kΩ</td>
<td>Can go higher if external 0.01μF capacitor is present on input pin.</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Total Absolute Error includes integral, differential, offset and gain errors.
2: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.
3: ADC VREF is from external VREF, VDD pin or FVR, whichever is selected as reference input.
4: When ADC is off, it will not consume any current other than leakage current. The power-down current specification includes any such leakage from the ADC module.
5: FVR voltage selected must be 2.048V.
TABLE 31-11: PIC12LF1840T39A A/D CONVERSION REQUIREMENTS

<table>
<thead>
<tr>
<th>Operating Conditions (unless otherwise stated)</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{DD} = 3.0V, T_A = 25°C$</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Sym.</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>AD130*</td>
<td>TAD</td>
<td>A/D Clock Period</td>
<td>1.0</td>
<td>—</td>
<td>9.0</td>
<td>µs</td>
<td>TOSC-based</td>
</tr>
<tr>
<td>AD130*</td>
<td>TAD</td>
<td>A/D Internal RC Oscillator Period</td>
<td>1.0</td>
<td>2</td>
<td>6.0</td>
<td>µs</td>
<td>ADCS&lt;1:0&gt; = 11 (ADRC mode)</td>
</tr>
</tbody>
</table>

| AD131     | TCNV | Conversion Time (not including Acquisition Time)$^{(1)}$ | —    | 11   | —    | TAD  | Set GO/DONE bit to conversion complete |

| AD132*    | TCQ  | Acquisition Time                             | —    | 5.0  | —    | µs    |                                   |

* These parameters are characterized but not tested.
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The ADRES register may be read on the following TCY cycle.

FIGURE 31-11: PIC12LF1840T39A A/D CONVERSION TIMING (NORMAL MODE)

Note 1: If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.
FIGURE 31-12: PIC12LF1840T39A A/D CONVERSION TIMING (SLEEP MODE)

Note 1: If the A/D clock source is selected as RC, a time of TCY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.
### TABLE 31-12: COMPARATOR SPECIFICATIONS

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>CM01</td>
<td>VIOFF</td>
<td>Input Offset Voltage</td>
<td>—</td>
<td>±7.5</td>
<td>±60</td>
<td>mV</td>
<td>CxSP = 1</td>
</tr>
<tr>
<td>CM02</td>
<td>VICM</td>
<td>Input Common Mode Voltage</td>
<td>0</td>
<td>—</td>
<td>VDD</td>
<td>V</td>
<td>VICM = VDD/2</td>
</tr>
<tr>
<td>CM03</td>
<td>CMRR</td>
<td>Common Mode Rejection Ratio</td>
<td>—</td>
<td>50</td>
<td>—</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>CM04A</td>
<td>TRESP(1)</td>
<td>Response Time Rising Edge</td>
<td>—</td>
<td>400</td>
<td>800</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>CM04B</td>
<td>TRESP(1)</td>
<td>Response Time Falling Edge</td>
<td>—</td>
<td>200</td>
<td>400</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>CM04C</td>
<td>TRESP(1)</td>
<td>Response Time Rising Edge</td>
<td>—</td>
<td>1200</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>CM04D</td>
<td>TRESP(1)</td>
<td>Response Time Falling Edge</td>
<td>—</td>
<td>550</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>CM05</td>
<td>TMC2OV</td>
<td>Comparator Mode Change to Output Valid*</td>
<td>—</td>
<td>—</td>
<td>10</td>
<td>µs</td>
<td></td>
</tr>
<tr>
<td>CM06</td>
<td>CHYSER</td>
<td>Comparator Hysteresis</td>
<td>—</td>
<td>45</td>
<td>—</td>
<td>mV</td>
<td>CxSHYS = 1, CxSP = 1 (Note 2)</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

**Note 1:** Response time measured with one comparator input at VDD/2, while the other input transitions from VSS to VDD.

2: Comparator Hysteresis is available when the CxHYS bit of the CMxCON0 register is enabled.

### TABLE 31-13: DIGITAL-TO-ANALOG CONVERTER (DAC) SPECIFICATIONS

<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>DAC01</td>
<td>CLSB</td>
<td>Step Size</td>
<td>—</td>
<td>—</td>
<td>Vdd/32</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>DAC02</td>
<td>CACC</td>
<td>Absolute Accuracy</td>
<td>—</td>
<td>—</td>
<td>±1/2</td>
<td>LSb</td>
<td>VDD = 3.0V, TA = 25°C</td>
</tr>
<tr>
<td>DAC03</td>
<td>CR</td>
<td>Unit Resistor Value (R)</td>
<td>—</td>
<td>5K</td>
<td>—</td>
<td>Ω</td>
<td></td>
</tr>
<tr>
<td>DAC04*</td>
<td>CST</td>
<td>Settling Time (Note 1)</td>
<td>—</td>
<td>—</td>
<td>10</td>
<td>µs</td>
<td></td>
</tr>
</tbody>
</table>

* Parameter(s) characterized but not tested.

**Note 1:** Settling time measured while DACR<4:0> transitions from '0000' to '1111'.

### FIGURE 31-13: USART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

*Note:* Refer to Figure 31-4 for load conditions.
### TABLE 31-14: USART SYNCHRONOUS TRANSMISSION REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>US120</td>
<td>TckH2Dtv</td>
<td>SYNC XMIT (Master and Slave)</td>
<td>3.0-3.6V</td>
<td>—</td>
<td>80 ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Clock high to data-out valid</td>
<td>1.8-3.6V</td>
<td>—</td>
<td>100 ns</td>
<td></td>
</tr>
<tr>
<td>US121</td>
<td>TckRF</td>
<td>Clock out rise time and fall time (Master mode)</td>
<td>3.0-3.6V</td>
<td>—</td>
<td>45 ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1.8-3.6V</td>
<td>—</td>
<td>50 ns</td>
<td></td>
</tr>
<tr>
<td>US122</td>
<td>TdTRF</td>
<td>Data-out rise time and fall time</td>
<td>3.0-3.6V</td>
<td>—</td>
<td>45 ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1.8-3.6V</td>
<td>—</td>
<td>50 ns</td>
<td></td>
</tr>
</tbody>
</table>

### FIGURE 31-14: USART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

Note: Refer to Figure 31-4 for load conditions.

### TABLE 31-15: USART SYNCHRONOUS RECEIVE REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>US125</td>
<td>TdTV2Ckl</td>
<td>SYNC RCV (Master and Slave)</td>
<td>10</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Data-hold before CK ↓ (DT hold time)</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>US126</td>
<td>TckL2Dtl</td>
<td>Data-hold after CK ↓ (DT hold time)</td>
<td>15</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
</tbody>
</table>
FIGURE 31-15: SPI MASTER MODE TIMING (CKE = 0, SMP = 0)

Note: Refer to Figure 31-4 for load conditions.

FIGURE 31-16: SPI MASTER MODE TIMING (CKE = 1, SMP = 1)

Note: Refer to Figure 31-4 for load conditions.
FIGURE 31-17: SPI SLAVE MODE TIMING (CKE = 0)

Note: Refer to Figure 31-4 for load conditions.

FIGURE 31-18: SPI SLAVE MODE TIMING (CKE = 1)

Note: Refer to Figure 31-4 for load conditions.
## TABLE 31-16: SPI MODE REQUIREMENTS

<table>
<thead>
<tr>
<th>Param No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP70*</td>
<td>TssL2sCh, TssL2sCl</td>
<td>SSx↓ to SCKx↓ or SCKx↑ input</td>
<td></td>
<td>Tcy</td>
<td></td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP71*</td>
<td>TscH</td>
<td>SCKx input high time (Slave mode)</td>
<td>Tcy + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP72*</td>
<td>TscL</td>
<td>SCKx input low time (Slave mode)</td>
<td>Tcy + 20</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP73*</td>
<td>TdvV2sCh, TdvV2sCl</td>
<td>Setup time of SDIx data input to SCKx edge</td>
<td>100</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP74*</td>
<td>TscH2dIL, TscL2dIL</td>
<td>Hold time of SDIx data input to SCKx edge</td>
<td>100</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP75*</td>
<td>TdoR</td>
<td>SDO data output rise time</td>
<td>3.0-3.6V</td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1.8-3.6V</td>
<td>—</td>
<td>25</td>
<td>50</td>
<td>ns</td>
</tr>
<tr>
<td>SP76*</td>
<td>TdoF</td>
<td>SDOx data output fall time</td>
<td></td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
</tr>
<tr>
<td>SP77*</td>
<td>TssH2oOZ</td>
<td>SSx↑ to SDOx output high-impedance</td>
<td>10</td>
<td>—</td>
<td>—</td>
<td>50</td>
<td>ns</td>
</tr>
<tr>
<td>SP78*</td>
<td>TscR</td>
<td>SCKx output rise time (Master mode)</td>
<td>3.0-3.6V</td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1.8-3.6V</td>
<td>—</td>
<td>25</td>
<td>50</td>
<td>ns</td>
</tr>
<tr>
<td>SP79*</td>
<td>TscF</td>
<td>SCKx output fall time (Master mode)</td>
<td></td>
<td>—</td>
<td>10</td>
<td>25</td>
<td>ns</td>
</tr>
<tr>
<td>SP80*</td>
<td>TscH2oOV, TscL2oOV</td>
<td>SDOx data output valid after SCKx edge</td>
<td>3.0-3.6V</td>
<td>—</td>
<td>—</td>
<td>50</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>1.8-3.6V</td>
<td>—</td>
<td>—</td>
<td>145</td>
<td>ns</td>
</tr>
<tr>
<td>SP81*</td>
<td>TdOV2sCh, TdOV2sCl</td>
<td>SDOx data output setup to SCKx edge</td>
<td>Tcy</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
<tr>
<td>SP82*</td>
<td>TssL2oOV</td>
<td>SDOx data output valid after SS↓ edge</td>
<td></td>
<td>—</td>
<td>—</td>
<td>50</td>
<td>ns</td>
</tr>
<tr>
<td>SP83*</td>
<td>TscH2oSSh, TscL2oSSh</td>
<td>SSx↑ after SCKx edge</td>
<td>1.5Tcy + 40</td>
<td>—</td>
<td>—</td>
<td>ns</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.
† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.
FIGURE 31-19: **I²C™ BUS START/STOP BITS TIMING**

![I²C™ BUS START/STOP BITS TIMING Diagram]

Note: Refer to Figure 31-4 for load conditions.

TABLE 31-17: **I²C™ BUS START/STOP BITS REQUIREMENTS**

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP90*</td>
<td>TSU:STA</td>
<td>Start condition</td>
<td>100 kHz mode</td>
<td>4700</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Setup time</td>
<td>400 kHz mode</td>
<td>600</td>
<td>—</td>
<td>—</td>
<td>Only relevant for repeated Start condition</td>
</tr>
<tr>
<td>SP91*</td>
<td>THD:STA</td>
<td>Start condition</td>
<td>100 kHz mode</td>
<td>4000</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Hold time</td>
<td>400 kHz mode</td>
<td>600</td>
<td>—</td>
<td>—</td>
<td>After this period, the first clock pulse is generated</td>
</tr>
<tr>
<td>SP92*</td>
<td>TSU:STO</td>
<td>Stop condition</td>
<td>100 kHz mode</td>
<td>4700</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Setup time</td>
<td>400 kHz mode</td>
<td>600</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
<tr>
<td>SP93*</td>
<td>THD:STO</td>
<td>Stop condition</td>
<td>100 kHz mode</td>
<td>4000</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Hold time</td>
<td>400 kHz mode</td>
<td>600</td>
<td>—</td>
<td>—</td>
<td></td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

FIGURE 31-20: **I²C™ BUS DATA TIMING**

![I²C™ BUS DATA TIMING Diagram]

Note: Refer to Figure 31-4 for load conditions.
### TABLE 31-18: I²C™ BUS DATA REQUIREMENTS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>SP100*</td>
<td>THIGH</td>
<td>Clock high time</td>
<td>100 kHz mode</td>
<td>4.0</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>0.6</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>SSPx module</td>
<td>1.5TCY</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>SP101*</td>
<td>TLOW</td>
<td>Clock low time</td>
<td>100 kHz mode</td>
<td>4.7</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>1.3</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>SSPx module</td>
<td>1.5TCY</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>SP102*</td>
<td>TR</td>
<td>SDAx and SCLx rise time</td>
<td>100 kHz mode</td>
<td>—</td>
<td>1000</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>20 + 0.1C₆</td>
<td>300</td>
<td>ns</td>
</tr>
<tr>
<td>SP103*</td>
<td>TF</td>
<td>SDAx and SCLx fall time</td>
<td>100 kHz mode</td>
<td>—</td>
<td>250</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>20 + 0.1C₆</td>
<td>250</td>
<td>ns</td>
</tr>
<tr>
<td>SP106*</td>
<td>THD:DAT</td>
<td>Data input hold time</td>
<td>100 kHz mode</td>
<td>0</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>0</td>
<td>0.9</td>
<td>μs</td>
</tr>
<tr>
<td>SP107*</td>
<td>TSU:DAT</td>
<td>Data input setup time</td>
<td>100 kHz mode</td>
<td>250</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>100</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td>SP109*</td>
<td>TAA</td>
<td>Output valid from clock</td>
<td>100 kHz mode</td>
<td>—</td>
<td>3500</td>
<td>ns</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>—</td>
<td>—</td>
<td>ns</td>
</tr>
<tr>
<td>SP110*</td>
<td>TBUF</td>
<td>Bus free time</td>
<td>100 kHz mode</td>
<td>4.7</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>400 kHz mode</td>
<td>1.3</td>
<td>—</td>
<td>μs</td>
</tr>
<tr>
<td>SP111</td>
<td>C₆</td>
<td>Bus capacitive loading</td>
<td>—</td>
<td>—</td>
<td>400</td>
<td>pF</td>
</tr>
</tbody>
</table>

* These parameters are characterized but not tested.

**Note 1:** As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCLx to avoid unintended generation of Start or Stop conditions.

**Note 2:** A Fast mode (400 kHz) I²C™ bus device can be used in a Standard mode (100 kHz) I²C bus system, but the requirement TSU:DAT ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the low period of the SCLx signal. If such a device does stretch the low period of the SCLx signal, it must output the next data bit to the SDAx line TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I²C bus specification), before the SCLx line is released.
### TABLE 31-19: CAP SENSE OSCILLATOR SPECIFICATIONS

<table>
<thead>
<tr>
<th>Param. No.</th>
<th>Symbol</th>
<th>Characteristic</th>
<th>Min.</th>
<th>Typ†</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>CS01</td>
<td>ISRC</td>
<td>Current Source</td>
<td>High</td>
<td>—</td>
<td>—</td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Medium</td>
<td>—</td>
<td>-1.5</td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Low</td>
<td>—</td>
<td>-0.3</td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td>CS02</td>
<td>ISNK</td>
<td>Current Sink</td>
<td>High</td>
<td>7.5</td>
<td>—</td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Medium</td>
<td>1.5</td>
<td>—</td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Low</td>
<td>0.25</td>
<td>—</td>
<td>μA</td>
<td></td>
</tr>
<tr>
<td>CS03</td>
<td>VCTH</td>
<td>Cap Threshold</td>
<td>—</td>
<td>0.8</td>
<td>—</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>CS04</td>
<td>VCTL</td>
<td>Cap Threshold</td>
<td>—</td>
<td>0.4</td>
<td>—</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>CS05</td>
<td>VCHyst</td>
<td>CAP HYSTERESIS (VCTH - VCTL)</td>
<td>High</td>
<td>525</td>
<td>—</td>
<td>mV</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Medium</td>
<td>375</td>
<td>—</td>
<td>mV</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>Low</td>
<td>300</td>
<td>—</td>
<td>mV</td>
<td></td>
</tr>
</tbody>
</table>

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

### FIGURE 31-21: CAP SENSE OSCILLATOR

![Cap Sense Oscillator Diagram](image-url)
32.0 DC AND AC CHARACTERISTICS GRAPHS AND CHARTS

The graphs and tables provided in this section are for design guidance and are not tested.

In some graphs or tables, the data presented are outside specified operating range (i.e., outside specified VDD range). This is for information only and devices are ensured to operate properly only within the specified range.

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

“Typical” represents the mean of the distribution at 25°C. “MAXIMUM”, “Max.”, “MINIMUM” or “Min.” represents (mean + 3σ) or (mean - 3σ) respectively, where σ is a standard deviation, over each temperature range.
FIGURE 32-1:  \( \text{I} \text{DD}, \text{LP OSCILLATOR, } \text{Fosc} = 32 \text{ kHz} \)

Max: 85°C + 3\( \sigma \)
Typical: 25°C

Max.
Typical
FIGURE 32-2: $I_{DD}$ TYPICAL, XT AND EXTRC OSCILLATOR

Typical: 25°C

4 MHz EXTRC

4 MHz XT

1 MHz XT

FIGURE 32-3: $I_{DD}$ MAXIMUM, XT AND EXTRC OSCILLATOR

Max: 85°C + 3σ

4 MHz EXTRC

4 MHz XT

1 MHz XT
FIGURE 32-4: IDD, EXTERNAL CLOCK (ECL), LOW-POWER MODE, Fosc = 32 kHz

Max: 85°C + 3σ  
Typical: 25°C

FIGURE 32-5: IDD, EXTERNAL CLOCK (ECL), LOW-POWER MODE, Fosc = 500 kHz,

Max: 85°C + 3σ  
Typical: 25°C
FIGURE 32-6: \( I_{DD} \) TYPICAL, EXTERNAL CLOCK (ECM), MEDIUM-POWER MODE

Typical: 25°C

4 MHz

VDD (V)

\( I_{DD} \) (µA)

50

100

150

200

250

300

350

1.6 1.8 2.0 2.2 2.4 2.6 2.8 3.0 3.2 3.4 3.6 3.8

FIGURE 32-7: \( I_{DD} \) MAXIMUM, EXTERNAL CLOCK (ECM), MEDIUM-POWER MODE

Max: 85°C + 3σ

4 MHz

1 MHz

VDD (V)

\( I_{DD} \) (µA)

0

50

100

150

200

250

300

350

400

1.6 1.8 2.0 2.2 2.4 2.6 2.8 3.0 3.2 3.4 3.6 3.8
FIGURE 32-8: \( I_{DD} \) TYPICAL, EXTERNAL CLOCK (ECH), HIGH-POWER MODE,

![Typical: 25°C](image)

FIGURE 32-9: \( I_{DD} \) MAXIMUM, EXTERNAL CLOCK (ECH), HIGH-POWER MODE,

![Max: 85°C + 3\( \sigma \)](image)
FIGURE 32-10: $I_{DD}$, LFINTOSC, $F_{OSC} = 31$ kHz

FIGURE 32-11: $I_{DD}$ MFINTOSC, $F_{OSC} = 500$ kHz
FIGURE 32-12: IDD TYPICAL, HFINTOSC

![Graph showing IDD versus Vdd for different frequencies (4 MHz, 8 MHz, 16 MHz, 32 MHz) at different temperatures (25°C, 85°C + 3σ)].

FIGURE 32-13: IDD MAXIMUM, HFINTOSC

![Graph showing IDD versus Vdd for different frequencies (4 MHz, 8 MHz, 16 MHz, 32 MHz) at different temperatures (Max: 85°C + 3σ)].
FIGURE 32-14:  \( I_{DD} \) TYPICAL, HS OSCILLATOR

FIGURE 32-15:  \( I_{DD} \) MAXIMUM, HS OSCILLATOR
FIGURE 32-16: \( I_{PD} \) BASE, SLEEP MODE

![Graph of \( I_{PD} \) vs. \( V_{DD} \)](image)

- **Max:** \( 85°C + 3\sigma \)
- **Typical:** \( 25°C \)

FIGURE 32-17: \( I_{PD} \), WATCHDOG TIMER (WDT)

![Graph of \( I_{PD} \) vs. \( V_{DD} \)](image)

- **Max:** \( 85°C + 3\sigma \)
- **Typical:** \( 25°C \)
FIGURE 32-18: IPD, FIXED VOLTAGE REFERENCE (FVR)

Max: 85°C + 3σ
Typical: 25°C

FIGURE 32-19: IPD, BROWN-OUT RESET (BOR), BORV = 1

Max: 85°C + 3σ
Typical: 25°C
FIGURE 32-20: IPD, SECONDARY OSCILLATOR, Fosc = 32 kHz

Max: 85°C + 3σ
Typical: 25°C

IPD (µA)

VDD (V)

0.0
1.6 1.8 2.0 2.2 2.4 2.6 2.8 3.0 3.2 3.4 3.6 3.8

Typical

Max.
FIGURE 32-21: $V_{OH}$ vs. $I_{OH}$ OVER TEMPERATURE, $V_{DD} = 3.0V$

![Graph of $V_{OH}$ vs. $I_{OH}$ with temperature variations]

Max: $125^\circ C + 3\sigma$
Typical: $25^\circ C$
Min: $-40^\circ C - 3\sigma$

FIGURE 32-22: $V_{OL}$ vs. $I_{OL}$, OVER TEMPERATURE, $V_{DD} = 3.0V$

![Graph of $V_{OL}$ vs. $I_{OL}$ with temperature variations]

Max: $125^\circ C + 3\sigma$
Typical: $25^\circ C$
Min: $-40^\circ C - 3\sigma$
FIGURE 32-23: \( V_{OH} \) vs. \( I_{OH} \) OVER TEMPERATURE, \( V_{DD} = 1.8V \)

![Graph showing \( V_{OH} \) vs. \( I_{OH} \)]

FIGURE 32-24: \( V_{OL} \) vs. \( I_{OL} \) OVER TEMPERATURE, \( V_{DD} = 1.8V \)

![Graph showing \( V_{OL} \) vs. \( I_{OL} \)]
FIGURE 32-25: POR RELEASE VOLTAGE

<table>
<thead>
<tr>
<th>Voltage (V)</th>
<th>Max.</th>
<th>Typical</th>
<th>Min.</th>
</tr>
</thead>
<tbody>
<tr>
<td>1.50</td>
<td>1.52</td>
<td>1.54</td>
<td>1.56</td>
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<tr>
<td>1.58</td>
<td>1.60</td>
<td>1.62</td>
<td>1.64</td>
</tr>
<tr>
<td>1.66</td>
<td>1.68</td>
<td>1.70</td>
<td></td>
</tr>
</tbody>
</table>

-60 -40 -20 0 20 40 60 80 100 120 140
Temperature (°C)

Max: Typical + 3σ
Typical: 25°C
Min: Typical - 3σ

FIGURE 32-26: POR REARM VOLTAGE, NORMAL-POWER MODE

<table>
<thead>
<tr>
<th>Voltage (V)</th>
<th>Max.</th>
<th>Typical</th>
<th>Min.</th>
</tr>
</thead>
<tbody>
<tr>
<td>0.60</td>
<td>0.70</td>
<td>0.80</td>
<td>0.90</td>
</tr>
<tr>
<td>1.00</td>
<td>1.10</td>
<td>1.20</td>
<td></td>
</tr>
<tr>
<td>1.30</td>
<td>1.40</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

-60 -40 -20 0 20 40 60 80 100 120 140
Temperature (°C)

Max: Typical + 3σ
Typical: 25°C
Min: Typical - 3σ
FIGURE 32-27: BROWN-OUT RESET VOLTAGE, BORV = 1

FIGURE 32-28: BROWN-OUT RESET HYSTERESIS, BORV = 1
FIGURE 32-29:  BROWN-OUT RESET VOLTAGE, BORV = 0

![Graph showing BROWN-OUT RESET VOLTAGE, BORV = 0 with temperature ranging from -60°C to 140°C on the x-axis and voltage ranging from 2.55V to 2.80V on the y-axis.](image)

Max: Typical + 3σ
Min: Typical - 3σ

FIGURE 32-30:  BROWN-OUT RESET HYSTERESIS, BORV = 0

![Graph showing BROWN-OUT RESET HYSTERESIS, BORV = 0 with temperature ranging from -60°C to 140°C on the x-axis and voltage ranging from 0mV to 90mV on the y-axis.](image)

Max: Typical + 3σ
Typical: 25°C
Min: Typical - 3σ
FIGURE 32-31: LOW-POWER BROWN-OUT RESET VOLTAGE, LPBOR = 0

![Low-power brown-out reset voltage graph]

Typical
Max: Typical + 3σ
Min: Typical - 3σ

FIGURE 32-32: LOW-POWER BROWN-OUT RESET HYSTERESIS, LPBOR = 0

![Low-power brown-out reset hysteresis graph]

Max: Typical + 3σ
Typical: 25°C
Min: Typical - 3σ
FIGURE 32-33: WDT TIME-OUT PERIOD

FIGURE 32-34: PWRT PERIOD
FIGURE 32-35: FVR STABILIZATION PERIOD

Note:
The FVR Stabilization Period applies when:
1) coming out of RESET or exiting Sleep mode for PIC12/16LFxxxx devices.
2) when exiting sleep mode with VREGPM = 1 for PIC12/16Fxxxx devices
In all other cases, the FVR is stable when released from RESET.

Max: Typical + 3σ
Typical: statistical mean @ 25°C
FIGURE 32-36: COMPARATOR HYSTERESIS, NORMAL-POWER MODE (CxSP = 1, CxHYS = 1)

FIGURE 32-37: COMPARATOR HYSTERESIS, LOW-POWER MODE (CxSP = 0, CxHYS = 1)
FIGURE 32-38: COMPARATOR RESPONSE TIME, NORMAL-POWER MODE (CxSP = 1)

FIGURE 32-39: COMPARATOR RESPONSE TIME OVER TEMPERATURE, NORMAL-POWER MODE (CxSP = 1)
FIGURE 32-40: LFINTOSC FREQUENCY OVER VDD AND TEMPERATURE

Max: Typical + 3σ (-40°C to +125°C)
Typical: statistical mean @ 25°C
Min: Typical - 3σ (-40°C to +125°C)
FIGURE 32-41: CAP SENSE CURRENT SINK/SOURCE CHARACTERISTICS,
FIXED VOLTAGE REFERENCE (CPSRM = 0),
HIGH CURRENT RANGE (CPSRNG = 11)

FIGURE 32-42: CAP SENSE CURRENT SINK/SOURCE CHARACTERISTICS,
FIXED VOLTAGE REFERENCE (CPSRM = 0),
MEDIUM CURRENT RANGE (CPSRNG = 10)
FIGURE 32-43: CAP SENSE CURRENT SINK/SOURCE CHARACTERISTICS, FIXED VOLTAGE REFERENCE (CPSRM = 0), LOW CURRENT RANGE (CPSRNG = 01)

<table>
<thead>
<tr>
<th>VDD (V)</th>
<th>IPIN (uA)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1.6</td>
<td>-1.0</td>
</tr>
<tr>
<td>1.8</td>
<td>-0.8</td>
</tr>
<tr>
<td>2.0</td>
<td>-0.6</td>
</tr>
<tr>
<td>2.2</td>
<td>-0.4</td>
</tr>
<tr>
<td>2.4</td>
<td>-0.2</td>
</tr>
<tr>
<td>2.6</td>
<td>0.0</td>
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<td>3.4</td>
<td>0.8</td>
</tr>
<tr>
<td>3.6</td>
<td>1.0</td>
</tr>
<tr>
<td>3.8</td>
<td>1.2</td>
</tr>
</tbody>
</table>

Max: Typical + 3σ
Typical:
Min: Typical - 3σ
33.0 DEVELOPMENT SUPPORT

The PIC® microcontrollers (MCU) and dsPIC® digital signal controllers (DSC) are supported with a full range of software and hardware development tools:

- Integrated Development Environment
  - MPLAB® X IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB XC Compiler
  - MPASMTM Assembler
  - MPLINK™ Object Linker/
    MPLIB™ Object Librarian
  - MPLAB Assembler/Linker/Librarian for
    Various Device Families
- Simulators
  - MPLAB X SIM Software Simulator
- Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers/Programmers
  - MPLAB ICD 3
  - PICkit™ 3
- Device Programmers
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards,
  Evaluation Kits and Starter Kits
- Third-party development tools

33.1 MPLAB X Integrated Development Environment Software

The MPLAB X IDE is a single, unified graphical user interface for Microchip and third-party software, and hardware development tool that runs on Windows®, Linux and Mac OS® X. Based on the NetBeans IDE, MPLAB X IDE is an entirely new IDE with a host of free software components and plug-ins for high-performance application development and debugging. Moving between tools and upgrading from software simulators to hardware debugging and programming tools is simple with the seamless user interface.

With complete project management, visual call graphs, a configurable watch window and a feature-rich editor that includes code completion and context menus, MPLAB X IDE is flexible and friendly enough for new users. With the ability to support multiple tools on multiple projects with simultaneous debugging, MPLAB X IDE is also suitable for the needs of experienced users.

Feature-Rich Editor:

- Color syntax highlighting
- Smart code completion makes suggestions and provides hints as you type
- Automatic code formatting based on user-defined rules
- Live parsing

User-Friendly, Customizable Interface:

- Fully customizable interface: toolbars, toolbar buttons, windows, window placement, etc.
- Call graph window

Project-Based Workspaces:

- Multiple projects
- Multiple tools
- Multiple configurations
- Simultaneous debugging sessions

File History and Bug Tracking:

- Local file history feature
- Built-in support for Bugzilla issue tracker
33.2 MPLAB XC Compilers

The MPLAB XC Compilers are complete ANSI C compilers for all of Microchip’s 8, 16, and 32-bit MCU and DSC devices. These compilers provide powerful integration capabilities, superior code optimization and ease of use. MPLAB XC Compilers run on Windows, Linux or MAC OS X.

For easy source level debugging, the compilers provide debug information that is optimized to the MPLAB X IDE.

The free MPLAB XC Compiler editions support all devices and commands, with no time or memory restrictions, and offer sufficient code optimization for most applications.

MPLAB XC Compilers include an assembler, linker and utilities. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. MPLAB XC Compiler uses the assembler to produce its object file. Notable features of the assembler include:

- Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility

33.3 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel® standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code, and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB X IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multipurpose source files
- Directives that allow complete control over the assembly process

33.4 MPLINK Object Linker/
MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

33.5 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC DSC devices. MPLAB XC Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility
33.6 MPLAB X SIM Software Simulator

The MPLAB X SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB X SIM Software Simulator fully supports symbolic debugging using the MPLAB XC Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

33.7 MPLAB REAL ICE In-Circuit Emulator System

The MPLAB REAL ICE In-Circuit Emulator System is Microchip’s next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs all 8, 16 and 32-bit MCU, and DSC devices with the easy-to-use, powerful graphical user interface of the MPLAB X IDE.

The emulator is connected to the design engineer’s PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ-11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB X IDE. MPLAB REAL ICE offers significant advantages over competitive emulators including full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, logic probes, a ruggedized probe interface and long (up to three meters) interconnection cables.

33.8 MPLAB ICD 3 In-Circuit Debugger System

The MPLAB ICD 3 In-Circuit Debugger System is Microchip’s most cost-effective, high-speed hardware debugger/programmer for Microchip Flash DSC and MCU devices. It debugs and programs PIC Flash microcontrollers and dsPIC DSCs with the powerful, yet easy-to-use graphical user interface of the MPLAB IDE.

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer’s PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

33.9 PICkit 3 In-Circuit Debugger/Programmer

The MPLAB PICkit 3 allows debugging and programming of PIC and dsPIC Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB IDE. The MPLAB PICkit 3 is connected to the design engineer’s PC using a full-speed USB interface and can be connected to the target via a Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the Reset line to implement in-circuit debugging and In-Circuit Serial Programming™ (ICSP™).

33.10 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages, and a modular, detachable socket assembly to support various package types. The ICSP cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices, and incorporates an MMC card for file storage and data applications.
33.11 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM™ and dsPICDEM™ demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELoo® security ICs, CAN, IrDA®, PowerSmart battery management, SEEVAL® evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

33.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent® and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika®
34.0 PACKAGING INFORMATION

34.1 Package Marking Information

Legend:

- **XX.X** Customer-specific information
- **Y** Year code (last digit of calendar year)
- **YY** Year code (last 2 digits of calendar year)
- **WW** Week code (week of January 1 is week '01')
- **NNN** Alphanumeric traceability code
- **(e3)** Pb-free JEDEC® designator for Matte Tin (Sn)
- **(*)** This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

**Note:** In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

* Standard PICmicro® device marking consists of Microchip part number, year code, week code and traceability code. For PICmicro device marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.
34.2 Package Details
The following sections give the technical details of the packages.

14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging
14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com-packaging

<table>
<thead>
<tr>
<th>Units</th>
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</tr>
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<tbody>
<tr>
<td>Dimension</td>
<td>MIN</td>
</tr>
<tr>
<td>Number of Pins</td>
<td>N</td>
</tr>
<tr>
<td>Pitch</td>
<td>e</td>
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<tr>
<td>Overall Height</td>
<td>A</td>
</tr>
<tr>
<td>Molded Package Thickness</td>
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<td>Standoff</td>
<td>A1</td>
</tr>
<tr>
<td>Overall Width</td>
<td>E</td>
</tr>
<tr>
<td>Molded Package Width</td>
<td>E1</td>
</tr>
<tr>
<td>Molded Package Length</td>
<td>D</td>
</tr>
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<td>c</td>
</tr>
<tr>
<td>Lead Width</td>
<td>b</td>
</tr>
</tbody>
</table>

Notes:
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
3. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.
   REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing No. C04-087C Sheet 2 of 2
14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at [http://www.microchip.com/packaging](http://www.microchip.com/packaging)

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**RECOMMENDED LAND PATTERN**

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<thead>
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<tbody>
<tr>
<td>Dimension Limits</td>
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<tr>
<td>Contact Pitch</td>
<td>E</td>
</tr>
<tr>
<td>Contact Pad Spacing</td>
<td>C1</td>
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<tr>
<td>Contact Pad Width (X14)</td>
<td>X1</td>
</tr>
<tr>
<td>Contact Pad Length (X14)</td>
<td>Y1</td>
</tr>
<tr>
<td>Distance Between Pads</td>
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</tr>
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</table>

Notes:
1. Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2087A
APPENDIX A: DATA SHEET

REVISION HISTORY

Revision A (05/2012)
Initial release of this data sheet.

Revision B (07/2014)
Updated the Low-Power Features section.

APPENDIX B: MIGRATING FROM OTHER PIC® DEVICES

This section provides comparisons when migrating from other similar PIC® devices to the PIC12LF1840T39A family of devices.

B.1 PIC12F675 to PIC12LF1840T39A

TABLE B-1: FEATURE COMPARISON

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<thead>
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<th>Feature</th>
<th>PIC12F675</th>
<th>PIC12LF1840T39A</th>
</tr>
</thead>
<tbody>
<tr>
<td>Max. Operating Speed</td>
<td>20 MHz</td>
<td>32 MHz</td>
</tr>
<tr>
<td>Max. Program Memory (Words)</td>
<td>1K</td>
<td>4K</td>
</tr>
<tr>
<td>Max. SRAM (Bytes)</td>
<td>64</td>
<td>256</td>
</tr>
<tr>
<td>Max. EEPROM (Bytes)</td>
<td>128</td>
<td>256</td>
</tr>
<tr>
<td>A/D Resolution</td>
<td>10-bit</td>
<td>10-bit</td>
</tr>
<tr>
<td>Timers (8/16-bit)</td>
<td>1/1</td>
<td>2/1</td>
</tr>
<tr>
<td>Brown-out Reset</td>
<td>Y</td>
<td>Y</td>
</tr>
<tr>
<td>Internal Pull-ups</td>
<td>GP&lt;5:4&gt;, GP&lt;2:0&gt;</td>
<td>RA&lt;5:0&gt;</td>
</tr>
<tr>
<td>Interrupt-on-change</td>
<td>GP&lt;5:0&gt;</td>
<td>RA&lt;5:0&gt;, Edge Selectable</td>
</tr>
<tr>
<td>Comparator</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>EUSART</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>Extended WDT</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>Software Control Option of WDT/BOR</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>INTOSC Frequencies</td>
<td>4 MHz</td>
<td>31 kHz - 32 MHz</td>
</tr>
<tr>
<td>Clock Switching</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>Capacitive Sensing</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>CCP/ECCP</td>
<td>0/0</td>
<td>0/1</td>
</tr>
<tr>
<td>Enhanced PIC16 CPU</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>MSSPx/SSPx</td>
<td>0/0</td>
<td>Y</td>
</tr>
<tr>
<td>Reference Clock</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>Data Signal Modulator</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>SR Latch</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>Voltage Reference</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>DAC</td>
<td>N</td>
<td>Y</td>
</tr>
<tr>
<td>RF Range (MHz)</td>
<td>290 - 350 (K) 850 - 930 (H)</td>
<td>418 - 443.92, 864 - 869.85</td>
</tr>
<tr>
<td>Modulation</td>
<td>OOK, FSK</td>
<td>OOK, FSK</td>
</tr>
<tr>
<td>RF Output Power (dBm)</td>
<td>-70, -12, -4, +2, +9</td>
<td>0, +10</td>
</tr>
</tbody>
</table>
Note 1: This device has been designed to perform to the parameters of its data sheet. It has been tested to an electrical specification designed to determine its conformance with these parameters. Due to process differences in the manufacture of this device, this device may have different performance characteristics than its earlier version. These differences may cause this device to perform differently in your application than the earlier version of this device.

Note 1: The user should verify that the device oscillator starts and performs as expected. Adjusting the loading capacitor values and/or the Oscillator mode may be required.
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<table>
<thead>
<tr>
<th>PART NO.</th>
<th>Device</th>
<th>Tape and Reel Option</th>
<th>Temperature Range</th>
<th>Package</th>
<th>Pattern</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td>PIC12LF1840T39A</td>
<td>Blank = Standard packaging (tube or tray)</td>
<td>I = -40°C to +85°C (Industrial)</td>
<td>ST = TSSOP</td>
<td>QTP, SQTP, Code or Special Requirements (blank otherwise)</td>
</tr>
</tbody>
</table>

Examples:

a) PIC12LF1840T39AT - I/ST 301 Tape and Reel, Industrial temperature, TSSOP package

Note 1: Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.
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ISBN: 978-1-63276-430-0

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