**Features**

- Low Supply Current: 80 µA (Max)
- Low Dropout Voltage: 140 mV (Typ.) @ 150 mA
- High-Output Voltage Accuracy: ±0.4% (Typ.)
- Standard or Custom Output Voltages
- Power-Saving Shutdown Mode
- Reference Bypass Input for Ultra Low-Noise Operation
- Fast Shutdown Response Time: 60 µsec (Typ.)
- Overcurrent and Overtemperature Protection
- Space-Saving 5-Pin SOT-23A Package
- Pin- Compatible Upgrades for Bipolar Regulators
- Wide Operating Temperature Range: -40°C to +125°C
- Standard Output Voltage Options: - 1.8V, 2.5V, 2.6V, 2.7V, 2.8V, 2.85V, 3.0V, 3.3V, 5.0V

**Applications**

- Battery-Operated Systems
- Portable Computers
- Medical Instruments
- Instrumentation
- Cellular/GSM/PHS Phones
- Linear Post-Regulator for SMPS
- Pagers

**Related Literature**

- Application Notes: AN765, AN766, AN776 and AN792

**Package Type**

5-Pin SOT-23A

---

**General Description**

The TC2014, TC2015 and TC2185 are high-accuracy (typically ±0.4%) CMOS upgrades for bipolar Low Drop-out Regulators (LDOs), such as the LP2980. Total supply current is typically 55 µA; 20 to 60 times lower than in bipolar regulators.

The key features of the device include low noise operation (plus bypass reference), low dropout voltage – typically 45 mV for the TC2014, 90 mV for the TC2015, and 140 mV for the TC2185, at full load – and fast response to step changes in load. Supply current is reduced to 0.5 µA (max) and VOUT falls to zero when the shutdown input is low. These devices also incorporate overcurrent and overtemperature protection.

The TC2014, TC2015 and TC2185 are stable with an output capacitor of 1 µF and have maximum output currents of 50 mA, 100 mA and 150 mA, respectively. For higher-output current versions, see the TC1107 (DS21356), TC1108 (DS21357) and TC1173 (DS21362) (IOUT = 300 mA) data sheets.

**Typical Application**

![Typical Application Diagram]
1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Input Voltage ......................................................... 7.0V
Output Voltage ........................................... (– 0.3) to (V IN + 0.3)
Operating Temperature ......................... – 40°C < TJ < 125°C
Storage Temperature................................ – 65°C to +150°C
Maximum Voltage on Any Pin ................ VIN +0.3V to – 0.3V
Maximum Junction Temperature ..................... 150°C

† Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operation sections of the specifications is not implied. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

Electrical Specifications: Unless otherwise specified, VIN = VR + 1V, IL = 100 µA, COUT = 3.3 µF, SHDN > VIH, TA = +25°C.

BOLDFACE type specifications apply for junction temperature of -40°C to +125°C.

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min.</th>
<th>Typ</th>
<th>Max.</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input Operating Voltage</td>
<td>VIN</td>
<td>2.7</td>
<td>6.0</td>
<td></td>
<td>V</td>
<td>Note 1</td>
</tr>
<tr>
<td>Maximum Output Current</td>
<td>IOUTMAX</td>
<td>50</td>
<td></td>
<td></td>
<td>mA</td>
<td>TC2014</td>
</tr>
<tr>
<td></td>
<td></td>
<td>100</td>
<td></td>
<td></td>
<td></td>
<td>TC2015</td>
</tr>
<tr>
<td></td>
<td></td>
<td>150</td>
<td></td>
<td></td>
<td></td>
<td>TC2185</td>
</tr>
<tr>
<td>Output Voltage</td>
<td>VOUT</td>
<td>VR – 2.0%</td>
<td>VR ± 0.4%</td>
<td>VR + 2.0%</td>
<td>V</td>
<td>Note 2</td>
</tr>
<tr>
<td>VOUT Temperature Coefficient</td>
<td>TCVOUT</td>
<td>—</td>
<td>20</td>
<td>—</td>
<td>ppm/°C</td>
<td>Note 3</td>
</tr>
<tr>
<td>Line Regulation</td>
<td>ΔVOUT/ΔVIN</td>
<td>—</td>
<td>0.05</td>
<td>0.5</td>
<td>%</td>
<td></td>
</tr>
<tr>
<td>Load Regulation (Note 4)</td>
<td>ΔVOUT/VOUT</td>
<td>-1.0</td>
<td>0.33</td>
<td>+1.0</td>
<td>%</td>
<td>TC2014,TC2015: IL = 0.1 mA to IOUTMAX</td>
</tr>
<tr>
<td></td>
<td></td>
<td>-2.0</td>
<td>0.43</td>
<td>+2.0</td>
<td></td>
<td>TC2185: IL = 0.1 mA to IOUTMAX (Note 4)</td>
</tr>
<tr>
<td>Dropout Voltage</td>
<td>VIN – VOUT</td>
<td>—</td>
<td>2</td>
<td>—</td>
<td>mV</td>
<td>Note 5</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
<td>45</td>
<td>70</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
<td>90</td>
<td>140</td>
<td></td>
<td>TC2015, TC2185: IL = 100 mA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>—</td>
<td>140</td>
<td>210</td>
<td></td>
<td>TC2185: IL = 150 mA</td>
</tr>
<tr>
<td>Supply Current</td>
<td>IIN</td>
<td>—</td>
<td>55</td>
<td>80</td>
<td>µA</td>
<td>SHDN = VIH, IL = 0</td>
</tr>
<tr>
<td>Shutdown Supply Current</td>
<td>IINSD</td>
<td>—</td>
<td>0.05</td>
<td>0.5</td>
<td>µA</td>
<td>SHDN = 0V</td>
</tr>
<tr>
<td>Power Supply Rejection Ratio</td>
<td>PSRR</td>
<td>—</td>
<td>55</td>
<td>—</td>
<td>dB</td>
<td>F ≤ 1 kHz, Cbypass = 0.01 µF</td>
</tr>
<tr>
<td>Output Short Circuit Current</td>
<td>IOUTSC</td>
<td>—</td>
<td>160</td>
<td>300</td>
<td>mA</td>
<td>VOUT = 0V</td>
</tr>
</tbody>
</table>

Note 1: The minimum VIN has to meet two conditions: VIN = 2.7V and VIN = VR + VDROPOUT.
Note 2: VR is the regulator output voltage setting. For example: VR = 1.8V, 2.7V, 2.8V, 2.85V, 3.0V, 3.3V.
Note 3: TCVOUT = \( \frac{(V_{OUTMAX} - V_{OUTMIN}) \times 10^{-6}}{V_{OUT} \times \Delta T} \)
Note 4: Load regulation is tested over a load range from 1.0 mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the Thermal Regulation specification.
Note 5: Dropout Voltage is defined as the input-to-output differential at which the output voltage drops 2% below its nominal value.
Note 6: Thermal Regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to IMAX at VIN = 6V for T = 10 ms.
Note 7: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction-to-air (i.e. \( T_A, T_J, \theta_{JA} \)).
Note 8: Time required for VOUT to reach 95% of VR (output voltage setting), after VSHDN is switched from 0 to VIN.
**TEMPERATURE CHARACTERISTICS**

**ELECTRICAL CHARACTERISTICS (CONTINUED)**

Electrical Specifications: Unless otherwise specified, \( V_{IN} = V_R + 1V, I_L = 100 \mu A, C_{OUT} = 3.3 \mu F \), \( SHDN > V_{IH}, T_A = +25°C \). BOLDFACE type specifications apply for junction temperature of -40°C to +125°C.

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Thermal Regulation</td>
<td>( \Delta V_{OUT}/\Delta P_D )</td>
<td>—</td>
<td>0.04</td>
<td>—</td>
<td>V/W</td>
<td>Note 6, Note 7</td>
</tr>
<tr>
<td>Thermal Shutdown Die Temperature</td>
<td>( T_{SD} )</td>
<td>—</td>
<td>160</td>
<td>—</td>
<td>°C</td>
<td></td>
</tr>
<tr>
<td>Output Noise</td>
<td>( eN )</td>
<td>—</td>
<td>200</td>
<td>—</td>
<td>nV/\sqrt{Hz} ( I_L = I_{OUTMAX}, F = 10 \text{ kHz} )</td>
<td></td>
</tr>
<tr>
<td>Response Time (from Shutdown Mode) (Note 8)</td>
<td>( T_R )</td>
<td>—</td>
<td>60</td>
<td>—</td>
<td>µs</td>
<td></td>
</tr>
</tbody>
</table>

**SHDN Input**

| SHDN Input High Threshold | \( V_{IH} \) | 60 | —   | —   | %\( V_{IN} \) |
| SHDN Input Low Threshold  | \( V_{IL} \) | —  | —   | 15  | %\( V_{IN} \) |

**Note 1:** The minimum \( V_{IN} \) has to meet two conditions: \( V_{IN} = 2.7V \) and \( V_{IN} = V_R + V_{DROPOUT} \).

**Note 2:** \( V_R \) is the regulator output voltage setting. For example: \( V_R = 1.8V, 2.7V, 2.8V, 2.85V, 3.0V, 3.3V \).

**Note 3:**

\[
TCV_{OUT} = \left( \frac{V_{OUTMAX} - V_{OUTMIN}}{V_{OUT} \times \Delta T} \right) \times 10^{-6}
\]

**Note 4:** Regulation is measured at a constant junction temperature using low duty cycle pulse testing. Load regulation is tested over a load range from 1.0 mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the Thermal Regulation specification.

**Note 5:** Dropout Voltage is defined as the input-to-output differential at which the output voltage drops 2% below its nominal value.

**Note 6:** Thermal Regulation is defined as the change in output voltage at a time \( T \) after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to \( I_{MAX} \) at \( V_{IN} = 6V \) for \( T = 10 \text{ ms} \).

**Note 7:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction-to-air (i.e. \( T_A, T_J, \theta_{JA} \)).

**Note 8:** Time required for \( V_{OUT} \) to reach 95% of \( V_R \) (output voltage setting), after \( V_{SHDN} \) is switched from 0 to \( V_{IN} \).

**TEMPERATURE CHARACTERISTICS**

Electrical Specifications: Unless otherwise noted, \( V_{DD} = +2.7V \) to +6.0V and \( V_{SS} = \text{GND} \).

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Sym</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
<th>Conditions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Temperature Ranges:</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Extended Temperature Range</td>
<td>( T_A )</td>
<td>-40</td>
<td>—</td>
<td>+125</td>
<td>°C</td>
<td></td>
</tr>
<tr>
<td>Operating Temperature Range</td>
<td>( T_A )</td>
<td>-40</td>
<td>—</td>
<td>+125</td>
<td>°C</td>
<td></td>
</tr>
<tr>
<td>Storage Temperature Range</td>
<td>( T_A )</td>
<td>-65</td>
<td>—</td>
<td>+150</td>
<td>°C</td>
<td></td>
</tr>
<tr>
<td>Thermal Package Resistances:</td>
<td>( \theta_{JA} )</td>
<td>—</td>
<td>255</td>
<td>—</td>
<td>°C/W</td>
<td></td>
</tr>
</tbody>
</table>

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2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise indicated, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_{OUT} = 3.3 \mu F$, $SHDN > V_{IH}$, $T_A = +25^\circ C$.

**FIGURE 2-1:** Supply Current vs. Junction Temperature.

**FIGURE 2-2:** Load Regulation vs. Supply Voltage.

**FIGURE 2-3:** Output Voltage vs. Junction Temperature.

**FIGURE 2-4:** Output Voltage vs. Junction Temperature.

**FIGURE 2-5:** Output Voltage vs. Supply Voltage.

**FIGURE 2-6:** Dropout Voltage vs. Junction Temperature.

Note: Dropout Voltage is not a tested parameter for 1.8V. $VIN_{(min)} = 2.7V$. 

![Graphs and tables]

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Note: Unless otherwise indicated, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_{OUT} = 3.3 \mu F$, $SHDN > V_{IH}$, $T_A = +25^\circ C$.

**FIGURE 2-7:** Supply Current vs. Junction Temperature.

**FIGURE 2-8:** Load Regulation vs. Supply Voltage.

**FIGURE 2-9:** Output Voltage vs. Junction Temperature.

**FIGURE 2-10:** Output Voltage vs. Junction Temperature.

**FIGURE 2-11:** Output Voltage vs. Supply Voltage.

**FIGURE 2-12:** Dropout Voltage vs. Junction Temperature.
Note: Unless otherwise indicated, \( V_{\text{IN}} = V_R + 1V, \ L = 100 \mu A, \ C_{\text{OUT}} = 3.3 \mu F, \ SHDN > V_{\text{IH}}, \ T_A = +25^\circ C. \)

**FIGURE 2-13:** Supply Current vs. Junction Temperature.

**FIGURE 2-14:** Output Voltage vs. Junction Temperature. (\( C_{\text{OUT}} = 1 \mu F \).)

**FIGURE 2-15:** Load Regulation vs. Junction Temperature.

**FIGURE 2-16:** Dropout Voltage vs. Junction Temperature.

**FIGURE 2-17:** Load Transient Response. (\( C_{\text{OUT}} = 1 \mu F \).)

**FIGURE 2-18:** Load Transient Response. (\( C_{\text{OUT}} = 10 \mu F \).)
Note: Unless otherwise indicated, $V_{IN} = V_R + 1V$, $I_L = 100 \mu A$, $C_{OUT} = 3.3 \mu F$, $\overline{SHDN} > V_{IH}$, $T_A = +25^\circ C$.

**FIGURE 2-19:** Line Transient Response. ($C_{OUT} = 1 \mu F$).

**FIGURE 2-20:** Load Transient Response in Dropout. ($C_{OUT} = 10 \mu F$).

**FIGURE 2-21:** Shutdown Delay Time.

**FIGURE 2-22:** Wake-Up Response.

**FIGURE 2-23:** PSRR vs. Frequency ($C_{OUT} = 1 \mu F$ Ceramic).

**FIGURE 2-24:** PSRR vs. Frequency ($C_{OUT} = 10 \mu F$ Ceramic).
Note: Unless otherwise indicated, \( V_{IN} = V_R + 1V, I_L = 100 \mu A, C_{OUT} = 3.3 \mu F, \text{SHDN} > V_{IH}, T_A = +25^\circ C. \)

**FIGURE 2-25:** PSRR vs. Frequency

\( C_{OUT} = 10 \mu F \text{Tantalum} \).

**FIGURE 2-26:** Output Noise vs. Frequency.

\( V_{IN} = 4.0V \)
\( V_{OUTAC} = 100 \text{mV} \)
\( V_{OUTDC} = 3.0V \)
\( I_{OUT} = 150 mA \)
\( CBYPASS = 0.01 \mu F \)
\( CBYPASS = 0 \mu F \)
\( C_{OUT} = 10 \mu F \text{Tantalum} \).

\( VIN = 4.0V \)
\( VOUTDC = 3.0V \)
\( IOUT = 100 \mu A \)
\( COUT = 1 \mu F \)
\( COUT = 10 \mu F \).
3.0 PIN DESCRIPTIONS

The descriptions of the pins are described in Table 3-1.

<table>
<thead>
<tr>
<th>Pin No.</th>
<th>Symbol</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>V_{IN}</td>
<td>Unregulated supply input</td>
</tr>
<tr>
<td>2</td>
<td>GND</td>
<td>Ground terminal</td>
</tr>
<tr>
<td>3</td>
<td>SHDN</td>
<td>Shutdown control input</td>
</tr>
<tr>
<td>4</td>
<td>Bypass</td>
<td>Reference bypass input</td>
</tr>
<tr>
<td>5</td>
<td>V_{OUT}</td>
<td>Regulated voltage output</td>
</tr>
</tbody>
</table>

3.1 Unregulated Supply Input (V_{IN})

Connect the unregulated input supply to the V_{IN} pin. If there is a large distance between the input supply and the LDO regulator, some input capacitance is necessary for proper operation. A 1 µF capacitor, connected from V_{IN} to ground, is recommended for most applications.

3.2 Ground Terminal (GND)

Connect the unregulated input supply ground return to GND. Also connect one side of the 1 µF typical input decoupling capacitor close to this pin and one side of the output capacitor C_{OUT} to this pin.

3.3 Shutdown Control Input (SHDN)

The regulator is fully enabled when a logic-high is applied to SHDN. The regulator enters shutdown when a logic-low is applied to this input. During shutdown, the output voltage falls to zero and the supply current is reduced to 0.5 µA (max).

3.4 Reference Bypass Input (Bypass)

Connecting a low-value ceramic capacitor to Bypass will further reduce output voltage noise and improve the Power Supply Ripple Rejection (PSRR) performance of the LDO. Typical values from 470 pF to 0.01 µF are suggested. While smaller and larger values can be used, these affect the speed at which the LDO output voltage rises when input power is applied. The larger the bypass capacitor, the slower the output voltage will rise.

3.5 Regulated Voltage Output (V_{OUT})

Connect the output load to V_{OUT} of the LDO. Also connect one side of the LDO output de-coupling capacitor as close as possible to the V_{OUT} pin.
4.0 DETAILED DESCRIPTION

The TC2014, TC2015 and TC2185 are precision fixed-output voltage regulators (if an adjustable version is needed, see the TC1070, TC1071 and TC1187 (DS21353) data sheet). Unlike bipolar regulators, the TC2014, TC2015 and TC2185 supply current does not increase with load current. In addition, the LDO’s output voltage is stable using 1 µF of ceramic or tantalum capacitance over the entire specified input voltage range and output current range.

Figure 4-1 shows a typical application circuit. The regulator is enabled anytime the shutdown input (SHDN) is at or above VIH, and disabled (shutdown) when SHDN is at or below VIL. SHDN may be controlled by a CMOS logic gate or I/O port of a microcontroller. If the SHDN input is not required, it should be connected directly to the input supply. While in shutdown, the supply current decreases to 0.05 µA (typical) and VOUT falls to zero volts.

4.1 Bypass Input

A 0.01 µF ceramic capacitor, connected from the Bypass input to ground, reduces noise present on the internal reference, which, in turn, significantly reduces output noise. If output noise is not a concern, this input may be left unconnected. Larger capacitor values may be used, but the result is a longer time period to rated output voltage when power is initially applied.

4.2 Output Capacitor

A 1 µF (min) capacitor from VOUT to ground is required. The output capacitor should have an Effective Series Resistance (ESR) of 0.01Ω to 5Ω for VOUT ≥ 2.5V, and 0.05Ω to 5Ω for VOUT < 2.5V. Ceramic, tantalum or aluminum electrolytic capacitors can be used. When using ceramic capacitors, X5R and X7R dielectric material are recommended due to their stable tolerance over temperature. However, other dielectrics can be used as long as the minimum output capacitance is maintained.

4.3 Input Capacitor

A 1 µF capacitor should be connected from VIN to GND if there is more than 10 inches of wire between the regulator and this AC filter capacitor, or if a battery is used as the power source. Aluminum electrolytic or tantalum capacitors can be used (since many aluminum electrolytic capacitors freeze at approximately -30°C, solid tantalum are recommended for applications operating below -25°C). When operating from sources other than batteries, supply-noise rejection and transient response can be improved by increasing the value of the input and output capacitors and employing passive filtering techniques.
5.0 THERMAL CONSIDERATIONS

5.1 Thermal Shutdown

Integrated thermal protection circuitry shuts the regulator off when the die temperature exceeds approximately 160°C. The regulator remains off until the die temperature cools to approximately 150°C.

5.2 Power Dissipation

The amount of power the regulator dissipates is primarily a function of input voltage, output voltage and output current.

The following equation is used to calculate worst-case power dissipation.

**EQUATION 5-1:**

\[ P_D = (V_{INMAX} - V_{OUTMIN})I_{LMAX} \]

Where:

- \( P_D \) = Worst-case actual power dissipation
- \( V_{INMAX} \) = Maximum voltage on \( V_{IN} \)
- \( V_{OUTMIN} \) = Minimum regulator output voltage
- \( I_{LMAX} \) = Maximum output (load) current

The maximum allowable power dissipation (\( P_{DMAX} \)) is a function of the maximum ambient temperature (\( T_{AMAX} \)), the maximum allowable die temperature (\( T_{JMAX} \)) (+125°C) and the thermal resistance from junction-to-air (\( \theta_{JA} \)). The 5-Pin SOT-23A package has a \( \theta_{JA} \) of approximately 220°C/Watt when mounted on a typical two-layer FR4 dielectric copper-clad PC board.

**EQUATION 5-2:**

\[ P_{DMAX} = \frac{T_{JMAX} - T_{AMAX}}{\theta_{JA}} \]

Where all terms are previously defined.

The \( P_D \) equation can be used in conjunction with the \( P_{DMAX} \) equation to ensure that regulator thermal operation is within limits. For example:

Given:

- \( V_{INMAX} = 3.0V +10\% \)
- \( V_{OUTMIN} = 2.7V - 2.5\% \)
- \( I_{LOADMAX} = 40 \text{ mA} \)
- \( T_{JMAX} = +125^\circ\text{C} \)
- \( T_{AMAX} = +55^\circ\text{C} \)

Find:

1. Actual power dissipation
2. Maximum allowable dissipation

Actual power dissipation:

\[ P_{D} = (V_{INMAX} - V_{OUTMIN})I_{LMAX} \]

\[ = [(3.0 \times 1.1) - (2.7 \times 0.975)] \times 40 \times 10^{-3} \]

\[ = 26.7 \text{ mW} \]

Maximum allowable power dissipation:

\[ P_{DMAX} = \frac{T_{JMAX} - T_{AMAX}}{\theta_{JA}} \]

\[ = \frac{125 - 55}{220} \]

\[ = 318 \text{ mW} \]

In this example, the TC2014 dissipates a maximum of only 26.7 mW; far below the allowable limit of 318 mW. In a similar manner, the \( P_D \) and \( P_{DMAX} \) equations can be used to calculate maximum current and/or input voltage limits.

5.3 Layout Considerations

The primary path of heat conduction out of the package is via the package leads. Therefore, layouts having a ground plane, wide traces at the pads and wide power supply bus lines combine to lower \( \theta_{JA} \) and, therefore, increase the maximum allowable power dissipation limit.
6.0 PACKAGING INFORMATION

6.1 Package Marking Information

TABLE 6-1: PART NUMBER CODE AND TEMPERATURE RANGE

<table>
<thead>
<tr>
<th>(V)</th>
<th>TC2014</th>
<th>TC2015</th>
<th>TC2185</th>
</tr>
</thead>
<tbody>
<tr>
<td>1.8</td>
<td>PA</td>
<td>RA</td>
<td>UA</td>
</tr>
<tr>
<td>2.5</td>
<td>PB</td>
<td>RB</td>
<td>UB</td>
</tr>
<tr>
<td>2.6</td>
<td>PH</td>
<td>RH</td>
<td>UH</td>
</tr>
<tr>
<td>2.7</td>
<td>PC</td>
<td>RC</td>
<td>UC</td>
</tr>
<tr>
<td>2.8</td>
<td>PD</td>
<td>RD</td>
<td>UD</td>
</tr>
<tr>
<td>2.85</td>
<td>PE</td>
<td>RE</td>
<td>UE</td>
</tr>
<tr>
<td>3.0</td>
<td>PF</td>
<td>RF</td>
<td>UF</td>
</tr>
<tr>
<td>3.3</td>
<td>PG</td>
<td>RG</td>
<td>UG</td>
</tr>
<tr>
<td>5.0</td>
<td>PJ</td>
<td>RJ</td>
<td>UJ</td>
</tr>
</tbody>
</table>

① & ② represents part number code + temperature range and voltage
③ represents year and 2-month period code
④ represents lot ID number

6.2 Taping Form

Component Taping Orientation for 5-Pin SOT-23A (EIAJ SC-74A) Devices

Carrier Tape, Number of Components Per Reel and Reel Size:

<table>
<thead>
<tr>
<th>Package</th>
<th>Carrier Width (W)</th>
<th>Pitch (P)</th>
<th>Part Per Full Reel</th>
<th>Reel Size</th>
</tr>
</thead>
<tbody>
<tr>
<td>5-Pin SOT-23A</td>
<td>8 mm</td>
<td>4 mm</td>
<td>3000</td>
<td>7 in.</td>
</tr>
</tbody>
</table>
5-Lead Plastic Small Outline Transistor (OT) (SOT23)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

<table>
<thead>
<tr>
<th>Units</th>
<th>INCHES*</th>
<th>MILLIMETERS</th>
</tr>
</thead>
<tbody>
<tr>
<td>Dimension Limits</td>
<td>MIN</td>
<td>NOM</td>
</tr>
<tr>
<td>Number of Pins</td>
<td>n</td>
<td>5</td>
</tr>
<tr>
<td>Pitch</td>
<td>P</td>
<td>0.038</td>
</tr>
<tr>
<td>Outside lead pitch (basic)</td>
<td>p1</td>
<td>0.075</td>
</tr>
<tr>
<td>Overall Height</td>
<td>A</td>
<td>0.035</td>
</tr>
<tr>
<td>Molded Package Thickness</td>
<td>A2</td>
<td>0.035</td>
</tr>
<tr>
<td>Standoff</td>
<td>A1</td>
<td>0.000</td>
</tr>
<tr>
<td>Overall Width</td>
<td>E</td>
<td>0.102</td>
</tr>
<tr>
<td>Molded Package Width</td>
<td>E1</td>
<td>0.059</td>
</tr>
<tr>
<td>Overall Length</td>
<td>D</td>
<td>0.110</td>
</tr>
<tr>
<td>Foot Length</td>
<td>L</td>
<td>0.014</td>
</tr>
<tr>
<td>Foot Angle</td>
<td>T</td>
<td>0</td>
</tr>
<tr>
<td>Lead Thickness</td>
<td>c</td>
<td>0.004</td>
</tr>
<tr>
<td>Lead Width</td>
<td>B</td>
<td>0.014</td>
</tr>
<tr>
<td>Mold Draft Angle Top</td>
<td>a</td>
<td>0</td>
</tr>
<tr>
<td>Mold Draft Angle Bottom</td>
<td>b</td>
<td>0</td>
</tr>
</tbody>
</table>

* Controlling Parameter

**Notes:**
- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.005" (0.127mm) per side.
- EIAJ Equivalent: SC-74A
- Drawing No. C04-091

Revised 09-12-05
APPENDIX A: REVISION HISTORY

Revision F (December 2012)
• Added a note to each package outline drawing.

Revision E (May 2006)
• Page 1: Added overtemperature to bullet for overcurrent protection in features and general description verbiage.
• Page 3: Added Thermal Shutdown die Temperature to electrical characteristics table.
• Page 3: Added Thermal Characteristics Table.
• Page 5: Added new section 5.1 and new verbiage.
• Page 13: Updated package outline drawing.

Revision D (November 2004)
• Page 2: Changed Absolute Maximum Ratings from 6.5V to 7.0V.
• Packaging Information: Added package codes for 2.6V and 5.0V options.
• Product Identification System: Added 2.6V and 5.0V to Output voltage options.

Revision C (December 2002)
• Numerous changes

Revision B (May 2002)
• Numerous changes

Revision A (May 2001)
• Original Release of this Document.
To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<table>
<thead>
<tr>
<th>PART NO.</th>
<th>Device</th>
<th>Output Voltage</th>
<th>Temperature Range</th>
<th>Package</th>
</tr>
</thead>
<tbody>
<tr>
<td>TC2014</td>
<td>50 mA LDO with Shutdown and $V_{REF}$ Bypass</td>
<td>XX = 1.8V</td>
<td>-40°C to +125°C</td>
<td>Plastic Small Outline Transistor (SOT-23), 5-lead, Tape and Reel</td>
</tr>
<tr>
<td>TC2015</td>
<td>100 mA LDO with Shutdown and $V_{REF}$ Bypass</td>
<td>XX = 2.5V</td>
<td>XX = 2.6V</td>
<td>Tape and Reel</td>
</tr>
<tr>
<td>TC2185</td>
<td>150 mA LDO with Shutdown and $V_{REF}$ Bypass</td>
<td>XX = 2.7V</td>
<td>XX = 2.8V</td>
<td>Tape and Reel</td>
</tr>
<tr>
<td></td>
<td></td>
<td>XX = 2.85V</td>
<td>XX = 3.0V</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>XX = 3.3V</td>
<td>XX = 5.0V</td>
<td></td>
</tr>
</tbody>
</table>

Examples:

a) TC2014-1.8VCTTR: 5LD SOT-23-A, 1.8V, Tape and Reel.
b) TC2014-2.85VCTTR: 5LD SOT-23-A, 2.85V, Tape and Reel.
c) TC2014-3.3VCTTR: 5LD SOT-23-A, 3.3V, Tape and Reel.

a) TC2015-1.8VCTTR: 5LD SOT-23-A, 1.8V, Tape and Reel.
b) TC2015-2.85VCTTR: 5LD SOT-23-A, 2.85V, Tape and Reel.
c) TC2015-3.0VCTTR: 5LD SOT-23-A, 3.0V, Tape and Reel.

a) TC2185-1.8VCTTR: 5LD SOT-23-A, 1.8V, Tape and Reel.
b) TC2185-2.8VCTTR: 5LD SOT-23-A, 2.8V, Tape and Reel.
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